

PIC18F87K22 Family Data Sheet

64/80-Pin, High-Performance, 1-Mbit Enhanced Flash Microcontrollers with 12-Bit A/D and nanoWatt XLP Technology

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64/80-Pin, High-Performance, 1-Mbit Enhanced Flash MCUs with 12-Bit A/D and nanoWatt XLP Technology

Low-Power Features:

- Power-Managed modes:
 - Run: CPU on, peripherals on
 - Idle: CPU off, peripherals on
 - Sleep: CPU off, peripherals off
- Two-Speed Oscillator Start-up
- Fail-Safe Clock Monitor
- Power-Saving Peripheral Module Disable (PMD)
- Ultra Low-Power Wake-up
- Fast Wake-up, 1 μs Typical
- · Low-Power WDT, 300 nA Typical
- Ultra Low 50 nA Input Leakage
- Run mode Currents Down to 5.5 μA, Typical
- Idle mode Currents Down to 1.7 μA Typical
- Sleep mode Currents Down to Very Low 20 nA, Typical
- RTCC Current Downs to Very Low 700 nA, Typical

Special Microcontroller Features:

- Operating Voltage Range: 1.8V to 5.5V
- On-Chip 3.3V Regulator
- · Operating Speed up to 64 MHz
- Up to 128 Kbytes On-Chip Flash Program Memory
- Data EEPROM of 1,024 Bytes
- 4K x 8 General Purpose Registers (SRAM)
- 10,000 Erase/Write Cycle Flash Program Memory, Minimum
- 1,000,000 Erase/write Cycle Data EEPROM Memory, Typical
- Flash Retention: 40 Years, Minimum
- Three Internal Oscillators: LF-INTRC (31 kHz), MF-INTOSC (500 kHz) and HF-INTOSC (16 MHz)
- Self-Programmable under Software Control
- · Priority Levels for Interrupts
- 8 x 8 Single-Cycle Hardware Multiplier
- Extended Watchdog Timer (WDT):
 - Programmable period from 4 ms to 4,194s (about 70 minutes)
- In-Circuit Serial Programming[™] (ICSP[™]) via Two Pins
- In-Circuit Debug via Two Pins
- Programmable:
 - BOR
 - LVD

	Prog	ram Memory	ata Memory		10 54	CCP/	MSSP			RT	tors	s t	Bus			
Device	Flash (bytes)	# Single-Word Instructions	SRAM (bytes)	EEPROM (bytes)	I/O	12-Bit A/D (ch)	ECCP (PWM)		SPI	Master I ² C™	EUSAR	Comparators	Timers 8/16-Bit	External I	CTMU	RTCC
PIC18F65K22	32K	16,383	2K	1K	53	16	5/3	2	Y	Y	2	3	4/4	Ν	Y	Y
PIC18F66K22	64K	32,768	4K	1K	53	16	7/3	2	Y	Y	2	3	6/5	Ν	Y	Y
PIC18F67K22	128K	65,536	4K	1K	53	16	7/3	2	Y	Y	2	3	6/5	Ν	Y	Y
PIC18F85K22	32K	16,383	2K	1K	69	24	5/3	2	Y	Y	2	3	4/4	Υ	Y	Y
PIC18F86K22	64K	32,768	4K	1K	69	24	7/3	2	Y	Y	2	3	6/5	Υ	Y	Υ
PIC18F87K22	128K	65,536	4K	1K	69	24	7/3	2	Y	Y	2	3	6/5	Υ	Y	Υ

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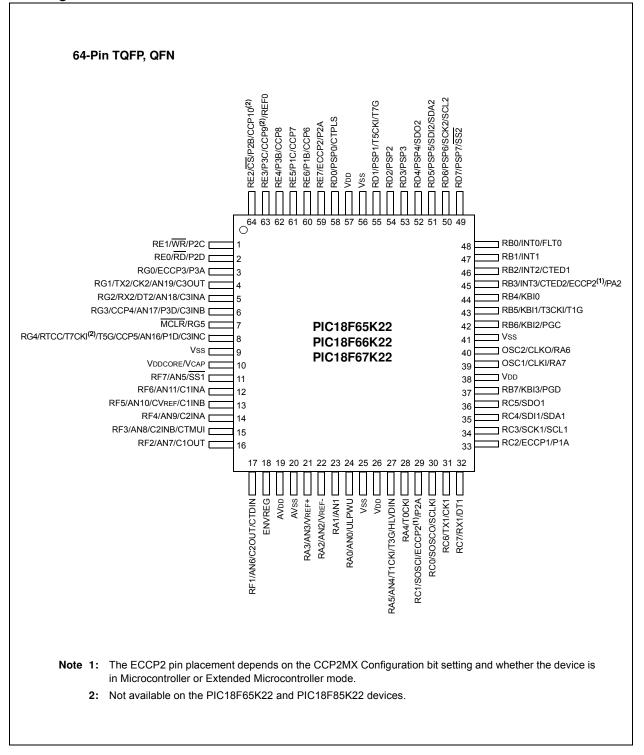
Peripheral Highlights:

- Up to Ten CCP/ECCP modules:
 - Up to seven Capture/Compare/PWM (CCP) modules
 - Three Enhanced Capture/Compare/PWM (ECCP) modules
- Up to Eleven 8/16-Bit Timer/Counter modules:
 - Timer0 8/16-bit timer/counter with 8-bit programmable prescaler
 - Timer1,3 16-bit timer/counter
 - Timer2,4,6,8 8-bit timer/counter
 - Timer5,7 16-bit timer/counter for 64k and 128k parts
 - Timer10,12 8-bit timer/counter for 64k and 128k parts
- Three Analog Comparators
- Configurable Reference Clock Output
- Hardware Real-Time Clock and Calendar (RTCC) module with Clock, Calendar and Alarm Functions

- Charge Time Measurement Unit (CTMU):
 - Capacitance measurement for mTouch™ sensing solution
 - Time measurement with 1 ns typical resolution
 - Integrated temperature sensor
- High-Current Sink/Source 25 mA/25 mA (PORTB and PORTC)
- · Up to Four External Interrupts
- Two Master Synchronous Serial Port (MSSP) modules:
 - 3/4-wire SPI (supports all four SPI modes)
 - I²C[™] Master and Slave modes
- Two Enhanced Addressable USART modules:
 - LIN/J2602 support
 - Auto-Baud Detect (ABD)
- 12-Bit A/D Converter with up to 24 Channels:
 - Auto-acquisition and Sleep operationDifferential input mode of operation
- Integrated Voltage Reference

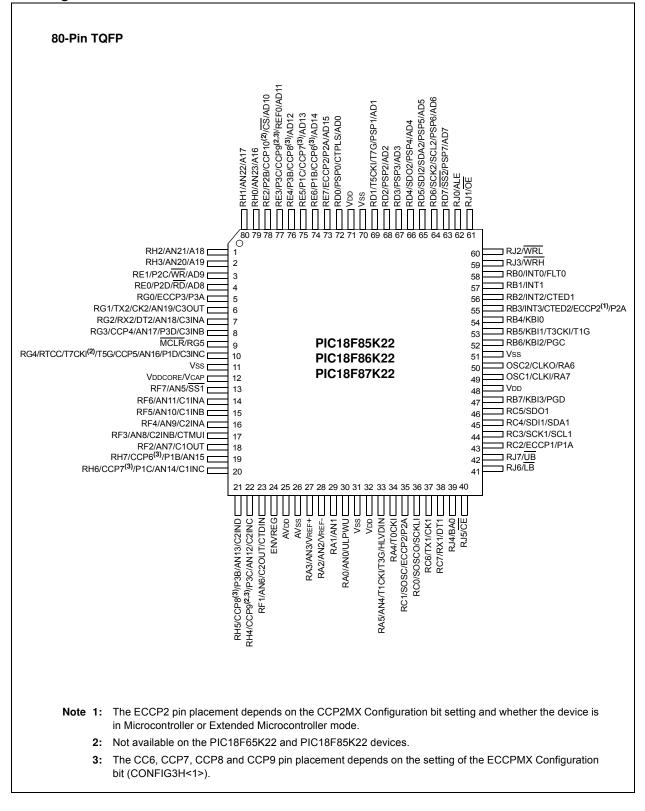
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Pin Diagrams – PIC18F6XK22



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Pin Diagrams – PIC18F8XK22



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1.0 DEVICE OVERVIEW

This document contains device-specific information for the following devices:

- PIC18F65K22 PIC18F85K22
- PIC18F66K22 PIC18F86K22
- PIC18F67K22 PIC18F87K22

This family combines the traditional advantages of all PIC18 microcontrollers – namely, high computational performance and a rich feature set – with an extremely competitive price point. These features make the PIC18F87K22 family a logical choice for many high-performance applications where price is a primary consideration.

1.1 Core Features

1.1.1 nanoWatt TECHNOLOGY

All of the devices in the PIC18F87K22 family incorporate a range of features that can significantly reduce power consumption during operation. Key items include:

- Alternate Run Modes: By clocking the controller from the Timer1 source or the Internal RC oscillator, power consumption during code execution can be reduced.
- **Multiple Idle Modes:** The controller can also run with its CPU core disabled but the peripherals still active. In these states, power consumption can be reduced even further.
- **On-the-Fly Mode Switching:** The power-managed modes are invoked by user code during operation, allowing the user to incorporate power-saving ideas into their application's software design.
- **nanoWatt XLP:** An extra low-power Sleep, BOR, RTCC and Watchdog Timer

1.1.2 OSCILLATOR OPTIONS AND FEATURES

All of the devices in the PIC18F87K22 family offer different oscillator options, allowing users a range of choices in developing application hardware. These include:

- External Resistor/Capacitor (RC); RA6 available
- External Resistor/Capacitor with Clock Out (RCIO)
- Three External Clock modes:
 - External Clock (EC); RA6 available
 - External Clock with Clock Out (ECIO)External Crystal (XT, HS, LP)
- A Phase Lock Loop (PLL) frequency multiplier, available to the External Oscillator modes, which allows clock speeds of up to 64 MHz. PLL can also be used with the internal oscillator.

- An internal oscillator block that provides a 16 MHz clock (±2% accuracy) and an INTRC source (approximately 31 kHz, stable over temperature and VDD)
 - Operates as HF-INTOSC or MF-INTOSC when block selected for 16 MHz or 500 kHz
 - Frees the two oscillator pins for use as additional general purpose I/O

The internal oscillator block provides a stable reference source that gives the family additional features for robust operation:

- Fail-Safe Clock Monitor: This option constantly monitors the main clock source against a reference signal provided by the internal oscillator. If a clock failure occurs, the controller is switched to the internal oscillator, allowing for continued low-speed operation or a safe application shutdown.
- **Two-Speed Start-up:** This option allows the internal oscillator to serve as the clock source from Power-on Reset, or wake-up from Sleep mode, until the primary clock source is available.

1.1.3 MEMORY OPTIONS

The PIC18F87K22 family provides ample room for application code, from 32 Kbytes to 128 Kbytes of code space. The Flash cells for program memory are rated to last up to 10,000 erase/write cycles. Data retention without refresh is conservatively estimated to be greater than 40 years.

The Flash program memory is readable and writable. During normal operation, the PIC18F87K22 family also provides plenty of room for dynamic application data with up to 3,862 bytes of data RAM.

1.1.4 EXTERNAL MEMORY BUS

Should 128 Kbytes of memory be inadequate for an application, the 80-pin members of the PIC18F87K22 family have an External Memory Bus (EMB) enabling the controller's internal Program Counter to address a memory space of up to 2 Mbytes. This is a level of data access that few 8-bit devices can claim and enables:

- Using combinations of on-chip and external memory of up to 2 Mbytes
- Using external Flash memory for reprogrammable application code or large data tables
- Using external RAM devices for storing large amounts of variable data

1.1.5 EXTENDED INSTRUCTION SET

The PIC18F87K22 family implements the optional extension to the PIC18 instruction set, adding eight new instructions and an Indexed Addressing mode. Enabled as a device configuration option, the extension has been specifically designed to optimize re-entrant application code originally developed in high-level languages, such as 'C'.

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1.1.6 EASY MIGRATION

All devices share the same rich set of peripherals except that the devices with 32 Kbytes of program memory (PIC18F65K22 and PIC18F85K22) have two less CCPs and three less timers. This provides a smooth migration path within the device family as applications evolve and grow.

The consistent pinout scheme, used throughout the entire family, also aids in migrating to the next larger device. This is true when moving between the 64-pin members, between the 80-pin members, or even jumping from 64-pin to 80-pin devices.

All of the devices in the family share the same rich set of peripherals, except for those with 32 Kbytes of program memory (PIC18F65K22 and PIC18F85K22). Those devices have two less CCPs and three less timers.

The PIC18F87K22 family is also largely pin compatible with other PIC18 families, such as the PIC18F8720 and PIC18F8722 and the PIC18F85J11. This allows a new dimension to the evolution of applications, allowing developers to select different price points within Microchip's PIC18 portfolio, while maintaining a similar feature set.

1.2 Other Special Features

- Communications: The PIC18F87K22 family incorporates a range of serial communication peripherals, including two Enhanced USARTs (EUSART) that support LIN/J2602, and two Master SSP modules, capable of both SPI and I²C[™] (Master and Slave) modes of operation.
- CCP Modules: PIC18F87K22 family devices incorporate up to seven Capture/Compare/PWM (CCP) modules. Up to six different time bases can be used to perform several different operations at once.
- ECCP Modules: The PIC18F87K22 family has three Enhanced CCP (ECCP) modules to maximize flexibility in control applications:
 - Up to eight different time bases for performing several different operations at once
 - Up to four PWM outputs for each module, for a total of 12 PWMs
 - Other beneficial features, such as polarity selection, programmable dead time, auto-shutdown and restart, and Half-Bridge and Full-Bridge Output modes

- 12-Bit A/D Converter: The PIC18F87K22 family has differential ADC. It incorporates programmable acquisition time, allowing for a channel to be selected and a conversion to be initiated without waiting for a sampling period, and thus, reducing code overhead.
- Charge Time Measurement Unit (CTMU): The CTMU is a flexible analog module that provides accurate differential time measurement between pulse sources, as well as asynchronous pulse generation.
- Together with other on-chip analog modules, the CTMU can precisely measure time, measure capacitance or relative changes in capacitance, or generate output pulses that are independent of the system clock.
- LP Watchdog Timer (WDT): This enhanced version incorporates a 22-bit prescaler, allowing an extended time-out range that is stable across operating voltage and temperature. See Section 31.0 "Electrical Characteristics" for time-out periods.
- Real-Time Clock and Calendar Module (RTCC): The RTCC module is intended for applications requiring that accurate time be maintained for extended periods of time with minimum to no intervention from the CPU.
- The module is a 100-year clock and calendar with automatic leap year detection. The range of the clock is from 00:00:00 (midnight) on January 1, 2000 to 23:59:59 on December 31, 2099.

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1.3 Details on Individual Family Members

Devices in the PIC18F87K22 family are available in 64-pin and 80-pin packages. Block diagrams for the two groups are shown in Figure 1-1 and Figure 1-2.

The devices are differentiated from each other in these ways:

- Flash Program Memory:
 - PIC18FX5K22 (PIC18F65K22 and PIC18F85K22) – 32 Kbytes
 - PIC18FX6K22 (PIC18F66K22 and PIC18F86K22) – 64 Kbytes
 - PIC18FX7K22 (PIC18F67K22 and PIC18F87K22) 128 Kbytes
- Data RAM:
 - All devices except PIC18FX5K22 4 Kbytes
 - PIC18FX5K22 2 Kbytes
- I/O Ports:
 - PIC18F6XK22 (64-pin devices) 7 bidirectional ports
 - PIC18F8XK22 (80-pin devices) 9 bidirectional ports

- CCP modules:
 - PIC18FX5K22 (PIC18F65K22 and PIC18F85K22) – 5 CCP modules
 - PIC18FX6K22 and PIC18FX7K22 (PIC18F66K22, PIC18F86K22, PIC18F67K22, and PIC18F87K22) – 7 CCP modules
- · Timer modules:
 - PIC18FX5K22 (PIC18F65K22 and PIC18F85K22) – Four 8-bit timer/counters and four 16-bit timer/counters
 - PIC18FX6K22 and PIC18FX7K22 (PIC18F66K22, PIC18F86K22, PIC18F67K22, and PIC18F87K22) – Six 8-bit timer/counters and five 16-bit timer/counters
- A/D Channels:
 - PIC18F6XK22 (64-pin devices) 24 channels
 - PIC18F8XK22 (80-pin devices) 16 channels

All other features for devices in this family are identical. These are summarized in Table 1-1 and Table 1-2.

The pinouts for all devices are listed in Table 1-3 and Table 1-4.

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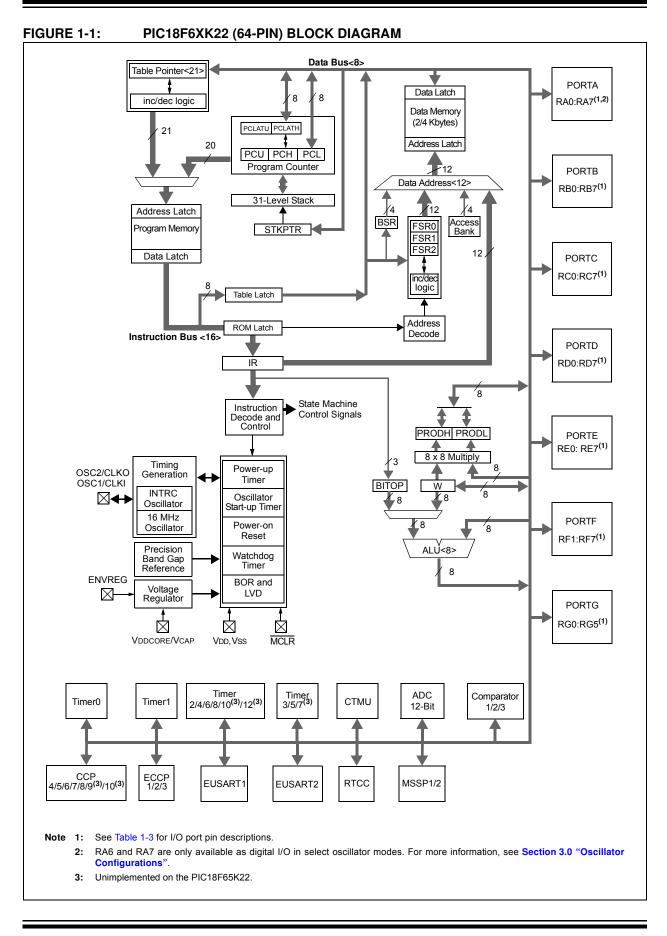
TABLE 1-1: DEVICE FEATURES FOR THE PIC18F6XK22 (64-PIN DEVICES)

Features	PIC18F65K22	PIC18F66K22	PIC18F67K22			
Operating Frequency		DC – 64 MHz	•			
Program Memory (Bytes)	32K	64K	128K			
Program Memory (Instructions)	16,384	32,768	65,536			
Data Memory (Bytes)	2K	4K	4K			
Interrupt Sources	42	4	48			
I/O Ports	Ports A, B, C, D, E, F, G					
Parallel Communications	Parallel Slave Port (PSP)					
Timers	8 11					
Comparators	3					
СТМИ	Yes					
RTCC		Yes				
Capture/Compare/PWM (CCP) Modules	5	7	7			
Enhanced CCP (ECCP) Modules		3				
Serial Communications	Two MSSPs and two Enhanced USARTs (EUSART)					
12-Bit Analog-to-Digital Module	16 Input Channels					
Resets (and Delays)	POR, BOR, RESET Instruction, Stack Full, Stack Underflow, MCLR, WDT (PWRT, OST)					
Instruction Set	75 Instructions, 83 with Extended Instruction Set Enabled					
Packages	64-Pin QFN, 64-Pin TQFP					

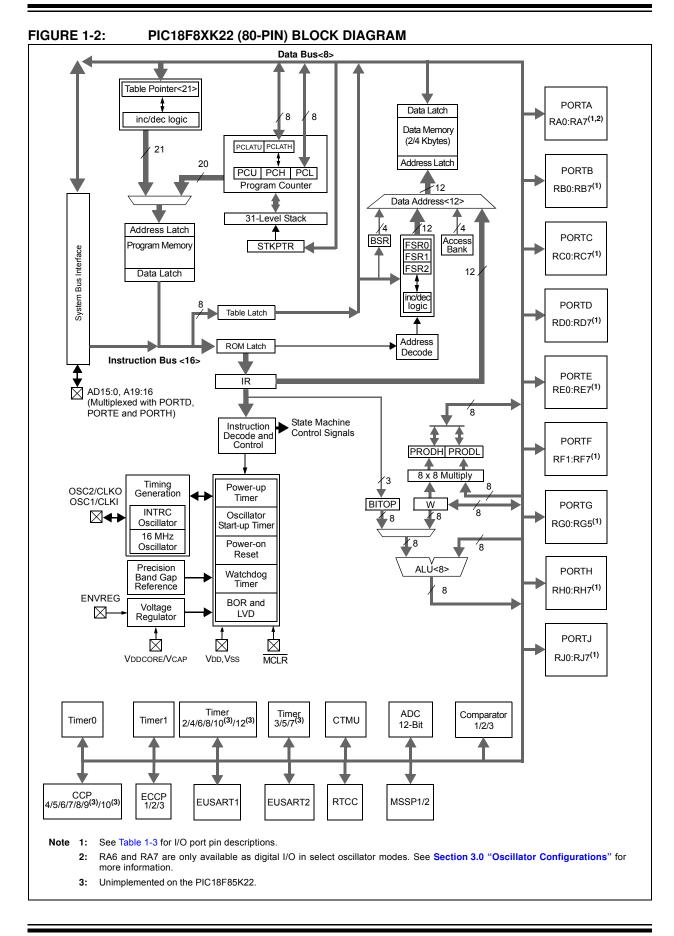
TABLE 1-2: DEVICE FEATURES FOR THE PIC18F8XK22 (80-PIN DEVICES)

Features	PIC18F85K22	PIC18F86K22	PIC18F87K22				
Operating Frequency	DC – 64 MHz						
Dragram Maman (Dutas)	32K	64K	128K				
Program Memory (Bytes)	(Up to	2 Mbytes with Extended M	lemory)				
Program Memory (Instructions)	16,384	32,768	65,536				
Data Memory (Bytes)	2K	4K	4K				
Interrupt Sources	42	4	48				
I/O Ports	l	Ports A, B, C, D, E, F, G, H	, J				
Parallel Communications	Parallel Slave Port (PSP)						
Timers	8 11						
Comparators	3						
СТМИ		Yes					
RTCC		Yes					
Capture/Compare/PWM (CCP) Modules	5	7	7				
Enhanced CCP (ECCP) Modules		3					
Serial Communications	Two MSSP	s and 2 Enhanced USART	ſs (EUSART)				
12-Bit Analog-to-Digital Module	24 Input Channels						
Resets (and Delays)	POR, BOR, RESET Ins	POR, BOR, RESET Instruction, Stack Full, Stack Underflow, MCLR, WDT					
Instruction Set	75 Instructions, 83 with Extended Instruction Set Enabled						
Packages		80-Pin TQFP					

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Din Nomo	Pin Number	Pin	Buffer	Description				
Pin Name	QFN/TQFP Type		Туре	Description				
	7			Master Clear (input) or programming voltage (input).				
MCLR/RG5								
MCLR		I	ST	This pin is an active-low Reset to the device.				
RG5		I	ST	General purpose, input only pin.				
OSC1/CLKI/RA7	39			Oscillator crystal or external clock input.				
OSC1		1	CMOS	Oscillator crystal input.				
CLKI		I	CMOS	External clock source input. Always associated				
				with pin function, OSC1. (See related OSC1/CLKI,				
				OSC2/CLKO pins.)				
RA7		I/O	TTL	General purpose I/O pin.				
OSC2/CLKO/RA6	40			Oscillator crystal or clock output.				
OSC2		0	—	Oscillator crystal output. Connects to crystal or				
				resonator in Crystal Oscillator mode.				
CLKO		0	—	In certain oscillator modes, OSC2 pin outputs CLKO,				
				which has 1/4 the frequency of OSC1 and denotes the				
				instruction cycle rate.				
RA6		I/O	TTL	General purpose I/O pin.				
Legend: TTL = TTL cor	npatible input			CMOS = CMOS compatible input or output				
ST = Schmitt Trigger input with CMOS levels Analog = Analog input								

TABLE 1-3: PIC18F6XK22 PINOUT I/O DESCRIPTIONS

ST 1

= Input Ρ

= Power $I^2C = I^2C^{TM}/SMBus$ OD

0

= Output

= Open-Drain (no P diode to VDD)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

- 2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.
- 3: Not available on PIC18F65K22 and PIC18F85K22 devices.
- 4: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

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Pin Name	Pin Number	Pin	Buffer	Description		
Pin Name	QFN/TQFP	Туре	Туре	Description		
				PORTA is a bidirectional I/O port.		
RA0/AN0/ULPWU RA0 AN0 ULPWU	24	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 0. Ultra Low-Power Wake-up input.		
RA1/AN1 RA1 AN1	23	I/O I	TTL Analog	Digital I/O. Analog Input 1.		
RA2/AN2/VREF- RA2 AN2 VREF-	22	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 2. A/D reference voltage (low) input.		
RA3/AN3/VREF+ RA3 AN3 VREF+	21	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 3. A/D reference voltage (high) input.		
RA4/T0CKI RA4 T0CKI	28	I/O I	ST ST	Digital I/O. Timer0 external clock input.		
RA5/AN4/T1CKI/T3G/ HLVDIN RA5 AN4 T1CKI T3G HLVDIN	27	I/O I I I	TTL Analog ST ST Analog	Digital I/O. Analog Input 4. Timer1 clock input. Timer3 external clock gate input. High/Low-Voltage Detect input.		
RA6				See the OSC2/CLKO/RA6 pin.		
RA7				See the OSC1/CLKI/RA7 pin.		
Legend: TTL = TTL cor ST = Schmitt I = Input P = Power $I^2C = I^2C^{TM}/S$	Trigger input w	vith CM	1OS levels	CMOS= CMOS compatible input or outputAnalog= Analog inputO= OutputOD= Open-Drain (no P diode to VDD)		

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

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TABLE 1-3: PIC18F	Pin Number			IPTIONS (CONTINUED)
Pin Name	QFN/TQFP	Pin Type	Buffer Type	Description
				PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.
RB0/INT0/FLTO RB0 INT0 FLT0	48	I/O I I	TTL ST ST	Digital I/O. External Interrupt 0. Enhanced PWM Fault input for ECCP1/2/3.
RB1/INT1 RB1 INT1	47	I/O I	TTL ST	Digital I/O. External Interrupt 1.
RB2/INT2/CTED1 RB2 INT2 CTED1	46	I/O I I	TTL ST ST	Digital I/O. External Interrupt 2. CTMU Edge 1 input.
RB3/INT3/CTED2/ ECCP2/P2A RB3 INT3 CTED2 ECCP2 P2A	45	I/O I I/O 0	TTL ST ST ST	Digital I/O. External Interrupt 3. CTMU Edge 2 input. Capture 2 input/Compare 2 output/PWM2. Enhanced PWM2 Output A.
RB4/KBI0 RB4 KBI0	44	I/O I	TTL TTL	Digital I/O. Interrupt-on-change pin.
RB5/KBI1/T3CKI/T1G RB5 KBI1 T3CKI T1G	43	I/O 	TTL TTL ST ST	Digital I/O. Interrupt-on-change pin. Timer3 clock input. Timer1 external clock gate input.
RB6/KBI2/PGC RB6 KBI2 PGC	42	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP™ programming clock pin.
RB7/KBI3/PGD RB7 KBI3 PGD	37	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP programming data pin.
Legend: TTL = TTL com ST = Schmitt [*] I = Input P = Power $I^2C = I^2C^{TM}/SN$	Trigger input w	∕ith C№	1OS levels	CMOS= CMOS compatible input or outputAnalog= Analog inputO= OutputOD= Open-Drain (no P diode to VDD)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

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Pin Name	Pin Number	Pin	Buffer	Description
Pin Name	QFN/TQFP	Туре	Туре	Description
				PORTC is a bidirectional I/O port.
RC0/SOSCO/SCLKI RC0 SOSCO SCLKI	30	I/O O I	ST — ST	Digital I/O. SOSC oscillator output. Digital SOSC input.
RC1/SOSCI/ECCP2/P2A RC1 SOSCI ECCP2 ⁽¹⁾ P2A	29	I/O I I/O O	ST CMOS ST	Digital I/O. SOSC oscillator input. Capture 2 input/Compare 2 output/PWM2 output. Enhanced PWM2 Output A.
RC2/ECCP1/P1A RC2 ECCP1 P1A	33	I/O I/O O	ST ST	Digital I/O. Capture 1 input/Compare 1 output/PWM1 output. Enhanced PWM1 Output A.
RC3/SCK1/SCL1 RC3 SCK1 SCL1 ⁽⁴⁾	34	I/O I/O I/O	ST ST I ² C	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I ² C™ mode.
RC4/SDI1/SDA1 RC4 SDI1 SDA1 ⁽⁴⁾	35	I/O I I/O	ST ST I ² C	Digital I/O. SPI data in. I ² C data I/O.
RC5/SDO1 RC5 SDO1	36	I/O O	ST —	Digital I/O. SPI data out.
RC6/TX1/CK1 RC6 TX1 CK1	31	I/O O I/O	ST — ST	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX1/DT1).
RC7/RX1/DT1 RC7 RX1 DT1	32	I/O I I/O	ST ST ST	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX1/CK1).
Legend: TTL = TTL con ST = Schmitt I = Input P = Power	npatible input Trigger input w	∕ith C№	10S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

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 $I^2C = I^2C^{TM}/SMBus$

Pin Name	Pin Number	Pin	Buffer	Description	
Pin Name	QFN/TQFP	Туре	Туре	Description	
				PORTD is a bidirectional I/O port.	
RD0/PSP0/CTPLS RD0 PSP0 CTPLS	58	I/O I/O O	ST TTL	Digital I/O. Parallel Slave Port data. CTMU pulse generator output.	
RD1/PSP1/T5CKI/T7G RD1 PSP1 T5CKI T7G	55	I/O I/O I	ST TTL ST ST	Digital I/O. Parallel Slave Port. Timer5 clock input. Timer7 external clock gate input.	
RD2/PSP2 RD2 PSP2	54	I/O O	ST TTL	Digital I/O. Parallel Slave Port.	
RD3/PSP3 RD3 PSP3	53	I/O I/O	ST TTL	Digital I/O. Parallel Slave Port.	
RD4/PSP4/SDO2 RD4 PSP4 SDO2	52	I/O I/O O	ST TTL	Digital I/O. Parallel Slave Port. SPI data out.	
RD5/PSP5/SDI2/SDA2 RD5 PSP5 SDI2 SDA2	51	I/O I/O I	ST TTL ST I ² C	Digital I/O. Parallel Slave Port. SPI data in. I ² C™ data I/O.	
RD6/PSP6/SCK2/SCL2 RD6 PSP6 SCK2 SCL2 ⁽⁴⁾	50	I/O I/O I/O I/O	ST TTL ST I ² C	Digital I/O. Parallel Slave Port. Synchronous serial clock. Synchronous serial clock I/O for I ² C mode.	
RD7/PSP7/SS2 RD7 PSP7 SS2	49	I/O I/O I	ST TTL TTL	Digital I/O. Parallel Slave Port. SPI slave select input.	
Legend: TTL = TTL con ST = Schmitt I = Input P = Power $I^2C = I^2C^{TM}/SI$	Trigger input w	vith CM	1OS levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)	

TABLE 1-3: PIC18F6XK22 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

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Pin Name	Pin Number	Pin	Buffer	Deparimine			
Pin Name	QFN/TQFP	Туре	Туре	Description			
				PORTE is a bidirectional I/O port.			
RE0/RD/P2D RE0 RD P2D	2	I/O I O	ST TTL	Digital I/O. Parallel Slave Port read strobe. EECP2 PWM Output D.			
RE1/WR/P2C RE1 WR P2C	1	I/O 0	ST TTL	Digital I/O. Parallel Slave Port write strobe. ECCP2 PWM Output C.			
RE2/CS/P2B/CCP10 RE2 CS P2B CCP10 ⁽³⁾	64	I/O I O I/O	ST TTL — S/T	Digital I/O. Parallel Slave Port chip select. ECCP2 PWM Output B. Capture 10 input/Compare 10 output/PWM10 output.			
RE3/P3C/CCP9/REFO RE3 P3C CCP9 ^(3,4) REFO	63	I/O O I/O O	ST S/T 	Digital I/O. ECCP3 PWM Output C. Capture 9 input/Compare 9 output/PWM9 output. Reference clock out.			
RE4/P3B/CCP8 RE4 P3B CCP8 ⁽⁴⁾	62	I/O O I/O	ST — S/T	Digital I/O. ECCP3 PWM Output B. Capture 8 input/Compare 8 output/PWM8 output.			
RE5/P1C/CCP7 RE5 P1C CCP7 ⁽⁴⁾	61	I/O O I/O	ST — S/T	Digital I/O. ECCP1 PWM Output C. Capture 7 input/Compare 7 output/PWM7 output.			
RE6/P1B/CCP6 RE6 P1B CCP6 ⁽⁴	60	I/O O I/O	ST — S/T	Digital I/O. ECCP1 PWM Output B. Capture 6 input/Compare 6 output/PWM6 output.			
RE7/ECCP2/P2A RE7 ECCP2 ⁽²⁾ P2A	59	I/O I/O O	ST ST	Digital I/O. Capture 2 input/Compare 2 output/PWM2 output. ECCP2 PWM Output A.			
Legend:TTL = TTL compatible inputCMOS= CMOS compatible input or outputST = Schmitt Trigger input with CMOS levelsAnalog= Analog inputI = InputO= OutputP = PowerOD= Open-Drain (no P diode to VDD) I^2C = $I^2C^{TM}/SMBus$ OD= Open-Drain (no P diode to VDD)							

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

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Din Nama	Pin Number	Pin		Description
Pin Name	QFN/TQFP	Туре		Description
				PORTF is a bidirectional I/O port.
RF1/AN6/C2OUT/CTDIN RF1 AN6 C2OUT CTDIN	17	I/O 0 	ST Analog — ST	Digital I/O. Analog Input 6. Comparator 2 output. CTMU pulse delay input.
RF2/AN7/C1OUT RF2 AN7 C1OUT	16	I/O I O	ST Analog —	Digital I/O. Analog Input 7. Comparator 1 output.
RF3/AN8/C2INB/CTMUI RF3 AN8 C2INB CTMUI	15	I/O 0	ST Analog Analog —	Digital I/O. Analog Input 8. Comparator 2 Input B. CTMU pulse generator charger for the C2INB comparator input.
RF4/AN9/C2INA RF4 AN9 C2INA	14	I/O I I	ST Analog Analog	Digital I/O. Analog Input 9. Comparator 2 Input A.
RF5/AN10/CVREF/C1INB RF5 AN10 CVREF C1INB	13	I/O 0 	ST Analog Analog Analog	Digital I/O. Analog Input 10. Comparator reference voltage output. Comparator 1 Input B.
RF6/AN11/C1INA RF6 AN11 C1INA	12	I/O 	ST Analog Analog	Digital I/O. Analog Input 11. Comparator 1 Input A.
RF7/AN5/SS1 RF7 <u>AN5</u> SS1	11	I/O O I	ST Analog TTL	Digital I/O. Analog Input 5. SPI1 slave select input.
Legend: TTL = TTL com ST = Schmitt I = Input P = Power $I^2C = I^2C^{TM}/SN$	Trigger input w //Bus			CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD) 2MX Configuration bit is set.

TABLE 1-3:	PIC18F6XK22 PINOUT I/O DESCRIPTIONS ((CONTINUED)

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

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Pin Name	Pin Number	Pin	Buffer	Description		
Pin Name	QFN/TQFP	Туре	Туре	Description		
				PORTG is a bidirectional I/O port.		
RG0/ECCP3/P3A RG0 ECCP3 P3A	3	I/O I/O O	ST ST	Digital I/O. Capture 3 input/Compare 3 output/PWM3 output. ECCP3 PWM Output A.		
RG1/TX2/CK2/AN19/ C3OUT RG1 TX2 CK2 AN19 C3OUT	4	I/O I/O I/O I O	ST — ST Analog —	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX2/DT2). Analog Input 19. Comparator 3 output.		
RG2/RX2/DT2/AN18/ C3INA RG2 RX2 DT2 AN18 C3INA	5	I/O I I/O I	ST ST ST Analog Analog	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX2/CK2). Analog Input 18. Comparator 3 Input A.		
RG3/CCP4/AN17/P3D/ C3INB RG3 CCP4 AN17 P3D C3INB	6	I/O I/O I O I	ST S/T Analog — Analog	Digital I/O. Capture 4 input/Compare 4 output/PWM4 output. Analog Input 18. ECCP3 PWM Output D. Comparator 3 Input B.		
RG4/RTCC/T7CKI/T5G/ CCP5/AN16/P1D/C3INC RG4 RTCC T7CKI ⁽³⁾ T5G CCP5 AN16 P1D C3INC	8	1/0 /0 0 	ST — ST ST Analog — Analog	Digital I/O. RTCC output Timer7 clock input. Timer5 external clock gate input. Capture 5 input/Compare 5 output/PWM5 output. Analog Input 16. ECCP1 PWM Output D. Comparator 3 Input C.		
RG5	7			See the MCLR/RG5 pin.		
Legend:TTL = TTL compatible inputCMOS= CMOS compatible input or outputST = Schmitt Trigger input with CMOS levelsAnalog= Analog inputI= InputO= OutputP= PowerOD= Open-Drain (no P diode to VDD) I^2C = $I^2C^{TM}/SMBus$ OD= Open-Drain (no P diode to VDD)						

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

- 2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.
- **3:** Not available on PIC18F65K22 and PIC18F85K22 devices.

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Pin Name	Pin Number	Pin Buffer Type Type	Buffer	Description			
FIII Naille	QFN/TQFP		Description				
Vss	9, 25, 41, 56	Р	_	Ground reference for logic and I/O pins.			
Vdd	26, 38, 57	Р	_	Positive supply for logic and I/O pins.			
AVss	20	Р	_	Ground reference for analog modules.			
AVDD	19	Р	_	Positive supply for analog modules.			
ENVREG	18	I	ST	Enable for on-chip voltage regulator.			
Vddcore/Vcap Vddcore Vcap	10	Core logic power or external filter capacitor connection. External filter capacitor connection (regulator enabled/disabled).					
Legend: TTL = TTL compatible input CMOS = CMOS compatible input or output ST = Schmitt Trigger input with CMOS levels Analog = Analog input							

= Input

= Power Ρ

 $I^2C = I^2C^{TM}/SMBus$

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

0

OD

= Output

= Open-Drain (no P diode to VDD)

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PIC18F8XK22 PINOUT I/O DESCRIPTIONS **TABLE 1-4:**

Pin Name	Pin Number	Pin	Buffer				
	TQFP	Туре	Туре	Description			
	9			Master Clear (input) or programming voltage (input).			
MCLR/RG5							
RG5		1	ST	This pin is an active-low Reset to the device.			
MCLR		I	ST	General purpose, input only pin.			
OSC1/CLKI/RA7	49			Oscillator crystal or external clock input.			
OSC1		1	CMOS	Oscillator crystal input.			
CLKI		1	CMOS	External clock source input. Always associated			
				with pin function, OSC1. (See related OSC1/CLKI,			
				OSC2/CLKO pins.)			
RA7		I/O	TTL	General purpose I/O pin.			
OSC2/CLKO/RA6	50			Oscillator crystal or clock output.			
OSC2		0	—	Oscillator crystal output. Connects to crystal or			
				resonator in Crystal Oscillator mode.			
CLKO		0	—	In certain oscillator modes, OSC2 pin outputs CLKO,			
				which has 1/4 the frequency of OSC1 and denotes the			
				instruction cycle rate.			
RA6 I/O TTL General purpose I/O pin.							
Legend: TTL = TTL comp	Legend: TTL = TTL compatible input CMOS = CMOS compatible input or output						
ST = Schmitt T	rigger input wit	h CMC	OS levels	Analog = Analog input			

- = Input L Ρ
- = Power
- $I^2C = I^2C^{TM}/SMBus$

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

- 2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.
- 3: Not available on PIC18F65K22 and PIC18F85K22 devices.
- 4: PSP is available only in Microcontroller mode.
- 5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

0

OD

= Output

= Open-Drain (no P diode to VDD)

Din Nama	Pin Number	Pin	Buffer	Description		
Pin Name	TQFP	Туре	Туре	Description		
				PORTA is a bidirectional I/O port.		
RA0/AN0/ULPWU RA0 AN0 ULPWU	30	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 0. Ultra Low-Power Wake-up input.		
RA1/AN1 RA1 AN1	29	I/O I	TTL Analog	Digital I/O. Analog Input 1.		
RA2/AN2/VREF- RA2 AN2 VREF-	28	I/O I I	TTL Analog Analog	Digital I/O. Analog Input 2. A/D reference voltage (low) input.		
RA3/AN3/VREF+ RA3 AN3 VREF+	27	I/O 	TTL Analog Analog	Digital I/O. Analog Input 3. A/D reference voltage (high) input.		
RA4/T0CKI RA4 T0CKI	34	I/O I	ST ST	Digital I/O. Timer0 external clock input.		
RA5/AN4/T1CKI/ T3G/HLVDIN RA5 AN4 T1CKI T3G HLVDIN	33	I/O 	TTL Analog ST ST Analog	Digital I/O. Analog Input 4. Timer1 clock input. Timer3 external clock gate input. High/Low-Voltage Detect input.		
RA6				See the OSC2/CLKO/RA6 pin.		
RA7				See the OSC1/CLKI/RA7 pin.		
Legend:TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I = InputCMOS = CMOS compatible input or output Analog = Analog inputP = Power $l^2C = l^2C^{TM}/SMBus$ OD = Open-Drain (no P diode to VDD)						

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

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Pin Name	Pin Number	Pin	Buffer	Description	
Fiir Naille	TQFP	Туре	Туре	Description	
				PORTB is a bidirectional I/O port. PORTB can be software programmed for internal weak pull-ups on all inputs.	
RB0/INT0/FLT0 RB0 INT0 FLT0	58	I/O I I	TTL ST ST	Digital I/O. External Interrupt 0. Enhanced PWM Fault input for ECCP1/2/3.	
RB1/INT1 RB1 INT1	57	I/O I	TTL ST	Digital I/O. External Interrupt 1.	
RB2/INT2/CTED1 RB2 INT2 CTED1	56	I/O I I	TTL ST ST	Digital I/O. External Interrupt 2. CTMU Edge 1 input.	
RB3/INT3/CTED2/ ECCP2/P2A RB3 INT3 CTED2 ECCP2 P2A	55	1/0 - /0 0	TTL ST ST ST ST	Digital I/O. External Interrupt 3. CTMU Edge 2 input. Capture 2 input/Compare 2 output/PWM2 output. Enhanced PWM2 Output A.	
RB4/KBI0 RB4 KBI0	54	I/O I	TTL TTL	Digital I/O. Interrupt-on-change pin.	
RB5/KBI1/T3CKI/T1G RB5 KBI1 T3CKI T1G	53	I/O 	TTL TTL ST ST	Digital I/O. Interrupt-on-change pin. Timer3 clock input. Timer1 external clock gate input.	
RB6/KBI2/PGC RB6 KBI2 PGC	52	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP™ programming clock pin.	
RB7/KBI3/PGD RB7 KBI3 PGD	47	I/O I I/O	TTL TTL ST	Digital I/O. Interrupt-on-change pin. In-Circuit Debugger and ICSP programming data pin.	
Legend:TTL = TTL compatible inputCMOS= CMOS compatible input or ouST = Schmitt Trigger input with CMOS levelsAnalog= Analog inputI= InputO= OutputP= PowerOD= Open-Drain (no P diode to VD l^2C = $l^2C^{TM}/SMBus$ O= Open-Drain (no P diode to VD					

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

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Pin Name	Pin Number	Pin	Buffer	
Pin Name	TQFP	Туре	Туре	Description
				PORTC is a bidirectional I/O port.
RC0/SOSCO/SCKLI RC0 SOSCO SCKLI	36	I/O O I	ST — ST	Digital I/O. SOSC oscillator output. Digital SOSC input.
RC1/SOSCI/ECCP2/P2A RC1 SOSCI ECCP2 ⁽¹⁾ P2A	35	I/O I I/O O	ST CMOS ST	Digital I/O. SOSC oscillator input. Capture 2 input/Compare 2 output/PWM2 output. Enhanced PWM2 Output A.
RC2/ECCP1/P1A RC2 ECCP1 P1A	43	I/O I/O O	ST ST	Digital I/O. Capture 1 input/Compare 1 output/PWM1 output. Enhanced PWM1 Output A.
RC3/SCK1/SCL1 RC3 SCK1 SCL1	44	I/O I/O I/O	ST ST I ² C	Digital I/O. Synchronous serial clock input/output for SPI mode. Synchronous serial clock input/output for I ² C™ mode.
RC4/SDI1/SDA1 RC4 SDI1 SDA1	45	I/O I I/O	ST ST I ² C	Digital I/O. SPI data in. I ² C data I/O.
RC5/SDO1 RC5 SDO1	46	I/O O	ST —	Digital I/O. SPI data out.
RC6/TX1/CK1 RC6 TX1 CK1	37	I/O O I/O	ST — ST	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX1/DT1).
RC7/RX1/DT1 RC7 RX1 DT1	38	I/O I I/O	ST ST ST	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX1/CK1).
Legend: TTL = TTL com ST = Schmitt T I = Input	patible input Frigger input wit	h CMC)S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output

P = Power

 $I^2C = I^2C^{TM}/SMBus$

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

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= Open-Drain (no P diode to VDD)

Pin Name	Pin Number	Pin	Buffer	Description
	TQFP	Туре	Туре	
				PORTD is a bidirectional I/O port.
RD0/PSP0/CTPLS/AD0 RD0 PSP0 ⁽⁴⁾ CTPLS AD0	72	I/O I/O O I/O	ST TTL ST TTL	Digital I/O Parallel Slave Port data CTMU pulse generator output External Memory Address/Data 0
RD1/T5CKI/T7G/PSP1/AD1 RD1 T5CKI T7G PSP1 ⁽⁴⁾ AD1	69	I/O /O I/O	ST ST ST TTL TTL	Digital I/O Timer5 clock input Timer7 external clock gate input Parallel Slave Port data External Memory Address/Data 1
RD2/PSP2/AD2 RD2 PSP2 ⁽⁴⁾ AD2	68	I/O I/O I/O	ST TTL TTL	Digital I/O. Parallel Slave Port data. External Memory Address/Data 2.
RD3/PSP3/AD3 RD3 PSP3 ⁽⁴⁾ AD3	67	I/O I/O I/O	ST TTL TTL	Digital I/O. Parallel Slave Port data. External Memory Address/Data 3.
RD4/SDO2/PSP4/AD4 RD4 SDO2 PSP4 ⁽⁴⁾ AD4	66	I/O O I/O I/O	ST — TTL TTL	Digital I/O. SPI data out. Parallel Slave Port data. External Memory Address/Data 4.
RD5/SDI2/SDA2/PSP5/	65			
AD5 RD5 SDI2 SDA2 PSP5 ⁽⁴⁾ AD5		I/O I I/O I/O I/O	ST ST I ² C TTL TTL	Digital I/O. SPI data in. I ² C™ data I/O. Parallel Slave Port data. External Memory Address/Data 5.
Legend: TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I = Input P = Power $I^2C = I^2C^{TM}/SMBus$				CMOS = CMOS compatible input or output Analog = Analog input O = Output OD = Open-Drain (no P diode to VDD)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

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= Open-Drain (no P diode to VDD)

Din Nomo	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
RD6/SCK2/SCL2/PSP6/	64			
AD6				
RD6		I/O	ST	Digital I/O.
SCK2		I/O	ST	Synchronous serial clock input/output for SPI mode.
SCL2		I/O	I ² C	Synchronous serial clock input/output for I ² C [™] mode.
PSP6 ⁽⁴⁾		I/O	TTL	Parallel Slave Port data.
AD6		I/O	TTL	External Memory Address/Data 6.
RD7/SS2/PSP7/AD7	63			
RD7		I/O	ST	Digital I/O.
SS2		I	TTL	SPI slave select input.
PSP7 ⁽⁴⁾		I/O	TTL	Parallel Slave Port data.
AD7		I/O	TTL	External Memory Address/Data 7.
Legend: TTL = TTL com	patible input	•	CMOS = CMOS compatible input or output	
ST = Schmitt	Frigger input wit	Analog = Analog input		
I = Input				O = Output

TABLE 1-4: PIC18F8XK22 PINOUT I/O DESCRIPTIONS (CONTINUED)

I²C = I²C™/SMBus

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

Ρ

= Power

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

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Pin Name	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре	Туре	Description
				PORTE is a bidirectional I/O port.
RE0/P2D/RD/AD8 RE0 P2D RD ⁽⁴⁾ AD8	4	I/O O I I/O	ST — TTL TTL	Digital I/O. ECCP2 PWM Output D. Parallel Slave Port read strobe. External Memory Address/Data 8.
RE1/P2C/WR/AD9 RE1 P2C WR ⁽⁴⁾ AD9	3	I/O O I I/O	ST — TTL TTL	Digital I/O. ECCP2 PWM Output C. Parallel Slave Port write strobe. External Memory Address/Data 9.
RE2/P2B/CCP10/ C S/ AD10 RE2 P2B <u>CCP10⁽³⁾ CS⁽⁴⁾ AD10</u>	78	1/0 1/0 1/0 1/0	ST ST ST TTL TTL	Digital I/O. ECCP2 PWM Output B. Capture 10 input/Compare 10 output/PWM10 output. Parallel Slave Port chip select. External Memory Address/Data 10.
RE3/P3C/CCP9/REFO AD11 RE3 P3C CCP9 ^(3,5) REFO AD11	77	I/O I/O 0/O I/O	ST — S/T — TTL	Digital I/O. ECCP3 PWM Output C. Capture 9 input/Compare 9 output/PWM9 output. Reference clock out. External Memory Address/Data 11.
RE4/P3B/CCP8/AD12 RE4 P3B CCP8 ⁽⁵⁾ AD12	76	I/O O I/O I/O	ST — ST TTL	Digital I/O. ECCP4 PWM Output B. Capture 8 input/Compare 8 output/PWM8 output. External Memory Address/Data 12.
RE5/P1C/CCP7/AD13 RE5 P1C CCP7 ⁽⁵⁾ AD13	75	I/O O I/O I/O	ST — ST TTL	Digital I/O. ECCP1 PWM Output C. Capture 7 input/Compare 7 output/PWM7 output. External Memory Address/Data 13.
Legend: TTL = TTL com ST = Schmitt I = Input	npatible input Trigger input wit	h CMC)S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output

P = Power

 $I^2C = I^2C^{TM}/SMBus$

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

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= Open-Drain (no P diode to VDD)

Pin Name	Pin Number		Buffer	Description
	TQFP		Туре	Description
RE6/P1B/CCP6/AD14	74			
RE6		I/O	ST	Digital I/O.
P1B		0	_	ECCP1 PWM Output B.
CCP6 ⁽⁵⁾		I/O	ST	Capture 6 input/Compare 6 output/PWM6 output.
AD14		I/O	ST	External Memory Address/Data 14.
RE7/ECCP2/P2A/AD15	73			
RE7		I/O	ST	Digital I/O.
ECCP2 ⁽²⁾		I/O	ST	Capture 2 input/Compare 2 output/PWM2 output.
P2A		0	—	ECCP2 PWM Output A.
AD15		I/O	ST	External Memory Address/Data 15.
Legend: TTL = TTL com	patible input	•		CMOS = CMOS compatible input or output
ST = Schmitt	Trigger input wit	h CMC)S levels	Analog = Analog input

= Input Р = Power $I^2C = I^2C^{TM}/SMBus$

Т

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

0

OD

= Output

= Open-Drain (no P diode to VDD)

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Pin Name	Pin Number	Pin	Buffer Type	Description
Pin Name	TQFP	Туре		
				PORTF is a bidirectional I/O port.
RF1/AN6/C2OUT/CTDIN RF1 AN6 C2OUT CTDIN	23	I/O I O I	ST Analog — ST	Digital I/O. Analog Input 6. Comparator 2 output. CTMU pulse delay input.
RF2/AN7/C1OUT RF2 AN7 C1OUT	18	I/O I O	ST Analog —	Digital I/O. Analog Input 7. Comparator 1 output.
RF3/AN8/C2INB/CTMUI RF3 AN8 C2INB CTMUI	17	I/O I I O	ST Analog Analog —	Digital I/O. Analog Input 8. Comparator 2 Input B. CTMU pulse generator charger for the C2INB comparator input.
RF4/AN9/C2INA RF4 AN9 C2INA	16	I/O I I	ST Analog Analog	Digital I/O. Analog Input 9. Comparator 2 Input A.
RF5/AN10/C1INB RF5 AN10 C1INB	15	I/O I I	ST Analog Analog	Digital I/O. Analog Input 10. Comparator 1 Input B.
RF6/AN11/C1INA RF6 AN11 C1INA	14	I/O I I	ST Analog Analog	Digital I/O. Analog Input 11. Comparator 1 Input A.
RF7/AN5/SS1 RF7 <u>AN5</u> SS1	13	I/O O I	ST Analog ST	Digital I/O. Analog Input 5. SPI slave select input.
Legend: TTL = TTL com ST = Schmitt T I = Input	patible input Trigger input wit	h CMC)S levels	CMOS = CMOS compatible input or output Analog = Analog input O = Output

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

DS39960D-page 32

= Open-Drain (no P diode to VDD)

Ρ

= Power

 $I^2C = I^2C^{TM}/SMBus$

Pin Name	Pin Number	Pin	Buffer	Description
Fill Name	TQFP	Туре	Туре	Description
RG0/ECCP3/P3A RG0 ECCP3 P3A	5	I/O I/O O	ST ST	PORTG is a bidirectional I/O port. Digital I/O. Capture 3 input/Compare 3 output/PWM3 output. ECCP3 PWM Output A.
RG1/TX2/CK2/AN19/ C3OUT RG1 TX2 CK2 AN19 C3OUT	6	1/0 1/0 1/0	ST — ST Analog —	Digital I/O. EUSART asynchronous transmit. EUSART synchronous clock (see related RX2/DT2). Analog Input 19. Comparator 3 output.
RG2/RX2/DT2/AN18/ C3INA RG2 RX2 DT2 AN18 C3INA	7	I/O I /O I /O I	ST ST ST Analog Analog	Digital I/O. EUSART asynchronous receive. EUSART synchronous data (see related TX2/CK2). Analog Input 18. Comparator 3 Input A.
RG3/CCP4/AN17/P3D/ C3INB RG3 CCP4 AN17 P3D C3INB	8	I/O I/O - O	ST ST Analog — Analog	Digital I/O. Capture 4 input/Compare 4 output/PWM4 output. Analog Input 17. ECCP3 PWM Output D. Comparator 3 Input B.
RG4/RTCC/T7CKI/T5G/ CCP5/AN16/P1D/C3INC RG4 RTCC T7CKI ⁽³⁾ T5G CCP5 AN16 P1D C3INC	10	l0 − − 0 − 0 −	ST — ST ST Analog — Analog	Digital I/O. RTCC output. Timer7 clock input. Timer5 external clock gate input. Capture 5 input/Compare 5 output/PWM5 output. Analog Input 16. ECCP1 PWM Output D. Comparator 3 Input C.
RG5	9		5	See the MCLR/RG5 pin.
Legend: TTL = TTL com ST = Schmitt T I = Input P = Power $I^2C = I^2C^{TM}/SM$	rigger input wit	h CMC)S levels	CMOS= CMOS compatible input or outputAnalog= Analog inputO= OutputOD= Open-Drain (no P diode to VDD)

TABLE 1-4: PIC18F8XK22 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

- **3:** Not available on PIC18F65K22 and PIC18F85K22 devices.
- 4: PSP is available only in Microcontroller mode.

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Din Nama	Pin Number	Pin	Buffer	Description
Pin Name	TQFP	Туре		Description
				PORTH is a bidirectional I/O port.
RH0/AN23/A16 RH0 AN23 A16	79	I/O I I/O	ST Analog TTL	Digital I/O. Analog Input 23. External Memory Address/Data 16.
RH1/AN22/A17 RH1 AN22 A17	80	I/O I I/O	ST Analog TTL	Digital I/O. Analog Input 22. External Address/Data 17.
RH2/AN21/A18 RH2 AN21 A18	1	I/O I I/O	ST Analog TTL	Digital I/O. Analog Input 21. External Address/Data 18.
RH3/AN20/A19 RH3 AN20 A19	2	I/O I I/O	ST Analog TTL	Digital I/O. Analog Input 20. External Address/Data 19.
RH4/CCP9/P3C/AN12/ C2INC RH4 CCP9 ^(3,5) P3C AN12 C2INC	22	I/O I/O O I	ST ST — Analog Analog	Digital I/O. Capture 9 input/Compare 9 output/PWM9 output. ECCP3 PWM Output C. Analog Input 12. Comparator 2 Input C.
RH5/CCP8/P3B/AN13/ C2IND RH5 CCP8 ⁽⁵⁾ P3B AN13 C2IND	21	I/O I/O O I I	ST ST — Analog Analog	Digital I/O. Capture 8 input/Compare 8 output/PWM8 output. ECCP3 PWM Output B. Analog Input 13. Comparator 1 Input D.
RH6/CCP7/P1C/AN14/ C1INC RH6 CCP7 ⁽⁵⁾ P1C AN14 C1INC	20	I/O I/O O I	ST ST — Analog Analog	Digital I/O. Capture 7 input/Compare 7 output/PWM7 output. ECCP1 PWM Output C. Analog Input 14. Comparator 1 Input C.

TABLE 1-4: PIC18F8XK22 PINOUT I/O DESCRIPTIONS (CONTINUED)

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

DS39960D-page 34

= Open-Drain (no P diode to VDD)

= Power

 $I^2C = I^2C^{\text{TM}}/\text{SMBus}$

Р

= Open-Drain (no P diode to VDD)

	TABLE 1-4:	PIC18F8XK22 PINOUT I/O DESCRIPTIONS ((CONTINUED))
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Pin Name	Pin Number	Pin Buffer	Description	
	TQFP	Туре Туре		Description
RH7/CCP6/P1B/AN15	19			
RH7		I/O	ST	Digital I/O.
CCP6 ⁽⁵⁾		I/O	ST	Capture 6 input/Compare 6 output/PWM6 output.
P1B		0	—	ECCP1 PWM Output B.
AN15		I	Analog	Analog Input 15.
Legend: TTL = TTL compatible input				CMOS = CMOS compatible input or output
ST = Schmitt Trigger input with CMOS levels				Analog = Analog input
I = Input				O = Output

= Power Ρ

 $I^2C = I^2C^{\text{TM}}/\text{SMBus}$

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

OD

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Pin Name	Pin Number	Pin	Buffer Type	Description	
	TQFP	Туре			
				PORTJ is a bidirectional I/O port.	
RJ0/ALE RJ0 ALE	62	I/O O	ST —	Digital I/O. External memory address latch enable.	
RJ1/OE RJ1 OE	61	I/O O	ST —	Digital I/O. External memory output enable.	
RJ2/WRL RJ2 WRL	60	I/O O	ST —	Digital I/O. External memory write low control.	
RJ3/WRH RJ3 WRH	59	I/O O	ST —	Digital I/O. External memory high control.	
RJ4/BA0 RJ4 BA0	39	I/O O	ST —	Digital I/O. External Memory Byte Address 0 control	
RJ5/CE RJ5 CE	40	I/O O	ST —	Digital I/O External memory chip enable control.	
RJ6/LB RJ6 LB	41	I/O O	ST —	Digital I/O. External memory low byte control.	
RJ7/UB RJ7 UB	42	I/O O	ST —	Digital I/O. External memory high byte control.	
Vss	11, 31, 51, 70	Р	—	Ground reference for logic and I/O pins.	
VDD	32, 48, 71	Р	—	Positive supply for logic and I/O pins.	
AVss	26	Р	—	Ground reference for analog modules.	
AVDD	25	Ρ	—	Positive supply for analog modules.	
ENVREG	24	Ι	ST	Enable for on-chip voltage regulator.	
VDDCORE/VCAP VDDCORE VCAP	12	Ρ	_	Core logic power or external filter capacitor connection. External filter capacitor connection (regulator enabled/disabled).	
Legend: TTL = TTL comp ST = Schmitt T I = Input P = Power I^2C = I^2C^{TM}/SM	rigger input wit	h CMC)S levels	CMOS= CMOS compatible input or outputAnalog= Analog inputO= OutputOD= Open-Drain (no P diode to VDD)	
 Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set. Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared 					

TABLE 1-4: PIC18F8XK22 PINOUT I/O DESCRIPTIONS (CONTINUED)

2: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared.

3: Not available on PIC18F65K22 and PIC18F85K22 devices.

4: PSP is available only in Microcontroller mode.

5: The CC6, CCP7, CCP8 and CCP9 pin placement depends on the setting of the ECCPMX Configuration bit (CONFIG3H<1>).

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2.0 GUIDELINES FOR GETTING STARTED WITH PIC18FXXKXX MICROCONTROLLERS

2.1 Basic Connection Requirements

Getting started with the PIC18F87K22 family family of 8-bit microcontrollers requires attention to a minimal set of device pin connections before proceeding with development.

The following pins must always be connected:

- All VDD and Vss pins (see Section 2.2 "Power Supply Pins")
- All AVDD and AVss pins, regardless of whether or not the analog device features are used (see Section 2.2 "Power Supply Pins")
- MCLR pin
 (see Section 2.3 "Master Clear (MCLR) Pin")
- ENVREG (if implemented) and VCAP/VDDCORE pins (see Section 2.4 "Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)")

These pins must also be connected if they are being used in the end application:

- PGC/PGD pins used for In-Circuit Serial Programming[™] (ICSP[™]) and debugging purposes (see Section 2.5 "ICSP Pins")
- OSCI and OSCO pins when an external oscillator source is used

(see Section 2.6 "External Oscillator Pins")

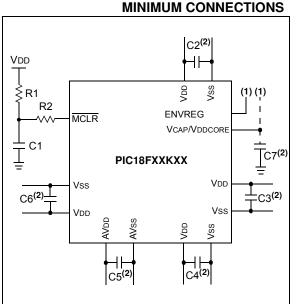
Additionally, the following pins may be required:

• VREF+/VREF- pins are used when external voltage reference for analog modules is implemented

Note: The AVDD and AVss pins must always be connected, regardless of whether any of the analog modules are being used.

The minimum mandatory connections are shown in Figure 2-1.

FIGURE 2-1: RECOMMENDED



Key (all values are recommendations):

C1 through C6: 0.1 $\mu\text{F},$ 20V ceramic R1: 10 k Ω

R2: 100Ω to 470Ω

- Note 1: See Section 2.4 "Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)" for explanation of ENVREG pin connections.
 - 2: The example shown is for a PIC18F device with five VDD/Vss and AVDD/AVss pairs. Other devices may have more or less pairs; adjust the number of decoupling capacitors appropriately.

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2.2 Power Supply Pins

2.2.1 DECOUPLING CAPACITORS

The use of decoupling capacitors on every pair of power supply pins, such as VDD, VSS, AVDD and AVSS, is required.

Consider the following criteria when using decoupling capacitors:

- Value and type of capacitor: A 0.1 μ F (100 nF), 10-20V capacitor is recommended. The capacitor should be a low-ESR device, with a resonance frequency in the range of 200 MHz and higher. Ceramic capacitors are recommended.
- Placement on the printed circuit board: The decoupling capacitors should be placed as close to the pins as possible. It is recommended to place the capacitors on the same side of the board as the device. If space is constricted, the capacitor can be placed on another layer on the PCB using a via; however, ensure that the trace length from the pin to the capacitor is no greater than 0.25 inch (6 mm).
- Handling high-frequency noise: If the board is experiencing high-frequency noise (upward of tens of MHz), add a second ceramic type capacitor in parallel to the above described decoupling capacitor. The value of the second capacitor can be in the range of 0.01 μ F to 0.001 μ F. Place this second capacitor next to each primary decoupling capacitor. In high-speed circuit designs, consider implementing a decade pair of capacitances as close to the power and ground pins as possible (e.g., 0.1 μ F in parallel with 0.001 μ F).
- Maximizing performance: On the board layout from the power supply circuit, run the power and return traces to the decoupling capacitors first, and then to the device pins. This ensures that the decoupling capacitors are first in the power chain. Equally important is to keep the trace length between the capacitor and the power pins to a minimum, thereby reducing PCB trace inductance.

2.2.2 TANK CAPACITORS

On boards with power traces running longer than six inches in length, it is suggested to use a tank capacitor for integrated circuits, including microcontrollers, to supply a local power source. The value of the tank capacitor should be determined based on the trace resistance that connects the power supply source to the device, and the maximum current drawn by the device in the application. In other words, select the tank capacitor so that it meets the acceptable voltage sag at the device. Typical values range from 4.7 μF to 47 μF .

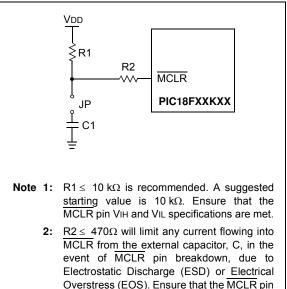
2.3 Master Clear (MCLR) Pin

The MCLR pin provides two specific device functions: Device Reset, and Device Programming and Debugging. If programming and debugging are not required in the end application, a direct connection to VDD may be all that is required. The addition of other components, to help increase the application's resistance to spurious Resets from voltage sags, may be beneficial. A typical configuration is shown in Figure 2-1. Other circuit designs may be implemented, depending on the application's requirements.

During programming and debugging, the resistance and capacitance that can be added to the pin must be considered. Device programmers and debuggers drive the MCLR pin. Consequently, specific voltage levels (VIH and VIL) and fast signal transitions must not be adversely affected. Therefore, specific values of R1 and C1 will need to be adjusted based on the application and PCB requirements. For example, it is recommended that the capacitor, C1, be isolated from the MCLR pin during programming and debugging operations by using a jumper (Figure 2-2). The jumper is replaced for normal run-time operations.

Any components associated with the $\overline{\text{MCLR}}$ pin should be placed within 0.25 inch (6 mm) of the pin.

FIGURE 2-2: EXAMPLE OF MCLR PIN CONNECTIONS



VIH and VIL specifications are met.

2.4 Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)

The on-chip voltage regulator enable pin, ENVREG, must always be connected directly to either a supply voltage or to ground. Tying ENVREG to VDD enables the regulator, while tying it to ground disables the regulator. Refer to **Section 28.3 "On-Chip Voltage Regulator"** for details on connecting and using the on-chip regulator.

When the regulator is enabled, a low-ESR (< 5 Ω) capacitor is required on the VCAP/VDDCORE pin to stabilize the voltage regulator output voltage. The VCAP/VDDCORE pin must not be connected to VDD and must use a capacitor of 10 μ F connected to ground. The type can be ceramic or tantalum. Suitable examples of capacitors are shown in Table 2-1. Capacitors with equivalent specifications can be used.

Designers may use Figure 2-3 to evaluate ESR equivalence of candidate devices.

It is recommended that the trace length not exceed 0.25 inch (6 mm). Refer to **Section 31.0** "**Electrical Characteristics**" for additional information.

When the regulator is disabled, the VCAP/VDDCORE pin must be tied to a voltage supply at the VDDCORE level. Refer to **Section 31.0 "Electrical Characteristics**" for information on VDD and VDDCORE. Some PIC18FXXKXX families, or some devices within a family, do not provide the option of enabling or disabling the on-chip voltage regulator:

- Some devices (with the name, PIC18LFXXKXX) permanently disable the voltage regulator. These devices do not have the ENVREG pin and require a 0.1 μ F capacitor on the VCAP/VDDCORE pin. The VDD level of these devices must comply with the "voltage regulator disabled" specification for Parameter D001, in Section 31.0 "Electrical Characteristics".
- Some devices permanently enable the voltage regulator.
 These devices also do not have the ENV/DEC picture.

These devices also do not have the ENVREG pin. The 10 μF capacitor is still required on the VCAP/VDDCORE pin.

FIGURE 2-3:

FREQUENCY vs. ESR PERFORMANCE FOR SUGGESTED VCAP

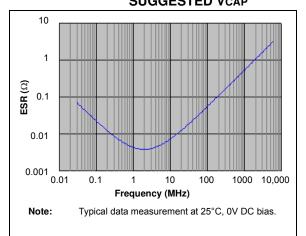


TABLE 2-1: SUITABLE CAPACITOR EQUIVALENTS

Make	Part #	Nominal Capacitance	Base Tolerance	Rated Voltage	Temp. Range
TDK	C3216X7R1C106K	10 µF	±10%	16V	-55 to 125°C
TDK	C3216X5R1C106K	10 µF	±10%	16V	-55 to 85°C
Panasonic	ECJ-3YX1C106K	10 µF	±10%	16V	-55 to 125°C
Panasonic	ECJ-4YB1C106K	10 µF	±10%	16V	-55 to 85°C
Murata	GRM32DR71C106KA01L	10 µF	±10%	16V	-55 to 125°C
Murata	GRM31CR61C106KC31L	10 µF	±10%	16V	-55 to 85°C

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2.4.1 CONSIDERATIONS FOR CERAMIC CAPACITORS

In recent years, large value, low-voltage, surface-mount ceramic capacitors have become very cost effective in sizes up to a few tens of microfarad. The low-ESR, small physical size and other properties make ceramic capacitors very attractive in many types of applications.

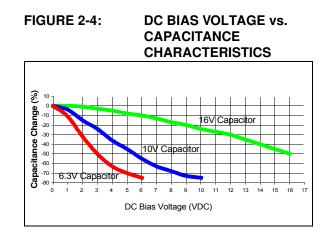
Ceramic capacitors are suitable for use with the internal voltage regulator of this microcontroller. However, some care is needed in selecting the capacitor to ensure that it maintains sufficient capacitance over the intended operating range of the application.

Typical low-cost, 10 μ F ceramic capacitors are available in X5R, X7R and Y5V dielectric ratings (other types are also available, but are less common). The initial tolerance specifications for these types of capacitors are often specified as ±10% to ±20% (X5R and X7R), or -20%/+80% (Y5V). However, the effective capacitance that these capacitors provide in an application circuit will also vary based on additional factors, such as the applied DC bias voltage and the temperature. The total in-circuit tolerance is, therefore, much wider than the initial tolerance specification.

The X5R and X7R capacitors typically exhibit satisfactory temperature stability (ex: $\pm 15\%$ over a wide temperature range, but consult the manufacturer's data sheets for exact specifications). However, Y5V capacitors typically have extreme temperature tolerance specifications of $\pm 22\%/-82\%$. Due to the extreme temperature tolerance, a 10 μ F nominal rated Y5V type capacitor may not deliver enough total capacitance to meet minimum internal voltage regulator stability and transient response requirements. Therefore, Y5V capacitors are not recommended for use with the internal regulator if the application must operate over a wide temperature range.

In addition to temperature tolerance, the effective capacitance of large value ceramic capacitors can vary substantially, based on the amount of DC voltage applied to the capacitor. This effect can be very significant, but is often overlooked or is not always documented.

A typical DC bias voltage vs. capacitance graph for X7R type and Y5V type capacitors is shown in Figure 2-4.



When selecting a ceramic capacitor to be used with the internal voltage regulator, it is suggested to select a high-voltage rating, so that the operating voltage is a small percentage of the maximum rated capacitor voltage. For example, choose a ceramic capacitor rated at 16V for the 2.5V core voltage. Suggested capacitors are shown in Table 2-1.

2.5 ICSP Pins

The PGC and PGD pins are used for In-Circuit Serial ProgrammingTM (ICSPTM) and debugging purposes. It is recommended to keep the trace length between the ICSP connector and the ICSP pins on the device as short as possible. If the ICSP connector is expected to experience an ESD event, a series resistor is recommended, with the value in the range of a few tens of ohms, not to exceed 100 Ω .

Pull-up resistors, series diodes and capacitors on the PGC and PGD pins are not recommended as they will interfere with the programmer/debugger communications to the device. If such discrete components are an application requirement, they should be removed from the circuit during programming and debugging. Alternatively, refer to the AC/DC characteristics and timing requirements information in the respective device Flash programming specification for information on capacitive loading limits, and pin input voltage high (VIH) and input low (VIL) requirements.

For device emulation, ensure that the "Communication Channel Select" (i.e., PGCx/PGDx pins), programmed into the device, matches the physical connections for the ICSP to the Microchip debugger/emulator tool.

For more information on available Microchip development tools connection requirements, refer to **Section 30.0 "Development Support"**.

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2.6 External Oscillator Pins

Many microcontrollers have options for at least two oscillators: a high-frequency primary oscillator and a low-frequency secondary oscillator (refer to **Section 3.0 "Oscillator Configurations**" for details).

The oscillator circuit should be placed on the same side of the board as the device. Place the oscillator circuit close to the respective oscillator pins with no more than 0.5 inch (12 mm) between the circuit components and the pins. The load capacitors should be placed next to the oscillator itself, on the same side of the board.

Use a grounded copper pour around the oscillator circuit to isolate it from surrounding circuits. The grounded copper pour should be routed directly to the MCU ground. Do not run any signal traces or power traces inside the ground pour. Also, if using a two-sided board, avoid any traces on the other side of the board where the crystal is placed.

Layout suggestions are shown in Figure 2-4. In-line packages may be handled with a single-sided layout that completely encompasses the oscillator pins. With fine-pitch packages, it is not always possible to completely surround the pins and components. A suitable solution is to tie the broken guard sections to a mirrored ground layer. In all cases, the guard trace(s) must be returned to ground.

In planning the application's routing and I/O assignments, ensure that adjacent port pins, and other signals in close proximity to the oscillator, are benign (i.e., free of high frequencies, short rise and fall times, and other similar noise).

For additional information and design guidance on oscillator circuits, please refer to these Microchip Application Notes, available at the corporate web site (www.microchip.com):

- AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[™] and PICmicro[®] Devices"
- AN849, "Basic PICmicro[®] Oscillator Design"
- AN943, "Practical PICmicro[®] Oscillator Analysis and Design"
- AN949, "Making Your Oscillator Work"

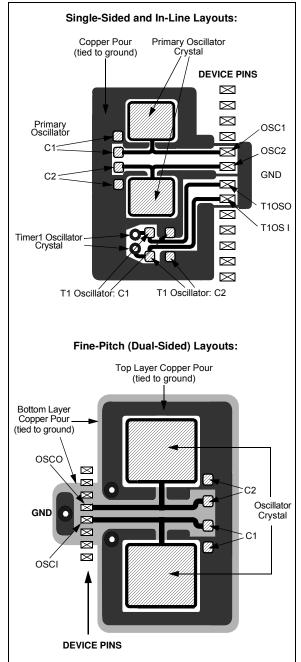
2.7 Unused I/Os

Unused I/O pins should be configured as outputs and driven to a logic low state. Alternatively, connect a 1 k Ω to 10 k Ω resistor to Vss on unused pins and drive the output to logic low.

FIGURE 2-5:

PLACEMENT OF THE OSCILLATOR CIRCUIT

SUGGESTED



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NOTES:

3.0 OSCILLATOR CONFIGURATIONS

3.1 Oscillator Types

The PIC18F87K22 family of devices can be operated in the following oscillator modes:

- EC External clock, RA6 available
- ECIO External clock, clock out RA6 (Fosc/4 on RA6)
- HS High-Speed Crystal/Resonator
- XT Crystal/Resonator
- LP Low-Power Crystal
- RC External Resistor/Capacitor, RA6 available
- RCIO External Resistor/Capacitor, clock out RA6 (Fosc/4 on RA6)
- INTIO2 Internal Oscillator with I/O on RA6 and RA7
- INTIO1 Internal Oscillator with Fosc/4 output on RA6 and I/O on RA7

There is also an option for running the 4xPLL on any of the clock sources in the input frequency range of 4 to 16 MHz.

The PLL is enabled by setting the PLLCFG bit (CONFIG1H<4>) or the PLLEN bit (OSCTUNE<6>).

For the EC and HS mode, the PLLEN (software) or PLLCFG (CONFIG) bit can be used to enable the PLL.

For the INTIOx modes (HF-INTOSC):

- Only the PLLEN can enable the PLL (PLLCFG is ignored).
- When the oscillator is configured for the internal oscillator (FOSC<3:0> = 100x), the PLL can be enabled only when the HF-INTOSC frequency is 8 or 16 MHz.

When the RA6 and RA7 pins are not used for an oscillator function or CLKOUT function, they are available as general purpose I/Os. To optimize power consumption when using EC/HS/ XT/LP/RC as the primary oscillator, the frequency input range can be configured to yield an optimized power bias:

- Low-Power Bias External frequency less than 160 kHz
- Medium Power Bias External frequency between 160 kHz and 16 MHz
- High-Power Bias External frequency greater than 16 MHz

All of these modes are selected by the user by programming the FOSC<3:0> Configuration bits (CONFIG1H<3:0>). In addition, PIC18F87K22 family devices can switch between different clock sources, either under software control or, under certain conditions, automatically. This allows for additional power savings by managing device clock speed in real time without resetting the application. The clock sources for the PIC18F87K22 family of devices are shown in Figure 3-1.

For the HS and EC mode, there are additional power modes of operation – depending on the frequency of operation.

HS1 is the Medium Power mode with a frequency range of 4 MHz to 16 MHz. HS2 is the High-Power mode, where the oscillator frequency can go from 16 MHz to 25 MHz. HS1 and HS2 are achieved by setting the CONFIG1H<3:0> correctly. (For details, see Register 28-2 on page 406.)

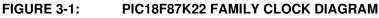
EC mode has these modes of operation:

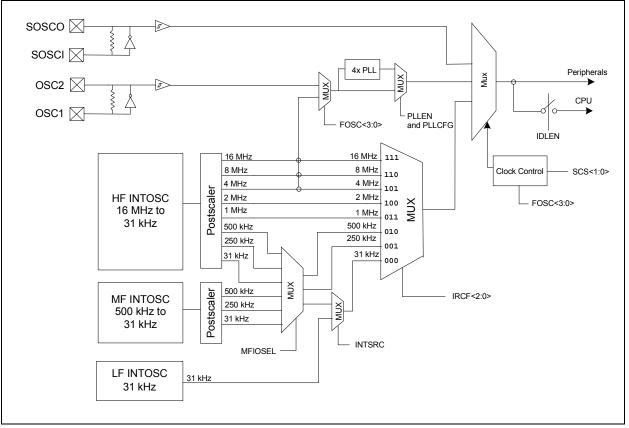
- EC1 For low power with a frequency range up to 160 kHz
- EC2 Medium power with a frequency range of 160 kHz to 16 MHz
- EC3 High power with a frequency range of 16 MHz to 64 MHz

EC1, EC2 and EC3 are achieved by setting the CONFIG1H<3:0> correctly. (For details, see Register 28-2 on page 406.)

Table 3-1 shows the HS and EC modes' frequencyrange and FOSC<3:0> settings.

Mode	Frequency Range	FOSC<3:0> Setting
EC1 (low power)	EC1 (low power)	
(EC1 & EC1IO)	DC-160 kHz	1100
EC2 (medium power)		1011
(EC2 & EC2IO)		1010
EC3 (high power)		0101
(EC3 & EC3IO)		0100
HS1 (medium power)	4 MHz-16 MHz	0011
HS2 (high power)	16 MHz-25 MHz	0010
XT	XT 100 kHz-4 MHz	
LP	31.25 kHz	0000
RC (External) 0-4 MHz		001x
INTIO	32 KHz-16 MHz	100x (and OSCCON, OSCCON2)





3.2 Control Registers

The OSCCON register (Register 3-1) controls the main aspects of the device clock's operation. It selects the oscillator type to be used, which of the power-managed modes to invoke and the output frequency of the INTOSC source. It also provides status on the oscillators.

The OSCTUNE register (Register 3-3) controls the tuning and operation of the internal oscillator block. It also implements the PLLEN bit which controls the operation of the Phase Locked Loop (PLL) (see Section 3.5.3 "PLL Frequency Multiplier").

REGISTER 3-1: OSCCON: OSCILLATOR CONTROL REGISTER⁽¹⁾

R/W-0	R/W-1	R/W-1	R/W-0	R ⁽¹⁾	R-0	R/W-0	R/W-0
IDLEN	IRCF2 ⁽²⁾	IRCF1 ⁽²⁾	IRCF0 ⁽²⁾	OSTS	HFIOFS	SCS1 ⁽⁴⁾	SCS0 ⁽⁴⁾
bit 7 b							bit 0
Legend:							
R = Readable bit W = Writable bit		bit	U = Unimplemented bit, read as '0'				
-n = Value at POR '1		'1' = Bit is set		'0' = Bit is cleared x =		x = Bit is unkr	nown

bit 7	IDLEN: Idle Enable bit
	1 = Device enters an Idle mode when a SLEEP instruction is executed
	 0 = Device enters Sleep mode when a SLEEP instruction is executed
bit 6-4	IRCF<2:0>: Internal Oscillator Frequency Select bits ⁽²⁾
	111 = HF-INTOSC output frequency is used (16 MHz)
	110 = HF-INTOSC/2 output frequency is used (8 MHz, default)
	101 = HF-INTOSC/4 output frequency is used (4 MHz)
	100 = HF-INTOSC/8 output frequency is used (2 MHz)
	011 = HF-INTOSC/16 output frequency is used (1 MHz)
	If INTSRC = 0 and MFIOSEL = 0 : ^(3,5)
	010 = HF-INTOSC/32 output frequency is used (500 kHz)
	001 = HF-INTOSC/64 output frequency is used (250 kHz)
	000 = LF-INTOSC output frequency is used (31.25 kHz) If INTSRC = 1 and MFIOSEL = 0: ^(3,5)
	010 = HF-INTOSC/32 output frequency is used (500 kHz)
	001 = HF-INTOSC/64 output frequency is used (250 kHz)
	000 = HF-INTOSC/512 output frequency is used (31.25 kHz)
	If INTSRC = 0 and MFIOSEL = 1: ^(3,5)
	010 = MF-INTOSC output frequency is used (500 kHz)
	001 = MF-INTOSC/2 output frequency is used (250 kHz)
	000 = LF-INTOSC output frequency is used (31.25 kHz) ⁽⁶⁾
	If INTSRC = 1 and MFIOSEL = 1: $(3,5)$
	010 = MF-INTOSC output frequency is used (500 kHz)
	001 = MF-INTOSC/2 output frequency is used (250 kHz)
	000 = MF-INTOSC/16 output frequency is used (31.25 kHz)
bit 3	OSTS: Oscillator Start-up Timer Time-out Status bit ⁽¹⁾
	1 = Oscillator Start-up Timer (OST) time-out has expired; primary oscillator is running, as defined by FOSC<3:0>
	 0 = Oscillator Start-up Timer (OST) time-out is running; primary oscillator is not ready – device is running from internal oscillator (HF-INTOSC, MF-INTOSC or LF-INTOSC)
Note 1:	The Reset state depends on the state of the IESO Configuration bit (CONFIG1H<7>).
2:	Modifying these bits will cause an immediate clock frequency switch if the internal oscillator is providing the device clocks.
3:	Source selected by the INTSRC bit (OSCTUNE<7>).
4:	Modifying these bits will cause an immediate clock source switch.
5:	INTSRC = OSCTUNE<7> and MFIOSEL = OSCCON2<0>.

6: Lowest power option for an internal source.

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REGISTER 3-1: OSCCON: OSCILLATOR CONTROL REGISTER⁽¹⁾ (CONTINUED)

- bit 2 **HFIOFS:** INTOSC Frequency Stable bit
 - 1 = HF-INTOSC oscillator frequency is stable
 - 0 = HF-INTOSC oscillator frequency is not stable
- bit 1-0 SCS<1:0>: System Clock Select bits⁽⁴⁾
 - 1x = Internal oscillator block (LF-INTOSC, MF-INTOSC or HF-INTOSC)
 - 01 = SOSC oscillator
 - 00 = Default primary oscillator (OSC1/OSC2 or HF-INTOSC with or without PLL; defined by the FOSC<3:0> Configuration bits, CONFIG1H<3:0>.)
- Note 1: The Reset state depends on the state of the IESO Configuration bit (CONFIG1H<7>).
 - 2: Modifying these bits will cause an immediate clock frequency switch if the internal oscillator is providing the device clocks.
 - 3: Source selected by the INTSRC bit (OSCTUNE<7>).
 - 4: Modifying these bits will cause an immediate clock source switch.
 - **5:** INTSRC = OSCTUNE<7> and MFIOSEL = OSCCON2<0>.
 - 6: Lowest power option for an internal source.

REGISTER 3-2: OSCCON2: OSCILLATOR CONTROL REGISTER 2

U-0	R-0	U-0	U-0	R/W-0	U-0	R-x	R/W-0
—	SOSCRUN	_	—	SOSCGO	—	MFIOFS	MFIOSEL
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as 'O'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	SOSCRUN: SOSC Run Status bit
	1 = System clock comes from a secondary SOSC
	0 = System clock comes from an oscillator other than SOSC
bit 5-4	Unimplemented: Read as '0'
bit 3	SOSCGO: Oscillator Start Control bit
	 1 = Oscillator is running, even if no other sources are requesting it 0 = Oscillator is shut off if no other sources are requesting it (When the SOSC is selected to run from a digital clock input, rather than an external crystal, this bit has no effect.)
bit 2	Unimplemented: Read as '0'
bit 1	MFIOFS: MF-INTOSC Frequency Stable bit
	1 = MF-INTOSC is stable
	0 = MF-INTOSC is not stable
bit 0	MFIOSEL: MF-INTOSC Select bit
	 1 = MF-INTOSC is used in place of HF-INTOSC frequencies of 500 kHz, 250 kHz and 31.25 kHz 0 = MF-INTOSC is not used

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
INTSRC	PLLEN	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0
bit 7	1						bit C
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	l as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	iown
bit 6	0 = 31 kHz d	levice clock der quency Multiplie nabled	ived from INT	6 MHz INTOSC RC 31 kHz oscil bit			1, HF-INTOSC
bit 5-0	011111 = Ma 000001 000000 = Ce 111111	aximum frequer • •	ncy ; fast RC oscill	Frequency Tunir	-	d frequency	

REGISTER 3-3: OSCTUNE: OSCILLATOR TUNING REGISTER

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3.3 Clock Sources and Oscillator Switching

Essentially, PIC18F87K22 family devices have these independent clock sources:

- Primary oscillators
- · Secondary oscillators
- Internal oscillator

The **primary oscillators** can be thought of as the main device oscillators. These are any external oscillators connected to the OSC1 and OSC2 pins, and include the External Crystal and Resonator modes and the External Clock modes. If selected by the FOSC<3:0> Configuration bits (CONFIG1H<3:0>), the internal oscillator block may be considered a primary oscillator. The internal oscillator block can be one of the following:

- 31 kHz LF-INTRC source
- 31 kHz to 500 kHz MF-INTOSC source
- · 31 kHz to 16 MHz HF-INTOSC source

The particular mode is defined by the FOSC Configuration bits. The details of these modes are covered in Section 3.5 "External Oscillator Modes".

The **secondary oscillators** are external clock sources that are not connected to the OSC1 or OSC2 pin. These sources may continue to operate, even after the controller is placed in a power-managed mode. PIC18F87K22 family devices offer the SOSC (Timer1/3/5/7) oscillator as a secondary oscillator source. This oscillator, in all power-managed modes, is often the time base for functions, such as a Real-Time Clock (RTC).

The SOSC can be enabled from any peripheral that requests it. There are eight ways the SOSC can be enabled: if the SOSC is selected as the source by any of the odd timers, which is done by each respective SOSCEN bit (TxCON<3>), if the SOSC is selected as the RTCC source by the RTCOSC Configuration bit (CONFIG3L<1>), if the SOSC is selected as the CPU clock source by the SCS bits (OSCCON<1:0>) or if the SOSCGO bit is set (OSCCON2<3>). The SOSCGO bit is used to warm up the SOSC so that it is ready before any peripheral requests it.

The secondary oscillator has three Run modes. The SOSCSEL<1:0> bits (CONFIG1L<4:3>) decide the SOSC mode of operation:

- 11 = High-power SOSC circuit
- 10 = Digital (SCLKI) mode
- 01 = Low-power SOSC circuit

If a secondary oscillator is not desired and digital I/O on port pins, RC0 and RC1, is needed, the SOSCSEL bits must be set to Digital mode.

In addition to being a primary clock source in some circumstances, the **internal oscillator** is available as a power-managed mode clock source. The LF-INTOSC source is also used as the clock source for several special features, such as the WDT and Fail-Safe Clock Monitor. The internal oscillator block is discussed in more detail in **Section 3.6** "Internal Oscillator Block".

The PIC18F87K22 family includes features that allow the device clock source to be switched from the main oscillator, chosen by device configuration, to one of the alternate clock sources. When an alternate clock source is enabled, various power-managed operating modes are available.

3.3.1 OSC1/OSC2 OSCILLATOR

The OSC1/OSC2 oscillator block is used to provide the oscillator modes and frequency ranges:

Mode	Design Operating Frequency		
LP	31.25-100 kHz		
ХТ	100 kHz to 4 MHz		
HS	4 MHz to 25 MHz		
EC 0 to 64 MHz (external clock			
EXTRC 0 to 4 MHz (external RC)			

The crystal-based oscillators (XT, HS and LP) have a built-in start-up time. The operation of the EC and EXTRC clocks is immediate.

3.3.2 CLOCK SOURCE SELECTION

The System Clock Select bits, SCS<1:0> (OSCCON2<1:0>), select the clock source. The available clock sources are the primary clock defined by the FOSC<3:0> Configuration bits, the secondary clock (SOSC oscillator) and the internal oscillator. The clock source changes after one or more of the bits is written to, following a brief clock transition interval.

The OSTS (OSCCON<3>) and SOSCRUN (OSCCON<6>) bits indicate which clock source is currently providing the device clock. The OSTS bit indicates that the Oscillator Start-up Timer (OST) has timed out and the primary clock is providing the device clock in primary clock modes. The SOSCRUN bit indicates when the SOSC oscillator (from Timer1/3/5/7) is providing the device clock in secondary clock modes. In power-managed modes, only one of these bits will be set at any time. If neither of these bits is set, the INTRC is providing the clock, or the internal oscillator has just started and is not yet stable.

The IDLEN bit (OSCCON<7>) determines if the device goes into Sleep mode or one of the Idle modes when the SLEEP instruction is executed.

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The use of the flag and control bits in the OSCCON register is discussed in more detail in **Section 4.0** "**Power-Managed Modes**".

- Note 1: The Timer1/3/5/7 oscillator must be enabled to select the secondary clock source. The Timerx oscillator is enabled by setting the SOSCEN bit in the Timerx Control register (TxCON<3>). If the Timerx oscillator is not enabled, then any attempt to select a secondary clock source when executing a SLEEP instruction will be ignored.
 - 2: It is recommended that the Timerx oscillator be operating and stable before executing the SLEEP instruction or a very long delay may occur while the Timerx oscillator starts.

3.3.2.1 System Clock Selection and Device Resets

Since the SCS bits are cleared on all forms of Reset, this means the primary oscillator, defined by the FOSC<3:0> Configuration bits, is used as the primary clock source on device Resets. This could either be the internal oscillator block by itself, or one of the other primary clock source (HS, EC, XT, LP, External RC and PLL-Enabled modes).

In those cases when the internal oscillator block, without PLL, is the default clock on Reset, the Fast RC oscillator (INTOSC) will be used as the device clock source. It will initially start at 8 MHz; the postscaler selection that corresponds to the Reset value of the IRCF<2:0> bits ('110').

Regardless of which primary oscillator is selected, INTRC will always be enabled on device power-up. It serves as the clock source until the device has loaded its configuration values from memory. It is at this point that the FOSC Configuration bits are read and the oscillator selection of the operational mode is made.

Note that either the primary clock source or the internal oscillator will have two bit setting options for the possible values of the SCS<1:0> bits, at any given time.

3.3.3 OSCILLATOR TRANSITIONS

PIC18F87K22 family devices contain circuitry to prevent clock "glitches" when switching between clock sources. A short pause in the device clock occurs during the clock switch. The length of this pause is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable.

Clock transitions are discussed in greater detail in Section 4.1.2 "Entering Power-Managed Modes".

3.4 RC Oscillator

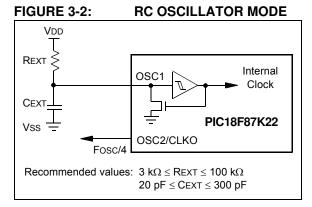
For timing-insensitive applications, the RC and RCIO Oscillator modes offer additional cost savings. The actual oscillator frequency is a function of several factors:

- · Supply Voltage
- Values of the External Resistor (REXT) and Capacitor (CEXT)
- Operating Temperature

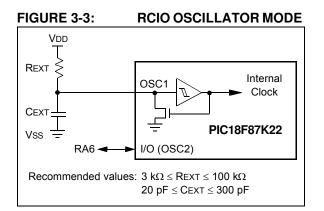
Given the same device, operating voltage and temperature, and component values, there will also be unit to unit frequency variations. These are due to factors, such as:

- Normal manufacturing variation
- Difference in lead frame capacitance between package types (especially for low CEXT values)
- Variations within the tolerance of limits of $\ensuremath{\mathsf{REXT}}$ and $\ensuremath{\mathsf{CEXT}}$

In the RC Oscillator mode, the oscillator frequency, divided by 4, is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure 3-2 shows how the R/C combination is connected.



The RCIO Oscillator mode (Figure 3-3) functions like the RC mode, except that the OSC2 pin becomes an additional general purpose I/O pin. The I/O pin becomes bit 6 of PORTA (RA6).



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3.5 External Oscillator Modes

3.5.1 CRYSTAL OSCILLATOR/CERAMIC RESONATORS (HS MODES)

In HS or HSPLL Oscillator modes, a crystal or ceramic resonator is connected to the OSC1 and OSC2 pins to establish oscillation. Figure 3-4 shows the pin connections.

The oscillator design requires the use of a crystal rated for parallel resonant operation.

Note: Use of a crystal rated for series resonant operation may give a frequency out of the crystal manufacturer's specifications.

TABLE 3-2: CAPACITOR SELECTION FOR CERAMIC RESONATORS

ed:
ed:

Mode	Freq.	OSC1	OSC2
HS	8.0 MHz	27 pF	27 pF
	16.0 MHz	22 pF	22 pF

Capacitor values are for design guidance only.

Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application. Refer to the following application notes for oscillator-specific information:

- AN588, "PIC[®] Microcontroller Oscillator Design Guide"
- AN826, "Crystal Oscillator Basics and Crystal Selection for rfPIC[®] and PIC[®] Devices"
- AN849, "Basic PIC[®] Oscillator Design"
- AN943, "Practical PIC[®] Oscillator Analysis and Design"
- · AN949, "Making Your Oscillator Work"

See the notes following Table 3-3 for additional information.

TABLE 3-3:CAPACITOR SELECTION FOR
CRYSTAL OSCILLATOR

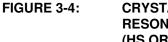
Osc Type	Crystal Freq.	Typical Capacitor Values Tested:			
	Fieq.	C1	C2		
HS	4 MHz	27 pF	27 pF		
	8 MHz	22 pF	22 pF		
	20 MHz	15 pF	15 pF		

Capacitor values are for design guidance only.

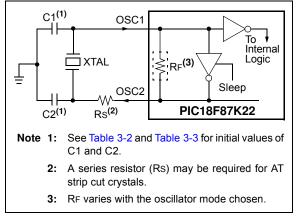
Different capacitor values may be required to produce acceptable oscillator operation. The user should test the performance of the oscillator over the expected VDD and temperature range for the application.

Refer to the Microchip application notes cited in Table 3-2 for oscillator-specific information. Also see the notes following this table for additional information.

- Note 1: Higher capacitance increases the stability of the oscillator but also increases the start-up time.
 - 2: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - 3: Rs may be required to avoid overdriving crystals with a low drive level specification.
 - 4: Always verify oscillator performance over the VDD and temperature range that is expected for the application.



CRYSTAL/CERAMIC RESONATOR OPERATION (HS OR HSPLL CONFIGURATION)

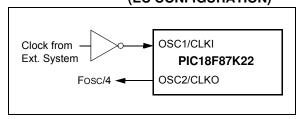


3.5.2 EXTERNAL CLOCK INPUT (EC MODES)

The EC and ECPLL Oscillator modes require an external clock source to be connected to the OSC1 pin. There is no oscillator start-up time required after a Power-on Reset or after an exit from Sleep mode.

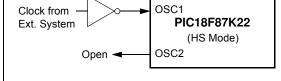
In the EC Oscillator mode, the oscillator frequency, divided by 4, is available on the OSC2 pin. This signal may be used for test purposes or to synchronize other logic. Figure 3-5 shows the pin connections for the EC Oscillator mode.

FIGURE 3-5: EXTERNAL CLOCK INPUT OPERATION (EC CONFIGURATION)



An external clock source may also be connected to the OSC1 pin in HS mode, as shown in Figure 3-6. In this configuration, the divide-by-4 output on OSC2 is not available. Current consumption in this configuration will be somewhat higher than EC mode, as the internal oscillator's feedback circuitry will be enabled (in EC mode, the feedback circuit is disabled).

FIGURE 3-6: EXTERNAL CLOCK INPUT OPERATION (HS OSC CONFIGURATION)



3.5.3 PLL FREQUENCY MULTIPLIER

A Phase Locked Loop (PLL) circuit is provided as an option for users who want to use a lower frequency oscillator circuit, or to clock the device up to its highest rated frequency from a crystal oscillator. This may be useful for customers who are concerned with EMI due to high-frequency crystals, or users who require higher clock speeds from an internal oscillator.

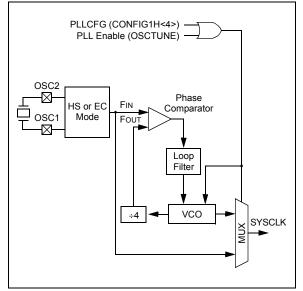
3.5.3.1 HSPLL and ECPLL Modes

The HSPLL and ECPLL modes provide the ability to selectively run the device at four times the external oscillating source to produce frequencies up to 64 MHz.

The PLL is enabled by setting the PLLEN bit (OSCTUNE<6>) or the PLLCFG bit (CONFIG1H<4>). The PLLEN bit provides a software control for the PLL, even if PLLCFG is set to '0'. The PLL is enabled only when the HS or EC oscillator frequency is within the 4 MHz to16 MHz input range.

This enables additional flexibility for controlling the application's clock speed in software. The PLLEN should be enabled in HS or EC Oscillator mode only if the input frequency is in the range of 4 MHz-16 MHz.

FIGURE 3-7: PLL BLOCK DIAGRAM



3.5.3.2 PLL and HF-INTOSC

The PLL is available to the internal oscillator block when the internal oscillator block is configured as the primary clock source. In this configuration, the PLL is enabled in software and generates a clock output of up to 64 MHz.

The operation of INTOSC with the PLL is described in **Section 3.6.2 "INTPLL Modes**". Care should be taken that the PLL is enabled only if the HF-INTOSC postscaler is configured for 8 MHz or 16 MHz.

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3.6 Internal Oscillator Block

The PIC18F87K22 family of devices includes an internal oscillator block which generates two different clock signals. Either clock can be used as the microcontroller's clock source, which may eliminate the need for an external oscillator circuit on the OSC1 and/or OSC2 pins.

The internal oscillator consists of three blocks, depending on the frequency of operation. They are HF-INTOSC, MF-INTOSC and LF-INTRC.

In HF-INTOSC mode, the internal oscillator can provide a frequency ranging from 31 kHz to 16 MHz, with the postscaler deciding the selected frequency (IRCF<2:0>).

The INTSRC bit (OSCTUNE<7>) and MFIOSEL bit (OSCCON2<0>) also decide which INTOSC provides the lower frequency (500 kHz to 31 KHz). For the HF-INTOSC to provide these frequencies, INTSRC = 1 and MFIOSEL = 0.

In HF-INTOSC, the postscaler (IRCF<2:0>) provides the frequency range of 31 kHz to 16 MHz. If HF-INTOSC is used with the PLL, the input frequency to the PLL should be 8 MHz or 16 MHz (IRCF<2:0> = 111 or 110).

For MF-INTOSC mode to provide a frequency range of 500 kHz to 31 kHz, INTSRC = 1 and MFIOSEL = 1. The postscaler (IRCF<2:0>), in this mode, provides the frequency range of 31 kHz to 500 kHz.

The LF-INTRC can provide only 31 kHz if INTSRC = 0.

The LF-INTRC provides 31 kHz and is enabled if it is selected as the device clock source. The mode is enabled automatically when any of the following are enabled:

- · Power-up Timer
- · Fail-Safe Clock Monitor
- · Watchdog Timer
- Two-Speed Start-up

These features are discussed in greater detail in Section 28.0 "Special Features of the CPU".

The clock source frequency (HF-INTOSC, MF-INTOSC or LF-INTRC direct) is selected by configuring the IRCF bits of the OSCCON register, as well the INTSRC and MFIOSEL bits. The default frequency on device Resets is 8 MHz.

3.6.1 INTIO MODES

Using the internal oscillator as the clock source eliminates the need for up to two external oscillator pins, which can then be used for digital I/O. Two distinct oscillator configurations, which are determined by the FOSC Configuration bits, are available:

- In INTIO1 mode, the OSC2 pin (RA6) outputs Fosc/4, while OSC1 functions as RA7 (see Figure 3-8) for digital input and output.
- In INTIO2 mode, OSC1 functions as RA7 and OSC2 functions as RA6 (see Figure 3-9). Both are available as digital input and output ports.

FIGURE 3-8: INTIO1 OSCILLATOR MODE

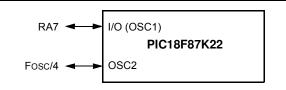
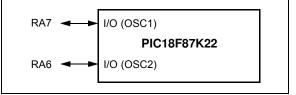


FIGURE 3-9: INTIO2 OSCILLATOR MODE



3.6.2 INTPLL MODES

The 4x Phase Lock Loop (PLL) can be used with the HF-INTOSC to produce faster device clock speeds than are normally possible with the internal oscillator sources. When enabled, the PLL produces a clock speed of 32 MHz or 64 MHz.

PLL operation is controlled through software. The control bit, PLLEN (OSCTUNE<6>), is used to enable or disable its operation. Additionally, the PLL will only function when the selected HF-INTOSC frequency is either 8 MHz or 16 MHz (OSCCON<6:4> = 111 or 110).

Like the INTIO modes, there are two distinct INTPLL modes available:

- In INTPLL1 mode, the OSC2 pin outputs Fosc/4, while OSC1 functions as RA7 for digital input and output. Externally, this is identical in appearance to INTIO1 (Figure 3-8).
- In INTPLL2 mode, OSC1 functions as RA7 and OSC2 functions as RA6, both for digital input and output. Externally, this is identical to INTIO2 (Figure 3-9).

3.6.3 INTERNAL OSCILLATOR OUTPUT FREQUENCY AND TUNING

The internal oscillator block is calibrated at the factory to produce an INTOSC output frequency of 16 MHz. It can be adjusted in the user's application by writing to TUN<5:0> (OSCTUNE<5:0>) in the OSCTUNE register (Register 3-3).

When the OSCTUNE register is modified, the INTOSC (HF-INTOSC and MF-INTOSC) frequency will begin shifting to the new frequency. The oscillator will require some time to stabilize. Code execution continues during this shift and there is no indication that the shift has occurred.

The LF-INTOSC oscillator operates independently of the HF-INTOSC or the MF-INTOSC source. Any changes in the HF-INTOSC or the MF-INTOSC source, across voltage and temperature, are not necessarily reflected by changes in LF-INTOSC or vice versa. The frequency of LF-INTOSC is not affected by OSCTUNE.

3.6.4 INTOSC FREQUENCY DRIFT

The INTOSC frequency may drift as VDD or temperature changes, and can affect the controller operation in a variety of ways. It is possible to adjust the INTOSC frequency by modifying the value in the OSCTUNE register. Depending on the device, this may have no effect on the LF-INTOSC clock source frequency.

Tuning INTOSC requires knowing when to make the adjustment, in which direction it should be made, and in some cases, how large a change is needed. Three compensation techniques are shown here.

3.6.4.1 Compensating with the EUSART

An adjustment may be required when the EUSART begins to generate framing errors or receives data with errors while in Asynchronous mode. Framing errors indicate that the device clock frequency is too high. To adjust for this, decrement the value in OSCTUNE to reduce the clock frequency. On the other hand, errors in data may suggest that the clock speed is too low. To compensate, increment OSCTUNE to increase the clock frequency.

3.6.4.2 Compensating with the Timers

This technique compares device clock speed to some reference clock. Two timers may be used; one timer is clocked by the peripheral clock, while the other is clocked by a fixed reference source, such as the SOSC oscillator.

Both timers are cleared, but the timer clocked by the reference source generates interrupts. When an interrupt occurs, the internally clocked timer is read and both timers are cleared. If the internally clocked timer value is much greater than expected, then the internal oscillator block is running too fast. To adjust for this, decrement the OSCTUNE register.

3.6.4.3 Compensating with the CCP Module in Capture Mode

A CCP module can use free-running Timer1 (or Timer3), clocked by the internal oscillator block and an external event with a known period (i.e., AC power frequency). The time of the first event is captured in the CCPRxH:CCPRxL registers and is recorded for use later. When the second event causes a capture, the time of the first event is subtracted from the time of the second event. Since the period of the external event is known, the time difference between events can be calculated.

If the measured time is much greater than the calculated time, the internal oscillator block is running too fast. To compensate, decrement the OSCTUNE register. If the measured time is much less than the calculated time, the internal oscillator block is running too slow. To compensate, increment the OSCTUNE register.

3.7 Reference Clock Output

In addition to the Fosc/4 clock output, in certain oscillator modes, the device clock in the PIC18F87K22 family can also be configured to provide a reference clock output signal to a port pin. This feature is available in all oscillator configurations and allows the user to select a greater range of clock submultiples to drive external devices in the application.

This reference clock output is controlled by the REFOCON register (Register 3-4). Setting the ROON bit (REFOCON<7>) makes the clock signal available on the REFO (RE3) pin. The RODIV<3:0> bits enable the selection of 16 different clock divider options.

The ROSSLP and ROSEL bits (REFOCON<5:4>) control the availability of the reference output during Sleep mode. The ROSEL bit determines if the oscillator on OSC1 and OSC2, or the current system clock source, is used for the reference clock output. The ROSSLP bit determines if the reference source is available on RE3 when the device is in Sleep mode.

To use the reference clock output in Sleep mode, both the ROSSLP and ROSEL bits must be set. The device clock must also be configured for an EC or HS mode. If not, the oscillator on OSC1 and OSC2 will be powered down when the device enters Sleep mode. Clearing the ROSEL bit allows the reference output frequency to change as the system clock changes during any clock switches.

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R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
ROON	<u> </u>	ROSSLP	ROSEL ⁽¹⁾	RODIV3	RODIV2	RODIV1	RODIV0			
bit 7							bit			
Legend:										
R = Reada	ble bit	W = Writable	bit	U = Unimplen	nented bit, rea	d as '0'				
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unki	nown			
bit 7	ROON: Refe	rence Oscillato	r Output Enabl	le bit						
		e oscillator out e oscillator out								
bit 6	Unimplemer	ted: Read as '	0'							
bit 5	ROSSLP: Re	eference Oscilla	tor Output Sto	p in Sleep bit						
		e oscillator con e oscillator is d								
bit 4	ROSEL: Refe	ROSEL: Reference Oscillator Source Select bit ⁽¹⁾								
		oscillator (EC or lock is used as				ck switching of	the device			
bit 3-0	RODIV<3:0>	RODIV<3:0>: Reference Oscillator Divisor Select bits								
		1111 = Base clock value divided by 32,768								
	1110 = Base clock value divided by 16,384									
		1101 = Base clock value divided by 8,192 1100 = Base clock value divided by 4,096								
		1011 = Base clock value divided by 4,090								
		1010 = Base clock value divided by 1,024								
		1001 = Base clock value divided by 512								
		1000 = Base clock value divided by 256								
		0111 = Base clock value divided by 128 0110 = Base clock value divided by 64								
	0101 = Base clock value divided by 64 0101 = Base clock value divided by 32									
		clock value div	•							
		clock value div								
		clock value div	ided by 4							
		clock value div clock value div	ided by 4							

REGISTER 3-4: REFOCON: REFERENCE OSCILLATOR CONTROL REGISTER

Note 1: For ROSEL (REVOCON<4>), the primary oscillator is available only when configured as the default via the FOSC settings. This is regardless of whether the device is in Sleep mode.

3.8 Effects of Power-Managed Modes on the Various Clock Sources

When PRI_IDLE mode is selected, the designated primary oscillator continues to run without interruption. For all other power-managed modes, the oscillator using the OSC1 pin is disabled. The OSC1 pin (and OSC2 pin if used by the oscillator) will stop oscillating.

In secondary clock modes (SEC_RUN and SEC_IDLE), the SOSC oscillator is operating and providing the device clock. The SOSC oscillator may also run in all power-managed modes if required to clock SOSC.

In RC_RUN and RC_IDLE modes, the internal oscillator provides the device clock source. The 31 kHz LF-INTOSC output can be used directly to provide the clock and may be enabled to support various special features, regardless of the power-managed mode (see Section 28.2 "Watchdog Timer (WDT)" through Section 28.5 "Fail-Safe Clock Monitor" for more information on WDT, Fail-Safe Clock Monitor and Two-Speed Start-up).

If Sleep mode is selected, all clock sources are stopped. Since all the transistor switching currents have been stopped, Sleep mode achieves the lowest current consumption of the device (only leakage currents).

Enabling any on-chip feature that will operate during Sleep will increase the current consumed during Sleep. The INTOSC is required to support WDT operation. The SOSC oscillator may be operating to support a Real-Time Clock (RTC). Other features may be operating that do not require a device clock source (i.e., MSSP slave, INTx pins and others). Peripherals that may add significant current consumption are listed in Section 31.2 "DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/ Extended)".

3.9 Power-up Delays

Power-up delays are controlled by two timers, so that no external Reset circuitry is required for most applications. The delays ensure that the device is kept in Reset until the device power supply is stable under normal circumstances and the primary clock is operating and stable. For additional information on power-up delays, see **Section 5.6 "Power-up Timer (PWRT)**".

The first timer is the Power-up Timer (PWRT), which provides a fixed delay on a power-up time of about 64 ms (Parameter 33, Table 31-13); it is always enabled.

The second timer is the Oscillator Start-up Timer (OST), intended to keep the chip in Reset until the crystal oscillator is stable (HS, XT or LP modes). The OST does this by counting 1,024 oscillator cycles before allowing the oscillator to clock the device.

There is a delay of interval, TCSD (Parameter 38, Table 31-13), following POR, while the controller becomes ready to execute instructions.

TABLE 3-4:OSC1 AND OSC2 PIN STATES IN SLEEP MODE

Oscillator Mode	OSC1 Pin	OSC2 Pin		
EC, ECPLL	Floating, pulled by external clock	At logic low (clock/4 output)		
HS, HSPLL	Feedback inverter is disabled at quiescent voltage level	Feedback inverter is disabled at quiescent voltage level		
INTOSC, INTPLL1/2	I/O pin, RA6, direction is controlled by TRISA<6>	I/O pin, RA6, direction is controlled by TRISA<7>		

Note: See Section 5.0 "Reset" for time-outs due to Sleep and MCLR Reset.

NOTES:

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4.0 POWER-MANAGED MODES

The PIC18F87K22 family of devices offers a total of seven operating modes for more efficient power management. These modes provide a variety of options for selective power conservation in applications where resources may be limited (such as battery-powered devices).

There are three categories of power-managed mode:

- Run modes
- Idle modes
- Sleep mode

There is an Ultra Low-Power Wake-up (ULPWU) for waking from the Sleep mode.

These categories define which portions of the device are clocked, and sometimes, at what speed. The Run and Idle modes may use any of the three available clock sources (primary, secondary or internal oscillator block). The Sleep mode does not use a clock source.

The ULPWU mode, on the RA0 pin, enables a slow falling voltage to generate a wake-up, even from Sleep, without excess current consumption. (See Section 4.7 "Ultra Low-Power Wake-up".)

The power-managed modes include several powersaving features offered on previous PIC[®] devices. One is the clock switching feature, offered in other PIC18 devices. This feature allows the controller to use the SOSC oscillator instead of the primary one. Another power-saving feature is Sleep mode, offered by all PIC devices, where all device clocks are stopped.

4.1 Selecting Power-Managed Modes

Selecting a power-managed mode requires two decisions:

- · Will the CPU be clocked or not
- · What will be the clock source

The IDLEN bit (OSCCON<7>) controls CPU clocking, while the SCS<1:0> bits (OSCCON<1:0>) select the clock source. The individual modes, bit settings, clock sources and affected modules are summarized in Table 4-1.

4.1.1 CLOCK SOURCES

The SCS<1:0> bits select one of three clock sources for power-managed modes. Those sources are:

- The primary clock as defined by the FOSC<3:0> Configuration bits
- The secondary clock (the SOSC oscillator)
- The internal oscillator block (for LF-INTOSC modes)

4.1.2 ENTERING POWER-MANAGED MODES

Switching from one power-managed mode to another begins by loading the OSCCON register. The SCS<1:0> bits select the clock source and determine which Run or Idle mode is used. Changing these bits causes an immediate switch to the new clock source, assuming that it is running. The switch may also be subject to clock transition delays. These considerations are discussed in Section 4.1.3 "Clock Transitions and Status Indicators" and subsequent sections.

Entering the power-managed Idle or Sleep modes is triggered by the execution of a SLEEP instruction. The actual mode that results depends on the status of the IDLEN bit.

Depending on the current and impending mode, a change to a power-managed mode does not always require setting all of the previously discussed bits. Many transitions can be done by changing the oscillator select bits, or changing the IDLEN bit, prior to issuing a SLEEP instruction. If the IDLEN bit is already configured as desired, it may only be necessary to perform a SLEEP instruction to switch to the desired mode.

Mode	oscco	OSCCON Bits		Clocking	Available Clock and Oscillator Source
Mode	IDLEN<7> ⁽¹⁾	SCS<1:0>	CPU	Peripherals	Available Clock and Oscillator Source
Sleep	0	N/A	Off	Off	None – All clocks are disabled
PRI_RUN	N/A	00	Clocked	Clocked	Primary – XT, LP, HS, EC, RC and PLL modes. This is the normal, Full-Power Execution mode.
SEC_RUN	N/A	01	Clocked	Clocked	Secondary – SOSC Oscillator
RC_RUN	N/A	1x	Clocked	Clocked	Internal oscillator block ⁽²⁾
PRI_IDLE	1	00	Off	Clocked	Primary – LP, XT, HS, RC, EC
SEC_IDLE	1	01	Off	Clocked	Secondary – SOSC oscillator
RC_IDLE	1	lx	Off	Clocked	Internal oscillator block ⁽²⁾

TABLE 4-1:POWER-MANAGED MODES

Note 1: IDLEN reflects its value when the **SLEEP** instruction is executed.

^{2:} Includes INTOSC (HF-INTOSC and MG-INTOSC) and INTOSC postscaler, as well as the LF-INTOSC source.

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4.1.3 CLOCK TRANSITIONS AND STATUS INDICATORS

The length of the transition between clock sources is the sum of two cycles of the old clock source and three to four cycles of the new clock source. This formula assumes that the new clock source is stable. The HF-INTOSC and MF-INTOSC are termed as INTOSC in this chapter.

Three bits indicate the current clock source and its status, as shown in Table 4-2. The three bits are:

- OSTS (OSCCON<3>)
- HFIOFS (OSCCON<2>)
- SOSCRUN (OSCCON2<6>)

TABLE 4-2: SYSTEM CLOCK INDICATOR

Main Clock Source	OSTS	HFIOFSor MFIOFS	SOSCRUN
Primary Oscillator	1	0	0
INTOSC (HF-INTOSC or MF-INTOSC)	0	1	0
Secondary Oscillator	0	0	1
MF-INTOSC or HF-INTOSC as Primary Clock Source	1	1	0
LF-INTOSC is Running or INTOSC is Not Yet Stable	0	0	0

When the OSTS bit is set, the primary clock is providing the device clock. When the HFIOFS or MFIOFS bit is set, the INTOSC output is providing a stable 16 MHz clock source to a divider that actually drives the device clock. When the SOSCRUN bit is set, the SOSC oscillator is providing the clock. If none of these bits are set, either the LF-INTOSC clock source is clocking the device or the INTOSC source is not yet stable.

If the internal oscillator block is configured as the primary clock source by the FOSC<3:0> Configuration bits (CONFIG1H<3:0>), then the OSTS and HFIOFS or MFIOFS bits can be set when in PRI_RUN or PRI_IDLE modes. This indicates that the primary clock (INTOSC output) is generating a stable 16 MHz output. Entering another INTOSC power-managed mode at the same frequency would clear the OSTS bit.

- Note 1: Caution should be used when modifying a single IRCF bit. At a lower VDD, it is possible to select a higher clock speed than is supportable by that VDD. Improper device operation may result if the VDD/ FOSC specifications are violated.
 - 2: Executing a SLEEP instruction does not necessarily place the device into Sleep mode. It acts as the trigger to place the controller into either the Sleep mode or one of the Idle modes, depending on the setting of the IDLEN bit.

4.1.4 MULTIPLE SLEEP COMMANDS

The power-managed mode that is invoked with the SLEEP instruction is determined by the setting of the IDLEN bit at the time the instruction is executed. If another SLEEP instruction is executed, the device will enter the power-managed mode specified by IDLEN at that time. If IDLEN has changed, the device will enter the new power-managed mode specified by the new setting.

4.2 Run Modes

In the Run modes, clocks to both the core and peripherals are active. The difference between these modes is the clock source.

4.2.1 PRI_RUN MODE

The PRI_RUN mode is the normal, Full-Power Execution mode of the microcontroller. This is also the default mode upon a device Reset, unless Two-Speed Start-up is enabled. (For details, see Section 28.4 "Two-Speed Start-up".) In this mode, the OSTS bit is set. The HFIOFS or MFIOFS bit may be set if the internal oscillator block is the primary clock source. (See Section 3.2 "Control Registers".)

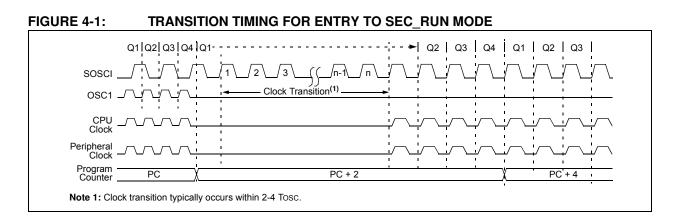
4.2.2 SEC_RUN MODE

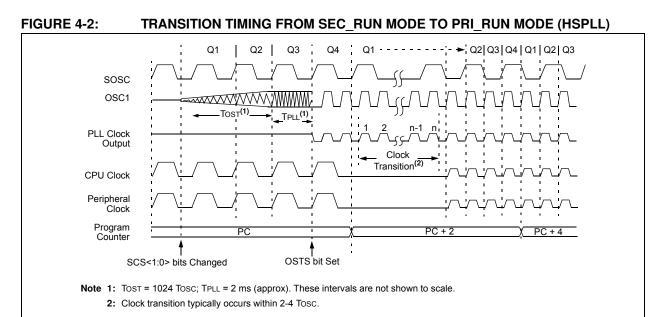
The SEC_RUN mode is the compatible mode to the "clock-switching" feature offered in other PIC18 devices. In this mode, the CPU and peripherals are clocked from the SOSC oscillator. This enables lower power consumption while retaining a high-accuracy clock source.

SEC_RUN mode is entered by setting the SCS<1:0> bits to '01'. The device clock source is switched to the SOSC oscillator (see Figure 4-1), the primary oscillator is shut down, the SOSCRUN bit (OSCCON2<6>) is set and the OSTS bit is cleared.

Note:	The SOSC oscillator can be enabled by setting the SOSCGO bit (OSCCON2<3>). If this bit is set, the clock switch to the SEC RUN mode can switch immediately
	once SCS<1:0> are set to '01'.

On transitions from SEC_RUN mode to PRI_RUN mode, the peripherals and CPU continue to be clocked from the SOSC oscillator while the primary clock is started. When the primary clock becomes ready, a clock switch back to the primary clock occurs (see Figure 4-2). When the clock switch is complete, the SOSCRUN bit is cleared, the OSTS bit is set and the primary clock is providing the clock. The IDLEN and SCS bits are not affected by the wake-up and the SOSC oscillator continues to run.





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4.2.3 RC_RUN MODE

In RC_RUN mode, the CPU and peripherals are clocked from the internal oscillator block using the INTOSC multiplexer. In this mode, the primary clock is shut down. When using the LF-INTOSC source, this mode provides the best power conservation of all the Run modes, while still executing code. It works well for user applications which are not highly timing-sensitive or do not require high-speed clocks at all times.

If the primary clock source is the internal oscillator block – either LF-INTOSC or INTOSC (MF-INTOSC or HF-INTOSC) – there are no distinguishable differences between the PRI_RUN and RC_RUN modes during execution. Entering or exiting RC_RUN mode, however, causes a clock switch delay. Therefore, if the primary clock source is the internal oscillator block, using RC_RUN mode is not recommended.

This mode is entered by setting the SCS1 bit to '1'. To maintain software compatibility with future devices, it is recommended that the SCS0 bit also be cleared, even though the bit is ignored. When the clock source is switched to the INTOSC multiplexer (see Figure 4-3),

the primary oscillator is shut down and the OSTS bit is cleared. The IRCF bits may be modified at any time to immediately change the clock speed.

Note:	Caution should be used when modifying a
	single IRCF bit. At a lower VDD, it is
	possible to select a higher clock speed
	than is supportable by that VDD. Improper
	device operation may result if the VDD/
	Fosc specifications are violated.

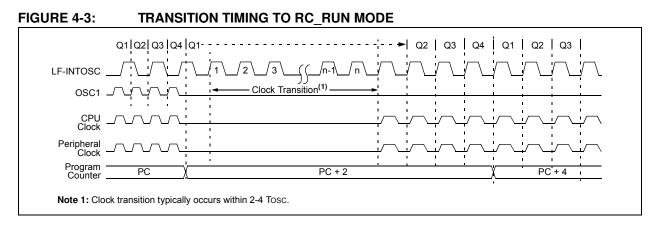
If the IRCF bits and the INTSRC bit are all clear, the INTOSC output (HF-INTOSC/MF-INTOSC) is not enabled and the HFIOFS and MFIOFS bits will remain clear. There will be no indication of the current clock source. The LF-INTOSC source is providing the device clocks.

If the IRCF bits are changed from all clear (thus, enabling the INTOSC output), or if INTSRC or MFIOSEL is set, the HFIOFS or MFIOFS bit is set after the INTOSC output becomes stable. For details, see Table 4-3.

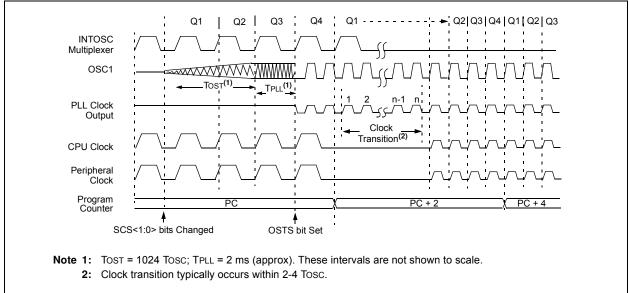
IRCF<2:0	>	INTSRC	MFIOSEL	Status of MFIOFS or HFIOFS when INTOSC is Stable
000		0	x	MFIOFS = 0, HFIOFS = 0 and clock source is LF-INTOSC
000		1	0	MFIOFS = 0, HFIOFS = 1 and clock source is HF-INTOSC
000		1	1	MFIOFS = 1, HFIOFS = 0 and clock source is MF-INTOSC
Non-Zero)	x	0	MFIOFS = 0, HFIOFS = 1 and clock source is HF-INTOSC
Non-Zero)	x	1	MFIOFS = 1, HFIOFS = 0 and clock source is MF-INTOSC

Clocks to the device continue while the INTOSC source stabilizes after an interval of TIOBST (Parameter 39, Table 31-13).

If the IRCF bits were previously at a non-zero value, or if INTSRC was set before setting SCS1 and the INTOSC source was already stable, the HFIOFS or MFIOFS bit will remain set. On transitions from RC_RUN mode to PRI_RUN mode, the device continues to be clocked from the INTOSC multiplexer while the primary clock is started. When the primary clock becomes ready, a clock switch to the primary clock occurs (see Figure 4-4). When the clock switch is complete, the HFIOFS or MFIOFS bit is cleared, the OSTS bit is set and the primary clock is providing the device clock. The IDLEN and SCS bits are not affected by the switch. The LF-INTOSC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.







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4.3 Sleep Mode

The power-managed Sleep mode in the PIC18F87K22 family of devices is identical to the legacy Sleep mode offered in all other PIC devices. It is entered by clearing the IDLEN bit (the default state on device Reset) and executing the SLEEP instruction. This shuts down the selected oscillator (Figure 4-5). All clock source status bits are cleared.

Entering Sleep mode from any other mode does not require a clock switch. This is because no clocks are needed once the controller has entered Sleep. If the WDT is selected, the LF-INTOSC source will continue to operate. If the SOSC oscillator is enabled, it will also continue to run.

When a wake event occurs in Sleep mode (by interrupt, Reset or WDT time-out), the device will not be clocked until the clock source selected by the SCS<1:0> bits becomes ready (see Figure 4-6). Alternately, the device will be clocked from the internal oscillator block if either the Two-Speed Start-up or the Fail-Safe Clock Monitor is enabled (see Section 28.0 "Special Features of the CPU"). In either case, the OSTS bit is set when the primary clock is providing the device clocks. The IDLEN and SCS bits are not affected by the wake-up.

4.4 Idle Modes

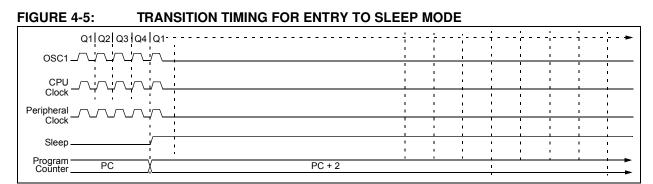
The Idle modes allow the controller's CPU to be selectively shut down while the peripherals continue to operate. Selecting a particular Idle mode allows users to further manage power consumption.

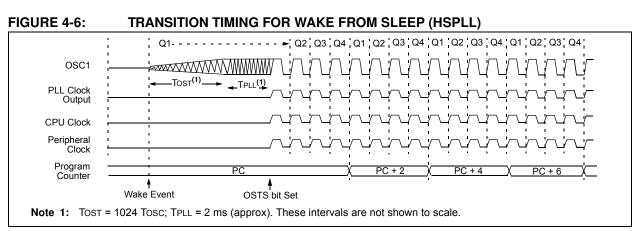
If the IDLEN bit is set to a '1' when a SLEEP instruction is executed, the peripherals will be clocked from the clock source selected using the SCS<1:0> bits. The CPU, however, will not be clocked. The clock source status bits are not affected. This approach is a quick method to switch from a given Run mode to its corresponding Idle mode.

If the WDT is selected, the LF-INTOSC source will continue to operate. If the SOSC oscillator is enabled, it will also continue to run.

Since the CPU is not executing instructions, the only exits from any of the Idle modes are by interrupt, WDT time-out or a Reset. When a wake event occurs, CPU execution is delayed by an interval of TCSD (Parameter 38, Table 31-13) while it becomes ready to execute code. When the CPU begins executing code, it resumes with the same clock source for the current Idle mode. For example, when waking from RC_IDLE mode, the internal oscillator block will clock the CPU and peripherals (in other words, RC_RUN mode). The IDLEN and SCS bits are not affected by the wake-up.

While in any Idle mode or Sleep mode, a WDT timeout will result in a WDT wake-up to the Run mode currently specified by the SCS<1:0> bits.





4.4.1 PRI_IDLE MODE

This mode is unique among the three low-power Idle modes, in that it does not disable the primary device clock. For timing-sensitive applications, this allows for the fastest resumption of device operation with its more accurate, primary clock source, since the clock source does not have to "warm-up" or transition from another oscillator.

PRI_IDLE mode is entered from PRI_RUN mode by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set IDLEN first, then clear the SCS bits and execute SLEEP. Although the CPU is disabled, the peripherals continue to be clocked from the primary clock source specified by the FOSC<3:0> Configuration bits. The OSTS bit remains set (see Figure 4-7).

When a wake event occurs, the CPU is clocked from the primary clock source. A delay of interval, TCSD (Parameter 39, Table 31-13), is required between the wake event and the start of code execution. This is required to allow the CPU to become ready to execute instructions. After the wake-up, the OSTS bit remains set. The IDLEN and SCS bits are not affected by the wake-up (see Figure 4-8).

4.4.2 SEC_IDLE MODE

In SEC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the SOSC oscillator. This mode is entered from SEC_RUN by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, set the IDLEN bit first, then set the SCS<1:0> bits to '01' and execute SLEEP. When the clock source is switched to the SOSC oscillator, the primary oscillator is shut down, the OSTS bit is cleared and the SOSCRUN bit is set.

When a wake event occurs, the peripherals continue to be clocked from the SOSC oscillator. After an interval of TCSD following the wake event, the CPU begins executing code being clocked by the SOSC oscillator. The IDLEN and SCS bits are not affected by the wakeup and the SOSC oscillator continues to run (see Figure 4-8).

FIGURE 4-7: TRANSITION TIMING FOR ENTRY TO IDLE MODE

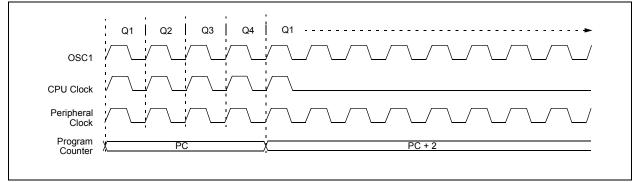
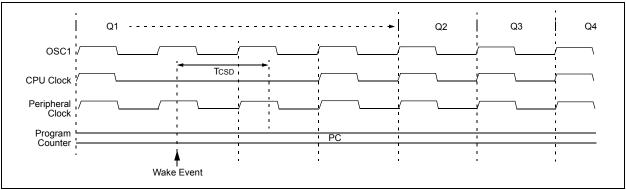


FIGURE 4-8: TRANSITION TIMING FOR WAKE FROM IDLE TO RUN MODE



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4.4.3 RC_IDLE MODE

In RC_IDLE mode, the CPU is disabled but the peripherals continue to be clocked from the internal oscillator block using the INTOSC multiplexer. This mode provides controllable power conservation during Idle periods.

From RC_RUN, this mode is entered by setting the IDLEN bit and executing a SLEEP instruction. If the device is in another Run mode, first set IDLEN, then set the SCS1 bit and execute SLEEP. To maintain software compatibility with future devices, it is recommended that SCS0 also be cleared, though its value is ignored. The INTOSC multiplexer may be used to select a higher clock frequency by modifying the IRCF bits before executing the SLEEP instruction. When the clock source is switched to the INTOSC multiplexer, the primary oscillator is shut down and the OSTS bit is cleared.

If the IRCF bits are set to any non-zero value, or the INTSRC/MFIOSEL bit is set, the INTOSC output is enabled. The HFIOFS/MFIOFS bits become set, after the INTOSC output becomes stable, after an interval of TIOBST (Parameter 38, Table 31-13). (For information on the HFIOFS/MFIOFS bits, see Table 4-3.)

Clocks to the peripherals continue while the INTOSC source stabilizes. The HFIOFS/MFIOFS bits will remain set if the IRCF bits were previously at a non-zero value or if INTSRC was set before the SLEEP instruction was executed and the INTOSC source was already stable. If the IRCF bits and INTSRC are all clear, the INTOSC output will not be enabled, the HFIOFS/MFIOFS bits will remain clear and there will be no indication of the current clock source.

When a wake event occurs, the peripherals continue to be clocked from the INTOSC multiplexer. After a delay of TCSD (Parameter 38, Table 31-13) following the wake event, the CPU begins executing code clocked by the INTOSC multiplexer. The IDLEN and SCS bits are not affected by the wake-up. The INTRC source will continue to run if either the WDT or the Fail-Safe Clock Monitor is enabled.

4.5 Selective Peripheral Module Control

Idle mode allows users to substantially reduce power consumption by stopping the CPU clock. Even so, peripheral modules still remain clocked, and thus, consume power. There may be cases where the application needs what this mode does not provide: the allocation of power resources to the CPU processing with minimal power consumption from the peripherals.

PIC18F87K22 family devices address this requirement by allowing peripheral modules to be selectively disabled, reducing or eliminating their power consumption. This can be done with two control bits:

- Peripheral Enable bit, generically named XXXEN Located in the respective module's main control register
- Peripheral Module Disable (PMD) bit, generically named, XXXMD – Located in one of the PMDx Control registers (PMD0, PMD1, PMD2 or PMD3)

Disabling a module by clearing its XXXEN bit disables the module's functionality, but leaves its registers available to be read and written to. This reduces power consumption, but not by as much as the second approach.

Most peripheral modules have an enable bit.

In contrast, setting the PMD bit for a module disables all clock sources to that module, reducing its power consumption to an absolute minimum. In this state, the control and status registers associated with the peripheral are also disabled, so writes to those registers have no effect and read values are invalid. Many peripheral modules have a corresponding PMD bit.

There are four PMD registers in the PIC18F87K22 family devices: PMD0, PMD1, PMD2 and PMD3. These registers have bits associated with each module for disabling or enabling a particular peripheral.

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REGISTER 4-1: PMD3: PERIPHERAL MODULE DISABLE REGISTER 3

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CCP10MD ⁽¹⁾	CCP9MD ⁽¹⁾	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD ⁽¹⁾
bit 7							bit C
Legend:							
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unk	nown
bit 7		MD CCP10 Er					
				enabled for CC	P10, disabling a	all of its clock :	sources
		lisabled for CC		(1)			
bit 6		ID CCP9 Enab			D0 diaabling al		
		lisabled for CC		enabled for CC	P9, disabling al	I OF ILS CIOCK SO	Durces
bit 5		ID CCP8 Enab					
	1 = Periphera	al Module Disa	ble (PMD) is e	enabled for CC	P8, disabling al	I of its clock so	ources
	0 = PMD is d	isabled for CC	P8 ` ´				
bit 4	CCP7MD: PN	ID CCP7 Enab	le/Disable bit				
				enabled for CC	P7, disabling al	I of its clock so	ources
1.11.0		lisabled for CC					
bit 3		ID CCP6 Enab			P6, disabling al	l of ito plank of	
	•	lisabled for CC	• •		Po, disability al	I OF ILS CIUCK SU	Juices
bit 2		ID CCP5 Enab					
	1 = Periphera	al Module Disa	ble (PMD) is e	enabled for CC	P5, disabling al	I of its clock so	ources
	0 = PMD is d	isabled for CC	P5		-		
bit 1	CCP4MD: PN	ID CCP4 Enab	le/Disable bit				
				enabled for CC	P4, disabling al	I of its clock so	ources
		lisabled for CC					
bit 0		MR12MD Disa					
		nabled and all lisabled and TM	•••••		e disabled		

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

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TMR10MD ⁽¹⁾ bit 7	TMR8MD	TMR7MD ⁽¹⁾	TMR6MD				
oit 7				TMR5MD	CMP3MD	CMP2MD	CMP1MD
							bit (
<u> </u>							
Legend:							
R = Readable		W = Writable bi	t	•	nented bit, read		
-n = Value at POR		'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
oit 7	TMR10MD: 1	MR10MD Disab	le bit ⁽¹⁾				
		al Module Disabl			TMR10MD clo	ck sources are	disabled
oit 6	TMR8MD: T	MR8MD Disable	bit				
		enabled and all T disabled and TMF			disabled		
oit 5	TMR7MD: T	MR7MD Disable	bit ⁽¹⁾				
		enabled and all T disabled and TMF			disabled		
oit 4	TMR6MD: T	MR6MD Disable	bit				
		enabled and all T disabled and TMF			disabled		
oit 3	TMR5MD: T	MR5MD Disable	bit				
		enabled and all T disabled and TMF			disabled		
oit 2	CMP3MD: PI	MD Comparator 3	3 Enable/Dis	able bit			
		enabled for Comp disabled for Com		abling all of its	clock sources		
pit 1	CMP2MD: PI	MD Comparator 3	3 Enable/Dis	able bit			
		enabled for Comp disabled for Com		abling all of its	clock sources		
oit 0	CMP1MD: PI	MD Comparator 3	3 Enable/Dis	able bit			
		enabled for Comp disabled for Com		abling all of its	clock sources		

REGISTER 4-2: PMD2: PERIPHERAL MODULE DISABLE REGISTER 2

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

REGISTER 4-3: PMD1: PERIPHERAL MODULE DISABLE REGISTER 1

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PSPMD	CTMUMD	RTCCMD ⁽¹⁾	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD
bit 7							bit (
Legend:							
R = Readab		W = Writable b	bit	-	nented bit, reac		
-n = Value a	t POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkn	iown
h:+ 7		inherel Medule I			Dischlahit		
bit 7		ipheral Module I nabled for PSP,	•	,			
		lisabled for PSP	uisabiiriy ali		ices		
bit 6	CTMUMD: P	MD CTMU Enal	ole/Disable bi	t			
	1 = PMD is e	enabled for CTM	IU, disabling	all of its clock s	sources		
		disabled for CTN	-				
bit 5		MD RTCC Enab					
		enabled for RTC		all of its clock s	ources		
hit 4		disabled for RTC MR4MD Disable	-				
bit 4		enabled and all	2.1	ck sources are	disabled		
	-	disabled and TM			uisableu		
bit 3	TMR3MD: T	MR3MD Disable	bit				
	1 = PMD is e	enabled and all	TMR3MD clo	ck sources are	disabled		
	0 = PMD is 0	disabled and TM	IR3MD is ena	bled			
bit 2		MR2MD Disable					
	-	enabled and all [·] disabled and TM			disabled		
bit 1		MR1MD Disable		ibieu			
		enabled and all		ck sources are	disabled		
		disabled and TM			alsablea		
bit 0	EMBMD: PM	1D EMB Enable/	Disable bit				
		nabled for EMB	•	of its clock sou	urces		
	0 = PMD is d	lisabled for EMB					

Clock and Calendar (RTCC)" for the unlock sequence (Example 18-1).

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R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD
bit 7				1			bit 0
Legend:							
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'			
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown	
bit 7	CCP3MD: PMD ECCP3 Enable/Disable bit 1 = Peripheral Module Disable (PMD) is enabled for ECCP3, disabling all of its clock sources 0 = PMD is disabled for ECCP3						
bit 6	CCP2MD: PMD ECCP2 Enable/Disable bit 1 = PMD is enabled for ECCP2, disabling all of its clock sources 0 = PMD is disabled for ECCP2						
bit 5	CCP1MD: PMD ECCP1 Enable/Disable bit 1 = PMD is enabled for ECCP1, disabling all of its clock sources 0 = PMD is disabled for ECCP1						
bit 4	UART2MD: PMD UART2 Enable/Disable bit 1 = PMD is enabled for UART2, disabling all of its clock sources 0 = PMD is disabled for UART2						
bit 3	UART1MD: PMD UART1 Enable/Disable bit 1 = PMD is enabled for UART1, disabling all of its clock sources 0 = PMD is disabled for UART1						
bit 2	SSP2MD: PMD MSSP2 Enable/Disable bit 1 = PMD is enabled for MSSP2, disabling all of its clock sources 0 = PMD is disabled for MSSP2						
bit 1	 SSP1MD: PMD MSSP1 Enable/Disable bit 1 = PMD is enabled for MSSP1, disabling all of its clock sources 0 = PMD is disabled for MSSP1 						
bit 0	ADCMD: PMD Analog/Digital Converter PMD Enable/Disable bit 1 = PMD is enabled for the Analog/Digital Converter, disabling all of its clock sources 0 = PMD is disabled for the Analog/Digital Converter						

REGISTER 4-4: PMD0: PERIPHERAL MODULE DISABLE REGISTER 0

4.6 Exiting Idle and Sleep Modes

An exit from Sleep mode or any of the Idle modes is triggered by an interrupt, a Reset or a WDT time-out. This section discusses the triggers that cause exits from power-managed modes. The clocking subsystem actions are discussed in each of the power-managed modes (see Section 4.2 "Run Modes", Section 4.3 "Sleep Mode" and Section 4.4 "Idle Modes").

4.6.1 EXIT BY INTERRUPT

Any of the available interrupt sources can cause the device to exit from an Idle mode or Sleep mode to a Run mode. To enable this functionality, an interrupt source must be enabled by setting its enable bit in one of the INTCONx or PIEx registers. The exit sequence is initiated when the corresponding interrupt flag bit is set.

On all exits from Idle or Sleep modes by interrupt, code execution branches to the interrupt vector if the GIE/ GIEH bit (INTCON<7>) is set. Otherwise, code execution continues or resumes without branching (see **Section 11.0 "Interrupts"**).

4.6.2 EXIT BY WDT TIME-OUT

A WDT time-out will cause different actions depending on which power-managed mode the device is in when the time-out occurs.

If the device is not executing code (all Idle modes and Sleep mode), the time-out will result in an exit from the power-managed mode (see Section 4.2 "Run Modes" and Section 4.3 "Sleep Mode"). If the device is executing code (all Run modes), the time-out will result in a WDT Reset (see Section 28.2 "Watchdog Timer (WDT)").

Executing a SLEEP or CLRWDT instruction clears the WDT timer and postscaler, loses the currently selected clock source (if the Fail-Safe Clock Monitor is enabled) and modifies the IRCF bits in the OSCCON register (if the internal oscillator block is the device clock source).

4.6.3 EXIT BY RESET

Normally, the device is held in Reset by the Oscillator Start-up Timer (OST) until the primary clock becomes ready. At that time, the OSTS bit is set and the device begins executing code. If the internal oscillator block is the new clock source, the HFIOFS/MFIOFS bits are set instead.

The exit delay time from Reset to the start of code execution depends on both the clock sources before and after the wake-up, and the type of oscillator if the new clock source is the primary clock. Exit delays are summarized in Table 4-4.

Code execution can begin before the primary clock becomes ready. If either the Two-Speed Start-up (see Section 28.4 "Two-Speed Start-up") or Fail-Safe Clock Monitor (see Section 28.5 "Fail-Safe Clock Monitor") is enabled, the device may begin execution as soon as the Reset source has cleared. Execution is clocked by the INTOSC multiplexer driven by the internal oscillator block. Execution is clocked by the internal oscillator block until either the primary clock becomes ready or a power-managed mode is entered before the primary clock becomes ready; the primary clock is then shut down.

4.6.4 EXIT WITHOUT AN OSCILLATOR START-UP DELAY

Certain exits from power-managed modes do not invoke the OST at all. The two cases are:

- When in PRI_IDLE mode, where the primary clock source is not stopped
- When the primary clock source is not any of the LP, XT, HS or HSPLL modes

In these instances, the primary clock source either does not require an oscillator start-up delay, since it is already running (PRI_IDLE), or normally does not require an oscillator start-up delay (RC, EC and INTIO Oscillator modes). However, a fixed delay of interval, TCSD, following the wake event, is still required when leaving Sleep and Idle modes to allow the CPU to prepare for execution. Instruction execution resumes on the first clock cycle following this delay.

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4.7 Ultra Low-Power Wake-up

The Ultra Low-Power Wake-up (ULPWU) on pin, RA0, allows a slow falling voltage to generate an interrupt without excess current consumption.

To use this feature:

- 1. Charge the capacitor on RA0 by configuring the RA0 pin to an output and setting it to '1'.
- 2. Stop charging the capacitor by configuring RA0 as an input.
- 3. Discharge the capacitor by setting the ULPEN and ULPSINK bits in the WDTCON register.
- 4. Configure Sleep mode.
- 5. Enter Sleep mode.

When the voltage on RA0 drops below VIL, the device wakes up and executes the next instruction.

This feature provides a low-power technique for periodically waking up the device from Sleep mode.

The time-out is dependent on the discharge time of the RC circuit on RA0.

When the ULPWU module wakes the device from Sleep mode, the ULPLVL bit (WDTCON<5>) is set. Software can check this bit upon wake-up to determine the wake-up source.

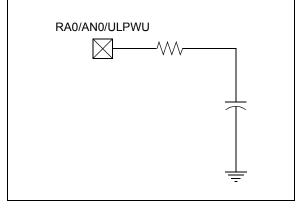
See Example 4-1 for initializing the ULPWU module.

EXAMPLE 4-1: ULTRA LOW-POWER WAKE-UP INITIALIZATION

```
//Charge the capacitor on RAO
       TRISAbits.TRISA0 = 0;
PORTAbits.RA0 = 1;
for(i = 0; i < 10000; i++) Nop();</pre>
       //Stop Charging the capacitor
       //on RAO
       TRISAbits.TRISA0 = 1;
       //Enable the Ultra Low Power
       //Wakeup module and allow
       //capacitor discharge
       WDTCONbits ULPEN = 1;
WDTCONbits.ULPSINK = 1;
       //For Sleep
OSCCONDits.IDLEN = 0;
       //Enter Sleep Mode
       11
Sleep();
       //for sleep, execution will
       //resume here
```

A series resistor, between RA0 and the external capacitor, provides overcurrent protection for the RA0/AN0/ ULPWU pin and enables software calibration of the time-out (see Figure 4-9).

FIGURE 4-9: ULTRA LOW-POWER WAKE-UP INITIALIZATION



A timer can be used to measure the charge time and discharge time of the capacitor. The charge time can then be adjusted to provide the desired delay in Sleep. This technique compensates for the affects of temperature, voltage and component accuracy. The peripheral can also be configured as a simple Programmable Low-Voltage Detect (LVD) or temperature sensor.

Note: For more information, see AN879, "Using the Microchip Ultra Low-Power Wake-up Module" (DS00879).

TABLE 4-4:EXIT DELAY ON WAKE-UP BY RESET FROM SLEEP MODE OR ANY IDLE MODE
(BY CLOCK SOURCES)

Power-Managed Mode	Clock Source ⁽⁵⁾	Exit Delay	Clock Ready Status Bits	
	LP, XT, HS		OSTS	
	HSPLL			
PRI_IDLE mode	EC, RC	Tcsp ⁽¹⁾		
	HF-INTOSC ⁽²⁾		HFIOFS	
	MF-INTOSC ⁽²⁾		MFIOFS	
	LF-INTOSC		None	
SEC_IDLE mode	SOSC	TCSD ⁽¹⁾	SOSCRUN	
	HF-INTOSC ⁽²⁾		HFIOFS	
RC_IDLE mode	MF-INTOSC ⁽²⁾	TCSD ⁽¹⁾	MFIOFS	
	LF-INTOSC		None	
	LP, XT, HS	Tost ⁽³⁾	OSTS	
	HSPLL	Tost + t _{rc} ⁽³⁾		
Sloop mode	EC, RC	TCSD ⁽¹⁾		
Sleep mode	HF-INTOSC ⁽²⁾	HF-INTOSC ⁽²⁾		
	MF-INTOSC ⁽²⁾	TIOBST ⁽⁴⁾	MFIOFS	
	LF-INTOSC		None	

Note 1: TCSD (Parameter 38, Table 31-13) is a required delay when waking from Sleep and all Idle modes, and runs concurrently with any other required delays (see Section 4.4 "Idle Modes").

2: Includes postscaler derived frequencies. On Reset, INTOSC defaults to HF-INTOSC at 8 MHz.

3: TOST is the Oscillator Start-up Timer (Parameter 32, Table 31-13). TRC is the PLL Lock-out Timer (Parameter F12, Table 31-7); it is also designated as TPLL.

- 4: Execution continues during TIOBST (Parameter 39, Table 31-13), the INTOSC stabilization period.
- **5:** The clock source is dependent upon the settings of the SCS (OSCCON<1:0>), IRCF (OSCCON<6:4>) and FOSC (CONFIG1H<3:0>) bits.

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NOTES:

5.0 RESET

The PIC18F87K22 family of devices differentiates between various kinds of Reset:

- a) Power-on Reset (POR)
- b) MCLR Reset during normal operation
- c) MCLR Reset during power-managed modes
- d) Watchdog Timer (WDT) Reset (during execution)
- e) Configuration Mismatch (CM) Reset
- f) Brown-out Reset (BOR)
- g) RESET Instruction
- h) Stack Full Reset
- i) Stack Underflow Reset

This section discusses Resets generated by MCLR, POR and BOR, and covers the operation of the various start-up timers. Stack Reset events are covered in **Section 6.1.3.4 "Stack Full and Underflow Resets"**. WDT Resets are covered in **Section 28.2 "Watchdog Timer (WDT)**".

A simplified block diagram of the on-chip Reset circuit is shown in Figure 5-1.

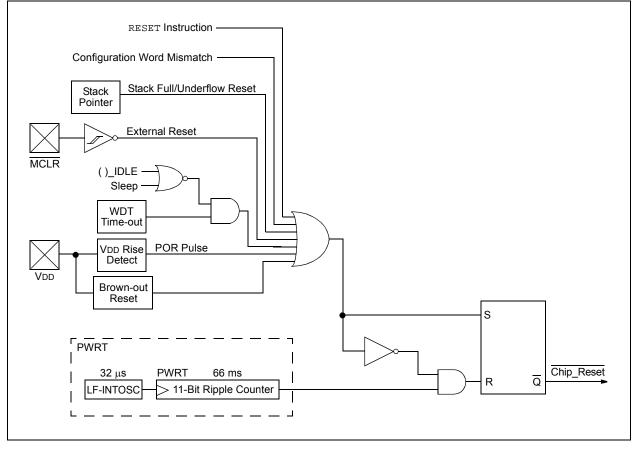
5.1 RCON Register

Device Reset events are tracked through the RCON register (Register 5-1). The lower five bits of the register indicate that a specific Reset event has occurred. In most cases, these bits can only be set by the event and must be cleared by the application after the event.

The state of these flag bits, taken together, can be read to indicate the type of Reset that just occurred. This is described in more detail in **Section 5.7** "**Reset State of Registers**".

The RCON register also has a control bit for setting interrupt priority (IPEN). Interrupt priority is discussed in **Section 11.0 "Interrupts**".

FIGURE 5-1: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



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R/W-1 R/W-1 R/W-1 R-1 R-1 R/W-0 R/W-0 R/W-0 CM RI TO PD POR **IPEN** SBOREN BOR bit 7 bit 0 Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7 IPEN: Interrupt Priority Enable bit 1 = Enable priority levels on interrupts 0 = Disable priority levels on interrupts (PIC16CXXX Compatibility mode) bit 6 SBOREN: BOR Software Enable bit If BOREN<1:0> = 01: 1 = BOR is enabled 0 = BOR is disabled If BOREN<1:0> = 00. 10 or 11: Bit is disabled and read as '0'. CM: Configuration Mismatch Flag bit bit 5 1 = A Configuration Mismatch Reset has not occurred 0 = A Configuration Mismatch Reset has occurred (must be set in software after a Configuration Mismatch Reset occurs) bit 4 RI: RESET Instruction Flag bit 1 = The RESET instruction was not executed (set by firmware only) 0 = The RESET instruction was executed causing a device Reset (must be set in software after a Brown-out Reset occurs) TO: Watchdog Time-out Flag bit bit 3 1 = Set by power-up, CLRWDT instruction or SLEEP instruction 0 = A WDT time-out occurred bit 2 PD: Power-Down Detection Flag bit 1 = Set by power-up or by the CLRWDT instruction 0 = Set by execution of the SLEEP instruction POR: Power-on Reset Status bit bit 1 1 = A Power-on Reset has not occurred (set by firmware only) 0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs) bit 0 BOR: Brown-out Reset Status bit 1 = A Brown-out Reset has not occurred (set by firmware only) 0 = A Brown-out Reset occurred (must be set in software after a Brown-out Reset occurs)

REGISTER 5-1: RCON: RESET CONTROL REGISTER

Note 1: It is recommended that the POR bit be set after a Power-on Reset has been detected, so that subsequent Power-on Resets may be detected.

2: Brown-out Reset is said to have occurred when BOR is '0' and POR is '1' (assuming that POR was set to '1' by software immediately after a Power-on Reset).

5.2 Master Clear (MCLR)

The MCLR pin provides a method for triggering a hard external Reset of the device. A Reset is generated by holding the pin low. PIC18 extended microcontroller devices have a noise filter in the MCLR Reset path which detects and ignores small pulses.

The $\overline{\text{MCLR}}$ pin is not driven low by any internal Resets, including the WDT.

5.3 Power-on Reset (POR)

A Power-on Reset condition is generated on-chip whenever VDD rises above a certain threshold. This allows the device to start in the initialized state when VDD is adequate for operation.

To take advantage of the POR circuitry, tie the $\overline{\text{MCLR}}$ pin through a resistor (1 k Ω to 10 k Ω) to VDD. This will eliminate external RC components usually needed to create a Power-on Reset delay. A minimum rise rate for VDD is specified (Parameter D004). For a slow rise time, see Figure 5-2.

When the device starts normal operation (exiting the Reset condition), device operating parameters (such as voltage, frequency and temperature) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met.

Power-on Reset events are captured by the \overrightarrow{POR} bit (RCON<1>). The state of the bit is set to '0' whenever a Power-on Reset occurs and does not change for any other Reset event. \overrightarrow{POR} is not reset to '1' by any hardware event. To capture multiple events, the user manually resets the bit to '1' in software following any Power-on Reset.

5.4 Brown-out Reset (BOR)

The PIC18F87K22 family has four BOR Power modes:

- High-Power BOR
- Medium Power BOR
- Low-Power BOR
- Zero-Power BOR

Each power Mode is selected by the BORPWR<1:0> bits setting (CONFIG2L<6:5>). For low, medium and high-power BOR, the module monitors the VDD depending on the BORV<1:0> setting (CONFIG1L<3:2>). A BOR event re-arms the Power-on Reset. It also causes a Reset, depending on which of the trip levels has been set: 1.8V, 2V, 2.7V or 3V.

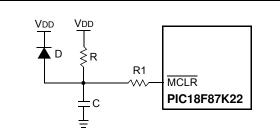
BOR is enabled by the BOREN<1:0> bits (CONFIG2L<2:1>) and the SBOREN bit (RCON<6>). Typical power consumption is listed as Parameter D022A in Section 31.0 "Electrical Characteristics".

In Zero-Power BOR (ZPBORMV), the module monitors the VDD voltage and re-arms the POR at about 2V. ZPBORMV does not cause a Reset, but re-arms the POR.

The BOR accuracy varies with its power level. The lower the power setting, the less accurate the BOR trip levels are. Therefore, the high-power BOR has the highest accuracy and the low-power BOR has the lowest accuracy. The trip levels (BVDD, Parameter D005), current consumption (Section 31.2 "DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/Extended)") and time required below BVDD (TBOR, Parameter 35) can all be found in Section 31.0 "Electrical Characteristics".

FIGURE 5-2:

EXTERNAL POWER-ON RESET CIRCUIT (FOR SLOW VDD POWER-UP)



- Note 1: External Power-on Reset circuit is required only if the VDD power-up slope is too slow. The diode, D, helps discharge the capacitor quickly when VDD powers down.
 - 2: $R < 40 \text{ k}\Omega$ is recommended to make sure that the voltage drop across R does not violate the device's electrical specification.

5.4.1 DETECTING BOR

The BOR bit always resets to '0' on any Brown-out Reset or Power-on Reset event. This makes it difficult to determine if a Brown-out Reset event has occurred just by reading the state of BOR alone. A more reliable method is to simultaneously check the state of both POR and BOR. This assumes that the POR bit is reset to '1' in software immediately after any Power-on Reset event. If BOR is '0' while POR is '1', it can be reliably assumed that a Brown-out Reset event has occurred.

LP-BOR cannot be detected with the $\overline{\text{BOR}}$ bit in the RCON register. LP-BOR can rearm the $\overline{\text{POR}}$ and can cause a Power-on Reset.

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5.5 Configuration Mismatch (CM)

The Configuration Mismatch (CM) Reset is designed to detect, and attempt to recover from, random, memory corrupting events. These include Electrostatic Discharge (ESD) events that can cause widespread, single bit changes throughout the device and result in catastrophic failure.

In PIC18F87K22 family Flash devices, the device Configuration registers (located in the configuration memory space) are continuously monitored during operation by comparing their values to complimentary shadow registers. If a mismatch is detected between the two sets of registers, a CM Reset automatically occurs. These events are captured by the CM bit (RCON<5>). The state of the bit is set to '0' whenever a CM event occurs and does not change for any other Reset event.

A CM Reset behaves similarly to a Master Clear Reset, RESET instruction, WDT time-out or Stack Event Reset. As with all hard and power Reset events, the device Configuration Words are reloaded from the Flash Configuration Words in program memory as the device restarts.

5.6 Power-up Timer (PWRT)

PIC18F87K22 family devices incorporate an on-chip Power-up Timer (PWRT) to help regulate the Power-on Reset process. The PWRT is enabled by setting the PWRTEN bit (CONFIG2L<0>). The main function is to ensure that the device voltage is stable before code is executed.

The Power-up Timer (PWRT) of the PIC18F87K22 family devices is a 13-bit counter that uses the LF-INTOSC source as the clock input. This yields an approximate time interval of 2,048 x 32 μ s = 66 ms. While the PWRT is counting, the device is held in Reset.

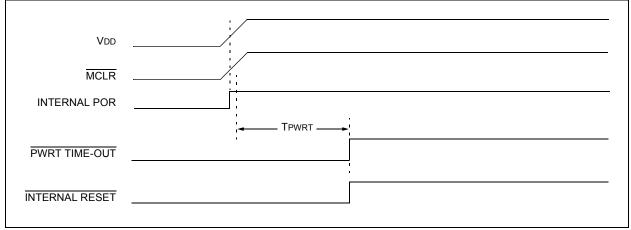
The power-up time delay depends on the LF-INTOSC clock and will vary from chip-to-chip due to temperature and process variation. See DC Parameter 33 for details.

5.6.1 TIME-OUT SEQUENCE

If enabled, the PWRT time-out is invoked after the POR pulse has cleared. The total time-out will vary based on the status of the PWRT. Figure 5-3, Figure 5-4, Figure 5-5 and Figure 5-6 all depict time-out sequences on power-up with the Power-up Timer enabled.

Since the time-outs occur from the POR pulse, if MCLR is kept low long enough, the PWRT will expire. Bringing MCLR high will begin execution immediately (Figure 5-5). This is useful for testing purposes or for synchronizing more than one PIC18 device operating in parallel.

FIGURE 5-3: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO VDD, VDD RISE < TPWRT)



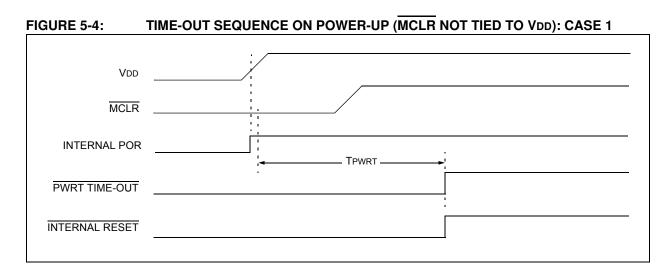


FIGURE 5-5: TIME-OUT SEQUENCE ON POWER-UP (MCLR NOT TIED TO VDD): CASE 2

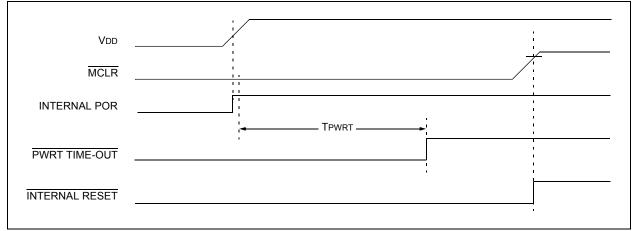
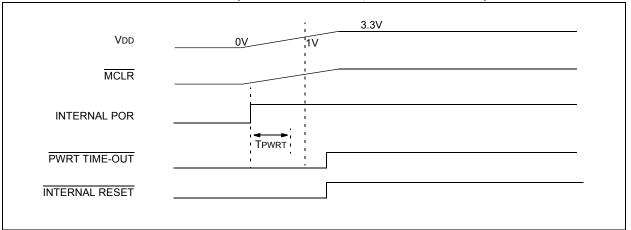


FIGURE 5-6: SLOW RISE TIME (MCLR TIED TO VDD, VDD RISE > TPWRT)



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5.7 Reset State of Registers

Most registers are unaffected by a Reset. Their status is unknown on POR and unchanged by all other Resets. The other registers are forced to a "Reset state" depending on the type of Reset that occurred.

Most registers are not affected by a WDT wake-up, since this is viewed as the resumption of normal operation. Status bits from the RCON register (\overline{CM} , \overline{RI} , \overline{TO} , \overline{PD} , \overline{POR} and \overline{BOR}) are set or cleared differently in

different Reset situations, as indicated in Table 5-1. These bits are used in software to determine the nature of the Reset.

Table 5-2 describes the Reset states for all of the Special Function Registers. These are categorized by Power-on and Brown-out Resets, Master Clear and WDT Resets, and WDT wake-ups.

TABLE 5-1:STATUS BITS, THEIR SIGNIFICANCE AND THE INITIALIZATION CONDITION FOR
RCON REGISTER

Oandition	Program			RCON I	Register	ər		STKPTR Register	
Condition	dition Counter ⁽¹⁾		RI	то	PD	POR	BOR	STKFUL	STKUNF
Power-on Reset	0000h	1	1	1	1	0	0	0	0
RESET instruction	0000h	u	0	u	u	u	u	u	u
Brown-out Reset	0000h	1	1	1	1	u	0	u	u
Configuration Mismatch Reset	0000h	0	u	u	u	u	u	u	u
MCLR Reset during power-managed Run modes	0000h	u	u	1	u	u	u	u	u
MCLR Reset during power- managed Idle modes and Sleep mode	0000h	u	u	1	0	u	u	u	u
MCLR Reset during full-power execution	0000h	u	u	u	u	u	u	u	u
Stack Full Reset (STVREN = 1)	0000h	u	u	u	u	u	u	1	u
Stack Underflow Reset (STVREN = 1)	0000h	u	u	u	u	u	u	u	1
Stack Underflow Error (not an actual Reset, STVREN = 0)	0000h	u	u	u	u	u	u	u	1
WDT time-out during full-power or power-managed Run modes	0000h	u	u	0	u	u	u	u	u
WDT time-out during power-managed Idle or Sleep modes	PC + 2	u	u	0	0	u	u	u	u
Interrupt exit from power-managed modes	PC + 2	u	u	u	0	u	u	u	u

Legend: u = unchanged

Note 1: When the wake-up is due to an interrupt and the GIEH or GIEL bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS						
Register Appli		e Devices	Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt	
TOSU	PIC18F6XK22	PIC18F8XK22	0 0000	0 0000	0 uuuu (1)	
TOSH	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu ⁽¹⁾	
TOSL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu (1)	
STKPTR	PIC18F6XK22	PIC18F8XK22	00-0 0000	uu-0 0000	uu-u uuuu (1)	
PCLATU	PIC18F6XK22	PIC18F8XK22	0 0000	0 0000	u uuuu	
PCLATH	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
PCL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	PC + 2 ⁽²⁾	
TBLPTRU	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu	
TBLPTRH	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
TBLPTRL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
TABLAT	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
PRODH	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu	
PRODL	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu	
INTCON	PIC18F6XK22	PIC18F8XK22	0000 000x	0000 000u	uuuu uuuu ⁽³⁾	
INTCON2	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu ⁽³⁾	
INTCON3	PIC18F6XK22	PIC18F8XK22	1100 0000	1100 0000	uuuu uuuu ⁽³⁾	
NDF0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
POSTINC0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
POSTDEC0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
PREINC0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
PLUSW0	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
FSR0H	PIC18F6XK22	PIC18F8XK22	0000	0000	uuuu	
FSR0L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu	
WREG	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	սսսս սսսս	
INDF1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
POSTINC1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
POSTDEC1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
PREINC1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
PLUSW1	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
FSR1H	PIC18F6XK22	PIC18F8XK22	0000	0000	uuuu	
FSR1L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu	
BSR	PIC18F6XK22	PIC18F8XK22	0000	0000	uuuu	

INITIAL IZATION CONDITIONS FOR ALL RECIPTERS

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

- 3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).
- 4: See Table 5-1 for Reset value for specific condition.

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TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)						
Register	Applicabl	le Devices Power-on Reset, Brown-out Reset		MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt	
INDF2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
POSTINC2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
POSTDEC2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
PREINC2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
PLUSW2	PIC18F6XK22	PIC18F8XK22	N/A	N/A	N/A	
FSR2H	PIC18F6XK22	PIC18F8XK22	xxxx	uuuu	uuuu	
FSR2L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu	
STATUS	PIC18F6XK22	PIC18F8XK22	x xxxx	u uuuu	u uuuu	
TMR0H	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu	
TMR0L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu	
TOCON	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu	
SPBRGH1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
OSCCON	PIC18F6XK22	PIC18F8XK22	0110 q000	0110 q000	uuuu quuu	
IPR5	PIC18F65K22	PIC18F85K22	1 -111	1 -111	u -uuu	
IPR5	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	1000 0000	1000 0000	սսսս սսսս	
WDTCON	PIC18F6XK22	PIC18F8XK22	0-x0 -000	0-x0 -000	u-uu -uuu	
RCON	PIC18F6XK22	PIC18F8XK22	0111 11qq	0uqq qquu	uuuu qquu	
TMR1H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu	
TMR1L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	uuuu uuuu	
T1CON	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu	
TMR2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
PR2	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu	
T2CON	PIC18F6XK22	PIC18F8XK22	-000 0000	-000 0000	-uuu uuuu	
SSP1BUF	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu	
SSP1ADD	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
SSP1STAT	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
SSP1CON1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
SSP1CON2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
ADRESH	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	սսսս սսսս	
ADRESL	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu	
ADCON0	PIC18F6XK22	PIC18F8XK22	-000 0000	-000 0000	-uuu uuuu	
ADCON1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
ADCON2	PIC18F6XK22	PIC18F8XK22	0-00 0000	0-00 0000	u-uu uuuu	

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

4: See Table 5-1 for Reset value for specific condition.

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)						
Register Applicable Dev		e Devices	Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt	
ECCP1AS	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
ECCP1DEL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
CCPR1H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu	
CCPR1L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCP1CON	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
PIR5	PIC18F65K22	PIC18F85K22	0 -000	0 -000	u -uuu	
PIR5	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	uuuu uuuu	
PIE5	PIC18F65K22	PIC18F85K22	0 0000	0 0000	u uuuu(1)	
PIE5	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 000	0000 0000	uuuu uuuu ⁽¹⁾	
IPR4	PIC18F65K22	PIC18F85K22	11 1111	11 1111	uu uuuu	
IPR4	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	1111 1111	1111 1111	uuuu uuuu	
PIR4	PIC18F65K22	PIC18F85K22	00 0000	00 0000	uu uuuu (1)	
PIR4	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	uuuu uuuu(1)	
PIE4	PIC18F65K22	PIC18F85K22	00 0000	00 0000	uu uuuu	
PIE4	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	սսսս սսսս	
CVRCON	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
CMSTAT	PIC18F6XK22	PIC18F8XK22	xxx	xxx	uuu	
TMR3H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu	
TMR3L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
T3CON	PIC18F6XK22	PIC18F8XK22	0000 0000	00x0 0x00	սսսս սսսս	
T3GCON	PIC18F6XK22	PIC18F8XK22	00x0 0x00	0000 0000	uuuu uuuu	
SPBRG1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
RCREG1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
TXREG1	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	XXXX XXXX	uuuu uuuu	
TXSTA1	PIC18F6XK22	PIC18F8XK22	0000 0010	0000 0010	սսսս սսսս	
RCSTA1	PIC18F6XK22	PIC18F8XK22	0000 000x	0000 000x	uuuu uuuu	
T1GCON	PIC18F6XK22	PIC18F8XK22	00x0 0x00	00x0 0x00	uuuu -uuu	
IPR6	PIC18F6XK22	PIC18F8XK22	1 -111	1 -111	u -uuu	
HLVDCON	PIC18F6XK22	PIC18F8XK22	0000 0101	0000 0101	սսսս սսսս	
PSPCON	PIC18F6XK22	PIC18F8XK22	0000	0000	uuuu	
PIR6	PIC18F6XK22	PIC18F8XK22	0 -000	0 -000	u -uuu	

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

4: See Table 5-1 for Reset value for specific condition.

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TABLE 5-2:	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)							
Register	Applicabl	e Devices	Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt			
IPR3	PIC18F6XK22	PIC18F8XK22	1-11 1111	1-11 1111	u-uu uuuu			
PIR3	PIC18F6XK22	PIC18F8XK22	0-00 0000	0-00 0000	u-uu uuuu			
PIE3	PIC18F6XK22	PIC18F8XK22	0-00 0000	0-00 0000	u-uu uuuu			
IPR2	PIC18F6XK22	PIC18F8XK22	1-11 1111	1-11 1111	u-uu uuuu			
PIR2	PIC18F6XK22	PIC18F8XK22	0-10 0000	0-10 0000	u-uu uuuu			
PIE2	PIC18F6XK22	PIC18F8XK22	0-00 0000	0-00 0000	u-uu uuuu			
IPR1	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu			
PIR1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu			
PIE1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu			
PSTR1CON	PIC18F6XK22	PIC18F8XK22	00-0 0001	00-0 0001	uu-u uuuu			
OSCTUNE	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu			
TRISJ	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu			
TRISH	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu			
TRISG	PIC18F6XK22	PIC18F8XK22	1 1111	1 1111	u uuuu			
TRISF	PIC18F6XK22	PIC18F8XK22	1111 111-	1111 111-	u uuuu			
TRISE	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu			
TRISD	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu			
TRISC	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu			
TRISB	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu			
TRISA	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	uuuu uuuu			
LATJ	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu			
LATH	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu			
LATG	PIC18F6XK22	PIC18F8XK22	x xxxx	u uuuu	u uuuu			
LATF	PIC18F6XK22	PIC18F8XK22	xxxx xxx-	uuuu uuu-	uuuu uuu-			
LATE	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu			
LATD	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu			
LATC	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu			
LATB	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu			
LATA	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu			
PORTJ	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	xxxx xxxx	սսսս սսսս			
PORTH	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	xxxx xxxx	uuuu uuuu			
PORTG	PIC18F6XK22	PIC18F8XK22	xx xxxx	xx xxxx	uu uuuu			
PORTF	PIC18F6XK22	PIC18F8XK22	xxxx xxx-	xxxx xxx-	uuuu uuu-			
PORTE	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	xxxx xxxx	uuuu uuuu			
PORTD	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	xxxx xxxx	นนนน นนนน			

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

4: See Table 5-1 for Reset value for specific condition.

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)						
Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt	
PORTC	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu	
PORTB	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	uuuu uuuu	
PORTA	PIC18F6XK22	PIC18F8XK22	xx0x 0000	uu0u 0000	uuuu uuuu	
EECON1	PIC18F6XK22	PIC18F8XK22	xx-0 x000	uu-0 u000	uu-u uuuu	
EECON2	PIC18F6XK22	PIC18F8XK22				
TMR5H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
TMR5L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
T5CON	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu	
T5GCON	PIC18F6XK22	PIC18F8XK22	00x0 0x00	uuuu uuuu	սսսս սսսս	
CCPR4H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCPR4L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCP4CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu	
CCPR5H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCPR5L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCP5CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu	
CCPR6H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCPR6L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCP6CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu	
CCPR7H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCPR7L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCP7CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu	
TMR4	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	xxxx xxxx	uuuu uuuu	
PR4	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	1111 1111	
T4CON	PIC18F6XK22	PIC18F8XK22	-111 1111	-111 1111	-uuu uuuu	
SSP2BUF	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
SSP2ADD	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
SSP2STAT	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
SSP2CON1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
SSP2CON2	PIC18F6XK22	PIC18F8XK22	0100 0000	0000 0000	սսսս սսսս	
BAUDCON1	PIC18F6XK22	PIC18F8XK22	0100 0-00	0100 0-00	uuuu u-uu	
OSCCON2	PIC18F6XK22	PIC18F8XK22	-0 0-x0	-0 0-u0	-u u-uu	
EEADRH	PIC18F6XK22	PIC18F8XK22	00	00	uu	
EEADR	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu	
EEDATA	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
PIE6	PIC18F6XK22	PIC18F8XK22	0 -000	0 -000	u -uuu	
	1					

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

4: See Table 5-1 for Reset value for specific condition.

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TABLE 5-2:	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)	

Register	Register Applicable Device		Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt	
RTCCFG	PIC18F6XK22	PIC18F8XK22	0-00 0000	u-uu uuuu	u-uu uuuu	
RTCCAL	PIC18F6XK22	PIC18F8XK22	0000 0000	սսսս սսսս	սսսս սսսս	
RTCVALH	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	սսսս սսսս	
RTCVALL	PIC18F6XK22	PIC18F8XK22	0000 0000	սսսս սսսս	սսսս սսսս	
ALRMCFG	PIC18F6XK22	PIC18F8XK22	0000 0000	uuuu uuuu	uuuu uuuu	
ALRMRPT	PIC18F6XK22	PIC18F8XK22	0000 0000	սսսս սսսս	սսսս սսսս	
ALRMVALH	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	սսսս սսսս	
ALRMVALL	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	սսսս սսսս	սսսս սսսս	
CTMUCONH	PIC18F6XK22	PIC18F8XK22	0-00 0000	0-00 0000	u-uu uuuu	
CTMUCONL	PIC18F6XK22	PIC18F8XK22	0000 00xx	0000 00xx	uuuu uuuu	
CTMUICONH	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
CM1CON	PIC18F6XK22	PIC18F8XK22	0001 1111	0001 1111	սսսս սսսս	
PADCFG1	PIC18F6XK22	PIC18F8XK22	0000-	uuuu-	uuuu-	
PADCFG1	PIC18F6XK22	PIC18F8XK22	00000-	uuuuu-	uuuuu-	
ECCP2AS	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
ECCP2DEL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
CCPR2H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	սսսս սսսս	
CCPR2L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	սսսս սսսս	
CCP2CON	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
ECCP3AS	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
ECCP3DEL	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
CCPR3H	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	սսսս սսսս	
CCPR3L	PIC18F6XK22	PIC18F8XK22	XXXX XXXX	uuuu uuuu	սսսս սսսս	
CCP3CON	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս	
CCPR8H	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu	
CCPR8L	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	սսսս սսսս	uuuu uuuu	
CCP8CON	PIC18F6XK22	PIC18F8XK22	00 0000	00 0000	uu uuuu	
CCPR9H	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	xxxx xxxx	uuuu uuuu	uuuu uuuu	
CCPR9L	PIC18F66K22 PIC18F67K22		XXXX XXXX	սսսս սսսս	uuuu uuuu	
CCP9CON	PIC18F66K22 PIC18F67K22		00 0000	00 0000	uu uuuu	
CCPR10H	PIC18F66K22 PIC18F67K22		xxxx xxxx	սսսս սսսս	uuuu uuuu	
CCPR10L	PIC18F66K22 PIC18F67K22		XXXX XXXX	սսսս սսսս	uuuu uuuu	

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

4: See Table 5-1 for Reset value for specific condition.

Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt
CCP10CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	00 0000	00 0000	uu uuuu
TMR7H	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	xxxx xxxx	սսսս սսսս	uuuu uuuu
TMR7L	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	XXXX XXXX	սսսս սսսս	uuuu uuuu
T7CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	սսսս սսսս	uuuu uuuu
T7GCON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0x00	0000 0x00	uuuu uuuu
TMR6	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PR6	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	սսսս սսսս
T6CON	PIC18F6XK22	PIC18F8XK22	-000 0000	-000 0000	-uuu uuuu
TMR8	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu
PR8	PIC18F6XK22	PIC18F8XK22	1111 1111	1111 1111	սսսս սսսս
T8CON	PIC18F6XK22	PIC18F8XK22	-000 0000	-000 0000	-uuu uuuu
TMR10	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	սսսս սսսս
PR10	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	1111 1111	1111 1111	սսսս սսսս
T10CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	-000 0000	-000 0000	-uuu uuuu
TMR12	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	uuuu uuuu
PR12	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	1111 1111	1111 1111	<u>uuuu</u> uuuu
T12CON	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	-000 0000	-000 0000	-uuu uuuu
CM2CON	PIC18F6XK22	PIC18F8XK22	0001 1111	0001 1111	uuuu uuuu
CM3CON	PIC18F6XK22	PIC18F8XK22	0001 1111	0001 1111	uuuu uuuu
CCPTMRS0	PIC18F6XK22	PIC18F8XK22	0000 0000	սսսս սսսս	uuuu uuuu
CCPTMRS1	PIC18F6XK22	PIC18F8XK22	00-0 -000	uu-u -uuu	uu-u -uuu
CCPTMRS2	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0 -000	u -uuu	u -uuu
CCPTMRS2	PIC18F65K22	PIC18F85K22	00	uu	uu
REFOCON	PIC18F6XK22	PIC18F8XK22	0-00 0000	u-uu uuuu	u-uu uuuu
ODCON1	PIC18F6XK22	PIC18F8XK22	0000	uuuu	uuuu

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

4: See Table 5-1 for Reset value for specific condition.

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TABLE 5-2:	INITIALIZA	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)						
Register	Applicable Devices		Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets, CM Resets	Wake-up via WDT or Interrupt			
ODCON2	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	սսսս սսսս	սսսս սսսս			
ODCON2	PIC18F65K22	PIC18F85K22	00 0000	uu uuuu	uu uuuu			
ODCON3	PIC18F6XK22	PIC18F8XK22	000	uuu	uuu			
MEMCON	PIC18F6XK22	PIC18F8XK22	0-0000	0-0000	u-uuuu			
ANCON0	PIC18F6XK22	PIC18F8XK22	1111 1111	սսսս սսսս	սսսս սսսս			
ANCON1	PIC18F6XK22	PIC18F8XK22	1111 1111	uuuu uuuu	uuuu uuuu			
ANCON2	PIC18F6XK22	PIC18F8XK22	1111 1111	սսսս սսսս	սսսս սսսս			
RCSTA2	PIC18F6XK22	PIC18F8XK22	0000 000x	0000 000x	սսսս սսսս			
TXSTA2	PIC18F6XK22	PIC18F8XK22	0000 0010	0000 0010	սսսս սսսս			
BAUDCON2	PIC18F6XK22	PIC18F8XK22	0100 0-00	0100 0-00	uuuu u-uu			
SPBRGH2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu			
SPBRG2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս			
RCREG2	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս			
TXREG2	PIC18F6XK22	PIC18F8XK22	xxxx xxxx	xxxx xxxx	uuuu uuuu			
PSTR2CON	PIC18F6XK22	PIC18F8XK22	00-0 0001	00-0 0001	uu-u uuuu			
PSTR3CON	PIC18F6XK22	PIC18F8XK22	00-0 0001	00-0 0001	uu-u uuuu			
PMD0	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	uuuu uuuu			
PMD1	PIC18F6XK22	PIC18F8XK22	0000 0000	0000 0000	սսսս սսսս			
PMD2	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	սսսս սսսս			
PMD2	PIC18F65K22	PIC18F85K22	-0-0 0000	-0-0 0000	-u-u uuuu			
PMD3	PIC18F66K22 PIC18F67K22	PIC18F86K22 PIC18F87K22	0000 0000	0000 0000	սսսս սսսս			
PMD3	PIC18F65K22	PIC18F85K22	00 000-	00 000-	uu uuu-			

TABLE 5-2: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: When the wake-up is due to an interrupt, and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

3: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

4: See Table 5-1 for Reset value for specific condition.

6.0 MEMORY ORGANIZATION

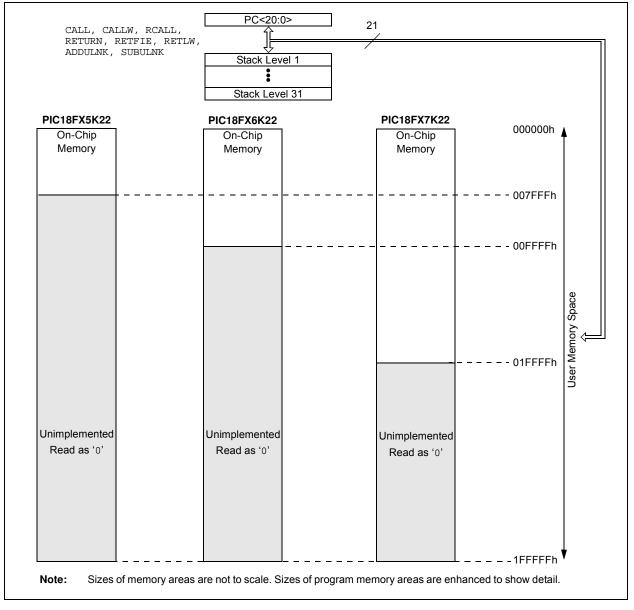
PIC18F87K22 family devices have these types of memory:

- Program Memory
- Data RAM
- Data EEPROM

As Harvard architecture devices, the data and program memories use separate busses. This enables concurrent access of the two memory spaces. The data EEPROM, for practical purposes, can be regarded as a peripheral device because it is addressed and accessed through a set of control registers.

Additional detailed information on the operation of the Flash program memory is provided in **Section 7.0 "Flash Program Memory"**. The data EEPROM is discussed separately in **Section 9.0 "Data EEPROM Memory"**.

FIGURE 6-1: MEMORY MAPS FOR PIC18F87K22 FAMILY DEVICES



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6.1 Program Memory Organization

PIC18 microcontrollers implement a 21-bit Program Counter that is capable of addressing a 2-Mbyte program memory space. Accessing a location between the upper boundary of the physically implemented memory and the 2-Mbyte address will return all '0's (a NOP instruction).

The entire PIC18F87K22 family offers a range of on-chip Flash program memory sizes, from 32 Kbytes (up to 16,384 single-word instructions) to 128 Kbytes (65,536 single-word instructions).

- PIC18F65K22 and PIC18F85K22 32 Kbytes of Flash memory, storing up to 16,384 single-word instructions
- PIC18F66K22 and PIC18F86K22 64 Kbytes of Flash memory, storing up to 32,768 single-word instructions
- PIC18F67K22 and PIC18F87K22 128 Kbytes of Flash memory, storing up to 65,536 single-word instructions

The program memory maps for individual family members are shown in Figure 6-1.

6.1.1 HARD MEMORY VECTORS

All PIC18 devices have a total of three hard-coded return vectors in their program memory space. The Reset vector address is the default value to which the Program Counter returns on all device Resets; it is located at 0000h.

PIC18 devices also have two interrupt vector addresses for handling high-priority and low-priority interrupts. The high-priority interrupt vector is located at 0008h and the low-priority interrupt vector is at 0018h. The locations of these vectors are shown, in relation to the program memory map, in Figure 6-2.

FIGURE 6-2: HARD VECTOR FOR PIC18F87K22 FAMILY DEVICES

	Reset Vector	0000h			
	High-Priority Interrupt Vector	0008h			
	Low-Priority Interrupt Vector	0018h			
	On-Chip Program Memory				
	Read '0'				
		1FFFFFh			
Le	egend: (Top of Memory) represents upper boundary of on-chip program memory space (see Figure 6-1 for device-specific values). Shaded area represents unimplemented memory. Areas are not shown to scale.				

6.1.2 PROGRAM COUNTER

The Program Counter (PC) specifies the address of the instruction to fetch for execution. The PC is 21 bits wide and contained in three separate 8-bit registers.

The low byte, known as the PCL register, is both readable and writable. The high byte, or PCH register, contains the PC<15:8> bits and is not directly readable or writable. Updates to the PCH register are performed through the PCLATH register. The upper byte is called PCU. This register contains the PC<20:16> bits; it is also not directly readable or writable. Updates to the PCU register are performed through the PCLATU register.

The contents of PCLATH and PCLATU are transferred to the Program Counter by any operation that writes PCL. Similarly, the upper two bytes of the Program Counter are transferred to PCLATH and PCLATU by an operation that reads PCL. This is useful for computed offsets to the PC (see Section 6.1.5.1 "Computed GOTO").

The PC addresses bytes in the program memory. To prevent the PC from becoming misaligned with word instructions, the Least Significant bit of PCL is fixed to a value of '0'. The PC increments by two to address sequential instructions in the program memory.

The CALL, RCALL, GOTO and program branch instructions write to the Program Counter directly. For these instructions, the contents of PCLATH and PCLATU are not transferred to the Program Counter.

6.1.3 RETURN ADDRESS STACK

The return address stack enables execution of any combination of up to 31 program calls and interrupts. The PC is pushed onto the stack when a CALL or RCALL instruction is executed or an interrupt is Acknowledged. The PC value is pulled off the stack on a RETURN, RETLW or a RETFIE instruction. The value also is pulled off the stack on ADDULNK and SUBULNK instructions, if the extended instruction set is enabled. PCLATU and PCLATH are not affected by any of the RETURN or CALL instructions.

The stack operates as a 31-word by 21-bit RAM and a 5-bit Stack Pointer, STKPTR. The stack space is not part of either program or data space. The Stack Pointer is readable and writable and the address on the top of the stack is readable and writable through the Top-of-Stack Special Function Registers. Data can also be pushed to, or popped from, the stack using these registers.

A CALL type instruction causes a push onto the stack. The Stack Pointer is first incremented and the location pointed to by the Stack Pointer is written with the contents of the PC (already pointing to the instruction following the CALL). A RETURN type instruction causes a pop from the stack. The contents of the location pointed to by the STKPTR are transferred to the PC and then the Stack Pointer is decremented.

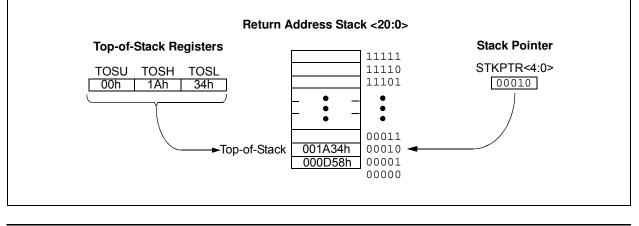
The Stack Pointer is initialized to '00000' after all Resets. There is no RAM associated with the location corresponding to a Stack Pointer value of '00000'; this is only a Reset value. Status bits indicate if the stack is full, has overflowed or has underflowed.

6.1.3.1 Top-of-Stack Access

Only the top of the return address stack (TOS) is readable and writable. A set of three registers, TOSU:TOSH:TOSL, holds the contents of the stack location pointed to by the STKPTR register (Figure 6-3). This allows users to implement a software stack, if necessary. After a CALL, RCALL or interrupt (or ADDULNK and SUBULNK instructions, if the extended instruction set is enabled), the software can read the pushed value by reading the TOSU:TOSH:TOSL registers. These values can be placed on a user-defined software stack. At return time, the software can return these values to TOSU:TOSH:TOSL and do a return.

While accessing the stack, users must disable the Global Interrupt Enable bits to prevent inadvertent stack corruption.





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6.1.3.2 Return Stack Pointer (STKPTR)

The STKPTR register (Register 6-1) contains the Stack Pointer value, the STKFUL (Stack Full) status bit and the STKUNF (Stack Underflow) status bits. The value of the Stack Pointer can be 0 through 31. The Stack Pointer increments before values are pushed onto the stack and decrements after values are popped off the stack. On Reset, the Stack Pointer value will be zero.

The user may read and write the Stack Pointer value. This feature can be used by a Real-Time Operating System (RTOS) for return stack maintenance.

After the PC is pushed onto the stack 31 times (without popping any values off the stack), the STKFUL bit is set. The STKFUL bit is cleared by software or by a POR.

What happens when the stack becomes full depends on the state of the STVREN (Stack Overflow Reset Enable) Configuration bit. (For a description of the device Configuration bits, see **Section 28.1 "Configuration Bits**".) If STVREN is set (default), the 31st push will push the (PC + 2) value onto the stack, set the STKFUL bit and reset the device. The STKFUL bit will remain set and the Stack Pointer will be set to zero.

If STVREN is cleared, the STKFUL bit will be set on the 31st push and the Stack Pointer will increment to 31. Any additional pushes will not overwrite the 31st push and the STKPTR will remain at 31.

When the stack has been popped enough times to unload the stack, the next pop will return a value of zero to the PC and set the STKUNF bit while the Stack Pointer remains at zero. The STKUNF bit will remain set until cleared by software or until a POR occurs.

Note:	Returning a value of zero to the PC on an
	underflow has the effect of vectoring the
	program to the Reset vector, where the
	stack conditions can be verified and
	appropriate actions can be taken. This is
	not the same as a Reset as the contents
	of the SFRs are not affected.

6.1.3.3 PUSH and POP Instructions

Since the Top-of-Stack (TOS) is readable and writable, the ability to push values onto the stack and pull values off of the stack, without disturbing normal program execution, is a desirable feature. The PIC18 instruction set includes two instructions, PUSH and POP, that permit the TOS to be manipulated under software control. TOSU, TOSH and TOSL can be modified to place data or a return address on the stack.

The PUSH instruction places the current PC value onto the stack. This increments the Stack Pointer and loads the current PC value onto the stack.

The POP instruction discards the current TOS by decrementing the Stack Pointer. The previous value pushed onto the stack then becomes the TOS value.

R/C-0 R/C-0 R/W-0 U-0 R/W-0 R/W-0 R/W-0 R/W-0 STKFUL⁽¹⁾ STKUNF⁽¹⁾ SP4 SP3 SP2 SP1 SP0 bit 7 bit 0 Legend: C = Clearable bit R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' '0' = Bit is cleared -n = Value at POR '1' = Bit is set x = Bit is unknown STKFUL: Stack Full Flag bit⁽¹⁾ bit 7 1 = Stack has become full or overflowed 0 = Stack has not become full or overflowed STKUNF: Stack Underflow Flag bit⁽¹⁾ bit 6 1 = Stack underflow has occurred 0 = Stack underflow did not occur bit 5 Unimplemented: Read as '0' bit 4-0 SP<4:0>: Stack Pointer Location bits **Note 1:** Bit 7 and bit 6 are cleared by user software or by a POR.

REGISTER 6-1: STKPTR: STACK POINTER REGISTER

6.1.3.4 Stack Full and Underflow Resets

Device Resets on stack overflow and stack underflow conditions are enabled by setting the STVREN bit (CONFIG4L<0>). When STVREN is set, a full or underflow condition will set the appropriate STKFUL or STKUNF bit and then cause a device Reset. When STVREN is cleared, a full or underflow condition will set the appropriate STKFUL or STKUNF bit, but not cause a device Reset. The STKFUL or STKUNF bits are cleared by the user software or a Power-on Reset.

6.1.4 FAST REGISTER STACK

A Fast Register Stack is provided for the STATUS, WREG and BSR registers to provide a "fast return" option for interrupts. This stack is only one level deep and is neither readable nor writable. It is loaded with the current value of the corresponding register when the processor vectors for an interrupt. All interrupt sources will push values into the Stack registers. The values in the registers are then loaded back into the working registers if the RETFIE, FAST instruction is used to return from the interrupt.

If both low and high-priority interrupts are enabled, the Stack registers cannot be used reliably to return from low-priority interrupts. If a high-priority interrupt occurs while servicing a low-priority interrupt, the Stack register values stored by the low-priority interrupt will be overwritten. In these cases, users must save the key registers in software during a low-priority interrupt.

If interrupt priority is not used, all interrupts may use the Fast Register Stack for returns from interrupt. If no interrupts are used, the Fast Register Stack can be used to restore the STATUS, WREG and BSR registers at the end of a subroutine call. To use the Fast Register Stack for a subroutine call, a CALL label, FAST instruction must be executed to save the STATUS, WREG and BSR registers to the Fast Register Stack. A RETURN, FAST instruction is then executed to restore these registers from the Fast Register Stack.

Example 6-1 shows a source code example that uses the Fast Register Stack during a subroutine call and return.

EXAMPLE 6-1: FAST REGISTER STACK CODE EXAMPLE

CALL SUB1, FAST	;STATUS, WREG, BSR
•	;SAVED IN FAST REGISTER
SUB1 •	;STACK
RETURN FAST	;RESTORE VALUES SAVED ;IN FAST REGISTER STACK

6.1.5 LOOK-UP TABLES IN PROGRAM MEMORY

There may be programming situations that require the creation of data structures, or look-up tables, in program memory. For PIC18 devices, look-up tables can be implemented in two ways:

- Computed GOTO
- Table Reads

6.1.5.1 Computed GOTO

A computed GOTO is accomplished by adding an offset to the Program Counter. An example is shown in Example 6-2.

A look-up table can be formed with an ADDWF PCL instruction and a group of RETLW nn instructions. The W register is loaded with an offset into the table before executing a call to that table. The first instruction of the called routine is the ADDWF PCL instruction. The next instruction executed will be one of the RETLW nn instructions that returns the value, 'nn', to the calling function.

The offset value (in WREG) specifies the number of bytes that the Program Counter should advance and should be multiples of two (LSb = 0).

In this method, only one data byte may be stored in each instruction location and room on the return address stack is required.

EXAMPLE 6-2: COMPUTED GOTO USING AN OFFSET VALUE

	MOVF	OFFSET, W
	CALL	TABLE
ORG	nn00h	
TABLE	ADDWF	PCL
	RETLW	nnh
	RETLW	nnh
	RETLW	nnh

6.1.5.2 Table Reads

A better method of storing data in program memory allows two bytes of data to be stored in each instruction location.

Look-up table data may be stored, two bytes per program word, while programming. The Table Pointer (TBLPTR) specifies the byte address and the Table Latch (TABLAT) contains the data that is read from the program memory. Data is transferred from program memory, one byte at a time.

The table read operation is discussed further in Section 7.1 "Table Reads and Table Writes".

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6.2 PIC18 Instruction Cycle

6.2.1 CLOCKING SCHEME

The microcontroller clock input, whether from an internal or external source, is internally divided by four to generate four non-overlapping, quadrature clocks (Q1, Q2, Q3 and Q4). Internally, the Program Counter is incremented on every Q1, with the instruction fetched from the program memory and latched into the Instruction Register (IR) during Q4.

The instruction is decoded and executed during the following Q1 through Q4. The clocks and instruction execution flow are shown in Figure 6-4.

6.2.2 INSTRUCTION FLOW/PIPELINING

An "Instruction Cycle" consists of four Q cycles, Q1 through Q4. The instruction fetch and execute are pipelined in such a manner that a fetch takes one instruction cycle, while the decode and execute take another instruction cycle. However, due to the pipelining, each instruction effectively executes in one cycle. If an instruction (such as GOTO) causes the Program Counter to change, two cycles are required to complete the instruction. (See Example 6-3.)

A fetch cycle begins with the Program Counter (PC) incrementing in Q1.

In the execution cycle, the fetched instruction is latched into the Instruction Register (IR) in cycle Q1. This instruction is then decoded and executed during the Q2, Q3 and Q4 cycles. Data memory is read during Q2 (operand read) and written during Q4 (destination write).

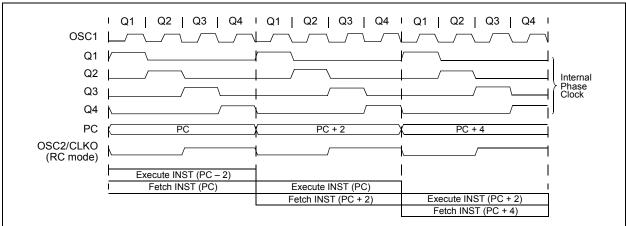
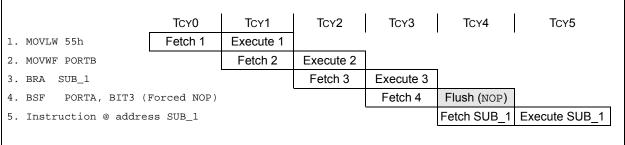


FIGURE 6-4: CLOCK/INSTRUCTION CYCLE

EXAMPLE 6-3: INSTRUCTION PIPELINE FLOW



All instructions are single cycle, except for any program branches. These take two cycles since the fetch instruction is "flushed" from the pipeline while the new instruction is being fetched and then executed.

6.2.3 INSTRUCTIONS IN PROGRAM MEMORY

The program memory is addressed in bytes. Instructions are stored as two or four bytes in program memory. The Least Significant Byte of an instruction word is always stored in a program memory location with an even address (LSB = 0). To maintain alignment with instruction boundaries, the PC increments in steps of two and the LSB will always read '0' (see Section 6.1.2 "Program Counter").

Figure 6-5 shows an example of how instruction words are stored in the program memory.

The CALL and GOTO instructions have the absolute program memory address embedded into the instruction. Since instructions are always stored on word boundaries, the data contained in the instruction is a word address. The word address is written to PC<20:1> which accesses the desired byte address in program memory. Instruction #2 in Figure 6-5 shows how the instruction, GOTO 0006h, is encoded in the program memory. Program branch instructions, which encode a relative address offset, operate in the same manner. The offset value stored in a branch instruction represents the number of single-word instructions that the PC will be offset by. For more details on the instruction set, see **Section 29.0 "Instruction Set Summary"**.

FIGURE 6-5:	INSTRUCTIONS IN PROGRAM MEMORY

			LSB = 1	LSB = 0	Word Address \downarrow
	Program M	lemory			000000h
	Byte Locat	ions \rightarrow			000002h
					000004h
					000006h
Instruction 1:	MOVLW	055h	0Fh	55h	000008h
Instruction 2:	GOTO	0006h	EFh	03h	00000Ah
			F0h	00h	00000Ch
Instruction 3:	MOVFF	123h, 456h	C1h	23h	00000Eh
			F4h	56h	000010h
					000012h
					000014h

6.2.4 TWO-WORD INSTRUCTIONS

The standard PIC18 instruction set has four, two-word instructions: CALL, MOVFF, GOTO and LSFR. In all cases, the second word of the instructions always has '1111' as its four Most Significant bits. The other 12 bits are literal data, usually a data memory address.

The use of '1111' in the 4 MSbs of an instruction specifies a special form of NOP. If the instruction is executed in proper sequence, immediately after the first word, the data in the second word is accessed and

used by the instruction sequence. If the first word is skipped, for some reason, and the second word is executed by itself, a NOP is executed instead. This is necessary for cases when the two-word instruction is preceded by a conditional instruction that changes the PC. Example 6-4 shows how this works.

Note: For information on two-word instructions in the extended instruction set, see Section 6.5 "Program Memory and the Extended Instruction Set".

CASE 1:		
Object Code	Source Code	
0110 0110 0000 0000	TSTFSZ REG1	; is RAM location 0?
1100 0001 0010 0011	MOVFF REG1, REG2	; No, skip this word
1111 0100 0101 0110		; Execute this word as a NOP
0010 0100 0000 0000	ADDWF REG3	; continue code
CASE 2:		
Object Code	Source Code	
0110 0110 0000 0000	TSTFSZ REG1	; is RAM location 0?
1100 0001 0010 0011	MOVFF REG1, REG2	; Yes, execute this word
1111 0100 0101 0110		; 2nd word of instruction
0010 0100 0000 0000	ADDWF REG3	; continue code

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6.3 Data Memory Organization

Note:	The operation of some aspects of data
	memory are changed when the PIC18
	extended instruction set is enabled. See
	Section 6.6 "Data Memory and the
	Extended Instruction Set" for more
	information.

The data memory in PIC18 devices is implemented as static RAM. Each register in the data memory has a 12-bit address, allowing up to 4,096 bytes of data memory. The memory space is divided into as many as 16 banks that contain 256 bytes each. PIC18FX6K22 and PIC18FX7K22 devices implement all 16 complete banks, for a total of 4 Kbytes. PIC18FX5K22 devices implement only the first eight complete banks, for a total of 2 Kbytes.

Figure 6-6 and Figure 6-7 show the data memory organization for the devices.

The data memory contains Special Function Registers (SFRs) and General Purpose Registers (GPRs). The SFRs are used for control and status of the controller and peripheral functions, while GPRs are used for data storage and scratchpad operations in the user's application. Any read of an unimplemented location will read as '0's.

The instruction set and architecture allow operations across all banks. The entire data memory may be accessed by Direct, Indirect or Indexed Addressing modes. Addressing modes are discussed later in this section.

To ensure that commonly used registers (select SFRs and select GPRs) can be accessed in a single cycle, PIC18 devices implement an Access Bank. This is a 256-byte memory space that provides fast access to select SFRs and the lower portion of GPR Bank 0 without using the Bank Select Register. For details on the Access RAM, see Section 6.3.2 "Access Bank".

6.3.1 BANK SELECT REGISTER

Large areas of data memory require an efficient addressing scheme to make it possible for rapid access to any address. Ideally, this means that an entire address does not need to be provided for each read or write operation. For PIC18 devices, this is accomplished with a RAM banking scheme. This divides the memory space into 16 contiguous banks of 256 bytes. Depending on the instruction, each location can be addressed directly by its full 12-bit address, or an 8-bit, low-order address and a 4-bit Bank Pointer.

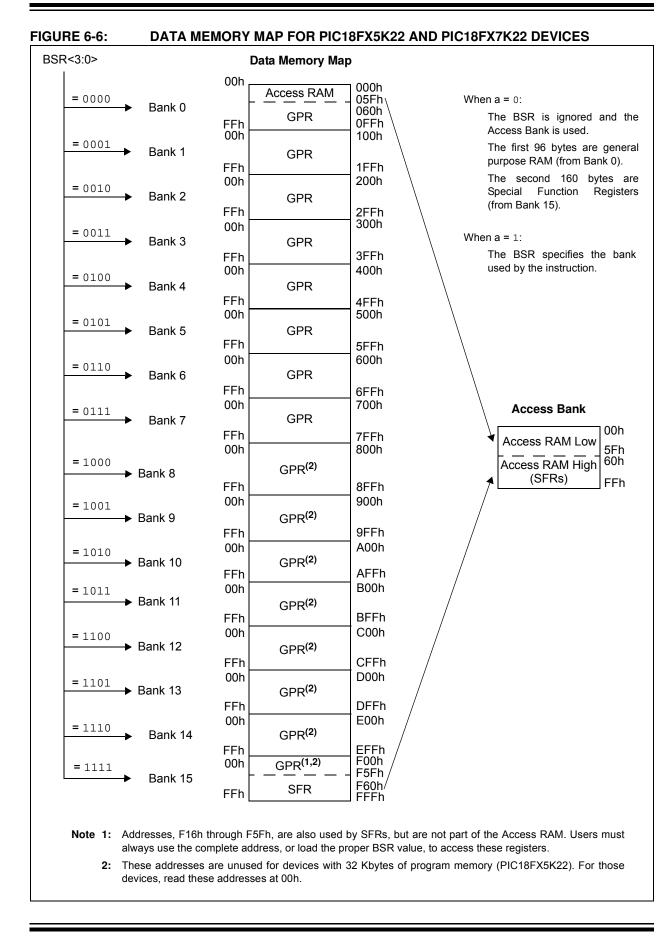
Most instructions in the PIC18 instruction set make use of the Bank Pointer, known as the Bank Select Register (BSR). This SFR holds the four Most Significant bits of a location's address. The instruction itself includes the eight Least Significant bits. Only the four lower bits of the BSR are implemented (BSR<3:0>). The upper four bits are unused, always read as '0' and cannot be written to. The BSR can be loaded directly by using the MOVLB instruction.

The value of the BSR indicates the bank in data memory. The eight bits in the instruction show the location in the bank and can be thought of as an offset from the bank's lower boundary. The relationship between the BSR's value and the bank division in data memory is shown in Figure 6-7.

Since up to 16 registers may share the same low-order address, the user must always be careful to ensure that the proper bank is selected before performing a data read or write. For example, writing what should be program data to an eight-bit address of F9h, while the BSR is 0Fh, will end up resetting the Program Counter.

While any bank can be selected, only those banks that are actually implemented can be read or written to. Writes to unimplemented banks are ignored, while reads from unimplemented banks will return '0's. Even so, the STATUS register will still be affected as if the operation was successful. The data memory map in Figure 6-6 indicates which banks are implemented.

In the core PIC18 instruction set, only the MOVFF instruction fully specifies the 12-bit address of the source and target registers. When this instruction executes, it ignores the BSR completely. All other instructions include only the low-order address as an operand and must use either the BSR or the Access Bank to locate their target registers.



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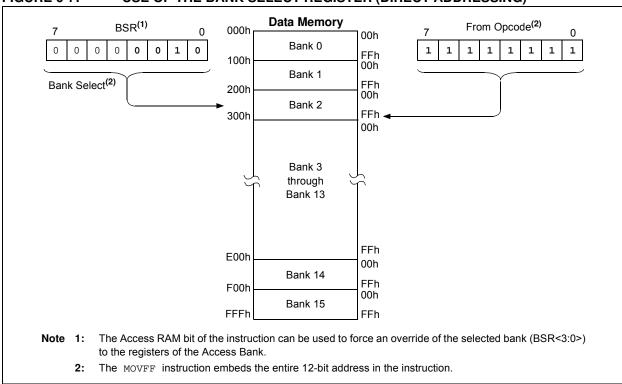


FIGURE 6-7: USE OF THE BANK SELECT REGISTER (DIRECT ADDRESSING)

6.3.2 ACCESS BANK

While the use of the BSR, with an embedded 8-bit address, allows users to address the entire range of data memory, it also means that the user must ensure that the correct bank is selected. If not, data may be read from, or written to, the wrong location. This can be disastrous if a GPR is the intended target of an operation, but an SFR is written to instead. Verifying and/or changing the BSR for each read or write to data memory can become very inefficient.

To streamline access for the most commonly used data memory locations, the data memory is configured with an Access Bank, which allows users to access a mapped block of memory without specifying a BSR. The Access Bank consists of the first 96 bytes of memory (00h-5Fh) in Bank 0 and the last 160 bytes of memory (60h-FFh) in Bank 15. The lower half is known as the "Access RAM" and is composed of GPRs. The upper half is where the device's SFRs are mapped. These two areas are mapped contiguously in the Access Bank and can be addressed in a linear fashion by an eight-bit address (Figure 6-6).

The Access Bank is used by core PIC18 instructions that include the Access RAM bit (the 'a' parameter in the instruction). When 'a' is equal to '1', the instruction uses the BSR and the 8-bit address included in the opcode for the data memory address. When 'a' is '0',

however, the instruction is forced to use the Access Bank address map. In that case, the current value of the BSR is ignored entirely.

Using this "forced" addressing allows the instruction to operate on a data address in a single cycle without updating the BSR first. For 8-bit addresses of 60h and above, this means that users can evaluate and operate on SFRs more efficiently. The Access RAM below 60h is a good place for data values that the user might need to access rapidly, such as immediate computational results or common program variables.

Access RAM also allows for faster and more code efficient context saving and switching of variables.

The mapping of the Access Bank is slightly different when the extended instruction set is enabled (XINST Configuration bit = 1). This is discussed in more detail in Section 6.6.3 "Mapping the Access Bank in Indexed Literal Offset Mode".

6.3.3 GENERAL PURPOSE REGISTER FILE

PIC18 devices may have banked memory in the GPR area. This is data RAM which is available for use by all instructions. GPRs start at the bottom of Bank 0 (address 000h) and grow upwards towards the bottom of the SFR area. GPRs are not initialized by a Power-on Reset and are unchanged on all other Resets.

6.3.4 SPECIAL FUNCTION REGISTERS

The Special Function Registers (SFRs) are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. SFRs start at the top of data memory (FFFh) and extend downward to occupy all of Bank 15 (F00h to FFFh) and the top part of Bank 14 (EF4h to EFFh).

A list of these registers is given in Table 6-1 and Table 6-2.

The SFRs can be classified into two sets: those associated with the "core" device functionality (ALU, Resets and interrupts) and those related to the peripheral functions. The Reset and Interrupt registers are described in their respective chapters, while the ALU's STATUS register is described later in this section. Registers related to the operation of the peripheral features are described in the chapter for that peripheral.

The SFRs are typically distributed among the peripherals whose functions they control. Unused SFR locations are unimplemented and read as '0's.

Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name ⁽⁴⁾
FFFh	TOSU	FDFh	INDF2 ⁽¹⁾	FBFh	ECCP1AS	F9Fh	IPR1	F7Fh	EECON1	F5Fh	RTCCFG
FFEh	TOSH	FDEh	POSTINC2 ⁽¹⁾	FBEh	ECCP1DEL	F9Eh	PIR1	F7Eh	EECON2	F5Eh	RTCCAL
FFDh	TOSL	FDDh	POSTDEC2 ⁽¹⁾	FBDh	CCPR1H	F9Dh	PIE1	F7Dh	TMR5H	F5Dh	RTCVALH
FFCh	STKPTR	FDCh	PREINC2 ⁽¹⁾	FBCh	CCPR1L	F9Ch	PSTR1CON	F7Ch	TMR5L	F5Ch	RTCVALL
FFBh	PCLATU	FDBh	PLUSW2 ⁽¹⁾	FBBh	CCP1CON	F9Bh	OSCTUNE	F7Bh	T5CON	F5Bh	ALRMCFG
FFAh	PCLATH	FDAh	FSR2H	FBAh	PIR5	F9Ah	TRISJ ⁽²⁾	F7Ah	T5GCON	F5Ah	ALRMRPT
FF9h	PCL	FD9h	FSR2L	FB9h	PIE5	F99h	TRISH ⁽²⁾	F79h	CCPR4H	F59h	ALRMVALH
FF8h	TBLPTRU	FD8h	STATUS	FB8h	IPR4	F98h	TRISG	F78h	CCPR4L	F58h	ALRMVALL
FF7h	TBLPTRH	FD7h	TMR0H	FB7h	PIR4	F97h	TRISF	F77h	CCP4CON	F57h	CTMUCONH
FF6h	TBLPTRL	FD6h	TMR0L	FB6h	PIE4	F96h	TRISE	F76h	CCPR5H	F56h	CTMUCONL
FF5h	TABLAT	FD5h	T0CON	FB5h	CVRCON	F95h	TRISD	F75h	CCPR5L	F55h	CTMUICONH
FF4h	PRODH	FD4h	SPBRGH1	FB4h	CMSTAT	F94h	TRISC	F74h	CCP5CON	F54h	CM1CON
FF3h	PRODL	FD3h	OSCCON	FB3h	TMR3H	F93h	TRISB	F73h	CCPR6H	F53h	PADCFG1
FF2h	INTCON	FD2h	IPR5	FB2h	TMR3L	F92h	TRISA	F72h	CCPR6L	F52h	ECCP2AS
FF1h	INTCON2	FD1h	WDTCON	FB1h	T3CON	F91h	LATJ ⁽²⁾	F71h	CCP6CON	F51h	ECCP2DEL
FF0h	INTCON3	FD0h	RCON	FB0h	T3GCON	F90h	LATH ⁽²⁾	F70h	CCPR7H	F50h	CCPR2H
FEFh	INDF0 ⁽¹⁾	FCFh	TMR1H	FAFh	SPBRG1	F8Fh	LATG	F6Fh	CCPR7L	F4Fh	CCPR2L
	POSTINC0 ⁽¹⁾	FCEh	TMR1L	FAEh	RCREG1	F8Eh	LATF	F6Eh	CCP7CON	F4Eh	CCP2CON
FEDh	POSTDEC0 ⁽¹⁾	FCDh	T1CON	FADh	TXREG1	F8Dh	LATE	F6Dh	TMR4	F4Dh	ECCP3AS
FECh	PREINC0 ⁽¹⁾	FCCh	TMR2	FACh	TXSTA1	F8Ch	LATD	F6Ch	PR4	F4Ch	ECCP3DEL
FEBh	PLUSW0 ⁽¹⁾	FCBh	PR2	FABh	RCSTA1	F8Bh	LATC	F6Bh	T4CON	F4Bh	CCPR3H
FEAh	FSR0H	FCAh	T2CON	FAAh	T1GCON	F8Ah	LATB	F6Ah	SSP2BUF	F4Ah	CCPR3L
FE9h	FSR0L	FC9h	SSP1BUF	FA9h	IPR6	F89h	LATA	F69h	SSP2ADD	F49h	CCP3CON
FE8h	WREG	FC8h	SSP1ADD	FA8h	HLVDCON	F88h	PORTJ ⁽²⁾	F68h	SSP2STAT	F48h	CCPR8H
FE7h	INDF1 ⁽¹⁾	FC7h	SSP1STAT	FA7h	PSPCON	F87h	PORTH ⁽²⁾	F67h	SSP2CON1	F47h	CCPR8L
FE6h	POSTINC1 ⁽¹⁾	FC6h	SSP1CON1	FA6h	PIR6	F86h	PORTG	F66h	SSP2CON2	F46h	CCP8CON
FE5h	POSTDEC1 ⁽¹⁾	FC5h	SSP1CON2	FA5h	IPR3	F85h	PORTF	F65h	BAUDCON1	F45h	CCPR9H ⁽³⁾
FE4h	PREINC1 ⁽¹⁾	FC4h	ADRESH	FA4h	PIR3	F84h	PORTE	F64h	OSCCON2	F44h	CCPR9L ⁽³⁾
FE3h	PLUSW1 ⁽¹⁾	FC3h	ADRESL	FA3h	PIE3	F83h	PORTD	F63h	EEADRH	F43h	CCP9CON ⁽³⁾
FE2h	FSR1H	FC2h	ADCON0	FA2h	IPR2	F82h	PORTC	F62h	EEADR	F42h	CCPR10H ⁽³⁾
FE1h	FSR1L	FC1h	ADCON1	FA1h	PIR2	F81h	PORTB	F61h	EEDATA	F41h	CCPR10L ⁽³⁾
FE0h	BSR	FC0h	ADCON2	FA0h	PIE2	F80h	PORTA	F60h	PIE6	F40h	CCP10CON ⁽³⁾

TABLE 6-1:SPECIAL FUNCTION REGISTER MAP FOR PIC18F87K22 FAMILY

Note 1: This is not a physical register.

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

3: This register is not available on devices with a program memory of 32 Kbytes (PIC18FX5K22).

4: Addresses, F16h through F5Fh, are also used by SFRs, but are not part of the Access RAM. To access these registers, users must always load the proper BSR value.

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TABLE 6-1: SPECIAL FUNCTION REGISTER MAP FOR PIC18F87K22 FAMILY (CONTINUED)

Addr.	Name	Addr.	Name	Addr.	Name	Addr.	Name
F3Fh	TMR7H ⁽³⁾	F32h	TMR12 ⁽³⁾	F25h	ANCON0	F18h	PMD1
F3Eh	TMR7L ⁽³⁾	F31h	PR12 ⁽³⁾	F24h	ANCON1	F17h	PMD2
F3Dh	T7CON ⁽³⁾	F30h	T12CON ⁽³⁾	F23h	ANCON2	F16h	PMD3
F3Ch	T7GCON ⁽³⁾	F2Fh	CM2CON	F22h	RCSTA2	_	
F3Bh	TMR6	F2Eh	CM3CON	F21h	TXSTA2		
F3Ah	PR6	F2Dh	CCPTMRS0	F20h	BAUDCON2		
F39H	T6CON	F2Ch	CCPTMRS1	F1Fh	SPBRGH2		
F38h	TMR8	F2Bh	CCPTMRS2	F1Eh	SPBRG2		
F37h	PR8	F2Ah	REFOCON	F1Dh	RCREG2		
F36h	T8CON	F29H	ODCON1	F1Ch	TXREG2		
F35h	TMR10 ⁽³⁾	F28h	ODCON2	F1Bh	PSTR2CON		
F34h	PR10 ⁽³⁾	F27h	ODCON3	F1Ah	PSTR3CON		
F33h	T10CON ⁽³⁾	F26h	MEMCON ⁽³⁾	F19h	PMD0		

Note 1: This is not a physical register.

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

- 3: This register is not available on devices with a program memory of 32 Kbytes (PIC18FX5K22).
- 4: Addresses, F16h through F5Fh, are also used by SFRs, but are not part of the Access RAM. To access these registers, users must always load the proper BSR value.

Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR	
FFFh	TOSU	_	_	_	Top-of-Stack U	pper Byte (TO	S<20:16>)			0 0000	
FFEh	TOSH	Top-of-Stack H	ligh Byte (TOS	S<15:8>)						0000 0000	
FFDh	TOSL	Top-of-Stack L	ow Byte (TOS	<7:0>)						0000 0000	
FFCh	STKPTR	STKFUL	STKUNF	_	Return Stack F	Pointer				uu-0 0000	
FFBh	PCLATU	_	— — Holding Register for PC<20:16>								
FFAh	PCLATH	Holding Regis	ter for PC<15:	8>						0000 0000	
FF9h	PCL	PC Low Byte	Low Byte (PC<7:0>)								
FF8h	TBLPTRU	_	_	bit 21	Program Mem	ory Table Point	ter Upper Byte	(TBLPTR<20:	16>)	00 0000	
FF7h	TBLPTRH	Program Merr	nory Table Poir	ter High Byte	(TBLPTR<15:8	>)				0000 0000	
FF6h	TBLPTRL	Program Merr	Program Memory Table Pointer Low Byte (TBLPTR<7:0>)								
FF5h	TABLAT	Program Merr	Program Memory Table Latch								
FF4h	PRODH	Product Regis	Product Register High Byte								
FF3h	PRODL	Product Regis	ter Low Byte							XXXX XXXX	
FF2h	INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	INT0IF	RBIF	0000 000x	
FF1h	INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP	1111 1111	
FF0h	INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF	1100 0000	
FEFh	INDF0	Uses contents	of FSR0 to ac	dress data m	emory – value o	f FSR0 not cha	anged (not a pl	nysical register)		
FEEh	POSTINC0	Uses contents	of FSR0 to ac	dress data m	emory – value o	f FSR0 post-in	cremented (no	t a physical reg	gister)		
FEDh	POSTDEC0	Uses contents	of FSR0 to ac	dress data m	emory – value o	f FSR0 post-de	ecremented (ne	ot a physical re	gister)		
FECh	PREINC0	Uses contents	Uses contents of FSR0 to address data memory – value of FSR0 pre-incremented (not a physical register)								
FEBh	PLUSW0	Uses contents of FSR0 to address data memory – value of FSR0 pre-incremented (not a physical register) – value of FSR0 offset by W									
FEAh	FSR0H	_	— — — Indirect Data Memory Address Pointer 0 High							0000	
FE9h	FSR0L	Indirect Data N	Indirect Data Memory Address Pointer 0 Low Byte							XXXX XXXX	
FE8h	WREG	Working Regis	ster							XXXX XXXX	
FE7h	INDF1	Uses contents	of FSR1 to ac	dress data m	emory – value o	f FSR1 not cha	anged (not a pl	nysical register)		

TABLE 6-2: PIC18F87K22 FAMILY REGISTER FILE SUMMARY

Note 1: This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented.

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

3: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
FE6h	POSTINC1	Uses contents	of FSR1 to ac	ldress data me	emory – value o	f FSR1 post-in	cremented (no	t a physical reg	gister)	
FE5h	POSTDEC1	Uses contents	of FSR1 to ac	ldress data me	emory – value o	f FSR1 post-de	ecremented (no	ot a physical re	gister)	
FE4h	PREINC1	Uses contents	of FSR1 to ac	ldress data me	emory – value o	f FSR1 pre-inc	remented (not	a physical regi	ster)	
FE3h	PLUSW1	Uses contents FSR1 offset by		dress data me	mory – value of	FSR1 pre-inci	remented (not a	a physical regis	ster) – value of	
FE2h	FSR1H	—	—	—	—	Indirect Data	Memory Addre	ss Pointer 1 H	igh	xxxx
FE1h	FSR1L	Indirect Data N	Memory Addre	ss Pointer 1 Lo	ow Byte					XXXX XXXX
FE0h	BSR	—	—		_	Bank Select F	Register			0000
FDFh	INDF2	Uses contents	of FSR2 to ac	ldress data me	emory – value o	f FSR2 not cha	anged (not a ph	nysical register)	
FDEh	POSTINC2	Uses contents	of FSR2 to ac	ldress data me	emory – value o	f FSR2 post-in	cremented (no	t a physical reg	jister)	
FDDh	POSTDEC2	Uses contents	of FSR2 to ac	ldress data me	emory – value o	f FSR2 post-de	ecremented (no	ot a physical re	gister)	
FDCh	PREINC2	Uses contents	of FSR2 to ac	ldress data me	emory – value o	f FSR2 pre-inc	remented (not	a physical regi	ster)	
FDBh	PLUSW2	Uses contents FSR2 offset by		ldress data me	mory – value of	FSR2 pre-incr	remented (not a	a physical regis	ster) – value of	
FDAh	FSR2H	_	_	_	_	Indirect Data	Memory Addre	ss Pointer 2 H	igh	xxxx
FD9h	FSR2L	Indirect Data M	Memory Addre	ss Pointer 2 Lo	ow Byte					xxxx xxxx
FD8h	STATUS	_	_	_	N	OV	Z	DC	С	x xxxx
FD7h	TMR0H	Timer0 Regist	er High Byte							0000 0000
FD6h	TMR0L	Timer0 Regist	er Low Byte							XXXX XXXX
FD5h	T0CON	TMR00N	T08BIT	TOCS	T0SE	PSA	TOPS2	TOPS1	TOPS0	1111 1111
FD4h	SPBRGH1	USART1 Bau	d Rate Genera	tor High Byte						0000 0000
FD3h	OSCCON	IDLEN	IRCF2	IRCF1	IRCF0	OSTS	HFIOFS	SCS1	SCS0	0110 q000
FD2h	IPR5	TMR7GIP ⁽³⁾	TMR12IP ⁽³⁾	TMR10IP ⁽³⁾	TMR8IP	TMR7IP ⁽³⁾	TMR6IP	TMR5IP	TMR4IP	1111 1111
FD1h	WDTCON	REGSLP	_	ULPLVL	SRETEN	_	ULPEN	ULPSINK	SWDTEN	0-x0 -000
FD0h	RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR	0111 11qq
FCFh	TMR1H	Timer1 Regist		0M	ru.	10	10	1 011	Dort	xxxx xxxx
FCEh	TMR1L	Timer1 Regist	• •							xxxx xxxx
FCDh	T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N	0000 0000
FCCh	TMR2	Timer2 Regist		TICKEST	TICKE 30	SUSCEN	TISTING	KD TO	TWICTON	0000 0000
FCBh	PR2	, , , , , , , , , , , , , , , , , , ,								
FCAh	T2CON	Timer2 Period	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0	1111 1111
FC9h	SSP1BUF		e Buffer/Trans		12001531	12001F30	TIVIRZON	1206531	1204530	-000 0000
	SSP1BUP				ode. SSP1 Bau	l Data Dalaad	Decister in 120	Maatar Mada		XXXX XXXX
FC8h			<u> </u>	_						0000 0000
FC7h	SSP1STAT	SMP	CKE	D/A	P	S	R/W	UA	BF	0000 0000
FC6h	SSP1CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000
FC5h	SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0000 0000
FC4h	ADRESH	A/D Result Register High Byte						XXXX XXXX		
FC3h	ADRESL	A/D Result Re	egister Low Byt			1			1	XXXX XXXX
FC2h	ADCON0	—	CHS4	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON	-000 0000
FC1h	ADCON1	TRIGSEL1	TRIGSEL0	VCFG1	VCFG0	VNCFG	CHSN2	CHSN1	CHSN0	0000 0000
FC0h	ADCON2	ADFM	—	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0	0-00 0000
FBFh	ECCP1AS	ECCP1ASE	ECCP1AS2	ECCP1AS1	ECCP1AS0	PSS1AC1	PSS1AC0	PSS1BD1	PSS1BD0	0000 0000
FBEh	ECCP1DEL	P1RSEN	P1DC6	P1DC5	P1DC4	P1DC3	P1DC2	P1DC1	P1DC0	0000 0000
FBDh	CCPR1H	Capture/Compare/PWM Register1 High Byte							XXXX XXXX	
FBCh	CCPR1L		bare/PWM Reg	gister1 Low By		1	1		1	XXXX XXXX
FBBh	CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0	0000 0000
FBAh	PIR5	TMR7GIF ⁽³⁾	TMR12IF ⁽³⁾	TMR10IF ⁽³⁾	TMR8IF	TMR7IF ⁽³⁾	TMR6IF	TMR5IF	TMR4IF	0000 0000
FB9h	PIE5	TMR7GIE ⁽³⁾	TMR12IE ⁽³⁾	TMR10IE ⁽³⁾	TMR8IE	TMR7IE ⁽³⁾	TMR6IE	TMR5IE	TMR4IE	0000 0000
FB8h	IPR4	CCP10IP ⁽³⁾	CCP9IP ⁽³⁾	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP	1111 1111
FB7h	PIR4	CCP10IF ⁽³⁾	CCP9IF ⁽³⁾	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF	0000 0000

PIC18E87K22 EAMILY REGISTER FILE SUMMARY (CONTINUED) **TARI F 6-2**

This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented. Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'. Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22). Note 1:

2: 3:

TABLE 6-2: PI	C18F87K22 FAMILY REGISTER FILE SUMMARY (CONTINUED)
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Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
FB6h	PIE4	CCP10IE ⁽³⁾	CCP9IE ⁽³⁾	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE	0000 0000
FB5h	CVRCON	CVREN	CVROE	CVRSS	CVR4	CVR3	CVR2	CVR1	CVR0	0000 0000
FB4h	CMSTAT	CMP3OUT	CMP2OUT	CMP1OUT	_	_	_	_	_	xxx
FB3h	TMR3H	Timer3 Regist	er High Byte							XXXX XXXX
FB2h	TMR3L	Timer3 Regist	er Low Byte							XXXX XXXX
FB1h	T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	SOSCEN	T3SYNC	RD16	TMR3ON	0000 0000
FB0h	T3GCON	TMR3GE	T3GPOL	T3GTM	T3GSPM	T3GGO/ T3DONE	T3GVAL	T3GSS1	T3GSS0	0000 0x00
FAFh	SPBRG1	USART1 Bau	d Rate Genera	tor						0000 0000
FAEh	RCREG1	USART1 Rece	eive Register							0000 0000
FADh	TXREG1	USART1 Tran	smit Register							XXXX XXXX
FACh	TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010
FABh	RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x
FAAh	T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/ T1DONE	T1GVAL	T1GSS1	T1GSS0	0000 0x00
FA9h	IPR6	—			EEIP		CMP3IP	CMP2IP	CMP1IP	1 -111
FA8h	HLVDCON	VDIRMAG	BGVST	IRVST	HLVDEN	HLVDL3	HLVDL2	HLVDL1	HLVDL0	0000 0000
FA7h	PSPCON	IBF	OBF	IBOV	PSPMODE	-	_	-	—	0000
FA6h	PIR6	_	_	_	EEIF	_	CMP3IF	CMP2IF	CMP1IF	0 -000
FA5h	IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	1-11 1111
FA4h	PIR3	TMR5GiF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	0-00 0000
FA3h	PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	0-00 0000
FA2h	IPR2	OSCFIP	_	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP	1-11 1111
FA1h	PIR2	OSCFIF	_	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF	0-00 0000
FA0h	PIE2	OSCFIE	_	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE	0-00 0000
F9Fh	IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP	1111 1111
F9Eh	PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF	0000 0000
F9Dh	PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE	0000 0000
F9Ch	PSTR1CON	CMPL1	CMPL0	_	STRSYNC	STRD	STRC	STRB	STRA	00-0 0001
F9Bh	OSCTUNE	INTSRC	PLLEN	TUN5	TUN4	TUN3	TUN2	TUN1	TUN0	0000 0000
F9Ah	TRISJ ⁽²⁾	TRISJ7	TRISJ6	TRISJ5	TRISJ4	TRISJ3	TRISJ2	TRISJ1	TRISJ0	1111 1111
F99h	TRISH ⁽²⁾	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0	1111 1111
F98h	TRISG	_	_	_	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0	1 1111
F97h	TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_	1111 111-
F96h	TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	_	1111 111-
F95h	TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	1111 1111
F94h	TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	1111 1111
F93h	TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	1111 1111
F92h	TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	1111 1111
F91h	LATJ ⁽²⁾	LATJ7	LATJ6	LATJ5	LATJ4	LATJ3	LATJ2	LATJ1	LATJ0	xxxx xxxx
F90h	LATH ⁽²⁾	LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0	xxxx xxxx
F8Fh	LATG	_	_	_	LATG4	LATG3	LATG2	LATG1	LATG0	x xxxx
F8Eh	LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	—	xxxx xxx-
F8Dh	LATE	LATE7	LATE6	LATE5	LATE4	LATE3	LATE2	LATE1	LATE0	xxxx xxxx
F8Ch	LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0	xxxx xxxx
F8Bh	LATC	LATC7	LATC6	LATC5	LATC4	LATC3	LATC2	LATC1	LATCO	xxxx xxxx
F8Ah	LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATBO	xxxx xxxx
F89h	LATA	LATA7	LATA6	LATA5	LATA4	LATA3	LATA2	LATA1	LATA0	XXXX XXXX
F88h	PORTJ ⁽²⁾	RJ7	RJ6	RJ5	RJ4	RJ3	RJ2	RJ1	RJ0	XXXX XXXX
F87h	PORTH ⁽²⁾	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0	XXXX XXXX

Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
F86h	PORTG	_	_	RG5 ⁽¹⁾	RG4	RG3	RG2	RG1	RG0	xx xxxx
F85h	PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	_	xxxx xxx-
F84h	PORTE	RE7	RE6	RE5	RE4	RE3	RE2	RE1	RE0	XXXX XXXX
F83h	PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0	XXXX XXXX
F82h	PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	XXXX XXXX
F81h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx
F80h	PORTA	RA7	RA6	RA5	RA4	RA3	RA2	RA1	RA0	XXXX XXXX
F7Fh	EECON1	EEPGD	CFGS	_	FREE	WRERR	WREN	WR	RD	xx-0 x000
F7Eh	EECON2	EEPROM Cor	EPROM Control Register 2 (not a physical register)							
F7Dh	TMR5H	Timer5 Regist	er High Byte							XXXX XXXX
F7Ch	TMR5L	Timer5 Regist	er Low Byte							XXXX XXXX
F7Bh	T5CON	TMR5CS1	TMR5CS0	T5CKPS1	T5CKPS0	SOSCEN	T5SYNC	RD16	TMR50N	0000 0000
F7Ah	T5GCON	TMR5GE	T5GPOL	T5GTM	T5GSPM	T5GGO/ T5DONE	T5GVAL	T5GSS1	T5GSS0	0000 0x00
F79h	CCPR4H	Capture/Com	bare/PWM Reg	gister 4 High B	yte					XXXX XXXX
F78h	CCPR4L	Capture/Com	bare/PWM Reg	gister 4 Low By	/te					XXXX XXXX
F77h	CCP4CON	_	—	DC4B1	DC4B0	CCP4M3	CCP4M2	CCP4M1	CCP4M0	00 0000
F76h	CCPR5H	Capture/Compare/PWM Register 5 High Byte						XXXX XXXX		
F75h	CCPR5L	Capture/Compare/PWM Register 5 Low Byte							XXXX XXXX	
F74h	CCP5CON	_	_	DC5B1	DC5B0	CCP5M3	CCP5M2	CCP5M1	CCP5M0	00 0000
F73h	CCPR6H	Capture/Com	bare/PWM Reg	gister 6 High B	yte					XXXX XXXX
F72h	CCPR6L	Capture/Com	Capture/Compare/PWM Register 6 Low Byte							xxxx xxxx
F71h	CCP6CON	_	_	DC6B1	DC6B0	CCP6M3	CCP6M2	CCP6M1	CCP6M0	00 0000
F70h	CCPR7H	Capture/Com	bare/PWM Reg	gister 7 High B	yte					XXXX XXXX
F6Fh	CCPR7L	Capture/Com	bare/PWM Reg	gister 7 Low By	/te					XXXX XXXX
F6Eh	CCP7CON	_	_	DC7B1	DC7B0	CCP7M3	CCP7M2	CCP7M1	CCP7M0	00 0000
F6Dh	TMR4	Timer4 Regist	er							XXXX XXXX
F6Ch	PR4	Timer4 Period	Register							1111 1111
F6Bh	T4CON	_	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR40N	T4CKPS1	T4CKPS0	-111 1111
F6Ah	SSP2BUF	MSSP Receiv	e Buffer/Trans	mit Register						XXXX XXXX
F69h	SSP2ADD	MSSP Addres	s Register in I	² C [™] Slave Mo	de. MSSP1 Ba	ud Rate Reloa	d Register in I ²	C Master Mod	e.	0000 0000
F68h	SSP2STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000 0000
F67h	SSP2CON1	WCOL	SSPOV	SSPEN	СКР	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000
F66h	SSP2CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0100 0000
F65h	BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	0100 0-00
F64h	OSCCON2	_	SOSCRUN	_	_	SOSCGO	_	MFIOFS	MFIOSEL	-0 0-x0
F63h	EEADRH	EEPROM Add	dress Register	High Byte						0000 0000
F62h	EEADR		dress Register	• •						0000 0000
F61h	EEDATA	EEPROM Dat	•							0000 0000
F60h	PIE6	_		_	EEIE	_	CMP3IE	CMP2IE	CMP1IE	0 -000
F5Fh	RTCCFG	RTCEN	_	RTCWREN	RTCSYNC	HALFSEC	RTCOE	RTCPTR1	RTCPTR0	0-00 0000
F5Eh	RTCCAL	CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0	0000 0000
F5Dh	RTCVALH	RTCC Value H	High Register V	Vindow Based	on RTCPTR<1	:0>				XXXX XXXX

PIC18F87K22 FAMILY REGISTER FILE SUMMARY (CONTINUED) **TABLE 6-2:**

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Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
F5Ch	RTCVALL	RTCC Value L	ow Register W	/indow Based	on RTCPTR<1	0>				0000 0000
F5Bh	ALRMCFG	ALRMEN	CHIME	AMASK3	AMASK2	AMASK1	AMASK0	ALRMPTR1	ALRMPTR0	0000 0000
F5Ah	ALRMRPT	ARPT7	ARPT6	ARPT5	ARPT4	ARPT3	ARPT2	ARPT1	ARPT0	0000 0000
F59h	ALRMVALH	Alarm Value H	ligh Register W	Vindow Based	on APTR<1:0>	•		•	•	xxxx xxxx
F58h	ALRMVALL	Alarm Value L	ow Register W	indow Based	on APTR<1:0>					xxxx xxxx
F57h	CTMUCONH	CTMUEN	_	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG	0-00 0000
F56h	CTMUCONL	EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT	0000 0000
F55h	CTMUICONH	ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1	ITRIM0	IRNG1	IRNG0	0000 0000
F54h	CM1CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0	0001 1111
F53h	PADCFG1	RDPU	REPU	rjpu ⁽²⁾	_	—	RTSECSEL1	RTSECSEL0	—	00000-
F52h	ECCP2AS	ECCP2ASE	ECCP2AS2	ECCP2AS1	ECCP2AS0	PSS2AC1	PSS2AC0	PSS2BD1	PSS2BD0	0000 0000
F51h	ECCP2DEL	P2RSEN	P2DC6	P2DC5	P2DC4	P2DC3	P2DC2	P2DC1	P2DC0	0000 0000
F50h	CCPR2H	Capture/Comp	pare/PWM Reg	gister 2 High B	yte					xxxx xxxx
F4Fh	CCPR2L	Capture/Comp	pare/PWM Reg	gister 2 Low By	/te					xxxx xxxx
F4Eh	CCP2CON	P2M1	P2M0	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0	0000 0000
F4Dh	ECCP3AS	ECCP3ASE	ECCP3AS2	ECCP3AS1	ECCP3AS0	PSS3AC1	PSS3AC0	PSS3BD1	PSS3BD0	0000 0000
F4Ch	ECCP3DEL	P3RSEN	P3DC6	P3DC5	P3DC4	P3DC3	P3DC2	P3DC1	P3DC0	0000 0000
F4Bh	CCPR3H	Capture/Comp	pare/PWM Reg	gister 3 High B	vte					xxxx xxxx
F4Ah	CCPR3L		pare/PWM Reg		-					xxxx xxxx
F49h	CCP3CON	P3M1	P3M0	DC3B1	DC3B0	CCP3M3	CCP3M2	CCP3M1	CCP3M0	0000 0000
F48h	CCPR8H								xxxx xxxx	
F47h	CCPR8L		pare/PWM Reg	, 0	,					xxxx xxxx
F46h	CCP8CON	_	_	DC8B1	DC8B0	CCP8M3	CCP8M2	CCP8M1	CCP8M0	00 0000
F45h	CCPR9H ⁽³⁾	Capture/Comr	are/PWM Reg							xxxx xxxx
F44h	CCPR9L ⁽³⁾		pare/PWM Reg							xxxx xxxx
F43h	CCP9CON ⁽³⁾		_	DC9B1	DC9B0	CCP9M3	CCP9M2	CCP9M1	CCP9M0	00 0000
F42h	CCPR10H ⁽³⁾	Capture/Comr	are/PWM Reg			00101110	00.0112	0010111	00.0110	xxxx xxxx
F41h	CCPR10L ⁽³⁾		pare/PWM Reg	, v						xxxx xxxx
F40h	CCP10CON ⁽³⁾		_	DC10B1	DC10B0	CCP10M3	CCP10M2	CCP10M1	CCP10M0	00 0000
F3Fh	TMR7H ⁽³⁾	Timer7 Regist	er High Byte	001001	001000					xxxx xxxx
F3Eh	TMR7L ⁽³⁾	Timer7 Regist								0000 0000
F3Dh	T7CON ⁽³⁾	TMR7CS1	TMR7CS0	T7CKPS1	T7CKPS0	SOSCEN	T7SYNC	RD16	TMR70N	0000 0000
F3Ch	T7GCON ⁽³⁾	TMR7GE	T7GPOL	T7GTM	T7GSPM	T7GGO/	T7GVAL	T7GSS1	T7GSS0	0000 0000 0000 0x00
1 3011	170001		TIONOL	17011	1700110	T7DONE	TTOVAL	170001	170000	0000 0100
F3Bh	TMR6	Timer6 Regist	er							0000 0000
F3Ah	PR6	Timer6 Period								1111 1111
F39h	T6CON	_	T6OUTPS3	T6OUTPS2	T6OUTPS1	T6OUTPS0	TMR6ON	T6CKPS1	T6CKPS0	-000 0000
F38h	TMR8	Timer8 Regist	er							0000 0000
F37h	PR8	Timer8 Period	Register							1111 1111
F36h	T8CON	_	T8OUTPS3	T8OUTPS2	T8OUTPS1	T8OUTPS0	TMR8ON	T8CKPS1	T8CKPS0	-000 0000
F35h	TMR10 ⁽³⁾	TMR10 Regis	ter							0000 0000
F34h	PR10 ⁽³⁾								1111 1111	
F33h	T10CON ⁽³⁾		T10OUTPS3	T10OUTPS2	T10OUTPS1	T10OUTPS0	TMR100N	T10CKPS1	T10CKPS0	-000 0000
F32h	TMR12 ⁽³⁾	TMR12 Regis								0000 0000
F31h	PR12 ⁽³⁾	-								1111 1111
F30h	T12CON ⁽³⁾	_	T12OUTPS3	T12OUTPS2	T12OUTPS1	T12OUTPS0	TMR12ON	T12CKPS1	T12CKPS0	-000 0000
	CM2CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0	0001 1111
FZFD				0.01				0.0111	0.0110	
F2Fh F2Eh	CM3CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0	0001 1111

PIC18E87K22 EAMILY REGISTER FILE SUMMARY (CONTINUED) TARI E 6-2.

IADLL	TABLE 0-2. FICTOFO/R22 FAMILT REGISTER FILE SUMMART (CONTINUED)									
Address	File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR, BOR
F2Ch	CCPTMRS1	C7TSEL1	C7TSEL0	-	C6TSEL0	—	C5TSEL0	C4TSEL1	C4TSEL0	00-0 -000
F2Bh	CCPTMRS2	—	_	—	C10TSEL0 ⁽³⁾	_	C9TSEL0 ⁽³⁾	C8TSEL1	C8TSEL0	0 -000
F2Ah	REFOCON	ROON	_	ROSSLP	ROSEL	RODIV3	RODIV2	RODIV1	RODIV0	0-00 0000
F29h	ODCON1	SSP10D	CCP2OD	CCP10D	_	_	_	_	SSP2OD	0000
F28h	ODCON2	CCP10OD ⁽³⁾	CCP9OD ⁽³⁾	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD	0000 0000
F27h	ODCON3	U2OD	U10D	_	_	_	—	_	CTMUDS	000
F26h	MEMCON ⁽²⁾	EBDIS	_	WAIT1	WAIT0	_	—	WM1	WM0	0-0000
F25h	ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0	1111 1111
F24h	ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8	1111 1111
F23h	ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16	1111 1111
F22h	RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x
F21h	TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	0000 0010
F20h	BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	0100 0-00
F1Fh	SPBRGH2	2 USART2 Baud Rate Generator High Byte								0000 0000
F1Eh	SPBRG2	USART2 Baud Rate Generator							0000 0000	
F1Dh	RCREG2	Receive Data	FIFO							0000 0000
F1Ch	TXREG2	Transmit Data	FIFO							xxxx xxxx
F1Bh	PSTR2CON	CMPL1	CMPL0	—	STRSYNC	STRD	STRC	STRB	STRA	00-0 0001
F1Ah	PSTR3CON	CMPL1	CMPL0	_	STRSYNC	STRD	STRC	STRB	STRA	00-0 0001
F19h	PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD	0000 0000
F18h	PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD	0000 0000
F17h	PMD2	TMR10MD ⁽³⁾	TMR8MD	TMR7MD ⁽³⁾	TMR6MD	TMR5MD	CMP3MD	CMP2MD	CMP1MD	0000 0000
F16h	PMD3	CCP10MD ⁽³⁾	CCP9MD ⁽³⁾	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD ⁽³⁾	0000 0000

TABLE 6-2: PIC18F87K22 FAMILY REGISTER FILE SUMMARY (CONTINUED)

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6.3.5 STATUS REGISTER

The STATUS register, shown in Register 6-2, contains the arithmetic status of the ALU. The STATUS register can be the operand for any instruction, as with any other register. If the STATUS register is the destination for an instruction that affects the Z, DC, C, OV or N bits, the write to these five bits is disabled.

These bits are set or cleared according to the device logic. Therefore, the result of an instruction with the STATUS register as destination may be different than intended. For example, CLRF STATUS will set the Z bit but leave the other bits unchanged. The STATUS register then reads back as '000u uluu'.

It is recommended, therefore, that only BCF, BSF, SWAPF, MOVFF and MOVWF instructions be used to alter the STATUS register because these instructions do not affect the Z, C, DC, OV or N bits in the STATUS register.

For other instructions not affecting any Status bits, see the instruction set summaries in Table 29-2 and Table 29-3.

Note: The C and DC bits operate, in subtraction, as borrow and digit borrow bits, respectively.

REGISTER 6-2: STATUS REGISTER

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x		
—	—		N	OV	Z	DC ⁽¹⁾	C ⁽²⁾		
bit 7							bit 0		
Lovendi									
Legend: R = Read	lable hit	W = Writable	hit	II – I Inimpler	nented bit, rea	ud as 'O'			
	e at POR	'1' = Bit is set		'0' = Bit is cle		x = Bit is unkr	NOWD		
			•		area		lowin		
bit 7-5	Unimplem	ented: Read as '	0'						
bit 4	N: Negative	e bit							
	This bit is u (ALU MSB	<pre>ised for signed ar = 1).</pre>	rithmetic (2's c	omplement). It	indicates whet	her the result wa	as negative		
		was negative was positive							
bit 3	OV: Overflo	ow bit							
		used for signed an which causes the				verflow of the se	even-bit		
		w occurred for signal for signal for signal for signal for the second se	gned arithmeti	c (in this arithm	etic operation))			
bit 2	Z: Zero bit								
		sult of an arithme sult of an arithme			o				
bit 1		arry/Borrow bit ⁽¹⁾ , ADDLW, SUBI		instructions:					
		-out from the 4th			urred				
		y-out from the 4t							
bit 0	C: Carry/Bo	orrow bit ⁽²⁾							
	For ADDWF, ADDLW, SUBLW and SUBWF instructions:								
		-out from the Mo							
	0 = No carr	y-out from the M	ost Significant	DIT OF THE RESULT	occurred				
Note 1:	For borrow, the operand.	polarity is reverse	ed. A subtraction	on is executed t	by adding the 2	's complement o	of the second		
2:	·								

6.4 Data Addressing Modes

Note:	The execution of some instructions in the
	core PIC18 instruction set are changed
	when the PIC18 extended instruction set is
	enabled. For more information, see
	Section 6.6 "Data Memory and the
	Extended Instruction Set".

While the program memory can be addressed in only one way, through the Program Counter, information in the data memory space can be addressed in several ways. For most instructions, the addressing mode is fixed. Other instructions may use up to three modes, depending on which operands are used and whether or not the extended instruction set is enabled.

The addressing modes are:

- Inherent
- Literal
- Direct
- Indirect

An additional addressing mode, Indexed Literal Offset, is available when the extended instruction set is enabled (XINST Configuration bit = 1). For details on this mode's operation, see Section 6.6.1 "Indexed Addressing with Literal Offset".

6.4.1 INHERENT AND LITERAL ADDRESSING

Many PIC18 control instructions do not need any argument at all. They either perform an operation that globally affects the device or they operate implicitly on one register. This addressing mode is known as Inherent Addressing. Examples of this mode include SLEEP, RESET and DAW.

Other instructions work in a similar way, but require an additional explicit argument in the opcode. This method is known as the Literal Addressing mode because the instructions require some literal value as an argument. Examples of this include ADDLW and MOVLW, which respectively, add or move a literal value to the W register. Other examples include CALL and GOTO, which include a 20-bit program memory address.

6.4.2 DIRECT ADDRESSING

Direct Addressing specifies all or part of the source and/or destination address of the operation within the opcode itself. The options are specified by the arguments accompanying the instruction.

In the core PIC18 instruction set, bit-oriented and byte-oriented instructions use some version of Direct Addressing by default. All of these instructions include some 8-bit literal address as their Least Significant Byte. This address specifies the instruction's data source as either a register address in one of the banks of data RAM (see Section 6.3.3 "General Purpose Register File") or a location in the Access Bank (see Section 6.3.2 "Access Bank").

The Access RAM bit, 'a', determines how the address is interpreted. When 'a' is '1', the contents of the BSR (Section 6.3.1 "Bank Select Register") are used with the address to determine the complete 12-bit address of the register. When 'a' is '0', the address is interpreted as being a register in the Access Bank. Addressing that uses the Access RAM is sometimes also known as Direct Forced Addressing mode.

A few instructions, such as MOVFF, include the entire 12-bit address (either source or destination) in their opcodes. In these cases, the BSR is ignored entirely.

The destination of the operation's results is determined by the destination bit, 'd'. When 'd' is '1', the results are stored back in the source register, overwriting its original contents. When 'd' is '0', the results are stored in the W register. Instructions without the 'd' argument have a destination that is implicit in the instruction, either the target register being operated on or the W register.

6.4.3 INDIRECT ADDRESSING

Indirect Addressing allows the user to access a location in data memory without giving a fixed address in the instruction. This is done by using File Select Registers (FSRs) as pointers to the locations to be read or written to. Since the FSRs are themselves located in RAM as Special Function Registers, they can also be directly manipulated under program control. This makes FSRs very useful in implementing data structures such as tables and arrays in data memory.

The registers for Indirect Addressing are also implemented with Indirect File Operands (INDFs) that permit automatic manipulation of the pointer value with auto-incrementing, auto-decrementing or offsetting with another value. This allows for efficient code using loops, such as the example of clearing an entire RAM bank in Example 6-5. It also enables users to perform Indexed Addressing and other Stack Pointer operations for program memory in data memory.

EXAMPLE 6-5: HOW TO CLEAR RAM (BANK 1) USING INDIRECT ADDRESSING

	LFSR	FSR0, 100h	;	
NEXT	CLRF	POSTINC0	;	Clear INDF
			;	register then
			;	inc pointer
	BTFSS	FSROH, 1	;	All done with
			;	Bank1?
	BRA	NEXT	;	NO, clear next
CONTIN	UE	;	YES, continue	

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6.4.3.1 FSR Registers and the INDF Operand

At the core of Indirect Addressing are three sets of registers: FSR0, FSR1 and FSR2. Each represents a pair of 8-bit registers: FSRnH and FSRnL. The four upper bits of the FSRnH register are not used, so each FSR pair holds a 12-bit value. This represents a value that can address the entire range of the data memory in a linear fashion. The FSR register pairs, then, serve as pointers to data memory locations.

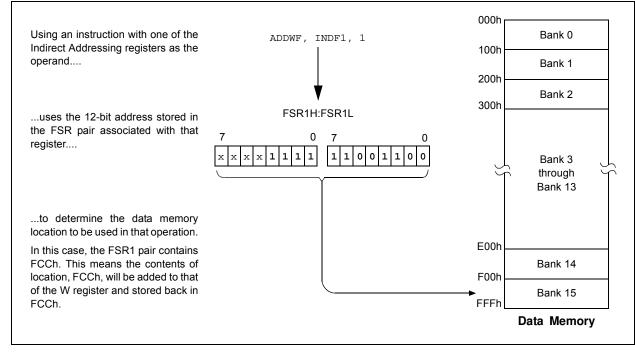
Indirect Addressing is accomplished with a set of Indirect File Operands, INDF0 through INDF2. These can be thought of as "virtual" registers. The operands are

FIGURE 6-8: INDIRECT ADDRESSING

mapped in the SFR space, but are not physically implemented. Reading or writing to a particular INDF register actually accesses its corresponding FSR register pair. A read from INDF1, for example, reads the data at the address indicated by FSR1H:FSR1L.

Instructions that use the INDF registers as operands actually use the contents of their corresponding FSR as a pointer to the instruction's target. The INDF operand is just a convenient way of using the pointer.

Because Indirect Addressing uses a full 12-bit address, data RAM banking is not necessary. Thus, the current contents of the BSR and the Access RAM bit have no effect on determining the target address.



6.4.3.2 FSR Registers and POSTINC, POSTDEC, PREINC and PLUSW

In addition to the INDF operand, each FSR register pair also has four additional indirect operands. Like INDF, these are "virtual" registers that cannot be indirectly read or written to. Accessing these registers actually accesses the associated FSR register pair, but also performs a specific action on its stored value.

These operands are:

- POSTDEC Accesses the FSR value, then automatically decrements it by '1' afterwards
- POSTINC Accesses the FSR value, then automatically increments it by '1' afterwards
- PREINC Increments the FSR value by '1', then uses it in the operation
- PLUSW Adds the signed value of the W register (range of -127 to 128) to that of the FSR and uses the new value in the operation

In this context, accessing an INDF register uses the value in the FSR registers without changing them. Similarly, accessing a PLUSW register gives the FSR value, offset by the value in the W register, with neither value actually changed in the operation. Accessing the other virtual registers changes the value of the FSR registers.

Operations on the FSRs with POSTDEC, POSTINC and PREINC affect the entire register pair. Rollovers of the FSRnL register, from FFh to 00h, carry over to the FSRnH register. On the other hand, results of these operations do not change the value of any flags in the STATUS register (for example, Z, N and OV bits).

The PLUSW register can be used to implement a form of Indexed Addressing in the data memory space. By manipulating the value in the W register, users can reach addresses that are fixed offsets from pointer addresses. In some applications, this can be used to implement some powerful program control structure, such as software stacks, inside of data memory.

6.4.3.3 Operations by FSRs on FSRs

Indirect Addressing operations that target other FSRs or virtual registers represent special cases. For example, using an FSR to point to one of the virtual registers will not result in successful operations.

As a specific case, assume that the FSR0H:FSR0L registers contain FE7h, the address of INDF1. Attempts to read the value of the INDF1, using INDF0 as an operand, will return 00h. Attempts to write to INDF1, using INDF0 as the operand, will result in a NOP.

On the other hand, using the virtual registers to write to an FSR pair may not occur as planned. In these cases, the value will be written to the FSR pair, but without any incrementing or decrementing. Thus, writing to INDF2 or POSTDEC2 will write the same value to the FSR2H:FSR2L.

Since the FSRs are physical registers mapped in the SFR space, they can be manipulated through all direct operations. Users should proceed cautiously when working on these registers, however, particularly if their code uses Indirect Addressing.

Similarly, operations by Indirect Addressing are generally permitted on all other SFRs. Users should exercise the appropriate caution, so that they do not inadvertently change settings that might affect the operation of the device.

6.5 Program Memory and the Extended Instruction Set

The operation of program memory is unaffected by the use of the extended instruction set.

Enabling the extended instruction set adds five additional two-word commands to the existing PIC18 instruction set: ADDFSR, CALLW, MOVSF, MOVSS and SUBFSR. These instructions are executed as described in Section 6.2.4 "Two-Word Instructions".

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6.6 Data Memory and the Extended Instruction Set

Enabling the PIC18 extended instruction set (XINST Configuration bit = 1) significantly changes certain aspects of data memory and its addressing. Using the Access Bank for many of the core PIC18 instructions introduces a new addressing mode for the data memory space. This mode also alters the behavior of Indirect Addressing using FSR2 and its associated operands.

What does not change is just as important. The size of the data memory space is unchanged, as well as its linear addressing. The SFR map remains the same. Core PIC18 instructions can still operate in both Direct and Indirect Addressing mode. Inherent and literal instructions do not change at all. Indirect Addressing with FSR0 and FSR1 also remains unchanged.

6.6.1 INDEXED ADDRESSING WITH LITERAL OFFSET

Enabling the PIC18 extended instruction set changes the behavior of Indirect Addressing using the FSR2 register pair and its associated file operands. Under the proper conditions, instructions that use the Access Bank – that is, most bit-oriented and byte-oriented instructions – can invoke a form of Indexed Addressing using an offset specified in the instruction. This special addressing mode is known as Indexed Addressing with Literal Offset or the Indexed Literal Offset mode.

When using the extended instruction set, this addressing mode requires the following:

- Use of the Access Bank ('a' = 0)
- A file address argument that is less than or equal to 5Fh

Under these conditions, the file address of the instruction is not interpreted as the lower byte of an address (used with the BSR in Direct Addressing) or as an 8-bit address in the Access Bank. Instead, the value is interpreted as an offset value to an Address Pointer specified by FSR2. The offset and the contents of FSR2 are added to obtain the target address of the operation.

6.6.2 INSTRUCTIONS AFFECTED BY INDEXED LITERAL OFFSET MODE

Any of the core PIC18 instructions that can use Direct Addressing are potentially affected by the Indexed Literal Offset Addressing mode. This includes all byte-oriented and bit-oriented instructions, or almost one-half of the standard PIC18 instruction set. Instructions that only use Inherent or Literal Addressing modes are unaffected.

Additionally, byte-oriented and bit-oriented instructions are not affected if they do not use the Access Bank (Access RAM bit = 1), or include a file address of 60h or above. Instructions meeting these criteria will continue to execute as before. A comparison of the different possible addressing modes when the extended instruction set is enabled is shown in Figure 6-9.

Those who desire to use byte-oriented or bit-oriented instructions in the Indexed Literal Offset mode should note the changes to assembler syntax for this mode. This is described in more detail in Section 29.2.1 "Extended Instruction Syntax".

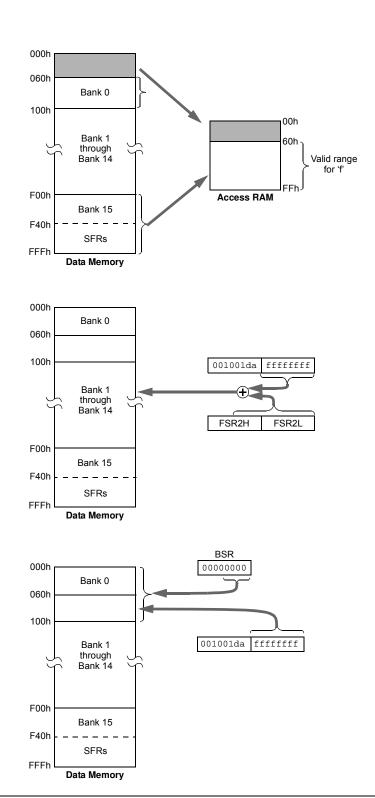
FIGURE 6-9: COMPARING ADDRESSING OPTIONS FOR BIT-ORIENTED AND BYTE-ORIENTED INSTRUCTIONS (EXTENDED INSTRUCTION SET ENABLED)

EXAMPLE INSTRUCTION: ADDWF, f, d, a (Opcode: 0010 01da ffff ffff)

When a = 0 and $f \ge 60h$:

The instruction executes in Direct Forced mode. 'f' is interpreted as a location in the Access RAM, between 060h and FFFh. This is the same as locations, F60h to FFFh (Bank 15), of data memory.

Locations below 060h are not available in this addressing mode.



When a = 0 and $f \le 5Fh$:

The instruction executes in Indexed Literal Offset mode. 'f' is interpreted as an offset to the address value in FSR2. The two are added together to obtain the address of the target register for the instruction. The address can be anywhere in the data memory space.

Note that in this mode, the correct syntax is now: ADDWF [k], d where 'k' is the same as 'f'.

When a = 1 (all values of f):

The instruction executes in Direct mode (also known as Direct Long mode). 'f' is interpreted as a location in one of the 16 banks of the data memory space. The bank is designated by the Bank Select Register (BSR). The address can be in any implemented bank in the data memory space.

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6.6.3 MAPPING THE ACCESS BANK IN INDEXED LITERAL OFFSET MODE

The use of Indexed Literal Offset Addressing mode effectively changes how the lower part of Access RAM (00h to 5Fh) is mapped. Rather than containing just the contents of the bottom part of Bank 0, this mode maps the contents from Bank 0 and a user-defined "window" that can be located anywhere in the data memory space.

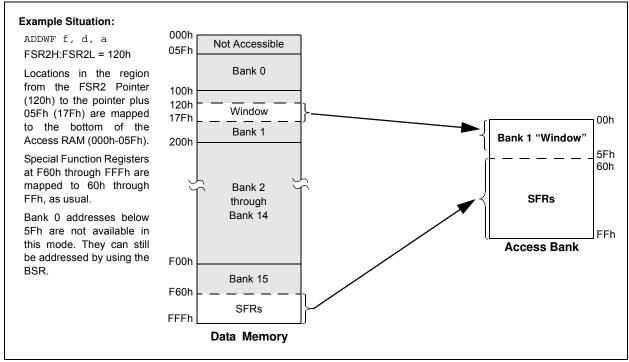
The value of FSR2 establishes the lower boundary of the addresses mapped into the window, while the upper boundary is defined by FSR2 plus 95 (5Fh). Addresses in the Access RAM above 5Fh are mapped as previously described. (See Section 6.3.2 "Access Bank".) An example of Access Bank remapping in this addressing mode is shown in Figure 6-10.

Remapping the Access Bank applies *only* to operations using the Indexed Literal Offset mode. Operations that use the BSR (Access RAM bit = 1) will continue to use Direct Addressing as before. Any Indirect or Indexed Addressing operation that explicitly uses any of the indirect file operands (including FSR2) will continue to operate as standard Indirect Addressing. Any instruction that uses the Access Bank, but includes a register address of greater than 05Fh, will use Direct Addressing and the normal Access Bank map.

6.6.4 BSR IN INDEXED LITERAL OFFSET MODE

Although the Access Bank is remapped when the extended instruction set is enabled, the operation of the BSR remains unchanged. Direct Addressing, using the BSR to select the data memory bank, operates in the same manner as previously described.

FIGURE 6-10: REMAPPING THE ACCESS BANK WITH INDEXED LITERAL OFFSET ADDRESSING



7.0 FLASH PROGRAM MEMORY

The Flash program memory is readable, writable and erasable during normal operation over the entire VDD range.

A read from program memory is executed on one byte at a time. For execution of a write to, or erasure of, program memory:

- Memory of 32 Kbytes and 64 Kbytes (PIC18FX5K22 and PIC18FX6K22 devices) – Blocks of 64 bytes
- Memory of 128 Kbytes (PIC18FX7K22 devices) Blocks of 128 bytes

Writing or erasing program memory will cease instruction fetches until the operation is complete. The program memory cannot be accessed during the write or erase, therefore, code cannot execute. An internal programming timer terminates program memory writes and erases.

A value written to program memory does not need to be a valid instruction. Executing a program memory location that forms an invalid instruction results in a NOP.

7.1 Table Reads and Table Writes

In order to read and write program memory, there are two operations that allow the processor to move bytes between the program memory space and the data RAM:

- Table Read (TBLRD)
- Table Write (TBLWT)

The program memory space is 16 bits wide, while the data RAM space is 8 bits wide. Table reads and table writes move data between these two memory spaces through an 8-bit register (TABLAT).

Table read operations retrieve data from program memory and place it into the data RAM space. Figure 7-1 shows the operation of a table read with program memory and data RAM.

Table write operations store data from the data memory space into holding registers in program memory. The procedure to write the contents of the holding registers into program memory is detailed in **Section 7.5** "Writing **to Flash Program Memory**". Figure 7-2 shows the operation of a table write with program memory and data RAM.

Table operations work with byte entities. A table block containing data, rather than program instructions, is not required to be word-aligned. Therefore, a table block can start and end at any byte address. If a table write is being used to write executable code into program memory, program instructions will need to be word-aligned.

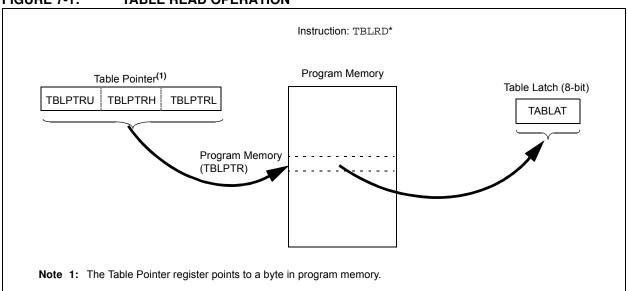
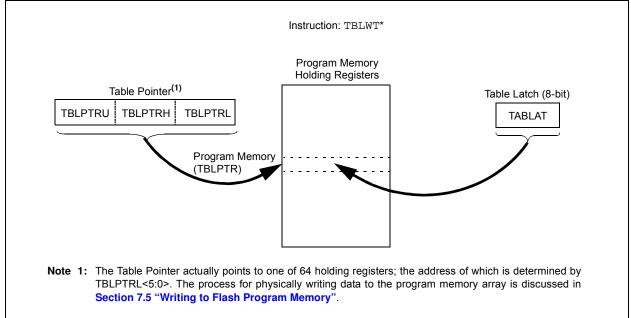


FIGURE 7-1: TABLE READ OPERATION

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FIGURE 7-2: TABLE WRITE OPERATION



7.2 Control Registers

Several control registers are used in conjunction with the ${\tt TBLRD}$ and ${\tt TBLWT}$ instructions. These include the:

- EECON1 register
- EECON2 register
- TABLAT register
- TBLPTR registers

7.2.1 EECON1 AND EECON2 REGISTERS

The EECON1 register (Register 7-1) is the control register for memory accesses. The EECON2 register, not a physical register, is used exclusively in the memory write and erase sequences. Reading EECON2 will read all '0's.

The EEPGD control bit determines if the access is a program or data EEPROM memory access. When clear, any subsequent operations operate on the data EEPROM memory. When set, any subsequent operations operate on the program memory.

The CFGS control bit determines if the access is to the Configuration/Calibration registers or to program memory/data EEPROM memory. When set, subsequent operations operate on Configuration registers regardless of EEPGD (see Section 28.0 "Special Features of the CPU"). When clear, memory selection access is determined by EEPGD.

The FREE bit, when set, allows a program memory erase operation. When FREE is set, the erase operation is initiated on the next WR command. When FREE is clear, only writes are enabled.

The WREN bit, when set, allows a write operation. On power-up, the WREN bit is clear. The WRERR bit is set in hardware when the WR bit is set and cleared when the internal programming timer expires and the write operation is complete.

Note:	During normal operation, the WRERR is									
	read as '1'. This can indicate that a write									
	operation was prematurely terminated by									
	a Reset or a write operation was									
	attempted improperly.									

The WR control bit initiates write operations. The bit cannot be cleared, only set, in software. It is cleared in hardware at the completion of the write operation.

Note: The EEIF interrupt flag bit (PIR6<4>) is set when the write is complete. It must be cleared in software.

REGISTER 7-1: EECON1: EEPROM CONTROL REGISTER 1

R/W-x	R/W-x	U-0	R/W-0	R/W-x	R/W-0	R/S-0	R/S-0
EEPGD	CFGS	—	FREE	WRERR ⁽¹⁾	WREN	WR	RD
bit 7							bit 0

Legend:	S = Settable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	EEPGD: Flash Program or Data EEPROM Memory Select bit
	1 = Access Flash program memory
	0 = Access data EEPROM memory
bit 6	CFGS: Flash Program/Data EEPROM or Configuration Select bit
	1 = Access Configuration registers
	0 = Access Flash program or data EEPROM memory
bit 5	Unimplemented: Read as '0'
bit 4	FREE: Flash Row Erase Enable bit
	1 = Erase the program memory row addressed by TBLPTR on the next WR command
	(cleared by completion of erase operation)
	0 = Perform write-only
bit 3	WRERR: Flash Program/Data EEPROM Error Flag bit ⁽¹⁾
	1 = A write operation is prematurely terminated (any Reset during self-timed programming in normal
	operation or an improper write attempt)
	0 = The write operation completed
bit 2	WREN: Flash Program/Data EEPROM Write Enable bit
	1 = Allows write cycles to Flash program/data EEPROM
	0 = Inhibits write cycles to Flash program/data EEPROM
bit 1	WR: Write Control bit
	1 = Initiates a data EEPROM erase/write cycle or a program memory erase cycle or write cycle
	(The operation is self-timed and the bit is cleared by hardware once the write is complete.
	The WR bit can only be set (not cleared) in software.)
	0 = Write cycle to the EEPROM is complete
bit 0	RD: Read Control bit
	1 = Initiates an EEPROM read (Read takes one cycle. RD is cleared in hardware. The RD bit can only
	be set (not cleared) in software. The RD bit cannot be set when EEPGD = 1 or CFGS = 1.)
	0 = Does not initiate an EEPROM read

Note 1: When a WRERR occurs, the EEPGD and CFGS bits are not cleared. This allows tracing of the error condition.

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7.2.2 TABLAT – TABLE LATCH REGISTER

The Table Latch (TABLAT) is an eight-bit register mapped into the SFR space. The Table Latch register is used to hold 8-bit data during data transfers between program memory and data RAM.

7.2.3 TBLPTR – TABLE POINTER REGISTER

The Table Pointer (TBLPTR) register addresses a byte within the program memory. The TBLPTR is comprised of three SFR registers: Table Pointer Upper Byte, Table Pointer High Byte and Table Pointer Low Byte (TBLPTRU:TBLPTRH:TBLPTRL). These three registers join to form a 22-bit wide pointer. The low-order 21 bits allow the device to address up to 2 Mbytes of program memory space. The 22nd bit allows access to the Device ID, the User ID and the Configuration bits.

The Table Pointer register, TBLPTR, is used by the TBLRD and TBLWT instructions. These instructions can update the TBLPTR in one of four ways, based on the table operation. These operations are shown in Table 7-1 and only affect the low-order 21 bits.

7.2.4 TABLE POINTER BOUNDARIES

The TBLPTR is used in reads, writes and erases of the Flash program memory.

When a TBLRD is executed, all 22 bits of the TBLPTR determine which byte is read from program memory into the TABLAT.

When a TBLWT is executed, the six LSbs of the Table Pointer register (TBLPTR<5:0>) determine which of the 64 program memory holding registers is written to. When the timed write to program memory begins (via the WR bit), the 16 MSbs of the TBLPTR (TBLPTR<21:6>) determine which program memory block of 64 bytes is written to. For more detail, see Section 7.5 "Writing to Flash Program Memory".

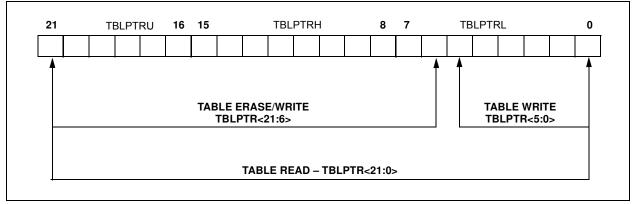
When an erase of program memory is executed, the 16 MSbs of the Table Pointer register (TBLPTR<21:6>) point to the 64-byte block that will be erased. The Least Significant bits (TBLPTR<5:0>) are ignored.

Figure 7-3 describes the relevant boundaries of the TBLPTR based on Flash program memory operations.

TABLE 7-1:	TABLE POINTER OPERATIONS WITH TBLRD AND TBLWT INSTRUCTIONS

Example	Operation on Table Pointer
TBLRD* TBLWT*	TBLPTR is not modified
TBLRD*+ TBLWT*+	TBLPTR is incremented after the read/write
TBLRD*- TBLWT*-	TBLPTR is decremented after the read/write
TBLRD+* TBLWT+*	TBLPTR is incremented before the read/write

FIGURE 7-3: TABLE POINTER BOUNDARIES BASED ON OPERATION



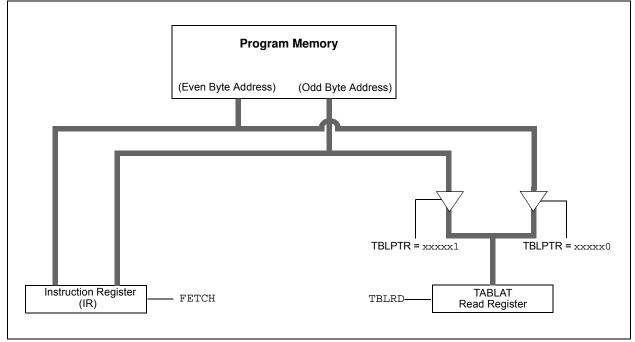
7.3 Reading the Flash Program Memory

The TBLRD instruction is used to retrieve data from program memory and places it into data RAM. Table reads from program memory are performed, one byte at a time.

TBLPTR points to a byte address in program space. Executing TBLRD places the byte pointed to into TABLAT. In addition, the TBLPTR can be modified automatically for the next table read operation.

The internal program memory is typically organized by words. The Least Significant bit of the address selects between the high and low bytes of the word. Figure 7-4 shows the interface between the internal program memory and the TABLAT.

FIGURE 7-4: READS FROM FLASH PROGRAM MEMORY



EXAMPLE 7-1: READING A FLASH PROGRAM MEMORY WORD

	BCF BSF MOVLW	EECON1, CFGS EECON1, EEPGD CODE ADDR UPPER	; point to Flash program memory ; access Flash program memory ; Load TBLPTR with the base
	MOVE	TBLPTRU	; address of the word
	MOVWF	CODE_ADDR_HIGH	/ address of the word
	MOVWF	TBLPTRH	
	MOVLW	CODE_ADDR_LOW	
	MOVWF	TBLPTRL	
READ_WORD			
	TBLRD*-	+	; read into TABLAT and increment
	MOVF	TABLAT, W	; get data
	MOVWF	WORD_EVEN	
	TBLRD*-	+	; read into TABLAT and increment
	MOVF	TABLAT, W	; get data
	MOVF	WORD_ODD	

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7.4 Erasing Flash Program Memory

The erase blocks are:

- PIC18FX5K22 and PIC18FX6K22 32 words or 64 bytes
- PIC18FX7K22 64 words or 128 bytes

Word erase in the Flash array is not supported.

When initiating an erase sequence from the microcontroller itself, a block of 64 or 128 bytes of program memory is erased. The Most Significant 16 bits of the TBLPTR<21:6> point to the block being erased. The TBLPTR<5:0> bits are ignored.

The EECON1 register commands the erase operation. The EEPGD bit must be set to point to the Flash program memory. The WREN bit must be set to enable write operations. The FREE bit is set to select an erase operation.

For protection, the write initiate sequence for EECON2 must be used.

A long write is necessary for erasing the internal Flash. Instruction execution is halted while in a long write cycle. The long write will be terminated by the internal programming timer.

7.4.1 FLASH PROGRAM MEMORY ERASE SEQUENCE

The sequence of events for erasing a block of internal program memory location is:

- 1. Load the Table Pointer register with the address of the row to be erased.
- 2. Set the EECON1 register for the erase operation:
 - · Set the EEPGD bit to point to program memory
 - Clear the CFGS bit to access program memory
 - · Set the WREN bit to enable writes
 - · Set the FREE bit to enable the erase
- 3. Disable the interrupts.
- 4. Write 0x55 to EECON2.
- 5. Write 0xAA to EECON2.
- Set the WR bit. This begins the row erase cycle. The CPU will stall for the duration of the erase for TIW. (See Parameter D133A.)
- 7. Re-enable interrupts.

	MOVLW MOVWF MOVLW MOVWF MOVLW MOVWF	CODE_ADDR_UPPER TBLPTRU CODE_ADDR_HIGH TBLPTRH CODE_ADDR_LOW TBLPTRL	; load TBLPTR with the base ; address of the memory block
ERASE_ROW			
	BSF	EECON1, EEPGD	; point to Flash program memory
	BCF	EECON1, CFGS	; access Flash program memory
	BSF	EECON1, WREN	; enable write to memory
	BSF	EECON1, FREE	; enable Row Erase operation
	BCF	INTCON, GIE	; disable interrupts
Required	MOVLW	0x55	
Sequence	MOVWF	EECON2	; write 55h
	MOVLW	0xAA	
	MOVWF	EECON2	; write OAAh
	BSF	EECON1, WR	; start erase (CPU stall)
	BSF	INTCON, GIE	; re-enable interrupts

EXAMPLE 7-2: ERASING A FLASH PROGRAM MEMORY ROW

7.5 Writing to Flash Program Memory

The programming blocks are:

- PIC18FX5K22 and PIC18FX6K22 32 words or 64 bytes
- PIC18FX7K22 64 words or 128 bytes

Word or byte programming is not supported.

Table writes are used internally to load the holding registers needed to program the Flash memory. The number of holding registers used for programming by the table writes are:

- PIC18FX5K22 and PIC18FX6K22 64
- PIC18FX7K22 128

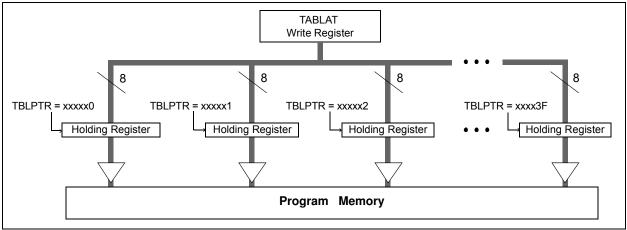
Since the Table Latch (TABLAT) is only a single byte, the TBLWT instruction may need to be executed 64 times for each programming operation. All of the table write operations will essentially be short writes because only the holding registers are written. At the end of updating the 64 or 128 holding registers, the EECON1 register must be written to in order to start the programming operation with a long write.

The long write is necessary for programming the internal Flash. Instruction execution is halted while in a long write cycle. The long write is terminated by the internal programming timer.

The EEPROM on-chip timer controls the write time. The write/erase voltages are generated by an on-chip charge pump, rated to operate over the voltage range of the device.

Note: The default value of the holding registers on device Resets, and after write operations, is FFh. A write of FFh to a holding register does not modify that byte. This means that individual bytes of program memory may be modified, provided that the change does not attempt to change any bit from a '0' to a '1'. When modifying individual bytes, it is not necessary to load all 64 or 128 holding registers before executing a write operation.





7.5.1 FLASH PROGRAM MEMORY WRITE SEQUENCE

The sequence of events for programming an internal program memory location should be:

- 1. Read the 64 or 128 bytes into RAM.
- 2. Update the data values in RAM as necessary.
- 3. Load the Table Pointer register with the address being erased.
- 4. Execute the row erase procedure.
- 5. Load the Table Pointer register with the address of the first byte being written.
- 6. Write the 64 or 128 bytes into the holding registers with auto-increment.
- 7. Set the EECON1 register for the write operation:
 - Set the EEPGD bit to point to program memory
 - · Clear the CFGS bit to access program memory
 - · Set the WREN to enable byte writes
- 8. Disable the interrupts.
- 9. Write 0x55 to EECON2.
- 10. Write 0xAA to EECON2.
- Set the WR bit. This will begin the write cycle. The CPU will stall for duration of the write for Tiw (see Parameter D133A).
- 12. Re-enable the interrupts.
- 13. Verify the memory (table read).

An example of the required code is shown in Example 7-3.

- Note: Before setting the WR bit, the Table Pointer address needs to be within the intended address range of the 64 or 128 bytes in the holding register.
- Note: Self-write execution to Flash and EEPROM memory cannot be done while running in LP Oscillator mode (Low-Power mode). Therefore, executing a self-write will put the device into High-Power mode.

EXAMPLE 7-3:	WRI	TING TO FLASH PROC	RAM MEMORY	
	MOVLW MOVWF	SIZE_OF_BLOCK COUNTER	; number of bytes in erase block	
	MOVLW MOVWF	BUFFER_ADDR_HIGH FSR0H	; point to buffer	
	MOVLW	BUFFER_ADDR_LOW		
	MOVLW	CODE_ADDR_UPPER TBLPTRU	; Load TBLPTR with the base ; address of the memory block	
	MOVLW	CODE_ADDR_HIGH	, address of the memory brock	
	MOVWF MOVLW	TBLPTRH CODE_ADDR_LOW		
DEAD DLOGK	MOVWF	TBLPTRL		
READ_BLOCK	TBLRD*+	-	; read into TABLAT, and inc	
	MOVF	TABLAT, W	; get data	
	MOVWF	POSTINC0	; store data	
	DECFSZ	COUNTER	; done?	
MODIFY_WORD	BRA	READ_BLOCK	; repeat	
NODIFI_WORD	MOVLW	DATA_ADDR_HIGH	; point to buffer	
	MOVWF	FSR0H		
	MOVLW MOVWF	DATA_ADDR_LOW FSR0L		
	MOVWF	NEW_DATA_LOW	; update buffer word	
	MOVWF	POSTINC0	, apadee baller word	
	MOVLW	NEW_DATA_HIGH		
	MOVWF	INDF0		
ERASE_BLOCK				
	MOVLW	CODE_ADDR_UPPER	; load TBLPTR with the base	
	MOVWF	TBLPTRU	; address of the memory block	
	MOVLW	CODE_ADDR_HIGH		
	MOVWF MOVLW	TBLPTRH CODE_ADDR_LOW		
	MOVWF	TBLPTRL		
	BSF	EECON1, EEPGD	; point to Flash program memory	
	BCF	EECON1, CFGS	; access Flash program memory	
	BSF	EECON1, WREN	; enable write to memory	
	BSF	EECON1, FREE	; enable Row Erase operation	
	BCF MOVLW	INTCON, GIE 0x55	; disable interrupts	
Required	MOVWF	EECON2	; write 55h	
Sequence	MOVLW	0xAA	· write Ozzh	
	MOVWF BSF	EECON2 EECON1, WR	; write OAAh ; start erase (CPU stall)	
	BSF	INTCON, GIE	; re-enable interrupts	
	TBLRD*-		; dummy read decrement	
	MOVLW	BUFFER_ADDR_HIGH	; point to buffer	
	MOVWF	FSROH		
	MOVLW	BUFFER_ADDR_LOW		
	MOVWF	FSROL		
WRITE_BUFFER_B		STAR OF BLOOK	; number of bytes in holding register	
	MOVLW MOVWF	SIZE_OF_BLOCK COUNTER	, number of pytes in notating register	
WRITE_BYTE_TO_		CONTER		
	MOVFF	POSTINC0, WREG	; get low byte of buffer data	
	MOVWF	TABLAT	; present data to table latch	
	TBLWT+*	÷	; write data, perform a short write	
			; to internal TBLWT holding register.	
		COUNTER	; loop until buffers are full	
	BRA	WRITE_BYTE_TO_HREGS		

EXAMPLE 7-3: WRITING TO FLASH PROGRAM MEMORY

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EXAMPLE 7-3: WRITING TO FLASH PROGRAM MEMORY (CONTINUED)

PROGRAM_MEMOR	Y			
	BSF	EECON1,	EEPGD	; point to Flash program memory
	BCF	EECON1,	CFGS	; access Flash program memory
	BSF	EECON1,	WREN	; enable write to memory
	BCF	INTCON,	GIE	; disable interrupts
	MOVLW	0x55		
Required	MOVWF	EECON2		; write 55h
Sequence	MOVLW	0xAA		
	MOVWF	EECON2		; write OAAh
	BSF	EECON1,	WR	; start program (CPU stall)
	BSF	INTCON,	GIE	; re-enable interrupts
	BCF	EECON1,	WREN	; disable write to memory

7.5.2 WRITE VERIFY

Depending on the application, good programming practice may dictate that the value written to the memory should be verified against the original value. This should be used in applications where excessive writes can stress bits near the specification limit.

7.5.3 UNEXPECTED TERMINATION OF WRITE OPERATION

If a write is terminated by an unplanned event, such as loss of power or an unexpected Reset, the memory location just programmed should be verified and reprogrammed if needed. If the write operation is interrupted by a MCLR Reset or a WDT Time-out Reset during normal operation, the user can check the WRERR bit and rewrite the location(s) as needed.

7.5.4 PROTECTION AGAINST SPURIOUS WRITES

To protect against spurious writes to Flash program memory, the write initiate sequence must also be followed. See **Section 28.0** "**Special Features of the CPU**" for more details.

7.6 Flash Program Operation During Code Protection

See Section 28.6 "Program Verification and Code Protection" for details on code protection of Flash program memory.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
TBLPTRU	bit 21 ⁽¹⁾ Program Memory Table Pointer Upper Byte (TBLPTR<20:16>)							
TBPLTRH	Program Me	mory Table Po	binter High	Byte (TBLPT	R<15:8>)			
TBLPTRL	Program Me	mory Table Po	pinter Low	Byte (TBLPT	R<7:0>)			
TABLAT	Program Me	mory Table La	atch					
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
EECON2	EEPROM Co	ontrol Registe	r 2 (not a p	hysical regist	er)			
EECON1	EEPGD	CFGS		FREE	WRERR	WREN	WR	RD
IPR6	—	—	_	EEIP	_	CMP3IP	CMP2IP	CMP1IP
PIR6	_	_		EEIF		CMP3IF	CMP2IF	CMP1IF
PIE6	_	_	_	EEIE	_	CMP3IE	CMP2IE	CMP1IE

TABLE 7-2: REGISTERS ASSOCIATED WITH PROGRAM FLASH MEMORY

Legend: — = unimplemented, read as '0'. Shaded cells are not used during Flash/EEPROM access.

Note 1: Bit 21 of the TBLPTRU allows access to the device Configuration bits.

8.0 EXTERNAL MEMORY BUS

Note:	The	External	Memory	Bus	is	not
	imple	mented on	64-pin dev	vices.		

The External Memory Bus (EMB) allows the device to access external memory devices (such as Flash, EPROM or SRAM) as program or data memory. It supports both 8 and 16-Bit Data Width modes and three address widths of up to 20 bits.

The bus is implemented with 28 pins, multiplexed across four I/O ports. Three ports (PORTD, PORTE and PORTH) are multiplexed with the address/data bus for a total of 20 available lines, while PORTJ is multiplexed with the bus control signals.

A list of the pins and their functions is provided in Table 8-1.

TABLE 8-1:	PIC18F87K22 FAMILY EXTERNAL BUS – I/O PORT FUNCTIONS
IADLE 0-1.	FIG 10F0/ K22 FAMILI EXTERINAL DUS - 1/0 FURT FUNCTIONS

Name	Port	Bit	External Memory Bus Function
RD0/AD0	PORTD	0	Address Bit 0 or Data Bit 0
RD1/AD1	PORTD	1	Address Bit 1 or Data Bit 1
RD2/AD2	PORTD	2	Address Bit 2 or Data Bit 2
RD3/AD3	PORTD	3	Address Bit 3 or Data Bit 3
RD4/AD4	PORTD	4	Address Bit 4 or Data Bit 4
RD5/AD5	PORTD	5	Address Bit 5 or Data Bit 5
RD6/AD6	PORTD	6	Address Bit 6 or Data Bit 6
RD7/AD7	PORTD	7	Address Bit 7 or Data Bit 7
RE0/AD8	PORTE	0	Address Bit 8 or Data Bit 8
RE1/AD9	PORTE	1	Address Bit 9 or Data Bit 9
RE2/AD10	PORTE	2	Address Bit 10 or Data Bit 10
RE3/AD11	PORTE	3	Address Bit 11 or Data Bit 11
RE4/AD12	PORTE	4	Address Bit 12 or Data Bit 12
RE5/AD13	PORTE	5	Address Bit 13 or Data Bit 13
RE6/AD14	PORTE	6	Address Bit 14 or Data Bit 14
RE7/AD15	PORTE	7	Address Bit 15 or Data Bit 15
RH0/A16	PORTH	0	Address Bit 16
RH1/A17	PORTH	1	Address Bit 17
RH2/A18	PORTH	2	Address Bit 18
RH3/A19	PORTH	3	Address Bit 19
RJ0/ALE	PORTJ	0	Address Latch Enable (ALE) Control pin
RJ1/OE	PORTJ	1	Output Enable (OE) Control pin
RJ2/WRL	PORTJ	2	Write Low (WRL) Control pin
RJ3/WRH	PORTJ	3	Write High (WRH) Control pin
RJ4/BA0	PORTJ	4	Byte Address Bit 0 (BA0)
RJ5/CE	PORTJ	5	Chip Enable (CE) Control pin
RJ6/LB	PORTJ	6	Lower Byte Enable (LB) Control pin
RJ7/UB	PORTJ	7	Upper Byte Enable (UB) Control pin

Note: For the sake of clarity, only I/O port and external bus assignments are shown here. One or more additional multiplexed features may be available on some pins.

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8.1 External Memory Bus Control

The operation of the interface is controlled by the MEMCON register (Register 8-1). This register is available in all program memory operating modes except Microcontroller mode. In this mode, the register is disabled and cannot be written to.

The EBDIS bit (MEMCON<7>) controls the operation of the bus and related port functions. Clearing EBDIS enables the interface and disables the I/O functions of the ports, as well as any other functions multiplexed to those pins. Setting the bit enables the I/O ports and other functions, but allows the interface to override everything else on the pins when an external memory operation is required. By default, the external bus is always enabled and disables all other I/O. The operation of the EBDIS bit is also influenced by the program memory mode being used. This is discussed in more detail in Section 8.5 "Program Memory Modes and the External Memory Bus".

The WAIT bits allow for the addition of Wait states to external memory operations. The use of these bits is discussed in **Section 8.3** "Wait States".

The WM bits select the particular operating mode used when the bus is operating in 16-Bit Data Width mode. These bits are discussed in more detail in **Section 8.6 "16-Bit Data Width Modes"**. These bits have no effect when an 8-Bit Data Width mode is selected.

REGISTER 8-1: MEMCON: EXTERNAL MEMORY BUS CONTROL REGISTER⁽¹⁾

R/W-0	U-0	R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0			
EBDIS	;	WAIT1	WAIT0	—	—	WM1	WM0			
bit 7							bit 0			
Legend:										
R = Read	able bit	W = Writable b	oit	U = Unimplem	nented bit, read	d as '0'				
-n = Value	e at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown			
bit 7	EBDIS: Ext	ernal Bus Disable	bit							
	1 = Externa	al bus is enabled	when microc	ontroller access	es external me	emory; otherwis	se, all external			
	bus dri	vers are mapped a	as I/O ports			•				
	0 = Externa	al bus is always er	nabled, I/O p	orts are disabled	1					
bit 6	Unimpleme	ented: Read as '0	,							
bit 5-4	WAIT<1:0>	WAIT<1:0>: Table Reads and Writes Bus Cycle Wait Count bits								
	11 = Table	reads and writes w	vill wait 0 Tch	(
	10 = Table	reads and writes w	vill wait 1 Ton	(
		reads and writes w								
		reads and writes v		(
bit 3-2	Unimpleme	ented: Read as '0	,							
bit 1-0	WM<1:0>:	TBLWT Operation	with 16-Bit D	ata Bus Width S	elect bits					
		1x = Word Write mode: TABLAT word output; WRH is active when TABLAT is written								
		Select mode: TABL								
	00 = Byte	Write mode: TABL	AT data is co	opied on both MS	SB and LSB; V	VRH or WRL wi	ill activate			
Note 1:	Unimplemented	d on 64-pin device	s (PIC18F6X	(K22) read as '0	,					

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

8.2 Address and Data Width

The PIC18F87K22 family of devices can be independently configured for different address and data widths on the same memory bus. Both address and data width are set by Configuration bits in the CONFIG3L register. As Configuration bits, this means that these options can only be configured by programming the device and are not controllable in software.

The BW bit selects an 8-bit or 16-bit data bus width. Setting this bit (default) selects a data width of 16 bits.

The ABW<1:0> bits determine both the program memory operating mode and the address bus width. The available options are 20-bit, 16-bit and 12-bit, as well as Microcontroller mode (external bus is disabled). Selecting a 16-bit or 12-bit width makes a corresponding number of high-order lines available for I/O functions. These pins are no longer affected by the setting of the EBDIS bit. For example, selecting a 16-Bit Addressing mode (ABW<1:0> = 01) disables A<19:16> and allows PORTH<3:0> to function without interruptions from the bus. Using the smaller address widths allows users to tailor the memory bus to the size of the external memory space for a particular design while freeing up pins for dedicated I/O operation.

Because the ABW bits have the effect of disabling pins for memory bus operations, it is important to always select an address width at least equal to the data width. If a 12-bit address width is used with a 16-bit data width, the upper four bits of data will not be available on the bus.

All combinations of address and data widths require multiplexing of address and data information on the same lines. The address and data multiplexing, as well as I/O ports made available by the use of smaller address widths, are summarized in Table 8-2.

8.2.1 ADDRESS SHIFTING ON THE EXTERNAL BUS

By default, the address presented on the external bus is the value of the PC. In practical terms, this means that addresses in the external memory device, below the top of on-chip memory, are unavailable to the microcontroller. To access these physical locations, the glue logic between the microcontroller and the external memory must somehow translate addresses.

To simplify the interface, the external bus offers an extension of Extended Microcontroller mode that automatically performs address shifting. This feature is controlled by the EASHFT Configuration bit. Setting this bit offsets addresses on the bus by the size of the microcontroller's on-chip program memory and sets the bottom address at 0000h. This allows the device to use the entire range of physical addresses of the external memory.

8.2.2 21-BIT ADDRESSING

As an extension of 20-bit address width operation, the External Memory Bus can also fully address a 2-Mbyte memory space. This is done by using the Bus Address Bit 0 (BA0) control line as the Least Significant bit of the address. The UB and LB control signals may also be used with certain memory devices to select the upper and lower bytes within a 16-bit wide data word.

This addressing mode is available in both 8-Bit and certain 16-Bit Data Width modes. Additional details are provided in Section 8.6.3 "16-Bit Byte Select Mode" and Section 8.7 "8-Bit Data Width Mode".

TABLE 0-2. AI	2. ADDRESS AND DATA LINES FOR DIFFERENT ADDRESS AND DATA WIDTHS								
Data Width	Address Width	Address Width Multiplexed Data and Address Lines (and Corresponding Ports)		Ports Available for I/O					
	12-bit		AD<11:8> (PORTE<3:0>)	PORTE<7:4>, All of PORTH					
8-bit	16-bit	AD<7:0> (PORTD<7:0>)	AD<15:8> (PORTE<7:0>)	All of PORTH					
	20-bit		A<19:16>, AD<15:8> (PORTH<3:0>, PORTE<7:0>)	-					
	16-bit	AD<15:0>	_	All of PORTH					
16-bit	20-bit	(PORTD<7:0>, PORTE<7:0>)	A<19:16> (PORTH<3:0>)	_					

TABLE 8-2: ADDRESS AND DATA LINES FOR DIFFERENT ADDRESS AND DATA WIDTHS

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8.3 Wait States

While it may be assumed that external memory devices will operate at the microcontroller clock rate, this is often not the case. In fact, many devices require longer times to write or retrieve data than the time allowed by the execution of table read or table write operations.

To compensate for this, the External Memory Bus can be configured to add a fixed delay to each table operation using the bus. Wait states are enabled by setting the WAIT Configuration bit. When enabled, the amount of delay is set by the WAIT<1:0> bits (MEMCON<5:4>). The delay is based on multiples of microcontroller instruction cycle time and is added following the instruction cycle when the table operation is executed. The range is from no delay to 3 Tcy (default value).

8.4 Port Pin Weak Pull-ups

With the exception of the upper address lines, A<19:16>, the pins associated with the External Memory Bus are equipped with weak pull-ups. The pull-ups are controlled by the upper three bits of the PADCFG1 register (PADCFG1<7:5>). They are named RDPU, REPU and RJPU, and control pull-ups on PORTD, PORTE and PORTJ, respectively. Setting one of these bits enables the corresponding pull-ups for that port. All pull-ups are disabled by default on all device Resets.

In Extended Microcontroller mode, the port pull-ups can be useful in preserving the memory state on the external bus while the bus is temporarily disabled (EBDIS = 1).

8.5 Program Memory Modes and the External Memory Bus

The PIC18F87K22 family of devices is capable of operating in one of two program memory modes, using combinations of on-chip and external program memory. The functions of the multiplexed port pins depend on the program memory mode selected, as well as the setting of the EBDIS bit.

In **Microcontroller Mode**, the bus is not active and the pins have their port functions only. Writes to the MEMCOM register are not permitted. The Reset value of EBDIS ('0') is ignored and the ABW pins behave as I/O ports.

In **Extended Microcontroller Mode**, the external program memory bus shares I/O port functions on the pins. When the device is fetching or doing table read/table write operations on the external program memory space, the pins will have the external bus function.

If the device is fetching and accessing internal program memory locations only, the EBDIS control bit will change the pins from external memory to I/O port

functions. When EBDIS = 0, the pins function as the external bus. When EBDIS = 1, the pins function as I/O ports.

If the device fetches or accesses external memory while EBDIS = 1, the pins will switch to the external bus. If the EBDIS bit is set by a program executing from external memory, the action of setting the bit will be delayed until the program branches into the internal memory. At that time, the pins will change from external bus to I/O ports.

If the device is executing out of internal memory when EBDIS = 0, the memory bus address/data and control pins will not be active. They will go to a state where the active address/data pins are tri-state, the \overline{CE} , \overline{OE} , \overline{WRH} , \overline{WRL} , \overline{UB} and \overline{LB} signals are '1', and ALE and BA0 are '0'. Note that only those pins associated with the current address width are forced to tri-state; the other pins continue to function as I/O. In the case of 16-bit address width, for example, only AD<15:0> (PORTD and PORTE) are affected; A<19:16> (PORTH<3:0>) continue to function as I/O.

In all external memory modes, the bus takes priority over any other peripherals that may share pins with it. This includes the Parallel Master Port (PMP) and serial communication modules which would otherwise take priority over the I/O port.

8.6 16-Bit Data Width Modes

In 16-Bit Data Width mode, the external memory interface can be connected to external memories in three different configurations:

- 16-Bit Byte Write
- 16-Bit Word Write
- 16-Bit Byte Select

The configuration to be used is determined by the WM<1:0> bits in the MEMCON register (MEMCON<1:0>). These three different configurations allow the designer maximum flexibility in using both 8-bit and 16-bit devices with 16-bit data.

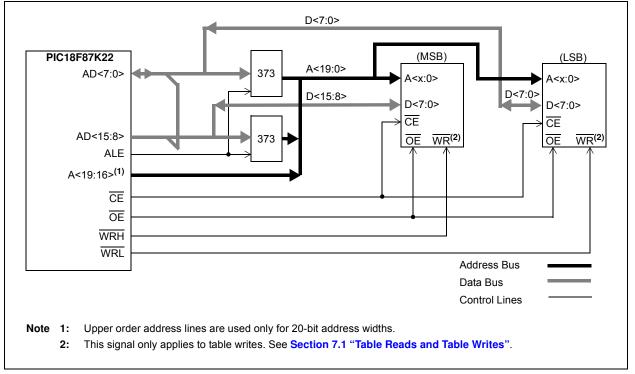
For all 16-bit modes, the Address Latch Enable (ALE) pin indicates that the Address bits, AD<15:0>, are available on the external memory interface bus. Following the address latch, the Output Enable (OE) signal will enable both bytes of program memory at once to form a 16-bit instruction word. The Chip Enable (CE signal) is active at any time that the microcontroller accesses external memory, whether reading or writing; it is inactive (asserted high) whenever the device is in Sleep mode.

In Byte Select mode, JEDEC standard Flash memories will require BA0 for the byte address line and one I/O line to select between Byte and Word mode. The other 16-bit modes do not need BA0. JEDEC standard static RAM memories will use the UB or LB signals for byte selection.

8.6.1 16-BIT BYTE WRITE MODE

Figure 8-1 shows an example of 16-Bit Byte Write mode for PIC18F87K22 family devices. This mode is used for two separate 8-bit memories connected for 16-bit operation. This generally includes basic EPROM and Flash devices. It allows table writes to byte-wide external memories. During a TBLWT instruction cycle, the TABLAT data is presented on the upper and <u>lower bytes</u> of the AD<15:0> bus. The appropriate WRH or WRL control line is strobed on the LSb of the TBLPTR.





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8.6.2 16-BIT WORD WRITE MODE

Figure 8-2 shows an example of 16-Bit Word Write mode for PIC18F87K22 family devices. This mode is used for word-wide memories, which includes some of the EPROM and Flash type memories. This mode allows opcode fetches and table reads from all forms of 16-bit memory, and table writes to any type of word-wide external memories. This method makes a distinction between TBLWT cycles to even or odd addresses.

During a TBLWT cycle to an even address (TBLPTR<0>= 0), the TABLAT data is transferred to a holding latch and the external address data bus is tri-stated for the data portion of the bus cycle. No write signals are activated.

During a TBLWT cycle to an odd address (TBLPTR<0> = 1), the TABLAT data is presented on the upper byte of the AD<15:0> bus. The contents of the holding latch are presented on the lower byte of the AD<15:0> bus.

<u>The WRH</u> signal is strobed for each write cycle; the WRL pin is unused. The signal on the BA0 pin indicates the LSb of the TBLPTR, but it is left unconnected. Instead, the UB and LB signals are active to select both bytes. The obvious limitation to this method is that the table write must be done in pairs on a specific word boundary to correctly write a word location.

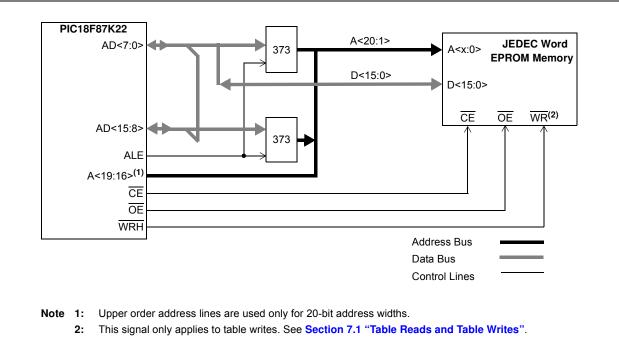


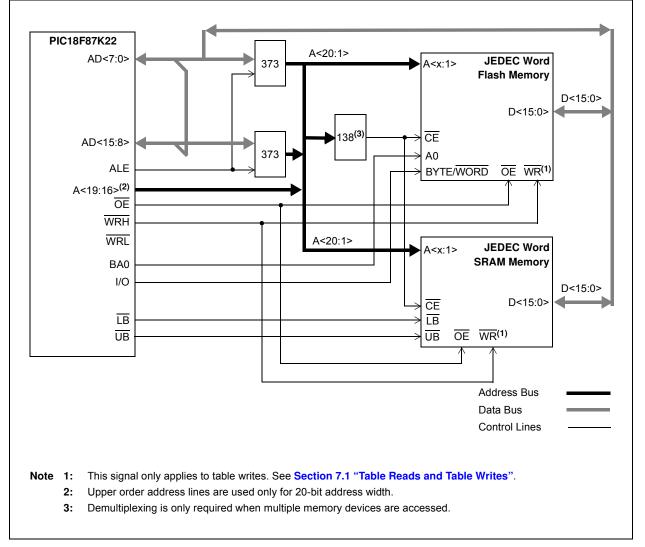
FIGURE 8-2: 16-BIT WORD WRITE MODE EXAMPLE

8.6.3 16-BIT BYTE SELECT MODE

Figure 8-3 shows an example of 16-Bit Byte Select mode. This mode allows table write operations to word-wide external memories with byte selection capability. This generally includes both word-wide Flash and SRAM devices.

During a TBLWT cycle, the TABLAT data is presented on the upper and lower byte of the AD<15:0> bus. The WRH signal is strobed for each write cycle; the WRL pin is not used. The BA0 or UB/LB signals are used to select the byte to be written, based on the Least Significant bit of the TBLPTR register. Flash and SRAM devices use different control signal combinations to implement Byte Select mode. JEDEC standard Flash memories require that a controller I/O port pin be connected to the memory's BYTE/WORD pin to provide the select signal. They also use the BA0 signal from the controller as a byte address. JEDEC standard static RAM memories, on the other hand, use the UB or LB signals to select the byte.





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8.6.4 16-BIT MODE TIMING

The presentation of control signals on the External Memory Bus is different for the various operating modes. Typical signal timing diagrams are shown in Figure 8-4 and Figure 8-5.

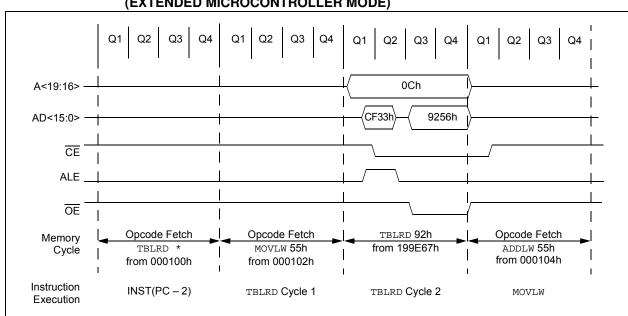
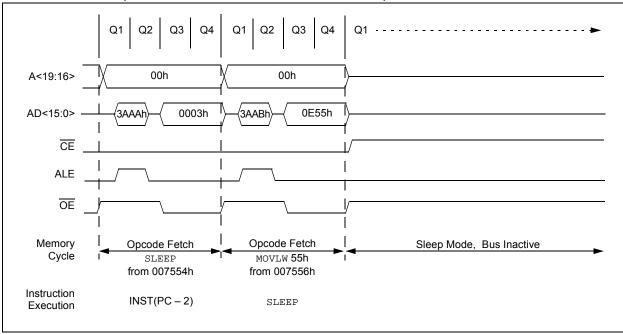


FIGURE 8-4: EXTERNAL MEMORY BUS TIMING FOR TBLRD (EXTENDED MICROCONTROLLER MODE)

FIGURE 8-5: EXTERNAL MEMORY BUS TIMING FOR SLEEP (EXTENDED MICROCONTROLLER MODE)



8.7 8-Bit Data Width Mode

In 8-Bit Data Width mode, the External Memory Bus operates only in Multiplexed mode; that is, data shares the 8 Least Significant bits of the address bus.

Figure 8-6 shows an example of 8-Bit Multiplexed mode for PIC18F8XK22 devices. This mode is used for a single, 8-bit memory connected for 16-bit operation. The instructions will be fetched as two 8-bit bytes on a shared data/address bus. The two bytes are sequentially fetched within one instruction cycle (TcY). Therefore, the designer must choose external memory devices according to timing calculations based on 1/2 TcY (2 times the instruction rate). For proper memory speed selection, glue logic propagation delay times must be considered, along with setup and hold times.

The Address Latch Enable (ALE) pin indicates that the Address bits, AD<15:0>, are available on the External Memory Bus interface. The Output Enable (OE) signal

will enable one byte of program memory for a portion of the instruction cycle, then BA0 will change and the second byte will be enabled to form the 16-bit instruction word. The Least Significant bit of the address, BA0, must be connected to the memory devices in this mode. The Chip Enable (CE) signal is active at any time that the microcontroller accesses external memory, whether reading or writing. It is inactive (asserted high) whenever the device is in Sleep mode.

This generally includes basic EPROM and Flash devices. It allows table writes to byte-wide external memories.

During a TBLWT instruction cycle, the TABLAT data is presented on the upper and lower bytes of the AD<15:0> bus. The appropriate level of the BA0 control line is strobed on the LSb of the TBLPTR.

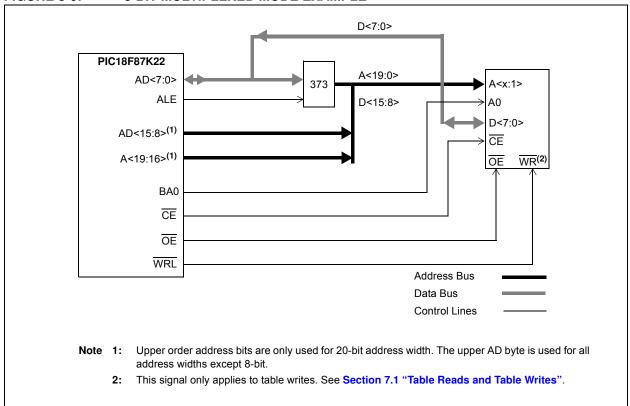


FIGURE 8-6: 8-BIT MULTIPLEXED MODE EXAMPLE

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8.7.1 8-BIT MODE TIMING

The presentation of control signals on the External Memory Bus is different for the various operating modes. Typical signal timing diagrams are shown in Figure 8-7 and Figure 8-8.

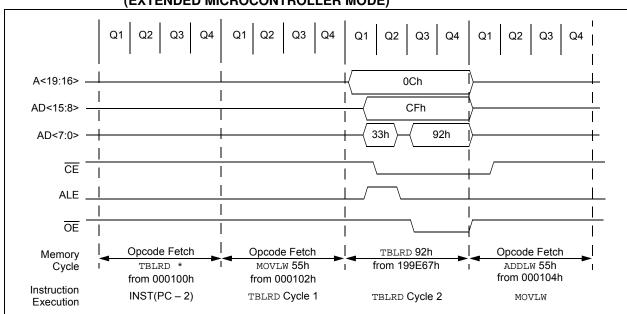
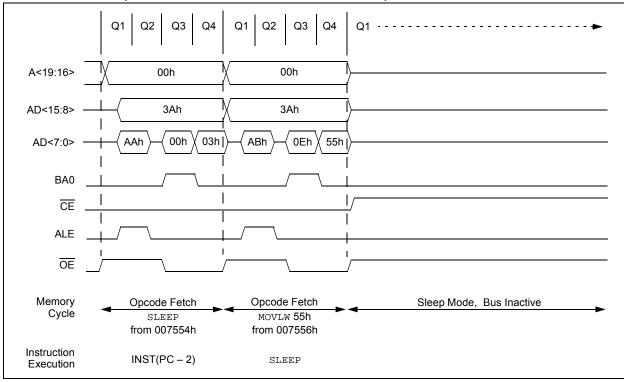


FIGURE 8-7: EXTERNAL MEMORY BUS TIMING FOR TBLRD (EXTENDED MICROCONTROLLER MODE)

FIGURE 8-8: EXTERNAL MEMORY BUS TIMING FOR SLEEP (EXTENDED MICROCONTROLLER MODE)



8.8 Operation in Power-Managed Modes

In alternate, power-managed Run modes, the external bus continues to operate normally. If a clock source with a lower speed is selected, bus operations will run at that speed. In these cases, excessive access times for the external memory may result if Wait states have been enabled and added to external memory operations. If operations in a lower power Run mode are anticipated, users should provide in their applications for adjusting memory access times at the lower clock speeds. In Sleep and Idle modes, the microcontroller core does not need to access data; bus operations are suspended. The state of the external bus is frozen, with the address/data pins, and most of the control pins, holding at the same state they were in when the mode was invoked. The only potential changes are to the \overline{CE} , \overline{LB} and \overline{UB} pins, which are held at logic high.

TABLE 8-3 :	REGISTERS ASSOCIATED WITH THE EXTERNAL MEMORY BUS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
MEMCON ⁽¹⁾	EBDIS	_	WAIT1	WAIT0		—	WM1	WM0
PADCFG1	RDPU	REPU	RJPU ⁽¹⁾	_	—	RTSECSEL1	RTSECSEL0	
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD

Legend: — = unimplemented, read as '0'. Shaded cells are not used during External Memory Bus access.

Note 1: Unimplemented in 64-pin devices (PIC18F6XK22), read as '0'.

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NOTES:

9.0 DATA EEPROM MEMORY

The data EEPROM is a nonvolatile memory array, separate from the data RAM and program memory, that is used for long-term storage of program data. It is not directly mapped in either the register file or program memory space, but is indirectly addressed through the Special Function Registers (SFRs). The EEPROM is readable and writable during normal operation over the entire VDD range.

Five SFRs are used to read and write to the data EEPROM, as well as the program memory. They are:

- EECON1
- EECON2
- EEDATA
- EEADR
- EEADRH

The data EEPROM allows byte read and write. When interfacing to the data memory block, EEDATA holds the 8-bit data for read/write and the EEADRH:EEADR register pair holds the address of the EEPROM location being accessed.

The EEPROM data memory is rated for high erase/write cycle endurance. A byte write automatically erases the location and writes the new data (erase-before-write). The write time is controlled by an on-chip timer; it will vary with voltage and temperature, as well as from chip-to-chip. Please refer to Parameter D122 (Table 31-1 in Section 31.0 "Electrical Characteristics") for exact limits.

9.1 EEADR and EEADRH Registers

The EEADRH:EEADR register pair is used to address the data EEPROM for read and write operations. EEADRH holds the two MSbs of the address; the upper 6 bits are ignored. The 10-bit range of the pair can address a memory range of 1024 bytes (00h to 3FFh).

9.2 EECON1 and EECON2 Registers

Access to the data EEPROM is controlled by two registers: EECON1 and EECON2. These are the same registers which control access to the program memory and are used in a similar manner for the data EEPROM.

The EECON1 register (Register 9-1) is the control register for data and program memory access. Control bit, EEPGD, determines if the access will be to program memory or data EEPROM memory. When clear, operations will access the data EEPROM memory. When set, program memory is accessed.

Control bit, CFGS, determines if the access will be to the Configuration registers or to program memory/data EEPROM memory. When set, subsequent operations access Configuration registers. When CFGS is clear, the EEPGD bit selects either program Flash or data EEPROM memory.

The WREN bit, when set, will allow a write operation. On power-up, the WREN bit is clear. The WRERR bit is set in hardware when the WREN bit is set, and cleared, when the internal programming timer expires and the write operation is complete.

Note:	During normal operation, the WRERR is			
	read as '1'. This can indicate that a write			
	operation was prematurely terminated by			
	a Reset or a write operation was			
	attempted improperly.			

The WR control bit initiates write operations. The bit cannot be cleared, only set, in software. It is cleared in hardware at the completion of the write operation.

Note:	The EEIF interrupt flag bit (PIR6<4>) is
	set when the write is complete; it must be
	cleared in software.

Control bits, RD and WR, start read and erase/write operations, respectively. These bits are set by firmware and cleared by hardware at the completion of the operation.

The RD bit cannot be set when accessing program memory (EEPGD = 1). Program memory is read using table read instructions. See **Section 7.1 "Table Reads and Table Writes"** regarding table reads.

The EECON2 register is not a physical register. It is used exclusively in the memory write and erase sequences. Reading EECON2 will read all '0's.

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R/W-x	R/W-x	U-0	R/W-0	R/W-x	R/W-0	R/S-0	R/S-0		
EEPGD	CFGS	—	FREE	WRERR ⁽¹⁾	WREN	WR	RD		
bit 7							bit (
Legend:		S = Settable	hit						
R = Readable	∍ hit	W = Writable		II = Unimplen	nented bit, rea	d as '0'			
-n = Value at		'1' = Bit is set		'0' = Bit is clea		x = Bit is unkr	nwn		
iii valao at			•		arou				
bit 7	EEPGD: FI	ash Program or I	Data EEPRO	M Memory Sele	ct bit				
		Flash program r		2					
	0 = Access	data EEPROM	memory						
bit 6		sh Program/Data		Configuration S	Select bit				
		Configuration re	•	OM momony					
bit 5		Flash program o ented: Read as '		Ow memory					
bit 4	•								
DIL 4	FREE: Flash Row Erase Enable bit 1 = Erase the program memory row addressed by TBLPTR on the next WR command (cleared by								
		etion of erase op							
	0 = Perforr	m write-only	,						
bit 3	WRERR: F	lash Program/Da	ata EEPROM	Error Flag bit ⁽¹⁾					
		operation is pre			set during self-	-timed programr	ning in norm		
		ion or an imprope ite operation has		pt)					
bit 2		•	•	/rite Enable bit					
	 WREN: Flash Program/Data EEPROM Write Enable bit 1 = Allows write cycles to Flash program/data EEPROM 								
		write cycles to F							
bit 1	WR: Write	Control bit							
		s a data EEPROI							
		peration is self-ti				e the write is co	mplete.		
	The WR bit can only be set (not cleared) in software.) 0 = Write cycle to the EEPROM is complete								
bit 0		BD: Read Control bit							
	1 = Initiate	s an EEPROM re	ead						
	-	takes one cycle.				•	not cleared) i		
		re. The RD bit ca not initiate an EE		/hen EEPGD =	1 or $CFGS = 1$	L.)			
	0 - 00631								
Note 1 · W	han a WRERE	R occurs the FEI	PGD and CE(3S hits are not o	leared This a	llows tracing of	the error		

REGISTER 9-1: EECON1: DATA EEPROM CONTROL REGISTER 1

Note 1: When a WRERR occurs, the EEPGD and CFGS bits are not cleared. This allows tracing of the error condition.

9.3 Reading the Data EEPROM Memory

To read a data memory location, the user must write the address to the EEADRH:EEADR register pair, clear the EEPGD control bit (EECON1<7>) and then set control bit, RD (EECON1<0>). The data is available in the EEDATA register after one cycle; therefore, it can be read after one NOP instruction. EEDATA will hold this value until another read operation or until it is written to by the user (during a write operation).

The basic process is shown in Example 9-1.

9.4 Writing to the Data EEPROM Memory

To write an EEPROM data location, the address must first be written to the EEADRH:EEADR register pair and the data written to the EEDATA register. The sequence in Example 9-2 must be followed to initiate the write cycle.

The write will not begin if this sequence is not exactly followed (write 0x55 to EECON2, write 0xAA to EECON2, then set WR bit) for each byte. It is strongly recommended that interrupts be disabled during this code segment.

Additionally, the WREN bit in EECON1 must be set to enable writes. This mechanism prevents accidental writes to data EEPROM due to unexpected code execution (i.e., runaway programs). The WREN bit should be kept clear at all times, except when updating the EEPROM. The WREN bit is not cleared by hardware. After a write sequence has been initiated, EECON1, EEADRH:EEADR and EEDATA cannot be modified. The WR bit will be inhibited from being set unless the WREN bit is set. The WREN bit must be set on a previous instruction. Both WR and WREN cannot be set with the same instruction.

At the completion of the write cycle, the WR bit is cleared in hardware and the EEPROM Interrupt Flag bit (EEIF) is set. The user may either enable this interrupt, or poll this bit. EEIF must be cleared by software.

9.5 Write Verify

Depending on the application, good programming practice may dictate that the value written to the memory should be verified against the original value. This should be used in applications where excessive writes can stress bits near the specification limit.

Note: Self-write execution to Flash and EEPROM memory cannot be done while running in LP Oscillator mode (Low-Power mode). Therefore, executing a self-write will put the device into High-Power mode.

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EXAMPLE 9-1: DATA EEPROM READ

MOVLW	DATA_EE_ADDRH	;
MOVWF	EEADRH	; Upper bits of Data Memory Address to read
MOVLW	DATA_EE_ADDR	;
MOVWF	EEADR	; Lower bits of Data Memory Address to read
BCF	EECON1, EEPGD	; Point to DATA memory
BCF	EECON1, CFGS	; Access EEPROM
BSF	EECON1, RD	; EEPROM Read
NOP		
MOVF	EEDATA, W	; $W = EEDATA$

EXAMPLE 9-2: DATA EEPROM WRITE

	MOVLW	DATA_EE_ADDRH	;
	MOVWF	EEADRH	; Upper bits of Data Memory Address to write
	MOVLW	DATA_EE_ADDR	;
	MOVWF	EEADR	; Lower bits of Data Memory Address to write
	MOVLW	DATA_EE_DATA	i
	MOVWF	EEDATA	; Data Memory Value to write
	BCF	EECON1, EEPGD	; Point to DATA memory
	BCF	EECON1, CFGS	; Access EEPROM
	BSF	EECON1, WREN	; Enable writes
	BCF	INTCON, GIE	; Disable Interrupts
	MOVLW	0x55	i
Required	MOVWF	EECON2	; Write 55h
Sequence	MOVLW	0xAA	i
	MOVWF	EECON2	; Write OAAh
	BSF	EECON1, WR	; Set WR bit to begin write
	BTFSC	EECON1, WR	; Wait for write to complete GOTO \$-2
	BSF	INTCON, GIE	; Enable Interrupts
			; User code execution
	BCF	EECON1, WREN	; Disable writes on write complete (EEIF set)

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9.6 Operation During Code-Protect

Data EEPROM memory has its own code-protect bits in the Configuration Words. External read and write operations are disabled if code protection is enabled.

The microcontroller itself can both read and write to the internal data EEPROM, regardless of the state of the code-protect Configuration bit. Refer to Section 28.0 "Special Features of the CPU" for additional information.

9.7 Protection Against Spurious Write

There are conditions when the device may not want to write to the data EEPROM memory. To protect against spurious EEPROM writes, various mechanisms have been implemented. On power-up, the WREN bit is cleared. In addition, writes to the EEPROM are blocked during the Power-up Timer period (TPWRT, Parameter 33 in Table 31-13).

The write initiate sequence, and the WREN bit together, help prevent an accidental write during brown-out, power glitch or software malfunction.

9.8 Using the Data EEPROM

The data EEPROM is a high-endurance, byte addressable array that has been optimized for the storage of frequently changing information (e.g., program variables or other data that is updated often). Frequently changing values will typically be updated more often than Specification D124. If this is the case, an array refresh must be performed. For this reason, variables that change infrequently (such as constants, IDs, calibration, etc.) should be stored in Flash program memory.

A simple data EEPROM refresh routine is shown in Example 9-3.

Note: If data EEPROM is only used to store constants and/or data that changes often, an array refresh is likely not required. See Specification D124.

	CLRF	EEADR EEADRH	; ;	Start at address 0
	BCF	EECON1, CFGS		Set for memory
	BCF			Set for Data EEPROM
	BCF	INTCON, GIE		Disable interrupts
				-
	BSF	EECON1, WREN		Enable writes
LOOP				Loop to refresh array
	BSF	EECON1, RD	;	Read current address
	MOVLW	0x55	;	
	MOVWF	EECON2	;	Write 55h
	MOVLW	0xAA	;	
	MOVWF	EECON2	;	Write OAAh
	BSF	EECON1, WR	;	Set WR bit to begin write
	BTFSC	EECON1, WR	;	Wait for write to complete
	BRA	\$-2		
	INCFSZ	EEADR, F	;	Increment address
	BRA	LOOP	;	Not zero, do it again
	INCFSZ	EEADRH, F	;	Increment the high address
	BRA	LOOP	;	Not zero, do it again
	BCF	EECON1, WREN	;	Disable writes
	BSF	INTCON, GIE		Enable interrupts
	DOF	INICON, GIE	'	Mable Incellupts

TABLE 9-1: REGISTERS ASSOCIATED WITH DATA EEPROM MEMORY

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
EEADRH	EEPROM A	EEPROM Address Register High Byte						
EEADR	EEPROM Address Register Low Byte							
EEDATA	EEPROM Data Register							
EECON2	EEPROM Control Register 2 (not a physical register)							
EECON1	EEPGD	CFGS	_	FREE	WRERR	WREN	WR	RD
IPR6	—	_	_	EEIP	_	CMP3IP	CMP2IP	CMP1IP
PIR6	—		_	EEIF	_	CMP3IF	CMP2IF	CMP1IF
PIE6	—	_	_	EEIE	_	CMP3IE	CMP2IE	CMP1IE

Legend: — = unimplemented, read as '0'. Shaded cells are not used during Flash/EEPROM access.

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10.0 8 x 8 HARDWARE MULTIPLIER

10.1 Introduction

All PIC18 devices include an 8 x 8 hardware multiplier as part of the ALU. The multiplier performs an unsigned operation and yields a 16-bit result that is stored in the product register pair, PRODH:PRODL. The multiplier's operation does not affect any flags in the STATUS register.

Making multiplication a hardware operation allows it to be completed in a single instruction cycle. This has the advantages of higher computational throughput and reduced code size for multiplication algorithms and allows PIC18 devices to be used in many applications previously reserved for digital-signal processors. A comparison of various hardware and software multiply operations, along with the savings in memory and execution time, is shown in Table 10-1.

10.2 Operation

Example 10-1 shows the instruction sequence for an 8 x 8 unsigned multiplication. Only one instruction is required when one of the arguments is already loaded in the WREG register.

Example 10-2 shows the sequence to do an 8 x 8 signed multiplication. To account for the sign bits of the arguments, each argument's Most Significant bit (MSb) is tested and the appropriate subtractions are done.

EXAMPLE 10-1: 8 x 8 UNSIGNED MULTIPLY ROUTINE

MOVF	ARG1,	W	;
MULWF	ARG2		; ARG1 * ARG2 ->
			; PRODH:PRODL

EXAMPLE 10-2: 8 x 8 SIGNED MULTIPLY ROUTINE

MOVF	ARG1, W		
MULWF	ARG2	; ARG1 * ARG2 -	>
		; PRODH:PRODL	
BTFSC	ARG2, SB	; Test Sign Bit	
SUBWF	PRODH, F	; PRODH = PRODH	
		; – ARG	1
MOVF	ARG2, W		
BTFSC	ARG1, SB	; Test Sign Bit	
SUBWF	PRODH, F	; PRODH = PRODH	
		; – ARG	2

TABLE 10-1: PERFORMANCE COMPARISON FOR VARIOUS MULTIPLY OPERATIONS								
		Program	Cycles	Time				
Routine	Multiply Method	Memory (Words)	(Max)	@ 64 MHz	@ 48 MHz	@ 10 MHz	@ 4 MHz	
8 x 8 Linsignod	Without Hardware Multiply	13	69	4.3 μ s	5.7 μs	27.6 μs	69 μs	
8 x 8 Unsigned	Hardware Multiply	1	1	62.5 ns	83.3 ns	400 ns	1 μs	
8 x 8 Signed	Without Hardware Multiply	33	91	5.6 μs	7.5 μs	36.4 μs	91 μs	
	Hardware Multiply	6	6	375 ns	500 ns	2.4 μs	6 μs	
16 x 16	Without Hardware Multiply	21	242	15.1 μs	20.1 μs	96.8 μs	242 μs	
Unsigned	Hardware Multiply	28	28	1.7 μs	2.3 μs	11.2 μs	28 μs	
10 × 10 Cinned	Without Hardware Multiply	52	254	15.8 μs	21.2 μs	101.6 μs	254 μs	
16 x 16 Signed	Hardware Multiply	35	40	2.5 μs	3.3 μs	16.0 μs	4 0 μs	

TABLE 10-1: PERFORMANCE COMPARISON FOR VARIOUS MULTIPLY OPERATIONS

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Example 10-3 shows the sequence to do a 16 x 16 unsigned multiplication. Equation 10-1 shows the algorithm that is used. The 32-bit result is stored in four registers (RES3:RES0).

EQUATION 10-1: 16 x 16 UNSIGNED MULTIPLICATION ALGORITHM

RES3:RES0	=	ARG1H:ARG1L • ARG2H:ARG2L
	=	$(ARG1H \bullet ARG2H \bullet 2^{16}) +$
		$(ARG1H \bullet ARG2L \bullet 2^8) +$
		$(ARG1L \bullet ARG2H \bullet 2^8) +$
		(ARG1L • ARG2L)

EXAMPLE 10-3: 16 x 16 UNSIGNED MULTIPLY ROUTINE

	MOVF	ARG1L, W	
	MULWF	ARG2L	; ARG1L * ARG2L->
			; PRODH:PRODL
	MOVFF	PRODH, RES1	;
	MOVFF	PRODL, RESO	;
;			
	MOVF	ARG1H, W	
	MULWF	ARG2H	; ARG1H * ARG2H->
			; PRODH:PRODL
	MOVFF	PRODH, RES3	;
	MOVFF	PRODL, RES2	;
;			
	MOVF	ARG1L, W	
	MULWF	ARG2H	; ARG1L * ARG2H->
			; PRODH:PRODL
	MOVF	PRODL, W	;
	ADDWF	RES1, F	; Add cross
		PRODH, W	; products
	ADDWFC	RES2, F	;
	CLRF	WREG	;
	ADDWFC	RES3, F	;
;			
		ARG1H, W	;
	MULWF	ARG2L	; ARG1H * ARG2L->
			; PRODH:PRODL
	MOVF		;
	ADDWF	RES1, F	; Add cross
			; products
	ADDWFC	RES2, F	;
	CLRF		;
	ADDWFC	RES3, F	;

Example 10-4 shows the sequence to do a 16 x 16 signed multiply. Equation 10-2 shows the algorithm used. The 32-bit result is stored in four registers (RES3:RES0). To account for the sign bits of the arguments, the MSb for each argument pair is tested and the appropriate subtractions are done.

EQUATION 10-2: 16 x 16 SIGNED MULTIPLICATION ALGORITHM

RES3:RES0=	ARG1H:ARG1L • ARG2H:ARG2L
=	$(ARG1H \bullet ARG2H \bullet 2^{16}) +$
	$(ARG1H \bullet ARG2L \bullet 2^8) +$
	$(ARG1L \bullet ARG2H \bullet 2^8) +$
	$(ARG1L \bullet ARG2L) +$
	$(-1 \bullet ARG2H < 7 > \bullet ARG1H: ARG1L \bullet 2^{16}) +$
	$(-1 \bullet ARG1H < 7 > \bullet ARG2H: ARG2L \bullet 2^{16})$

EXAMPLE 10-4: 16 x 16 SIGNED MULTIPLY ROUTINE

	MOVF	ARG1L, W		
	MULWF		;	ARG1L * ARG2L ->
	поши	III(021		PRODH:PRODL
	MOVEE			FRODITERODE
		PRODH, RES1		
	MOVFF	PRODL, RESO	,	
;				
	MOVF	ARG1H, W		
	MULWF	ARG2H	;	ARG1H * ARG2H ->
			;	PRODH:PRODL
	MOVFF	PRODH, RES3	;	
	MOVFF	PRODL, RES2	;	
;				
	MOVE	ARG1L, W		
	MULWF	ARG2H	:	ARG1L * ARG2H ->
	HOLMI	AROZII		PRODH:PRODL
	MOVE	DDODI W		PRODH:PRODL
	MOVF	PRODL, W	;	
		RES1, F		Add cross
	MOVF		;	products
	ADDWFC	RES2, F	;	
	CLRF	WREG	;	
	ADDWFC	RES3, F	;	
;				
	MOVF	ARG1H, W	;	
	MULWF	ARG2L	;	ARG1H * ARG2L ->
				PRODH:PRODL
	MOVF	PRODL, W	;	
	ADDWF	RES1, F		Add cross
		PRODH, W		
	MOVF			products
	ADDWFC		;	
	CLRF	WREG	;	
	ADDWFC	RES3, F	;	
;				
	BTFSS	ARG2H, 7	;	ARG2H:ARG2L neg?
	BRA	SIGN_ARG1	;	no, check ARG1
	MOVF	ARG1L, W	;	
	SUBWF	RES2	;	
	MOVF	ARG1H, W	;	
	SUBWFB	RES3	;	SIGN_ARG1
	BTFSS			ARG1H:ARG1L neg?
	BRA	CONT_CODE		
		ARG2L, W	;	no, done
	MOVF		΄.	
	SUBWF	RES2	'	
	MOVF	ARG2H, W	;	
	SUBWFB	RES3		
;				
CC	NT_CODE			
	:			

11.0 INTERRUPTS

Members of the PIC18F87K22 family of devices have multiple interrupt sources and an interrupt priority feature that allows most interrupt sources to be assigned a high-priority level or a low-priority level. The high-priority interrupt vector is at 0008h and the low-priority interrupt vector is at 0018h. High-priority interrupt events will interrupt any low-priority interrupts that may be in progress.

The registers for controlling interrupt operation are:

- RCON
- INTCON
- INTCON2
- INTCON3
- PIR1, PIR2, PIR3
- PIE1, PIE2, PIE3
- IPR1, IPR2, IPR3

It is recommended that the Microchip header files supplied with MPLAB[®] IDE be used for the symbolic bit names in these registers. This allows the assembler/compiler to automatically take care of the placement of these bits within the specified register.

In general, interrupt sources have three bits to control their operation. They are:

- Flag bit Indicating that an interrupt event occurred
- Enable bit Enabling program execution to branch to the interrupt vector address when the flag bit is set
- Priority bit Specifying high priority or low priority

The interrupt priority feature is enabled by setting the IPEN bit (RCON<7>). When interrupt priority is enabled, there are two bits that enable interrupts globally. Setting the GIEH bit (INTCON<7>) enables all interrupts that have the priority bit set (high priority). Setting the GIEL bit (INTCON<6>) and GIEH bit (INTCON<7>) enables all interrupts that have the priority bit cleared (low priority). When the interrupt flag, enable bit and appropriate Global Interrupt Enable bit are set, the interrupt will vector immediately to address, 0008h or 0018h, depending on the priority bit setting. Individual interrupts can be disabled through their corresponding enable bits.

When the IPEN bit is cleared (default state), the interrupt priority feature is disabled and interrupts are compatible with PIC[®] mid-range devices. In Compatibility mode, the interrupt priority bits for each source have no effect. INTCON<6> is the PEIE bit that enables/disables all peripheral interrupt sources. INTCON<7> is the GIE bit that enables/disables all interrupt sources. All interrupts branch to address, 0008h, in Compatibility mode.

When an interrupt is responded to, the Global Interrupt Enable bit is cleared to disable further interrupts. If the IPEN bit is cleared, this is the GIE bit. If interrupt priority levels are used, this will be either the GIEH or GIEL bit. High-priority interrupt sources can interrupt a low-priority interrupt. Low-priority interrupts are not processed while high-priority interrupts are in progress.

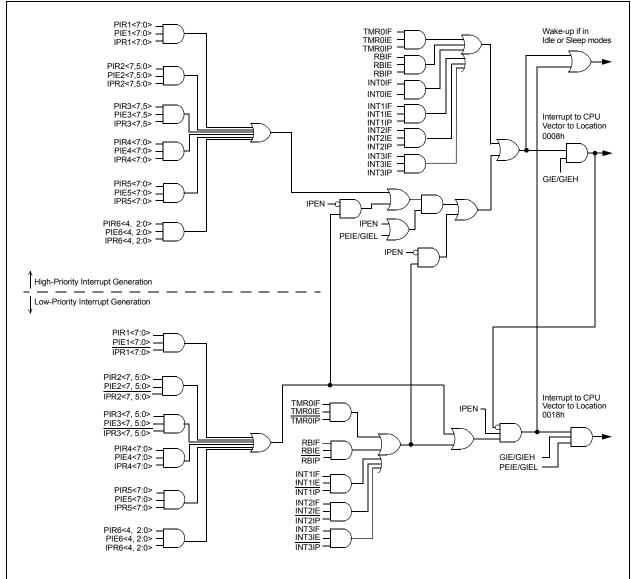
The return address is pushed onto the stack and the PC is loaded with the interrupt vector address (0008h or 0018h). Once in the Interrupt Service Routine (ISR), the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bits must be cleared in software, before re-enabling interrupts, to avoid recursive interrupts.

The "return from interrupt" instruction, RETFIE, exits the interrupt routine and sets the GIE bit (GIEH or GIEL if priority levels are used) that re-enables interrupts.

For external interrupt events, such as the INTx pins or the PORTB input change interrupt, the interrupt latency will be three to four instruction cycles. The exact latency is the same for one or two-cycle instructions. Individual interrupt flag bits are set regardless of the status of their corresponding enable bit or the GIE bit.

Note: Do not use the MOVFF instruction to modify any of the Interrupt Control registers while **any** interrupt is enabled. Doing so may cause erratic microcontroller behavior.

FIGURE 11-1: PIC18F87K22 FAMILY INTERRUPT LOGIC



11.1 INTCON Registers

The INTCON registers are readable and writable registers that contain various enable, priority and flag bits.

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

REGISTER 11-1: INTCON: INTERRUPT CONTROL REGISTER

GIE/GIEH PEIE/GIEL TMR0IE INT0IE RBIE TMR0IF INT0IF RBIF ⁽¹⁾ bit 7 bit 0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x
bit 7 bit 0	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF ⁽¹⁾
	bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	GIE/GIEH: Global Interrupt Enable bit <u>When IPEN = 0:</u> 1 = Enables all unmasked interrupts 0 = Disables all interrupts <u>When IPEN = 1:</u> 1 = Enables all high-priority interrupts 0 = Disables all interrupts
bit 6	PEIE/GIEL: Peripheral Interrupt Enable bit
	<u>When IPEN = 0:</u> 1 = Enables all unmasked peripheral interrupts 0 = Disables all peripheral interrupts
	<u>When IPEN = 1:</u> 1 = Enables all low-priority peripheral interrupts 0 = Disables all low-priority peripheral interrupts
bit 5	TMR0IE: TMR0 Overflow Interrupt Enable bit
	1 = Enables the TMR0 overflow interrupt0 = Disables the TMR0 overflow interrupt
bit 4	INTOIE: INTO External Interrupt Enable bit
	 1 = Enables the INT0 external interrupt 0 = Disables the INT0 external interrupt
bit 3	RBIE: RB Port Change Interrupt Enable bit
	 1 = Enables the RB port change interrupt 0 = Disables the RB port change interrupt
bit 2	TMR0IF: TMR0 Overflow Interrupt Flag bit
	 1 = TMR0 register has overflowed (must be cleared in software) 0 = TMR0 register did not overflow
bit 1	INT0IF: INT0 External Interrupt Flag bit
	 1 = The INT0 external interrupt occurred (must be cleared in software) 0 = The INT0 external interrupt did not occur
bit 0	RBIF: RB Port Change Interrupt Flag bit ⁽¹⁾
	 1 = At least one of the RB<7:4> pins changed state (must be cleared in software) 0 = None of the RB<7:4> pins have changed state
Note 1:	A mismatch condition will continue to set this bit. Reading PORTB, and then waiting one additional instruction cycle, will end the mismatch condition and allow the bit to be cleared.

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R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP
bit 7							bit
Legend:							
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'			
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown	
bit 7		R Pull-un Enst	ole hit				
	RBPU: PORTB Pull-up Enable bit 1 = All PORTB pull-ups are disabled						
	0 = PORTB pull-ups are enabled by individual TRIS register values						
bit 6	INTEDG0: External Interrupt 0 Edge Select bit						
	1 = Interrupt on rising edge						
	0 = Interrupt on falling edge						
bit 5	INTEDG1: External Interrupt 1 Edge Select bit 1 = Interrupt on rising edge						
	0 = Interrupt on falling edge						
bit 4	INTEDG2: External Interrupt 2 Edge Select bit						
	1 = Interrupt on rising edge						
	0 = Interrupt	on falling edge					
bit 3 bit 2	INTEDG3: External Interrupt 3 Edge Select bit						
		on rising edge on falling edge					
	•	R0 Overflow Int	errunt Priority	bit			
511 2	1 = High prio		chupt i nonty	bit			
	0 = Low prior						
bit 1	INT3IP: INT3 External Interrupt Priority bit						
	1 = High prio	•					
	0 = Low prio	•					
bit 0		rt Change Inter	rupt Priority bit	[
	1 = High prio 0 = Low prio						
	p	- 2					
Note: In	nterrupt flag bits	are set when	n intorrunt co	ndition occurs	rogardlass of	the state of ite	orrospondir
	nable bit or the (•		•		•

are clear prior to enabling an interrupt. This feature allows for software polling.

REGISTER 11-2: INTCON2: INTERRUPT CONTROL REGISTER 2

REGISTER 11-3: INTCON3: INTERRUPT CONTROL REGISTER 3

R/W-1	R/W-1	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF
bit 7							bit C
Legend:			L :4			-l (0)	
R = Reada		W = Writable		-	nented bit, read		
-n = Value	alPOR	'1' = Bit is set		'0' = Bit is cle	areo	x = Bit is unkr	IOWN
bit 7	INT2IP: INT2	External Interr	upt Priority bit				
	1 = High prio	rity					
	0 = Low prior	ity					
bit 6	INT1IP: INT1	External Interr	upt Priority bit				
	1 = High prio 0 = Low prior	,					
bit 5	•	External Interr	unt Enable bit				
bit 5		the INT3 extern	•				
		the INT3 exter					
bit 4	INT2IE: INT2	External Interr	upt Enable bit				
		the INT2 exterr					
	0 = Disables	the INT2 exter	nal interrupt				
bit 3		External Interr	•				
		the INT1 extern					
bit 2		the INT1 exter External Interr	•				
DIL Z				must be cleare	d in software)		
		external interr			a in soltware)		
bit 1	INT2IF: INT2	External Interr	upt Flag bit				
	1 = The INT2	external interr	upt occurred (must be cleared	d in software)		
	0 = The INT2	external interr	upt did not occ	cur			
bit 0		External Interr					
				must be cleared	d in software)		
	0 = The IN11	external interr	upt ala not occ	Cur			
Note:	Interrupt flag bits						
	enable bit or the C are clear prior to			er software sho	buid ensure the	appropriate int	errupt flag bits

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11.2 PIR Registers

The PIR registers contain the individual flag bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are six Peripheral Interrupt Request (Flag) registers (PIR1 through PIR6).

- Note 1: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the Global Interrupt Enable bit, GIE (INTCON<7>).
 - 2: User software should ensure the appropriate interrupt flag bits are cleared prior to enabling an interrupt and after servicing that interrupt.

REGISTER 11-4: PIR1: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 1

R/W-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
bit 7							bit 0

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7	PSPIF: Parallel Slave Port Read/Write Interrupt Flag bit
	 1 = A read or write operation has taken place (must be cleared in software) 0 = No read or write operation has occurred
bit 6	ADIF: A/D Converter Interrupt Flag bit
	1 = An A/D conversion completed (must be cleared in software)0 = The A/D conversion is not complete
bit 5	RC1IF: EUSART Receive Interrupt Flag bit
	 1 = The EUSART receive buffer, RCREG1, is full (cleared when RCREG1 is read) 0 = The EUSART receive buffer is empty
bit 4	TX1IF: EUSART Transmit Interrupt Flag bit
	 1 = The EUSART transmit buffer, TXREG1, is empty (cleared when TXREG1 is written) 0 = The EUSART transmit buffer is full
bit 3	SSP1IF: Master Synchronous Serial Port Interrupt Flag bit
	1 = The transmission/reception is complete (must be cleared in software)0 = Waiting to transmit/receive
bit 2	TMR1GIF: Timer1 Gate Interrupt Flag bit
	1 = Timer gate interrupt occurred (must be cleared in software)0 = No timer gate interrupt occurred
bit 1	TMR2IF: TMR2 to PR2 Match Interrupt Flag bit
	1 = TMR2 to PR2 match occurred (must be cleared in software)0 = No TMR2 to PR2 match occurred
bit 0	TMR1IF: TMR1 Overflow Interrupt Flag bit
	1 = TMR1 register overflowed (must be cleared in software)0 = TMR1 register did not overflow

REGISTER 11-5: PIR2: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 2

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
OSCFIF		SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 OSCFIF: Osc	illator Fail Interrupt Flag bit
	scillator failed, clock input has changed to INTOSC (must be cleared in software) ock is operating
bit 6 Unimplemen	ted: Read as '0'
bit 5 SSP2IF: Mas	ter Synchronous Serial Port Interrupt Flag bit
	smission/reception has been completed (must be cleared in software) o transmit/receive
bit 4 BCL2IF: Bus	Collision Interrupt Flag bit
	lision occurred (must be cleared in software) ollision occurred
bit 3 BCL1IF: Bus	Collision Interrupt Flag bit
	lision occurred (must be cleared in software) ollision occurred
bit 2 HLVDIF: High	/Low-Voltage Detect Interrupt Flag bit
	tage condition occurred (must be cleared in software) ce voltage is above the regulator's low-voltage trip point
bit 1 TMR3IF: TMF	R3 Overflow Interrupt Flag bit
	gister overflowed (must be cleared in software) gister did not overflow
bit 0 TMR3GIF: TN	/R3 Gate Interrupt Flag bit
•	te interrupt occurred (must be cleared in software) gate interrupt occurred

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R/W-0	U-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0
TMR5GIF		RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
bit 7							bit
Legend:							
R = Readabl	le bit	W = Writable		•	mented bit, read		
-n = Value at	t POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unki	nown
bit 7	TMR5GIF:	imer5 Gate Inte	rrupt Flag bit				
	1 = Timer g	ate interrupt occ r gate interrupt o	urred (must b	e cleared in sof	ftware)		
bit 6		nted: Read as '					
bit 5	RC2IF: EUS	SART Receive In	terrupt Flag b	it			
		SART receive be SART receive be		2, is full (cleare	d when RCREC	62 is read)	
bit 4		ART Transmit In		it			
		SART transmit b SART transmit b		2, is empty (cle	eared when TXF	REG2 is written)
bit 3	CTMUIF: C	TMU Interrupt Fla	ag bit				
		nterrupt occurre IU interrupt occu		ared in softwar	re)		
bit 2	CCP2IF: EC	CP2 Interrupt Fl	ag bit				
		<u>de:</u> register capture R register capture		st be cleared in	software)		
	Compare me 1 = A TMR	-	e match occur		eared in softwa	re)	
	<u>PWM mode</u> Unused in th	<u>.</u>		ined			
bit 1		CP1 Interrupt Fl	aq bit				
	<u>Capture mo</u> 1 = A TMR		occurred (mu	st be cleared in	i software)		
		<u>ode:</u> register compare R register compa			eared in softwa	re)	
	<u>PWM mode</u> Unused in th						
bit 0	RTCCIF: RT	CC Interrupt Fla	ıg bit				
		nterrupt occurred C interrupt occu		ared in softwar	e)		

REGISTER 11-6: PIR3: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 3

REGISTER 11-7: PIR4: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 4

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CCP10IF ⁽¹⁾	CCP9IF ⁽¹⁾	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
bit 7	- -						bit 0
d.							
Legend:	L :1		L:1			-l (O'	
R = Readable		W = Writable		•	nented bit, rea		
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7-1 bit 0	Capture Mode 1 = A TMR re 0 = No TMR Compare Mod 1 = A TMR re 0 = No TMR <u>PWM Mode</u> : Not used in P	egister capture register captur <u>de:</u> egister compar register compa	occurred (mu e occurred e match occu re match occ	ist be cleared in rred (must be c	n software) cleared in softw	are)	
	Capture Mode 1 = A TMR re 0 = No TMR Compare Mod 1 = A TMR re	<u>e:</u> egister capture register captur <u>de:</u> egister compare register compa	occurred (mu e occurred e match occur		n software) cleared in softw	are)	

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

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					、 ,					
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
TMR7GIF ⁽¹⁾	TMR12IF ⁽¹⁾	TMR10IF ⁽¹⁾	TMR8IF	TMR7IF ⁽¹⁾	TMR6IF	TMR5IF	TMR4IF			
bit 7	·	•		•		·	bit			
Legend:										
R = Readable	bit	W = Writable bi	t	U = Unimplem	ented bit. rea	nd as '0'				
-n = Value at F	POR	'1' = Bit is set	-	'0' = Bit is clea		x = Bit is unkr	nown			
bit 7		/IR7 Gate Interru	nt Elag bita	(1)						
					offwara)					
		 1 = TMR gate interrupt occurred (bit must be cleared in software) 0 = No TMR gate interrupt occurred 								
bit 6		IR12 to PR12 Ma		ot Flag bit ⁽¹⁾						
		o PR12 match or			software)					
	0 = No TMR1	= No TMR12 to PR12 match occurred								
bit 5	TMR10IF: TM	IR10 to PR10 Ma	atch Interru	ot Flag bit ⁽¹⁾						
	 1 = TMR10 to PR10 match occurred (must be cleared in software) 0 = No TMR10 to PR10 match occurred 									
bit 4		R8 to PR8 Match	•	•	(1					
		PR8 match occu to PR8 match o		be cleared in so	mware)					
bit 3				it(1)						
bit o	TMR7IF: TMR7 Overflow Interrupt Flag bit ⁽¹⁾ 1 = TMR7 register overflowed (must be cleared in software)									
		gister did not ove			,					
bit 2	TMR6IF: TMF	R6 to PR6 Match	Interrupt F	lag bit						
	1 = TMR6 to PR6 match occurred (must be cleared in software)									
		6 to PR6 match c								
bit 1		TMR5IF: TMR5 Overflow Interrupt Flag bit								
		gister overflowed gister did not ove		leared in softwa	ire)					
bit 0		R4 to PR4 Match		lag hit						
		PR4 match occu	•	•	ftware)					
		to PR4 match c								
Note 1: Uni	implemented on	n devices with a r	roaram me	mory of 32 Khy	Les (PIC12FY	5K22)				

REGISTER 11-8: PIR5: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 5

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

REGISTER 11-9: PIR6: PERIPHERAL INTERRUPT REQUEST (FLAG) REGISTER 6

U-0	U-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
—	—	—	EEIF	—	CMP3IF	CMP2IF	CMP1IF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-5	Unimplemented: Read as '0'
bit 4	EEIF: Data EEDATA/Flash Write Operation Interrupt Flag bit
	 1 = The write operation is complete (must be cleared in software) 0 = The write operation is not complete or has not been started
bit 3	Unimplemented: Read as '0'
bit 2	CMP3IF: CMP3 Interrupt Flag bit
	 1 = CMP3 interrupt occurred (must be cleared in software) 0 = No CMP3 interrupt occurred
bit 1	CMP2IF: CMP2 Interrupt Flag bit
	 1 = CMP2 interrupt occurred (must be cleared in software) 0 = No CMP2 interrupt occurred
bit 0	CMP1IF: CM1 Interrupt Flag bit
	 1 = CMP1 interrupt occurred (must be cleared in software) 0 = No CMP1 interrupt occurred

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11.3 PIE Registers

The PIE registers contain the individual enable bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are six Peripheral Interrupt Enable registers (PIE1 through PIE6). When IPEN (RCON<7>) = 0, the PEIE bit must be set to enable any of these peripheral interrupts.

REGISTER 11-10: PIE1: PERIPHERAL INTERRUPT ENABLE REGISTER 1

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE			
bit 7							bit 0			
Legend:										
R = Readab	ole bit	W = Writable	bit	U = Unimplen	nented bit, read	1 as '0'				
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea		x = Bit is unkr	nown			
bit 7	PSPIE: Paral	lel Slave Port F	Read/Write Inte	errupt Enable bi	t					
		he PSP read/w the PSP read/w								
bit 6		onverter Interru								
		he A/D interrup								
	0 = Disables	the A/D interrup	ot							
bit 5	RC1IE: EUSA	ART Receive In	terrupt Enable	e bit						
		the EUSART receive interrupt the EUSART receive interrupt								
bit 4		RT Transmit Ir	•							
DIL 4		he EUSART tra	•							
		the EUSART tr								
bit 3	SSP1IE: Mas	ter Synchronou	is Serial Port I	nterrupt Enable	bit					
		he MSSP interi								
		the MSSP inter	•							
bit 2		MR1 Gate Inter	rupt Enable bi	t						
	1 = Enables t 0 = Disabled	0								
bit 1		R2 to PR2 Mate	h Interrupt En	able bit						
		he TMR2 to PF	•							
		the TMR2 to PI								
bit 0		R1 Overflow Int		bit						
		he TMR1 overf								
	0 = Disables	the TMR1 over	now interrupt							

REGISTER 11-11: PIE2: PERIPHERAL INTERRUPT ENABLE REGISTER 2

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0						
OSCFIE	_	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE						
bit 7	•						bit 0						
Legend:													
R = Readable		W = Writable	oit	U = Unimplem		d as '0'							
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown						
bit 7		cillator Fail Inter	rupt Enable bi	t									
	1 = Enabled 0 = Disabled												
bit 6	Unimplemer	nimplemented: Read as '0'											
bit 5	5 SSP2IE: Master Synchronous Serial Port 2 Interrupt Enable bit												
		the MSSP inter the MSSP inter											
bit 4	BCL2IE: Bus	Collision Interr	upt Enable bit										
		the bus collision the bus collision											
bit 3	BCL1IE: Bus	BCL1IE: Bus Collision Interrupt Enable bit											
	1 = Enabled 0 = Disabled	1											
bit 2	HLVDIE: Hig	h/Low-Voltage [Detect Interrup	t Enable bit									
	1 = Enabled 0 = Disabled												
bit 1	TMR3IE: TM	R3 Overflow Int	errupt Enable	bit									
	1 = Enabled 0 = Disabled												
bit 0	TMR3GIE: T	imer3 Gate Inte	rrupt Enable b	it									
	1 = Enabled 0 = Disabled												

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R/W-0	U-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0				
TMR5GIE		RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE				
bit 7							bit 0				
Lovende											
Legend: R = Readable	- hit	W = Writable	hit	II – Unimplor	nented bit, read	d ac 'O'					
-n = Value at		'1' = Bit is set		'0' = Bit is clea		x = Bit is unkr					
	TOR				arcu		101111				
bit 7	TMR5GIE: Tir	mer5 Gate Inter	rupt Enable bi	t							
	1 = Enabled		·								
	0 = Disabled										
bit 6	Unimplemented: Read as '0'										
bit 5		RC2IE: EUSART Receive Interrupt Enable bit									
	1 = Enabled										
bit 4		RT Transmit In	terrupt Enable	e bit							
	1 = Enabled 0 = Disabled										
bit 3		MU Interrupt Er	nable bit								
	1 = Enabled										
	0 = Disabled										
bit 2	CCP2IE: ECO	CP2 Interrupt E	nable bit								
	1 = Enabled										
	0 = Disabled										
bit 1		CP1 Interrupt E	nable bit								
	1 = Enabled 0 = Disabled										
bit 0		CC Interrupt En	able bit								
	1 = Enabled										
	0 = Disabled										

REGISTER 11-12: PIE3: PERIPHERAL INTERRUPT ENABLE REGISTER 3

REGISTER 11-13: PIE4: PERIPHERAL INTERRUPT ENABLE REGISTER 4

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CCP10IE ⁽¹⁾	CCP9IE ⁽¹⁾	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
bit 7							bit 0
Legend:							
R = Readable I	bit	W = Writable I	oit	U = Unimpler	nented bit, read	as '0'	
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cleared x = Bit is unknown			
bit 7-0	CCP<10:3>IE	: CCP<10:3> I	nterrupt Enat	ble bits ⁽¹⁾			
	1 - Enabled						

1 = Enabled 0 = Disabled

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

REGISTER 11-14: PIE5: PERIPHERAL INTERRUPT ENABLE REGISTER 5

	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0				
) TMR12IE ⁽¹⁾	TMR10IE ⁽¹⁾	TMR8IE	TMR7IE ⁽¹⁾	TMR6IE	TMR5IE	TMR4IE				
						bit C				
e bit	W = Writable I	oit	U = Unimplem	nented bit, read	d as '0'					
POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	iown				
	MR7 Gate Interi	rupt Enable b	it ⁽¹⁾							
		Aatob Intorru	ot Encolo bit(1)							
TMR10IE: TM	IR10 to PR10 N	/latch Interru	ot Enable bit ⁽¹⁾							
1 = Enables	1 = Enables the TMR10 to PR10 match interrupt									
0 = Disables	the TMR10 to I	PR10 match i	nterrupt							
		•								
			•							
0 = Disables the TMR7 overflow interrupt										
TMR6IE: TM	R6 to PR6 Matc	h Interrupt E	nable bit							
	·									
0 = Disables	the TMR6 to P	R6 match inte	errupt							
		•								
		•								
			•							
	1 = Enabled 0 = Disabled TMR12IE: TM 1 = Enables 0 = Disables TMR10IE: TM 1 = Enables 0 = Disables TMR8IE: TMI 1 = Enables 0 = Disables TMR6IE: TMI 1 = Enables 0 = Disables TMR5IE: TMI 1 = Enables 0 = Disables TMR5IE: TMI 1 = Enables 0 = Disables TMR4IE: TMI 1 = Enables	POR '1' = Bit is set TMR7GIE: TMR7 Gate Internant 1 = Enabled 0 = Disabled TMR12IE: TMR12 to PR12 M 1 = Enables the TMR12 to PR12 M 1 = Enables the TMR12 to PR10 M 1 = Enables the TMR12 to PR10 M 1 = Enables the TMR10 to PR10 M 1 = Enables the TMR8 to PR 0 = Disables the TMR8 to PR 0 = Disables the TMR7 overflow Int 1 = Enables the TMR7 overflow Int 1 = Enables the TMR6 to PR 0 = Disables the TMR6 to PR 0 = Disables the TMR6 to PR 0 = Disables the TMR5 overflow Int 1 = Enables the TMR4 to PR4 Mato 1 = Enables the	POR '1' = Bit is set TMR7GIE: TMR7 Gate Interrupt Enable b 1 = Enabled 0 = Disabled TMR12IE: TMR12 to PR12 Match Interrupt 1 = Enables the TMR12 to PR12 match in 0 = Disables the TMR12 to PR12 match in 0 = Disables the TMR12 to PR12 match in 0 = Disables the TMR10 to PR10 match in 0 = Disables the TMR10 to PR10 match in 0 = Disables the TMR10 to PR10 match in 0 = Disables the TMR8 to PR8 match interrupt Enables 1 = Enables the TMR8 to PR8 match interrupt 0 = Disables the TMR7 overflow Interrupt Enables 1 = Enables the TMR7 overflow interrupt 0 = Disables the TMR7 overflow interrupt 1 = Enables the TMR7 overflow interrupt 0 = Disables the TMR7 overflow interrupt 0 = Disables the TMR7 overflow interrupt 1 = Enables the TMR6 to PR6 match interrupt 0 = Disables the TMR6 to PR6 match interrupt 1 = Enables the TMR5 overflow interrupt <td>POR'1' = Bit is set'0' = Bit is clearTMR7GIE: TMR7 Gate Interrupt Enable bit(1)1 = Enabled0 = DisabledTMR12IE: TMR12 to PR12 Match Interrupt Enable bit(1)1 = Enables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interrupt1 = Enables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interrupt1 = Enables the TMR10 to PR10 Match Interrupt Enable bit(1)1 = Enables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt1 = Enables the TMR8 to PR8 Match Interrupt Enable bit1 = Enables the TMR8 to PR8 match interrupt0 = Disables the TMR8 to PR8 match interrupt1 = Enables the TMR7 Overflow Interrupt Enable bit(1)1 = Enables the TMR7 overflow interrupt</td> <td>POR'1' = Bit is set'0' = Bit is clearedTMR7GIE: TMR7 Gate Interrupt Enable bit(1)1 = Enabled0 = DisabledTMR12IE: TMR12 to PR12 Match Interrupt Enable bit(1)1 = Enables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interruptTMR10IE: TMR10 to PR10 Match Interrupt Enable bit(1)1 = Enables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR8 match InterruptTMR8IE: TMR8 to PR8 Match Interrupt Enable bit1 = Enables the TMR8 to PR8 match interrupt0 = Disables the TMR8 to PR8 match interrupt0 = Disables the TMR7 overflow Interrupt Enable bit(1)1 = Enables the TMR7 overflow interrupt0 = Disables the TMR7 overflow interrupt0 = Disables the TMR7 overflow interrupt1 = Enables the TMR6 to PR6 match interrupt0 = Disables the TMR6 to PR6 match interrupt1 = Enables the TMR5 overflow Interrupt Enable bit1 = Enables the TMR5 overflow interrupt0 = Disables the TMR5 overflow interrupt0 = Disables the TMR5 overflow interrupt1 = Enables the</td> <td>POR'1' = Bit is set'0' = Bit is clearedx = Bit is unknownTMR7GIE: TMR7 Gate Interrupt Enable bit11 = Enabled0 = DisabledTMR12IE: TMR12 to PR12 Match Interrupt Enable bit1 = Enables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interrupt0 = Disables the TMR10 to PR10 Match Interrupt Enable bit1 = Enables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt0 = Disables the TMR8 to PR8 match interrupt0 = Disables the TMR6 to PR6 match interrupt0 = Disables the TMR7 overflow interrupt1 = Enables the TMR7 overflow interrupt0 = Disables the TMR6 to PR6 match interrupt0 = Disables the TMR6 to PR6 match interrupt0 = Disables the TMR6 to PR6 match interrupt1 = Enables the TMR5 overflow interrupt1 = Enables the TMR4 to PR4 match Interrupt1 = Enables the TMR4 to PR4 match interrupt</td>	POR'1' = Bit is set'0' = Bit is clearTMR7GIE: TMR7 Gate Interrupt Enable bit(1)1 = Enabled0 = DisabledTMR12IE: TMR12 to PR12 Match Interrupt Enable bit(1)1 = Enables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interrupt1 = Enables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interrupt1 = Enables the TMR10 to PR10 Match Interrupt Enable bit(1)1 = Enables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt1 = Enables the TMR8 to PR8 Match Interrupt Enable bit1 = Enables the TMR8 to PR8 match interrupt0 = Disables the TMR8 to PR8 match interrupt1 = Enables the TMR7 Overflow Interrupt Enable bit(1)1 = Enables the TMR7 overflow interrupt	POR'1' = Bit is set'0' = Bit is clearedTMR7GIE: TMR7 Gate Interrupt Enable bit(1)1 = Enabled0 = DisabledTMR12IE: TMR12 to PR12 Match Interrupt Enable bit(1)1 = Enables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interruptTMR10IE: TMR10 to PR10 Match Interrupt Enable bit(1)1 = Enables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR8 match InterruptTMR8IE: TMR8 to PR8 Match Interrupt Enable bit1 = Enables the TMR8 to PR8 match interrupt0 = Disables the TMR8 to PR8 match interrupt0 = Disables the TMR7 overflow Interrupt Enable bit(1)1 = Enables the TMR7 overflow interrupt0 = Disables the TMR7 overflow interrupt0 = Disables the TMR7 overflow interrupt1 = Enables the TMR6 to PR6 match interrupt0 = Disables the TMR6 to PR6 match interrupt1 = Enables the TMR5 overflow Interrupt Enable bit1 = Enables the TMR5 overflow interrupt0 = Disables the TMR5 overflow interrupt0 = Disables the TMR5 overflow interrupt1 = Enables the	POR'1' = Bit is set'0' = Bit is clearedx = Bit is unknownTMR7GIE: TMR7 Gate Interrupt Enable bit11 = Enabled0 = DisabledTMR12IE: TMR12 to PR12 Match Interrupt Enable bit1 = Enables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interrupt0 = Disables the TMR12 to PR12 match interrupt0 = Disables the TMR10 to PR10 Match Interrupt Enable bit1 = Enables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt0 = Disables the TMR10 to PR10 match interrupt0 = Disables the TMR8 to PR8 match interrupt0 = Disables the TMR6 to PR6 match interrupt0 = Disables the TMR7 overflow interrupt1 = Enables the TMR7 overflow interrupt0 = Disables the TMR6 to PR6 match interrupt0 = Disables the TMR6 to PR6 match interrupt0 = Disables the TMR6 to PR6 match interrupt1 = Enables the TMR5 overflow interrupt1 = Enables the TMR4 to PR4 match Interrupt1 = Enables the TMR4 to PR4 match interrupt				

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

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U-0	U-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0			
_	_	—	EEIE	_	CMP3IE	CMP2IE	CMP1IE			
bit 7							bit 0			
Legend:										
R = Readable bit W = Writable bit U = Unimplemented bit, read a						d as '0'				
-n = Value at PO	DR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown			
bit 7-5	Unimplemen	ted: Read as ')'							
bit 4	EEIE: Data EEDATA/Flash Write Operation Enable bit									
	1 = Interrupt is enabled									
	0 = interrupt	is disabled								
bit 3	Unimplemen	ted: Read as ')'							
bit 2	CMP3IE: CM	P3 Enable bit								
	1 = Interrupt is enabled									
	0 = interrupt	is disabled								
bit 1	CMP2E: CMF	CMP2E: CMP2 Enable bit								
	1 = Interrupt is enabled									
	0 = interrupt i									
	CMP1IE: CMI									
	1 = Interrupt	is enabled								

REGISTER 11-15: PIE6: PERIPHERAL INTERRUPT ENABLE REGISTER 6

1 = Interrupt is enabled0 = interrupt is disabled

11.4 IPR Registers

The IPR registers contain the individual priority bits for the peripheral interrupts. Due to the number of peripheral interrupt sources, there are six Peripheral Interrupt Priority registers (IPR1 through IPR6). Using the priority bits requires that the Interrupt Priority Enable (IPEN) bit (RCON<7>) be set.

REGISTER 11-16: IPR1: PERIPHERAL INTERRUPT PRIORITY REGISTER 1

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	
PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP	
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimplemented bit, read as '0'				
-n = Value at POR '1' = Bit is s		'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	nown	

bit 7	PSPIP: Parallel Slave Port Read/Write Interrupt Priority bit 1 = High priority 0 = Low priority
bit 6	ADIP: A/D Converter Interrupt Priority bit 1 = High priority 0 = Low priority
bit 5	RC1IP: EUSART Receive Interrupt Priority bit 1 = High priority 0 = Low priority
bit 4	TX1IP: EUSART Transmit Interrupt Priority bit 1 = High priority 0 = Low priority
bit 3	SSP1IP: Master Synchronous Serial Port Interrupt Priority bit 1 = High priority 0 = Low priority
bit 2	TMR1GIP: Timer1 Gate Interrupt Priority bit 1 = High priority 0 = Low priority
bit 1	TMR2IP: TMR2 to PR2 Match Interrupt Priority bit 1 = High priority 0 = Low priority
bit 0	TMR1IP: TMR1 Overflow Interrupt Priority bit 1 = High priority 0 = Low priority

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			-									
R/W-1	U-0	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1					
OSCFIP	—	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP					
bit 7							bit (
Legend:												
Legend: R = Readabl	o hit	W = Writable	hit	II = I Inimplen	nented bit, read	1 as '0'						
-n = Value at		'1' = Bit is set	'0' = Bit is clea		x = Bit is unk	nown						
					areu		nown					
bit 7	OSCFIP: Osc	cillator Fail Inter	rrupt Priority bi	t								
		= High priority										
	0 = Low prio	rity										
oit 6	Unimplemen	Jnimplemented: Read as '0'										
oit 5	SSP2IP: Master Synchronous Serial Port 2 Interrupt Priority bit											
		1 = High priority										
	0 = Low prio	0 = Low priority										
oit 4		CL2IP: Bus Collision Interrupt priority bit (MSSP)										
	1 = High priority											
-:+ 0	•	0 = Low priority										
bit 3	BCL1IP: Bus Collision Interrupt Priority bit											
	1 = High priority 0 = Low priority											
bit 2	-	•	Detect Interrup	t Priority bit								
	HLVDIP: High/Low-Voltage Detect Interrupt Priority bit 1 = High priority											
	0 = Low prio											
bit 1	TMR3IP: TMR3 Overflow Interrupt Priority bit											
	1 = High pric	ority										
	0 = Low prior	rity										
bit 0		MR3 Gate Inter	rupt Priority bit	:								
	1 = High price	•										
	0 = Low prior	rity										

REGISTER 11-17: IPR2: PERIPHERAL INTERRUPT PRIORITY REGISTER 2

REGISTER 11-18: IPR3: PERIPHERAL INTERRUPT PRIORITY REGISTER 3

R/W-1 TMR5GIP bit 7 Legend:	U-0 —	R-1 RC2IP	R-1 TX2IP	R/W-1 CTMUIP	R/W-1 CCP2IP	R/W-1 CCP1IP	R/W-1 RTCCIP					
bit 7	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP					
							111001					
Legend:							bit (
Legend:												
R = Readable	bit	W = Writable	oit	U = Unimplem	ented bit, read	d as '0'						
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	own					
bit 7	TMR5GIP: Tir	ner5 Gate inter	rupt Priority b	it								
	1 = High prior											
	0 = Low prior	•										
		Jnimplemented: Read as '0'										
	RC2IP: EUSART Receive Priority Flag bit											
	1 = High prior											
	0 = Low prior	,										
		RT Transmit In	terrupt Priority	/ DIT								
	1 = High prior 0 = Low prior											
	•	//U Interrupt Pr	iority hit									
	1 = High prior	•	ionty bit									
	0 = Low priority											
	-	P2 Interrupt P	riority bit									
	1 = High prior	rity										
	0 = Low prior	ity										
bit 1	CCP1IP: ECC	P1 Interrupt P	riority bit									
	1 = High prior											
	0 = Low prior	-										
		C Interrupt Pri	ority bit									
	1 = High prior	•										
	0 = Low prior	цу										

REGISTER 11-19: IPR4: PERIPHERAL INTERRUPT PRIORITY REGISTER 4

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
CCP10IP ⁽¹⁾	CCP9IP ⁽¹⁾	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
bit 7							bit 0

Legend:					
R = Readable bit	W = Writable bit	/ = Writable bit U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-0 CCP<10:3>IP: CCP<10:3> Interrupt Priority bits⁽¹⁾

1 = High priority

0 = Low priority

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

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R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1		
TMR7GIP ⁽¹⁾	TMR12IP ⁽¹⁾	TMR10IP ⁽¹⁾	TMR8IP	TMR7IP ⁽¹⁾	TMR6IP	TMR5IP	TMR4IP		
bit 7		·		· ·			bit C		
Legend:									
Legend. R = Readable	hit	W = Writable b	i+	U = Unimplem	opted bit rea	d aa '0'			
-n = Value at F		'1' = Bit is set	it.	'0' = Bit is clea		x = Bit is unkr			
	OK	i – Dit is set			areu				
bit 7	TMR7GIP: TM	IR7 Gate Interru	upt Priority b	it(1)					
	1 = High prior								
	0 = Low priori	ty							
bit 6	TMR12IP: TM	R12 to PR12 M	atch Interrup	ot Priority bit ⁽¹⁾					
	1 = High prior	•							
	0 = Low priori	•							
bit 5		FMR10IP: TMR10 to PR10 Match Interrupt Priority bit ⁽¹⁾							
	1 = High prior 0 = Low priori	•							
bit 4	•	8 to PR8 Match	n Interrupt Pi	riority bit					
	1 = High prior		i inton apri i	lionty sit					
	0 = Low priori	•							
bit 3	TMR7IP: TMR	7 Overflow Inte	rrupt Priority	/ bit ⁽¹⁾					
	1 = High prior								
	0 = Low priori	•							
bit 2		6 to PR6 Match	n Interrupt Pi	riority bit					
	1 = High prior								
bit 1	•	 0 = Low priority TMR5IP: TMR5 Overflow Interrupt Priority bit 							
	1 = High prior		nuptinonty	, bit					
	0 = Low priori								
bit 0	-	4 to PR4 Match	n Interrupt Pi	riority bit					
	1 = High prior		-	-					
	0 = Low priori	tv							

REGISTER 11-20: IPR5: PERIPHERAL INTERRUPT PRIORITY REGISTER 5

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

REGISTER 11-21: IPR6: PERIPHERAL INTERRUPT PRIORITY REGISTER 6

U-0	U-0	U-0	R/W-1	U-0	R/W-1	R/W-1	R/W-1		
_	_	_	EEIP	_	CMP3IP	CMP2IP	CMP1IP		
bit 7							bit 0		
Legend:									
R = Readab	le bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'			
-n = Value a	It POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown		
bit 7-5	Unimplement	ted: Read as '	0'						
bit 4	EEIP: EE Inte	rrupt Priority b	it						
	1 = High prior	rity							
	0 = Low prior	ity							
bit 3	Unimplement	ted: Read as '	0'						
bit 2	CMP3IP: CM	P3 Interrupt Pr	iority bit						
	1 = High prior	rity							
	0 = Low prior	ity							
bit 1	CMP2IP: CM	P2 Interrupt Pr	iority bit						
		1 = High priority							
	0 = Low prior	ity							
bit 0	CMP1IP: CM	P1 Interrupt Pr	iority bit						
	1 = High prior								
	0 = Low prior	ity							

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11.5 RCON Register

The RCON register contains the bits used to determine the cause of the last Reset, or wake-up from Idle or Sleep modes. RCON also contains the bit that enables interrupt priorities (IPEN).

REGISTER 11-22: RCON: RESET CONTROL REGISTER

R/W-0	R/W-1	R/W-1	R/W-1	R-1	R-1	R/W-0	R/W-0
IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	IPEN: Interrupt Priority Enable bit
	1 = Enable priority levels on interrupts
	0 = Disable priority levels on interrupts (PIC16CXXX Compatibility mode)
bit 6	SBOREN: Software BOR Enable bit
	For details of bit operation, see Register 5-1.
bit 5	CM: Configuration Mismatch Flag bit
	1 = A Configuration Mismatch Reset has not occurred
	0 = A Configuration Mismatch Reset has occurred (must be subsequently set in software)
bit 4	RI: RESET Instruction Flag bit
	For details of bit operation, see Register 5-1.
bit 3	TO: Watchdog Timer Time-out Flag bit
	For details of bit operation, see Register 5-1.
bit 2	PD: Power-Down Detection Flag bit
	For details of bit operation, see Register 5-1.
bit 1	POR: Power-on Reset Status bit
	For details of bit operation, see Register 5-1.
bit 0	BOR: Brown-out Reset Status bit
	For details of bit operation, see Register 5-1.

11.6 INTx Pin Interrupts

External interrupts on the RB0/INT0, RB1/INT1, RB2/INT2 and RB3/INT3 pins are edge-triggered. If the corresponding INTEDGx bit in the INTCON2 register is set (= 1), the interrupt is triggered by a rising edge. If that bit is clear, the trigger is on the falling edge.

When a valid edge appears on the RBx/INTx pin, the corresponding flag bit, INTxIF, is set. This interrupt can be disabled by clearing the corresponding enable bit, INTxIE. Before re-enabling the interrupt, the flag bit (INTxIF) must be cleared in software in the Interrupt Service Routine.

All external interrupts (INT0, INT1, INT2 and INT3) can wake up the processor from the power-managed modes if bit, INTxIE, was set prior to going into the power-managed modes. If the Global Interrupt Enable bit (GIE) is set, the processor will branch to the interrupt vector following wake-up.

The interrupt priority for INT1, INT2 and INT3 is determined by the value contained in the Interrupt Priority bits, INT1IP (INTCON3<6>), INT2IP (INTCON3<7>) and INT3IP (INTCON2<1>).

There is no priority bit associated with INTO. It is always a high-priority interrupt source.

11.7 TMR0 Interrupt

In 8-bit mode (the default), an overflow in the TMR0 register (FFh \rightarrow 00h) will set flag bit, TMR0IF. In 16-bit mode, an overflow in the TMR0H:TMR0L register pair (FFFFh \rightarrow 0000h) will set TMR0IF.

The interrupt can be enabled/disabled by setting/clearing enable bit, TMR0IE (INTCON<5>). Interrupt priority for Timer0 is determined by the value contained in the interrupt priority bit, TMR0IP (INTCON2<2>). For further details on the Timer0 module, see **Section 13.0 "Timer0 Module**".

11.8 PORTB Interrupt-on-Change

An input change on PORTB<7:4> sets flag bit, RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON<3>).

Interrupt priority for PORTB interrupt-on-change is determined by the value contained in the interrupt priority bit, RBIP (INTCON2<0>).

11.9 Context Saving During Interrupts

During interrupts, the return PC address is saved on the stack. Additionally, the WREG, STATUS and BSR registers are saved on the Fast Return Stack.

If a fast return from interrupt is not used (see **Section 6.3 "Data Memory Organization**"), the user may need to save the WREG, STATUS and BSR registers on entry to the Interrupt Service Routine (ISR). Depending on the user's application, other registers may also need to be saved.

Example 11-1 saves and restores the WREG, STATUS and BSR registers during an Interrupt Service Routine.

EXAMPLE 11-1:	SAVING STATUS, WREG AND BSR REGISTERS IN RAM	
---------------	--	--

MOVWF W_TEMP MOVFF STATUS, STATUS_TEMP MOVFF BSR, BSR TEMP	<pre>% W_TEMP is in virtual bank % STATUS_TEMP located anywhere % BSR TMEP located anywhere</pre>
;	, BSR_IMEP located anywhere
; USER ISR CODE	
;	
MOVFF BSR_TEMP, BSR	; Restore BSR
MOVF W_TEMP, W	; Restore WREG
MOVFF STATUS_TEMP, STATUS	; Restore STATUS

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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP
INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF
PIR1	PSPIP	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIR2	OSCFIF	—	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF
PIR3	TMR5GIF	—	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIR4	CCP10IF ⁽¹⁾	CCP9IF ⁽¹⁾	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
PIR5	TMR7GIF ⁽¹⁾	TMR12IF ⁽¹⁾	TMR10IF ⁽¹⁾	TMR8IF	TMR7IF ⁽¹⁾	TMR6IF	TMR5IF	TMR4IF
PIR6		—	_	EEIF	—	CMP3IF	CMP2IF	CMP1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
PIE2	OSCFIE	—	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE
PIE3	TMR5GIE	—	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
PIE4	CCP10IE ⁽¹⁾	CCP9IE ⁽¹⁾	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
PIE5	TMR7GIE ⁽¹⁾	TMR12IE ⁽¹⁾	TMR10IE ⁽¹⁾	TMR8IE	TMR7IE ⁽¹⁾	TMR6IE	TMR5IE	TMR4IE
PIE6	_	—	_	EEIE	—	CMP3IE	CMP2IE	CMP1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
IPR2	OSCFIP	—	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP
IPR3	TMR5GIP	—	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
IPR4	CCP10IP ⁽¹⁾	CCP9IP ⁽¹⁾	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
IPR5	TMR7GIP ⁽¹⁾	TMR12IP ⁽¹⁾	TMR10IP ⁽¹⁾	TMR8IP	TMR7IP ⁽¹⁾	TMR6IP	TMR5IP	TMR4IP
IPR6		_		EEIP		CMP3IP	CMP2IP	CMP1IP
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR

TABLE 11-1: SUMMARY OF REGISTERS ASSOCIATED WITH INTERRUPTS

Legend: Shaded cells are not used by the interrupts.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

12.0 I/O PORTS

Depending on the device selected and features enabled, there are up to nine ports available. Some pins of the I/O ports are multiplexed with an alternate function from the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

Each port has three memory mapped registers for its operation:

- TRIS register (Data Direction register)
- PORT register (reads the levels on the pins of the device)
- LAT register (Output Latch register)

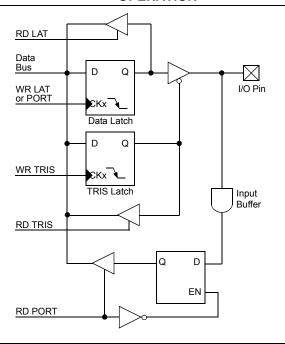
Reading the PORT register reads the current status of the pins, whereas writing to the PORT register writes to the Output Latch (LAT) register.

Setting a TRIS bit (= 1) makes the corresponding port pin an input (putting the corresponding output driver in a High-Impedance mode). Clearing a TRIS bit (= 0) makes the corresponding port pin an output (i.e., puts the contents of the corresponding LAT bit on the selected pin).

The Output Latch (LAT register) is useful for read-modify-write operations on the value that the I/O pins are driving. Read-modify-write operations on the LAT register read and write the latched output value for the PORT register.

A simplified model of a generic I/O port, without the interfaces to other peripherals, is shown in Figure 12-1.

FIGURE 12-1: GENERIC I/O PORT OPERATION



12.1 I/O Port Pin Capabilities

When developing an application, the capabilities of the port pins must be considered. Outputs on some pins have higher output drive strength than others. Similarly, some pins can tolerate higher than VDD input levels.

All of the digital ports are 5.5V input tolerant. The analog ports have the same tolerance – having clamping diodes implemented internally.

12.1.1 PIN OUTPUT DRIVE

When used as digital I/O, the output pin drive strengths vary, according to the pins' grouping, to meet the needs for a variety of applications. In general, there are two classes of output pins, in terms of drive capability:

- Outputs designed to drive higher current loads, such as LEDs:
 - PORTA PORTB
 - PORTC
- Outputs with lower drive levels, but capable of driving normal digital circuit loads with a high input impedance. Able to drive LEDs, but only those with smaller current requirements:
 - PORTD PORTE
 - PORTE PORTG
 - PORTH^(†) PORTJ^(†)
 - † These ports are not available on 64-pin devices.

12.1.2 PULL-UP CONFIGURATION

Four of the I/O ports (PORTB, PORTD, PORTE and PORTJ) implement configurable weak pull-ups on all pins. These are internal pull-ups that allow floating digital input signals to be pulled to a consistent level without the use of external resistors.

The pull-ups are enabled with a single bit for each of the ports: RBPU (INTCON2<7>) for PORTB, and RDPU, REPU and RJPU (PADCFG1<7:5>) for the other ports.

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REGISTER 12-1: PADCFG1: PAD CONFIGURATION REGISTER

R/W-0	R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	U-0		
RDPU	REPU	RJPU ⁽²⁾	_	_	RTSECSEL1 ⁽¹⁾	RTSECSEL0 ⁽¹⁾	_		
bit 7							bit 0		
-									
Legend:									
R = Readal	ble bit	W = Writable	bit	U = Unimple	emented bit, read a	s '0'			
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is cl	eared	x = Bit is unknown			
bit 7	RDPU: POF	RTD Pull-up Ena	able bit						
				-	al port latch values				
		TD pull-up resis	tors are disa	abled					
bit 6	REPU: POR	RTE Pull-up Ena	ble bit						
		pull-up resistors			al port latch values				
bit 5		TJ Pull-up Enal							
		•		d by individua	l port latch values				
		TJ pull-up resist			•				
bit 4-3	Unimpleme	nted: Read as	'0'						
bit 2-1	RTSECSEL	<1:0>: RTCC S	econds Clo	ck Output Sel	ect bits ⁽¹⁾				
	11 = Reserv	ved; do not use							
		10 = RTCC source clock is selected for the RTCC pin (the pin can be LF-INTOSC or SOSC, depending on the RTCOSC (CONFIG3L<1>) bit setting)							
		seconds clock i	,	• /	bin				
		alarm pulse is s							
bit 0	Unimpleme	nted: Read as	'0'						
	-								

- Note 1: To enable the actual RTCC output, the RTCOE (RTCCFG<2>) bit must be set.
 - 2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

12.1.3 OPEN-DRAIN OUTPUTS

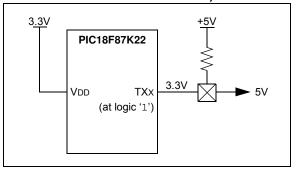
The output pins for several peripherals are also equipped with a configurable, open-drain output option. This allows the peripherals to communicate with external digital logic, operating at a higher voltage level, without the use of level translators.

The open-drain option is implemented on port pins specifically associated with the data and clock outputs of the USARTs, the MSSP module (in SPI mode) and the CCP modules. This option is selectively enabled by setting the open-drain control bits in the registers, ODCON1, ODCON2 and ODCON3.

When the open-drain option is required, the output pin must also be tied through an external pull-up resistor provided by the user to a higher voltage level, up to 5V (Figure 12-2). When a digital logic high signal is output, it is pulled up to the higher voltage level.

FIGURE 12-2:

USING THE OPEN-DRAIN OUTPUT (USART SHOWN AS EXAMPLE)



REGISTER 12-2: ODCON1: PERIPHERAL OPEN-DRAIN CONTROL REGISTER 1

R/W-0	R/W-0	R/W-0	U-0	U-0	U-0	U-0	R/W-0
SSP10D	CCP2OD	CCP10D	_	—	—	_	SSP2OD
bit 7 bit 0							

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	SSP10D: MSSP1 Open-Drain Output Enable bit 1 = Open-drain capability is enabled 0 = Open-drain capability is disabled
bit 6	CCP2OD: ECCP2 Open-Drain Output Enable bit
	1 = Open-drain capability is enabled0 = Open-drain capability is disabled
bit 5	CCP10D: ECCP1 Open-Drain Output Enable bit
	1 = Open-drain capability is enabled0 = Open-drain capability is disabled
bit 4-1	Unimplemented: Read as '0'
bit 0	SSP2OD: MSSP2 Open-Drain Output Enable bit
	 1 = Open-drain capability is enabled 0 = Open-drain capability is disabled

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R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	
CCP100D ⁽¹⁾	CCP9OD ⁽¹⁾	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD	
bit 7							bit	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimpler	nented bit, rea	d as '0'		
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown	
=								
bit 7	CCP100D: C	•		able bit "				
	1 = Open-dra 0 = Open-dra							
bit 6	CCP9OD: CC			ole bit ⁽¹⁾				
	1 = Open-dra	-	-					
	0 = Open-dra	in capability is	disabled					
bit 5	CCP8OD: CC		•	ole bit				
	1 = Open-dra 0 = Open-dra							
bit 4	CCP7OD: CC	, ,		lo hit				
	1 = Open-dra	•	•					
	0 = Open-dra							
bit 3	CCP6OD: CC	P6 Open-Drai	n Output Enab	ole bit				
	1 = Open-dra							
	0 = Open-dra	in capability is	disabled					
bit 2	CCP5OD: CC		•	ole bit				
	1 = Open-dra 0 = Open-dra							
bit 1	•	, ,		ole hit				
	CCP4OD: CCP4 Open-Drain Output Enable bit 1 = Open-drain capability is enabled							
	0 = Open-dra	• •						
bit 0	CCP3OD: EC	CP3 Open-Dra	ain Output Ena	able bit				
	1 = Open-dra							
	0 = Open-dra	in capability is	disabled					

REGISTER 12-3: ODCON2: PERIPHERAL OPEN-DRAIN CONTROL REGISTER 2

Note 1: Not implemented on devices with 32-byte program memory (PIC18FX5K22).

REGISTER 12-4: ODCON3: PERIPHERAL OPEN-DRAIN CONTROL REGISTER 3

R/W-0	R/W-0	U-0	U-0	U-0	U-0	U-0	R/W-0			
U2OD	U10D	_	_	—	—	_	CTMUDS			
bit 7										
Legend:										
R = Reada	able bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'				
-n = Value at POR '1' = Bit is set				'0' = Bit is cleared x = Bit is unknown						
bit 7	it 7 U2OD: EUSART2 Open-Drain Output Enable bit									
	1 = Open-	drain capability is	enabled							
	0 = Open-	drain capability is	disabled							
bit 6	bit 6 U10D: EUSART1 Open-Drain Output Enable bit									
	1 = Open-drain capability is enabled									
	0 = Open-drain capability is disabled									
bit 5-1	Unimplemented: Read as '0'									
1.1.0	07141000									

bit 0 CTMUDS: CTMU Pulse Delay Enable bit

- 1 = Pulse delay input for CTMU is enabled on pin, RF1
- 0 = Pulse delay input for CTMU is disabled on pin, RF1

12.1.4 ANALOG AND DIGITAL PORTS

Many of the ports multiplex analog and digital functionality, providing a lot of flexibility for hardware designers. PIC18F87K22 family devices can make any analog pin analog or digital, depending on an application's needs. The ports' analog/digital functionality is controlled by registers: ANCON0, ANCON1 and ANCON2. Setting these registers makes the corresponding pins analog and clearing the registers makes the ports digital. For details on these registers, see Section 23.0 "12-Bit Analog-to-Digital Converter (A/D) Module".

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12.2 PORTA, TRISA and LATA Registers

PORTA is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISA and LATA.

RA4/T0CKI is a Schmitt Trigger input. All other PORTA pins have TTL input levels and full CMOS output drivers.

RA5 and RA<3:0> are multiplexed with analog inputs for the A/D Converter.

The operation of the analog inputs as A/D Converter inputs is selected by clearing or setting the ANSEL control bits in the ANCON1 register. The corresponding TRISA bits control the direction of these pins, even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs.

Note:	RA5 and RA<3:0> are configured as
	analog inputs on any Reset and are read
	as '0'. RA4 is configured as a digital input.

OSC2/CLKO/RA6 and OSC1/CLKI/RA7 normally serve as the external circuit connections for the external (primary) oscillator circuit (HS Oscillator modes), or the external clock input and output (EC Oscillator modes). In these cases, RA6 and RA7 are not available as digital I/O and their corresponding TRIS and LAT bits are read as '0'. When the device is configured to use HF-INTOSC, MF-INTOSC or LF-INTOSC as the default oscillator mode, RA6 and RA7 are automatically configured as digital I/O; the oscillator and clock in/clock out functions are disabled.

RA5 has additional functionality for Timer1 and Timer3. It can be configured as the Timer1 clock input or the Timer3 external clock gate input.

CLRF	PORTA	; Initialize PORTA by
		; clearing output latches
CLRF	LATA	; Alternate method to
		; clear output data latches
BANKSEL	ANCON1	; Select bank with ANCON1 register
MOVLW	00h	; Configure A/D
MOVWF	ANCON1	; for digital inputs
BANKSEL	TRISA	; Select bank with TRISA register
MOVLW	0BFh	; Value used to initialize
		; data direction
MOVWF	TRISA	; Set RA<7, 5:0> as inputs,
		; RA<6> as output

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TADLE 12-1.	FUNIAF				
Pin Name	Function	TRIS Setting	I/O	I/O Туре	Description
RA0/AN0/ULPWU	RA0	0	0	DIG	LATA<0> data output; not affected by analog input.
		1	I	TTL	PORTA<0> data input; disabled when analog input is enabled.
	AN0	1	I	ANA	A/D Input Channel 0. Default input configuration on POR; does not affect digital output.
	ULPWU	1	I	ANA	Ultra Low-Power Wake-up input.
RA1/AN1	RA1	0	0	DIG	LATA<1> data output; not affected by analog input.
		1	I	TTL	PORTA<1> data input; disabled when analog input is enabled.
	AN1	1	I	ANA	A/D Input Channel 1. Default input configuration on POR; does not affect digital output.
RA2/AN2/VREF-	RA2	0	0	DIG	LATA<2> data output; not affected by analog input.
		1	Ι	TTL	PORTA<2> data input; disabled when analog functions are enabled.
	AN2	1	Ι	ANA	A/D Input Channel 2. Default input configuration on POR.
	VREF-	1	-	ANA	A/D and comparator low reference voltage input.
RA3/AN3/VREF+	RA3	0	0	DIG	LATA<3> data output; not affected by analog input.
		1	Ι	TTL	PORTA<3> data input; disabled when analog input is enabled.
	AN3	1	Ι	ANA	A/D Input Channel 3. Default input configuration on POR.
	VREF+	1	-	ANA	A/D and comparator high reference voltage input.
RA4/T0CKI	RA4	0	0	DIG	LATA<4> data output.
		1	-	ST	PORTA<4> data input. Default configuration on POR.
	TOCKI	x	Ι	ST	Timer0 clock input.
RA5/AN4/T1CKI/	RA5	0	0	DIG	LATA<5> data output; not affected by analog input.
T3G/HLVDIN		1	-	TTL	PORTA<5> data input; disabled when analog input is enabled.
	AN4	1	Ι	ANA	A/D Input Channel 4. Default configuration on POR.
	T1CKI	x	-	ST	Timer1 clock input.
	T3G	х	-	ST	Timer3 external clock gate input.
	HLVDIN	1	Ι	ANA	High/Low-Voltage Detect (HLVD) external trip point input.
OSC2/CLKO/RA6	OSC2	x	0	ANA	Main oscillator feedback output connection (HS, XT and LP modes).
	CLKO	х	0	DIG	System cycle clock output (Fosc/4, EC and INTOSC modes).
	RA6	0	0	DIG	LATA<6> data output; disabled when FOSC2 Configuration bit is set.
		1	Ι	TTL	PORTA<6> data input; disabled when FOSC2 Configuration bit is set.
OSC1/CLKI/RA7	OSC1	х	I	ANA	Main oscillator input connection (HS, XT and LP modes).
	CLKI	х	-	ANA	Main external clock source input (EC modes).
	RA7	0	0	DIG	LATA<7> data output; disabled when FOSC2 Configuration bit is set.
		1	I	TTL	PORTA<7> data input; disabled when FOSC2 Configuration bit is set.

TABLE 12-1: PORTA FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 12-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTA	RA7 ⁽¹⁾	RA6 ⁽¹⁾	RA5	RA4	RA3	RA2	RA1	RA0
LATA	LATA7 ⁽¹⁾	LATA6 ⁽¹⁾	LATA5	LATA4	LATA3	LATA2	LATA1	LATA0
TRISA	TRISA7 ⁽¹⁾	TRISA6 ⁽¹⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTA.

Note 1: These bits are enabled depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read as 'x'.

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12.3 PORTB, TRISB and LATB Registers

PORTB is an eight-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISB and LATB. All pins on PORTB are digital only.

EXAMPLE 12-2:	INITIALIZING PORTB
---------------	--------------------

CLRF PORTE	; Initialize PORTB by ; clearing output ; data latches
CLRF LATB	; Alternate method
-	; to clear output
	; data latches
MOVLW 0CFh	; Value used to
	; initialize data
	; direction
MOVWF TRISE	; Set RB<3:0> as inputs
	; RB<5:4> as outputs
	; RB<7:6> as inputs

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by clearing bit, RBPU (INTCON2<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.

Four of the PORTB pins (RB<7:4>) have an interrupt-on-change feature. Only pins configured as inputs can cause this interrupt to occur. Any RB<7:4> pin configured as an output will be excluded from the interrupt-on-change comparison.

Comparisons with the input pins (of RB<7:4>) are made with the old value latched on the last read of PORTB. The "mismatch" outputs of RB<7:4> are ORed together to generate the RB Port Change Interrupt with Flag bit, RBIF (INTCON<0>).

This interrupt can wake the device from power-managed modes. To clear the interrupt in the Interrupt Service Routine:

- a) Any read or write of PORTB (except with the MOVFF (ANY), PORTB instruction). This will end the mismatch condition.
- b) Wait one instruction cycle (such as executing a NOP instruction).
- c) Clear flag bit, RBIF.

A mismatch condition will continue to set flag bit, RBIF. Reading PORTB will end the mismatch condition and allow flag bit, RBIF, to be cleared after one TCY delay.

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

The RB<3:2> pins are multiplexed as CTMU edge inputs. RB5 has an additional function for Timer3 and Timer1. It can be configured for Timer3 clock input or Timer1 external clock gate input.

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RB0/INT0/FLT0	RB0	0	0	DIG	LATB<0> data output.
		1	I	TTL	PORTB<0> data input; weak pull-up when RBPU bit is cleared.
	INT0	1	I	ST	External Interrupt 0 input.
	FLT0	x	Ι	ST	Enhanced PWM Fault input for ECCPx.
RB1/INT1	RB1	0	0	DIG	LATB<1> data output.
		1	Ι	TTL	PORTB<1> data input; weak pull-up when RBPU bit is cleared.
	INT1	1	Ι	ST	External Interrupt 1 input.
RB2/INT2/CTED1	RB2	0	0	DIG	LATB<2> data output.
		1	Ι	TTL	PORTB<2> data input; weak pull-up when RBPU bit is cleared.
	INT2	1	Ι	ST	External Interrupt 2 input.
	CTED1	x	Ι	ST	CTMU Edge 1 input.

TABLE 12-3: PORTB FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Extended Microcontroller mode.

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RB3/INT3/CTED2/	RB3	0	0	DIG	LATB<3> data output.
ECCP2/P2A		1	I	TTL	PORTB<3> data input; weak pull-up when RBPU bit is cleared.
	INT3	1	I	ST	External Interrupt 3 input.
	CTED2	x	Ι	ST	CTMU Edge 2 input.
	ECCP2 ⁽¹⁾	0	0	DIG	ECCP2 compare output and ECCP2 PWM output. Takes priority over port data.
		1	I	ST	ECCP2 capture input.
	P2A	0	0	DIG	ECCP2 Enhanced PWM output, Channel A. May be configured for tri-state during Enhanced PWM shutdown events. Takes priority over port data.
RB4/KBI0	RB4	0	0	DIG	LATB<4> data output.
		1	Ι	TTL	PORTB<4> data input; weak pull-up when RBPU bit is cleared.
	KBI0	1	Ι	TTL	Interrupt-on-pin change.
RB5/KBI1/T3CKI/	RB5	0	0	DIG	LATB<5> data output.
T1G		1	I	TTL	PORTB<5> data input; weak pull-up when $\overline{\text{RBPU}}$ bit is cleared.
	KBI1	1	I	TTL	Interrupt-on-pin change.
	T3CKI	x	Ι	ST	Timer3 clock input.
	T1G	x	Ι	ST	Timer1 external clock gate input.
RB6/KBI2/PGC	RB6	0	0	DIG	LATB<6> data output.
		1	I	TTL	PORTB<6> data input; weak pull-up when $\overline{\text{RBPU}}$ bit is cleared.
	KBI2	1	I	TTL	Interrupt-on-pin change.
	PGC	x	Ι	ST	Serial execution (ICSP™) clock input for ICSP and ICD operations.
RB7/KBI3/PGD	RB7	0	0	DIG	LATB<7> data output.
		1	Ι	TTL	PORTB<7> data input; weak pull-up when RBPU bit is cleared.
	KBI3	1	Ι	TTL	Interrupt-on-pin change.
	PGD	x	0	DIG	Serial execution data output for ICSP and ICD operations.
		x	Ι	ST	Serial execution data input for ICSP and ICD operations.

TABLE 12-3:	PORTB FUNCTIONS	(CONTINUED)
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Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Extended Microcontroller mode.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0
LATB	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
INTCON2	RBPU	INTEDG0	INTEDG1	INTEDG2	INTEDG3	TMR0IP	INT3IP	RBIP
INTCON3	INT2IP	INT1IP	INT3IE	INT2IE	INT1IE	INT3IF	INT2IF	INT1IF
ODCON1	SSP10D	CCP2OD	CCP10D		_	_		SSP2OD

Legend: Shaded cells are not used by PORTB.

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12.4 PORTC, TRISC and LATC Registers

PORTC is an eight-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISC and LATC. Only PORTC pins, RC2 through RC7, are digital only pins.

PORTC is multiplexed with ECCP, MSSP and EUSART peripheral functions (Table 12-5). The pins have Schmitt Trigger input buffers. The pins for ECCP, SPI and EUSART are also configurable for open-drain output whenever these functions are active. Open-drain configuration is selected by setting the SPIOD, CCPxOD and U1OD control bits in the registers, ODCON1 and ODCON3.

RC1 is normally configured as the default peripheral pin for the ECCP2 module. The assignment of ECCP2 is controlled by Configuration bit, CCP2MX (default state, CCP2MX = 1).

When enabling peripheral functions, use care in defining TRIS bits for each PORTC pin. Some peripherals can override the TRIS bit to make a pin an output or input. Consult the corresponding peripheral section for the correct TRIS bit settings.

Note:	These pins are configured as digital inputs
	on any device Reset.

The contents of the TRISC register are affected by peripheral overrides. Reading TRISC always returns the current contents, even though a peripheral device may be overriding one or more of the pins.

EXAMPLE 12-3:	INITIALIZING PORTC

CLRF	PORTC	; Initialize PORTC by ; clearing output
CLRF	LATC	; data latches ; Alternate method ; to clear output
MOVLW	0CFh	; data latches ; Value used to ; initialize data
MOVWF	TRISC	<pre>; direction ; Set RC<3:0> as inputs ; RC<5:4> as outputs</pre>
MOVWF	TRISC	; Set RC<3:0> as inputs

Pin Name	Function	TRIS Setting	I/O	I∕O Type	Description
RC0/SOSCO/	RC0	0	0	DIG	LATC<0> data output.
SCLKI/		1	Ι	ST	PORTC<0> data input.
	SOSCO	1	Ι	ST	SOSC oscillator output.
	SCLKI	1	Ι	ST	Digital clock input; enabled when SOSC oscillator is disabled.
RC1/SOSCI/	RC1	0	0	DIG	LATC<1> data output.
ECCP2/P2A		1	Ι	ST	PORTC<1> data input.
	SOSCI	х	Ι	ANA	SOSC oscillator input.
	ECCP2 ⁽¹⁾ 0 O DIG ECC		DIG	ECCP2 compare output and ECCP2 PWM output; takes priority over port data	
		1	Ι	ST	ECCP2 capture input.
	P2A	0	0	DIG	ECCP2 Enhanced PWM output, Channel A. May be configured for tri-state during Enhanced PWM shutdown events; takes priority over port data.
RC2/ECCP1/	RC2	0	0	DIG	LATC<2> data output.
P1A		1	Ι	ST	PORTC<2> data input.
	ECCP1	0	0	DIG	ECCP1 compare output and ECCP1 PWM output; takes priority over port data.
		1	Ι	ST	ECCP1 capture input.
	P1A	0	0	DIG	ECCP1 Enhanced PWM output, Channel A. May be configured for tri-state during Enhanced PWM shutdown events; takes priority over port data.

TABLE 12-5:PORTC FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, $I^2C = I^2C^{TM}$ /SMBus Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description
RC3/SCK1/	RC3	0	0	DIG	LATC<3> data output.
SCL1		1	Ι	ST	PORTC<3> data input.
	SCK1	0	0	DIG	SPI clock output (MSSP module); takes priority over port data.
		1	Ι	ST	SPI clock input (MSSP module).
	SCL1	0	0	DIG	I ² C clock output (MSSP module); takes priority over port data.
		1	Ι	l ² C	I ² C clock input (MSSP module); input type depends on module setting.
RC4/SDI1/	RC4	0	0	DIG	LATC<4> data output.
SDA1		1	Ι	ST	PORTC<4> data input.
	SDI1		Ι	ST	SPI data input (MSSP module).
	SDA1	1	0	DIG	I ² C data output (MSSP module); takes priority over port data.
		1	Ι	l ² C	I ² C data input (MSSP module); input type depends on module setting.
RC5/SDO1 RC5 0 O DIG LATC<5>		DIG	LATC<5> data output.		
		1	Ι	ST	PORTC<5> data input.
	SDO1	0	0	DIG	SPI data output (MSSP module).
RC6/TX1/CK1	RC6	0	0	DIG	LATC<6> data output.
		1		ST	PORTC<6> data input.
	TX1	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.
	CK1	1	0	DIG	Synchronous serial data input (EUSART module); user must configure as an input.
		1	Ι	ST	Synchronous serial clock input (EUSART module).
RC7/RX1/DT1	RC7	0	0	DIG	LATC<7> data output.
		1	Ι	ST	PORTC<7> data input.
	RX1	1	Ι	ST	Asynchronous serial receive data input (EUSART module).
	DT1	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.
		1	Ι	ST	Synchronous serial data input (EUSART module); user must configure as an input.

TABLE 12-5: PORTC FUNCTIONS (CONTINUED)

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, TTL = TTL Buffer Input, $I^2C = I^2C^{TM}$ /SMBus Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Default assignment for ECCP2 when the CCP2MX Configuration bit is set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0
LATC	LATC7	LATBC6	LATC5	LATCB4	LATC3	LATC2	LATC1	LATC0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
ODCON1	SSP10D	CCP2OD	CCP10D	_	_	_	_	SSP2OD
ODCON3	U2OD	U10D	_	_	_	_	_	CTMUDS

Legend: Shaded cells are not used by PORTC.

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12.5 PORTD, TRISD and LATD Registers

PORTD is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISD and LATD.

All pins on PORTD are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note:	These pins are configured as digital inputs
	on any device Reset.

Each of the PORTD pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by setting bit, RDPU (PADCFG1<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on all device Resets.

On 80-pin devices, PORTD is multiplexed with the system bus as part of the external memory interface. The I/O port and other functions are only available when the interface is disabled by setting the EBDIS bit (MEMCON<7>). When the interface is enabled, PORTD is the low-order byte of the multiplexed address/data bus (AD<7:0>). The TRISD bits are also overridden.

PORTD can also be configured as an 8-bit wide microprocessor port (Parallel Slave Port) by setting control bit, PSPMODE (TRISE<4>). In this mode, the input buffers are TTL. For additional information, see **Section 12.11 "Parallel Slave Port"**.

The PORTD also has the I²C and SPI functionality on RD4, RD5 and RD6. The pins for SPI are also configurable for open-drain output. Open-drain configuration is selected by setting bit, SSP2OD (ODCON1<0>).

RD0 has a CTMU functionality. RD1 has the functionality for the Timer5 clock input and Timer7 external clock gate input.

INITIALIZING PORTD

EXAMPLE 12-4:

CLRF	PORTD	; Initialize PORTD by
		; clearing output
		; data latches
CLRF	LATD	; Alternate method
		; to clear output
		; data latches
MOVLW	OCFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISD	; Set RD<3:0> as inputs
		; RD<5:4> as outputs
		; RD<7:6> as inputs

Function	TRIS Setting	I/O	I∕O Type	Description						
RD0	0	0	DIG	LATD<0> data output.						
	1	Ι	ST	PORTD<0> data input.						
PSP0 ⁽¹⁾	х	I/O	TTL	Parallel Slave Port data.						
AD0 ⁽²⁾	x	I/O	TTL	External Memory Address/Data 0.						
CTPLS	х	0	DIG	CTMU pulse generator output.						
RD1	0	0	DIG	LATD<1> data output.						
	1	Ι	ST	PORTD<1> data input.						
PSP1 ⁽¹⁾	х	I/O	TTL	Parallel Slave Port data.						
AD1 ⁽²⁾	х	I/O	TTL	External Memory Address/Data 1.						
T5CKI	x	Ι	ST	Timer5 clock input.						
T7G	x	Ι	ST	Timer7 external clock gate input.						
RD2	0	0	DIG	LATD<2> data output.						
	1	Ι	ST	PORTD<2> data input.						
PSP2 ⁽¹⁾	x	I/O	TTL	Parallel Slave Port data.						
AD2 ⁽²⁾	x	I/O	TTL	External Memory Address/Data 2.						
	RD0 PSP0 ⁽¹⁾ AD0 ⁽²⁾ CTPLS RD1 PSP1 ⁽¹⁾ AD1 ⁽²⁾ T5CKI T7G RD2 PSP2 ⁽¹⁾ AD2 ⁽²⁾	Function Setting RD0 0 PSP0 ⁽¹⁾ x AD0 ⁽²⁾ x CTPLS x RD1 0 PSP1 ⁽¹⁾ x AD1 ⁽²⁾ x TSCKI x T7G x RD2 0 PSP2 ⁽¹⁾ x AD2 ⁽²⁾ x	Function Setting I/O RD0 0 0 RD0 1 1 PSP0 ⁽¹⁾ x I/O AD0 ⁽²⁾ x I/O CTPLS x O RD1 0 O RD1 1 I PSP1 ⁽¹⁾ x I/O AD1 ⁽²⁾ x I/O FSCKI x I RD2 0 O I I I PSP2 ⁽¹⁾ x I/O AD2 ⁽²⁾ x I/O	Function Setting I/O Type RD0 0 OIG DIG 1 I ST PSP0 ⁽¹⁾ x I/O TTL AD0 ⁽²⁾ x I/O TTL CTPLS x O DIG RD1 0 O DIG RD1 0 O DIG RD1 0 O DIG RD1 1 ST ST PSP1 ⁽¹⁾ x I/O TTL AD1 ⁽²⁾ x I/O TTL AD1 ⁽²⁾ x I/O TTL T5CKI x I ST RD2 0 O DIG 1 I ST ST RD2 0 O DIG 1 I ST ST PSP2 ⁽¹⁾ x I/O TTL						

TABLE 12-7: PORTD FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, I²C = I²C[™]/SMBus Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: The Parallel Slave Port (PSP) is available only in Microcontroller mode.

2: This feature is available only on PIC18F8XK22 devices.

IABLE 12-7:	PORTD FUNCTIONS (CONTINUED)				
Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RD3/PSP3/AD3	RD3	0	0	DIG	LATD<3> data output.
		1	I	ST	PORTD<3> data input.
	PSP3 ⁽¹⁾	x	I/O	TTL	Parallel Slave Port data.
	AD3 ⁽²⁾	x	I/O	TTL	External Memory Address/Data 3.
RD4/PSP4/	RD4	0	0	DIG	LATD<4> data output.
AD4/SDO2		1	Ι	ST	PORTD<4> data input.
	PSP4 ⁽¹⁾	x	I/O	TTL	Parallel Slave Port data.
-	AD4 ⁽²⁾	x	I/O	TTL	External Memory Address/Data 4.
	SDO2	0	Р	DOG	SPI data output (MSSP module).
RD5/PSP5/	RD5	0	0	DIG	LATD<5> data output.
AD5/SDI2/		1	I	ST	PORTD<5> data input.
SDA2	PSP5 ⁽¹⁾	x	I/O	TTL	Parallel Slave Port data.
	AD5 ⁽²⁾	x	I/O	TTL	External Memory Address/Data 5.
	SDI2	1	-	ST	SPI data input (MSSP module).
	SDA2	0	0	l ² C	I ² C data input (MSSP module). Input type depends on module setting
RD6/PSP6/	RD6	0	0	DIG	LATD<6> data output.
AD6/SCK2/ SCL2		1	Ι	ST	PORTD<6> data input.
SULZ	PSP6 ⁽¹⁾	x	I/O	TTL	Parallel Slave Port data.
	AD6 ⁽²⁾	x	I/O	TTL	External Memory Address/Data 6.
	SCK2	0	0	DIG	SPI clock output (MSSP module); takes priority over port data.
		1	-	ST	SPI clock input (MSSP module).
	SCL2	0	0	DIG	I ² C clock output (MSSP module); takes priority over port data.
		1	I	l ² C	I ² C clock input (MSSP module). Input type depends on module setting.
RD7/ <u>PSP</u> 7/	RD7	0	0	DIG	LATD<7> data output.
AD7/SS2		1	Ι	ST	PORTD<7> data input.
	PSP7 ⁽¹⁾	x	I/O	TTL	Parallel Slave Port data.
	AD7 ⁽²⁾	x	I/O	TTL	External Memory Address/Data 7.
	SS2	1	I	TTL	Slave select input for MSSP module.

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, $I^2C = I^2C^{TM}/SMBus$ Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: The Parallel Slave Port (PSP) is available only in Microcontroller mode.

2: This feature is available only on PIC18F8XK22 devices.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0
LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0
PADCFG1	RDPU	REPU	RJPU ⁽¹⁾		_	RTSECSEL1	RESECSEL0	
ODCON1	SSP10D	CCP2OD	CCP10D		_	_	—	SSP2OD

Legend: Shaded cells are not used by PORTD.

Note 1: Unimplemented on PIC18F6XK22 devices, read as '0'.

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12.6 PORTE, TRISE and LATE Registers

PORTE is an eight-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISE and LATE.

All pins on PORTE are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output. The RE7 pin is also configurable for open-drain output when ECCP2 is active on this pin. Open-drain configuration is selected by setting the CCP2OD control bit (ODCON1<6>)

Note:	These pins are configured as digital inputs
	on any device Reset.

Each of the PORTE pins has a weak internal pull-up. A single control bit can turn off all the pull-ups. This is performed by setting bit, REPU (PADCFG1<6>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

PORTE is also multiplexed with Enhanced PWM Outputs, B and C for ECCP1 and ECCP3, for Outputs, B, C and D for ECCP2. For all devices, their default assignments are on PORTE<6:0>.

On 80-pin devices, the multiplexing for the outputs of ECCP1 and ECCP3 is controlled by the ECCPMX Configuration bit. Clearing this bit re-assigns the P1B/P1C and P3B/P3C outputs to PORTH.

For devices operating in Microcontroller mode, the RE7 pin can be configured as the alternate peripheral pin for the ECCP2 module and Enhanced PWM Output 2A. This is done by clearing the CCP2MX Configuration bit. PORTE is also multiplexed with the Parallel Slave Port address lines. RE1 and RE0 are multiplexed with the control signals, WR and RD.

RE3 can also be configured as the Reference Clock Output (REFO) from the system clock. For further details, see **Section 3.7 "Reference Clock Output"**.

CLRF	PORTE	; Initialize PORTE by
		; clearing output
		; data latches
CLRF	LATE	; Alternate method
		; to clear output
		; data latches
MOVLW	03h	; Value used to
		; initialize data
		; direction
MOVWF	TRISE	; Set RE<1:0> as inputs
		; RE<7:2> as outputs

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description	
RE0/RD/P2D	RE0	0	0	DIG	LATE<0> data output.	
AD8		1	I	ST	PORTE<0> data input.	
	RD	x	0	DIG	Parallel Slave Port read strobe pin.	
		x	I	TTL	Parallel Slave Port read pin.	
	P2D	0	0	—	ECCP2 PWM Output D. May be configured for tri-state during Enhanced PWM shutdown events.	
	AD8 ⁽²⁾	x	0	DIG	External memory interface, Data Bit 8 output.	
		x	Ι	TTL	External memory interface, Data Bit 8 input.	
RE1/P2C/WR/	RE1	0	0	DIG	LATE<1> data output.	
AD9		1	I	ST	PORTE<1> data input.	
	P2C	0	0	—	ECCP2 PWM Output C. May be configured for tri-state during Enhanced PWM shutdown events.	
	WR	x	0	DIG	Parallel Slave Port write strobe pin.	
		x	Ι	TTL	Parallel Slave Port write pin.	
	AD9 ⁽²⁾	x	0	DIG	External memory interface, Data Bit 9 output.	
		x	Ι	TTL	External memory interface, Data Bit 9 input.	
Legend: O = Output I = Input ANA = Analog Signal DIG = Digital Output ST = Schmitt Trigger Buffer Input						

TABLE 12-9: PORTE FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Microcontroller mode.

2: This feature is only available on PIC18F8XKXX devices.

TABLE 12-9:	PORTE FUNCTIONS (CONTINUED)					
Pin Name	Function	TRIS Setting	I/O	I/O Type	Description	
RE2/CS/P2B/	RE2	0	0	DIG	LATE<2> data output.	
CCP10/AD10		1	-	ST	PORTE<2> data input.	
	CS	x	I	TTL	Parallel Slave Port chip select.	
	P2B	0	0	—	ECCP2 PWM Output B. May be configured for tri-state during Enhanced PWM shutdown events.	
	CCP10	1	I/O	ST	Capture 10 input/Compare 10 output/PWM10 output.	
	AD10 ⁽²⁾	x	0	DIG	External memory interface, Address/Data Bit 10 output.	
		x	Ι	TTL	External memory interface, Data Bit 10 input.	
RE3/P3C/	RE3	0	0	DIG	LATE<3> data output.	
CCP9/REFO/		1	I	ST	PORTE<3> data input.	
AD11	P3C	0	0	—	ECCP3 PWM Output C. May be configured for tri-state during Enhanced PWM shutdown events.	
	CCP9	0	0	DIG	CCP9 Compare/PWM output; takes priority over port data.	
-		1	Ι	ST	CCP9 capture input.	
	REFO	x	0	DIG	Reference output clock.	
	AD11 ⁽²⁾	x	0	DIG	External memory interface, Address/Data Bit 11 output.	
		x	I	TTL	External memory interface, Data Bit 11 input.	
RE4/P3B/ CCP8/AD12	RE4	0	0	DIG	LATE<4> data output.	
		1	I	ST	PORTE<4> data input.	
	P3B	0	0	—	ECCP3 PWM Output B. May be configured for tri-state during Enhanced PWM shutdown events.	
	CCP8	0	0	DIG	CCP8 compare/PWM output; takes priority over port data.	
		1	I	ST	CCP8 capture input.	
	AD12 ⁽²⁾	x	0	DIG	External memory interface, Address/Data Bit 12 output.	
		x	I	TTL	External memory interface, Data Bit 12 input.	
RE5/P1C/	RE5	0	0	DIG	LATE<5> data output.	
CCP7/AD13		1	I	ST	PORTE<5> data input.	
	P1C	0	0	—	ECCP1 PWM Output C. May be configured for tri-state during Enhanced PWM shutdown events.	
	CCP7	0	0	DIG	CCP7 compare/PWM output; takes priority over port data.	
		1	I	ST	CCP7 capture input.	
	AD13 ⁽²⁾	x	0	DIG	External memory interface, Address/Data Bit 13 output.	
		x	I	TTL	External memory interface, Data Bit 13 input.	
RE6/P1B/	RE6	0	0	DIG	LATE<6> data output.	
CCP6/AD14		1	I	ST	PORTE<6> data input.	
	P1B	0	0	—	ECCP1 PWM Output B. May be configured for tri-state during Enhanced PWM shutdown events.	
	CCP6	0	0	DIG	CCP6 compare/PWM output; takes priority over port data.	
		1	I	ST	CCP9 capture input.	
	AD14 ⁽²⁾	x	0	DIG	External memory interface, Address/Data Bit 14 output.	
		x	Ι	TTL	External memory interface, Data Bit 14 input.	
Legend: O = 0	Outout I = Ino	ut ANA = A	halog 9		G = Digital Output, ST = Schmitt Trigger Buffer Input,	

TABLE 12-9: PORTE FUNCTIONS (CONTINUED)

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Microcontroller mode.

2: This feature is only available on PIC18F8XKXX devices.

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TABLE 12-9:	PORTE FUNCTIONS (CONTINUED)
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Pin Name	Function	TRIS Setting	I/O	l/O Type	Description
RE7/ECCP2/	RE7	0	0	DIG	LATE<7> data output.
P2A/AD15		1	Ι	ST	PORTE<7> data input.
	ECCP2 ⁽¹⁾	0	0	DIG	ECCP2 compare/PWM output; takes priority over port data.
		1	Ι	ST	ECCP2 capture input.
	P2A	0	0		ECCP2 PWM Output A. May be configured for tri-state during Enhanced PWM shutdown event.
	AD15 ⁽²⁾	x	0	DIG	External memory interface, Address/Data Bit 15 output.
		x	I	TTL	External memory interface, Data Bit 15 input.

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Note 1: Alternate assignment for ECCP2 when the CCP2MX Configuration bit is cleared and in Microcontroller mode.

2: This feature is only available on PIC18F8XKXX devices.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTE	RE7	RE6	RE5	RE4	RE3	RE2	RE1	RE0
LATE	LATE7	LATE6	LATE5	LATE4	LATE3	LATE2	LATE1	LATE0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0
PADCFG1	RDPU	REPU	RJPU ⁽²⁾	_	_	RTSECSEL1	RTSECSEL0	—
ODCON1	SSP10D	CCP2OD	CCP10D	_	_	—	—	SSP2OD
ODCON2	CCP100D ⁽¹⁾	CCP9OD ⁽¹⁾	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD

Legend: Shaded cells are not used by PORTE.

Note 1: Unimplemented on PIC18FX5K22 devices, read as '0'.

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

12.7 PORTF, LATF and TRISF Registers

PORTF is a 7-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISF and LATF. All pins on PORTF are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Pins, RF1 through RF6, may be used as comparator inputs or outputs by setting the appropriate bits in the CMCON register. To use RF<7:1> as digital inputs, it is also necessary to turn off the comparators.

- **Note 1:** On device Resets, pins, RF<7:1>, are configured as analog inputs and are read as '0'.
 - To configure PORTF as a digital I/O, turn off the comparators and clear ANCON1 and ANCON2 to digital.

EXAMPLE 12-6: INITIALIZING PORTF

CLRF	PORTF	; Initialize PORTF by ; clearing output ; data latches
CLRF	LATF	
BANKSEL	ANCON1	
	1Fh	_
MOVWF	ANCON1	
MOVLW	OFh	; Make AN8, AN9, AN10 and AN11 digital
MOVWF	ANCON	; Set PORTF as digital I/O
BANKSEL	TRISF	; Select bank with TRISF register
MOVLW	0CEh	; Value used to
		; initialize data
		; direction
MOVWF	TRISF	; Set RF3:RF1 as inputs
		; RF5:RF4 as outputs
		; RF7:RF6 as inputs

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description				
RF1/AN6/C2OUT/	RF1	0	0	DIG	LATF<1> data output; not affected by analog input.				
CTDIN		1	I	ST	PORTF<1> data input; disabled when analog input is enabled.				
	AN6	1	Ι	ANA	A A/D Input Channel 6. Default configuration on POR.				
	C2OUT	0	0	DIG	Comparator 2 output; takes priority over port data.				
	CTDIN	1	Ι	ST	CTMU pulse delay input.				
RF2/AN7/C1OUT	RF2	0	0	DIG	LATF<2> data output; not affected by analog input.				
		1	Ι	ST	PORTF<2> data input; disabled when analog input is enabled.				
	AN7	1	Ι	ANA	A/D Input Channel 7. Default configuration on POR.				
	C10UT	0	0	DIG	Comparator 1 output; takes priority over port data.				
RF3/AN8/C2INB/	RF3	0	0	DIG	LATF<3> data output; not affected by analog input.				
CTMUI		1	Ι	ST	PORTF<3> data input; disabled when analog input is enabled.				
	AN8	1	I	ANA	A/D Input Channel 8 and Comparator C2+ input. Default input configuration on POR; not affected by analog output.				
	C2INB	1	Ι	ANA	Comparator 2 Input B.				
	CTMUI	х	0	—	CTMU pulse generator charger for the C2INB comparator input.				
RF4/AN9/C2INA	RF4	0	0	DIG	LATF<4> data output; not affected by analog input.				
		1	Ι	ST	PORTF<4> data input; disabled when analog input is enabled.				
	AN9	1	Ι	ANA	A/D Input Channel 9 and Comparator C2- input. Default input configuration on POR; does not affect digital output.				
	C2INA	1	Ι	ANA	Comparator 2 Input A.				
RF5/AN10/CVREF/ C1INB	RF5	0	0	DIG	LATF<5> data output; not affected by analog input. Disabled when CVREF output is enabled.				
		1	I	ST	PORTF<5> data input; disabled when analog input is enabled. Disabled when CVREF output is enabled.				
	AN10	1	I	ANA	A/D Input Channel 10 and Comparator C1+ input. Default input configuration on POR.				
	CVREF	х	0	ANA	Comparator voltage reference output. Enabling this feature disables digital I/O.				
	C1INB	1	Ι	ANA	Comparator 1 Input B.				

TABLE 12-11: PORTF FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

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TABLE 12-11:	PORTF	FUNCTI	ONS	(CONTIN	VUED)

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RF6/AN11/C1INA	RF6	0	0	DIG	LATF<6> data output; not affected by analog input.
		1	I	ST	PORTF<6> data input; disabled when analog input is enabled.
	AN11	1	I	ANA	A/D Input Channel 11 and Comparator C1- input. Default input configuration on POR; does not affect digital output.
	C1INA	1	I	ANA	Comparator 1 Input A.
RF7/AN5/SS1	RF7	0	0	DIG	LATF<7> data output; not affected by analog input.
		1	I	ST	PORTF<7> data input; disabled when analog input is enabled.
	AN5	1	Ι	ANA	A/D Input Channel 5. Default configuration on POR.
	SS1	1	I	TTL	Slave select input for MSSP module.

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

TTL = TTL Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 12-12: SUMMARY OF REGISTERS ASSOCIATED WITH PORTF

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	—
LATF	LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	—
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTF.

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12.8 PORTG, TRISG and LATG Registers

PORTG is a 5-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISG and LATG.

PORTG is multiplexed with the AUSART and CCP, ECCP, Analog, Comparator, RTCC and Timer input functions (Table 12-13). When operating as I/O, all PORTG pins have Schmitt Trigger input buffers. The open-drain functionality for the CCPx and UART can be configured using ODCONx.

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTG pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. The user should refer to the corresponding peripheral section for the correct TRIS bit settings. The pin override value is not loaded into the TRIS register. This allows read-modify-write of the TRIS register without concern due to peripheral overrides.

EXAMPLE 12-7: INITIALIZING PORTG

CLRF	PORTG		Initialize PORTG by clearing output
		;	data latches
BCF	CM1CON, CON	;	disable
		;	comparator 1
CLRF	LATG	;	Alternate method
		;	to clear output
		;	data latches
BANKSEL	ANCON2	;	Select bank with ACON2 register
MOVLW	OFOh	;	make AN16 to AN19
		;	digital
MOVWF	ANCON2		
BANKSEL	TRISG	;	Select bank with TRISG register
MOVLW	04h	;	Value used to
		;	initialize data
		;	direction
MOVWF	TRISG	;	Set RG1:RG0 as
		;	outputs
		;	RG2 as input
		;	RG4:RG3 as inputs

TABLE 12-13: PORTG FUNCTIONS

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RG0/ECCP3/	RG0	0	0	DIG	LATG<0> data output.
P3A		1	Ι	ST	PORTG<0> data input.
	ECCP3	0	0	DIG	ECCP3 compare output and ECCP3 PWM output; takes priority over port data.
		1	I	ST	ECCP3 capture input.
	P3A	0	0	_	ECCP3 PWM Output A. May be configured for tri-state during Enhanced PWM shutdown events.
RG1/TX2/CK2/	RG1	0	0	DIG	LATG<1> data output.
AN19/C3OUT		1	I	ST	PORTG<1> data input.
	TX2	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.
	CK2	1	0	DIG	Synchronous serial data input (EUSART module); user must configure as an input.
		1	Ι	ST	Synchronous serial clock input (EUSART module).
	AN19	1	Ι	ANA	A/D Input Channel 19. Default input configuration on POR. Does not affect digital output.
	C3OUT	x	0	DIG	Comparator 3 output.

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

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TABLE 12-13: PORTG FUNCTIONS (CONTINUED)

Pin Name	Function	TRIS Setting	I/O	l/O Type	Description				
RG2/RX2/DT2/	RG2	0	0	DIG	LATG<2> data output.				
AN18/C3INA		1	I	ST	PORTG<2> data input.				
Ĭ	RX2	1	Ι	ST	Asynchronous serial receive data input (EUSART module).				
	DT2	1	0	DIG	Synchronous serial data output (EUSART module); takes priority over port data.				
		1	Ι	ST	Synchronous serial data input (EUSART module); user must configure as an input.				
	AN18	1	-	ANA	A/D Input Channel 18. Default input configuration on POR; does not affect digital output.				
	C3INA	х	Ι	ANA	Comparator 3 Input A.				
RG3/CCP4/AN17/	RG3	0	0	DIG	LATG<3> data output.				
P3D/C3INB		1	Ι	ST	PORTG<3> data input.				
	CCP4	0	0	DIG	CCP4 compare/PWM output; takes priority over port data.				
		1	Ι	ST	CCP4 capture input.				
	AN17	1	I	ANA	A/D Input Channel 17. Default input configuration on POR; does not affect digital output.				
	P3D	0	0	—	ECCP3 PWM Output D. May be configured for tri-state during Enhanced PWM.				
	C3INB	х	Ι	ANA	Comparator 3 Input B.				
RG4/RTCC/	RG4	0	0	DIG	LATG<4> data output.				
T7CKI/T5G/		1	I	ST	PORTG<4> data input.				
CCP5/AN16/ P1D/C3INC	RTCC	x	0	DIG	RTCC output.				
	T7CKI	x	Ι	ST	Timer7 clock input.				
	T5G	x	I	ST	Timer5 external clock gate input.				
	CCP5	0	0	DIG	CCP5 compare/PWM output; takes priority over port data.				
		1	Ι	ST	CCP5 capture input.				
	AN16	1	Ι	ANA	A/D Input Channel 17. Default input configuration on POR; does not affect digital output.				
	P1D	0	0	—	ECCP1 PWM Output D. May be configured for tri-state during Enhanced PWM.				
Ì	C3INC	x	I	ANA	Comparator 3 Input C.				

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

TABLE 12-14: SUMMARY OF REGISTERS ASSOCIATED WITH PORTG

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTG	_	_	RG5 ⁽²⁾	RG4	RG3	RG2	RG1	RG0
TRISG	—	_	_	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0
ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16
ODCON1	SSP10D	CCP2OD	CCP10D	_	_			SSP2OD
ODCON2	CCP100D ⁽¹⁾	CCP9OD ⁽¹⁾	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD
ODCON3	U2OD	U10D	_	_		—	—	CTMUDS

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PORTG.

Note 1: Unimplemented on PIC18FX5K22 devices, read as '0'.

2: This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented.

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12.9 PORTH, LATH and TRISH Registers

Note:	PORTH is	available	only	on	the	80-pin
	devices.					

PORTH is an 8-bit wide, bidirectional I/O port. The corresponding Data Direction and Output Latch registers are TRISH and LATH.

All pins on PORTH are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

EXAMPLE 12-8: INITIALIZING PORTH

CLRF	PORTH	;	Initialize PORTH by
		;	clearing output
		;	data latches
CLRF	LATH	;	Alternate method
		;	to clear output
		;	data latches
BANKSEL	ANCON2	;	Select bank with ANCON2 register
MOVLW	0Fh	;	Configure PORTH as
MOVWF	ANCON2	;	digital I/O
MOVLW	0Fh	;	Configure PORTH as
MOVWF	ANCON1	;	digital I/O
BANKSEL	TRISH	;	Select bank with TRISH register
MOVLW	0CFh	;	Value used to
		;	initialize data
		;	direction
MOVWF	TRISH	;	Set RH3:RH0 as inputs
		;	RH5:RH4 as outputs
		;	RH7:RH6 as inputs

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RH0/AN23/A16	RH0	0	0	DIG	LATH<0> data output.
		1	I	ST	PORTH<0> data input.
	AN23	1	I	ANA	A/D Input Channel 23. Default input configuration on POR; does not affect digital input.
	A16	x	0	DIG	External memory interface, Address Line 16; takes priority over port data.
RH1/AN22/A17	RH1	0	0	DIG	LATH<1> data output.
		1	Ι	ST	PORTH<1> data input.
	AN22	1	I	ANA	A/D Input Channel 22. Default input configuration on POR; does not affect digital input.
	A17	x	0	DIG	External memory interface, Address Line 17; takes priority over port data.
RH2/AN21/A18	RH2	0	0	DIG	LATH<2> data output.
		1	I	ST	PORTH<2> data input.
	AN21	1	I	ANA	A/D Input Channel 21. Default input configuration on POR; does not affect digital input.
	A18	x	0	DIG	External memory interface, Address Line 18; takes priority over port data.
RH3/AN20/A19	RH3	0	0	DIG	LATH<3> data output.
		1	I	ST	PORTH<3> data input.
	AN20	1	I	ANA	A/D Input Channel 20. Default input configuration on POR; does not affect digital input.
	A19	x	0	DIG	External memory interface, Address Line 19; takes priority over port data.

TABLE 12-15: PORTH FUNCTIONS

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

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TABLE 12-15: PORTH FUNCTIONS (CONTINUED)

Pin Name	Function	TRIS Setting	I/O	I/O Type	Description			
RH4/CCP9/	RH4	0	0	DIG	LATH<4> data output.			
P3C/AN12/		1	Ι	ST	PORTH<4> data input.			
C2INC	CCP9	0	0	DIG	CCP9 compare/PWM output; takes priority over port data.			
		1	I	ST	CCP9 capture input.			
	P3C	0	0	_	ECCP3 PWM Output C. May be configured for tri-state during Enhanced PWM.			
	AN12	1	Ι	ANA	A/D Input Channel 12. Default input configuration on POR; does not affect digital input.			
	C2INC	x	Ι	ANA	Comparator 2 Input C.			
RH5/CCP8/	RH5	0	0	DIG	LATH<5> data output.			
P3B/AN13/		1	Ι	ST	PORTH<5> data input.			
C2IND	CCP8	0	0	DIG	CCP8 compare/PWM output; takes priority over port data.			
		1	Ι	ST	CCP8 capture input.			
	P3B	0	0	—	ECCP3 PWM Output B. May be configured for tri-state during Enhanced PWM.			
-	AN13	1	I	ANA	A/D Input Channel 13. Default input configuration on POR; does not affect digital input.			
	C2IND	x	Ι	ANA	Comparator 2 Input D.			
RH6/CCP7/	RH6	0	0	DIG	LATH<6> data output.			
P1C/AN14/		1	I	ST	PORTH<6> data input.			
C1INC	CCP7	0	0	DIG	CCP7 compare/PWM output; takes priority over port data.			
		1	Ι	ST	CCP7 capture input.			
	P1C	0	0	—	ECCP1 PWM Output C. May be configured for tri-state during Enhanced PWM.			
	AN14	1	I	ANA	A/D Input Channel 14. Default input configuration on POR; does not affect digital input.			
	C1INC	x	I	ANA	Comparator 1 Input C.			
RH7/CCP6/	RH7	0	0	DIG	LATH<7> data output.			
P1B/AN15		1	Ι	ST	PORTH<7> data input.			
	CCP6	0	0	DIG	CCP6 compare/PWM output; takes priority over port data.			
		1	Ι	ST	CCP6 capture input.			
	P1B	0	0	—	ECCP1 PWM Output B. May be configured for tri-state during Enhanced PWM.			
	AN15	1	I	ANA	A/D Input Channel 15. Default input configuration on POR; does not affect digital input.			

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input, x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

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Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTH ⁽¹⁾	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0
LATH ⁽¹⁾	LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0
TRISH ⁽¹⁾	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8
ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16
ODCON2	CCP100D ⁽²⁾	CCP9OD ⁽²⁾	CCP8OD	CCP7OD	CCP6OD	CCP5OD	CCP4OD	CCP3OD

TABLE 12-16: SUMMARY OF REGISTERS ASSOCIATED WITH PORTH

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

2: Unimplemented on PIC18FX5K22 devices, read as '0'.

12.10 PORTJ, TRISJ and LATJ Registers

Note: PORTJ is available only on 80-pin devices.

PORTJ is an 8-bit wide, bidirectional port. The corresponding Data Direction and Output Latch registers are TRISJ and LATJ.

All pins on PORTJ are implemented with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

Note:	These pins are configured as digital inputs
	on any device Reset.

When the external memory interface is enabled, all of the PORTJ pins function as control outputs for the interface. This occurs automatically when the interface is enabled by clearing the EBDIS control bit (MEMCON<7>). The TRISJ bits are also overridden. Each of the PORTJ pins has a weak internal pull-up. The pull-ups are provided to keep the inputs at a known state for the external memory interface while powering up. A single control bit can turn off all the pull-ups. This is performed by clearing bit, RJPU (PADCFG1<5>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on any device Reset.

CLRF	PORTJ	; Initialize PORTJ by
		; clearing output latches
CLRF	LATJ	; Alternate method
		; to clear output latches
MOVLW	0CFh	; Value used to
		; initialize data
		; direction
MOVWF	TRISJ	; Set RJ3:RJ0 as inputs
		; RJ5:RJ4 as output
		; RJ7:RJ6 as inputs

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Pin Name	Function	TRIS Setting	I/O	I/O Type	Description
RJ0/ALE	RJ0	0	0	DIG	LATJ<0> data output.
		1	-	ST	PORTJ<0> data input.
	ALE	x	0	DIG	External memory interface address latch enable control output; takes priority over digital I/O.
RJ1/OE	RJ1	0	0	DIG	LATJ<1> data output.
		1	Ι	ST	PORTJ<1> data input.
	ŌĒ	x	0	DIG	External memory interface output enable control output; takes priority over digital I/O.
RJ2/WRL	RJ2	0	0	DIG	LATJ<2> data output.
		1	Ι	ST	PORTJ<2> data input.
	WRL	WRL x O DIG External Mo digital I/O.		DIG	External Memory Bus write low byte control; takes priority over digital I/O.
RJ3/WRH	RJ3 0		0	DIG	LATJ<3> data output.
		1	Ι	ST	PORTJ<3> data input.
	WRH	x	0	DIG	External memory interface write high-byte control; takes priority over digital I/O.
RJ4/BA0	RJ4	0	0	DIG	LATJ<4> data output.
		1	I	ST	PORTJ<4> data input.
	BA0	x	0	DIG	External Memory Interface Byte Address 0 control output; takes priority over digital I/O.
RJ5/CE	RJ5	0	0	DIG	LATJ<5> data output.
		1	I	ST	PORTJ<5> data input.
	CE	x	0	DIG	External memory interface chip enable control output; takes priority over digital I/O.
RJ6/LB	RJ6	0	0	DIG	LATJ<6> data output.
		1	I	ST	PORTJ<6> data input.
	LB	x	0	DIG	External memory interface lower byte enable control output; takes priority over digital I/O.
RJ7/UB	RJ7	0	0	DIG	LATJ<7> data output.
		1	I	ST	PORTJ<7> data input.
	UB	x	0	DIG	External memory interface upper byte enable control output; takes priority over digital I/O.

Legend: O = Output, I = Input, ANA = Analog Signal, DIG = Digital Output, ST = Schmitt Trigger Buffer Input,

x = Don't care (TRIS bit does not affect port direction or is overridden for this option).

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTJ ⁽¹⁾	RJ7	RJ6	RJ5	RJ4	RJ3	RJ2	RJ1	RJ0
LATJ ⁽¹⁾	LATJ7	LATJ6	LATJ5	LATJ4	LATJ3	LATJ2	LATJ1	LATJ0
TRISJ ⁽¹⁾	TRISJ7	TRISJ6	TRISJ5	TRISJ4	TRISJ3	TRISJ2	TRISJ1	TRISJ0
PADCFG1	RDPU	REPU	rjpu ⁽¹⁾	_	_	RTSECSEL1	RTSECSEL0	_

Legend: Shaded cells are not used by PORTJ.

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

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12.11 Parallel Slave Port

PORTD can function as an 8-bit-wide Parallel Slave Port (PSP), or microprocessor port, when control bit, PSPMODE (PSPCON<4>), is set. The port is asynchronously readable and writable by the external world through the RD control input pin (RE0/P2D/RD/AD8) and WR control input pin (RE1/P2C/WR/AD9).

Note:	The Parallel Slave Port is available only in
	Microcontroller mode.

The PSP can directly interface to an 8-bit microprocessor data bus. The external microprocessor can read or write the PORTD latch as an eight-bit latch.

Setting bit, PSPMODE, enables port pin, RE0/P2D/RD/AD8, to be the RD input, RE1/P2C/WR/AD9 to be the WR input and RE2/P2B/CCP10/CS/AD10 to be the CS (Chip Select) input. For this functionality, the corresponding data direction bits of the TRISE register (TRISE<2:0>) must be configured as inputs (= 111).

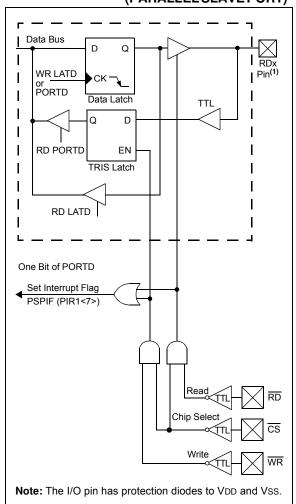
A write to the PSP occurs when both the \overline{CS} and \overline{WR} lines are first detected low and ends when either are detected high. The PSPIF and IBF flag bits (PIR1<7> and PSPCON<7>, respectively) are set when the write ends.

A read from the PSP occurs when both the \overline{CS} and \overline{RD} lines are first detected low. The data in PORTD is read out and the OBF bit (PSPCON<6>) is set. If the user writes new data to PORTD to set OBF, the data is immediately read out, but the OBF bit is not set.

When either the \overline{CS} or \overline{RD} line is detected high, the PORTD pins return to the input state and the PSPIF bit is set. User applications should wait for PSPIF to be set before servicing the PSP. When this happens, the IBF and OBF bits can be polled and the appropriate action taken.

The timing for the control signals in Write and Read modes is shown in Figure 12-4 and Figure 12-5, respectively.

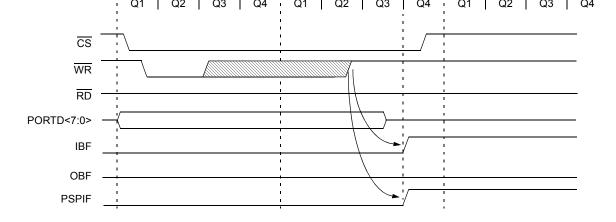
FIGURE 12-3: PORTD AND PORTE BLOCK DIAGRAM (PARALLEL SLAVE PORT)



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REGISTER 12-5: PSPCON: PARALLEL SLAVE PORT CONTROL REGISTER

R-0	R-0	R/W-0	R/W-0	U-0	U-0	U-0	U-0
IBF	OBF	IBOV	PSPMODE	_	—	—	_
bit 7							bit
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimpler	mented bit, read	as '0'	
-n = Value at	POR	'1' = Bit is se	t	'0' = Bit is cle	ared	x = Bit is unkn	own
bit 7			ed and is waiti	ng to be read	by the CPU		
bit 6	1 = The outpu	Buffer Full Sta ut buffer still ho ut buffer has b	olds a previous	ly written word	i		
bit 5	•			out word had n	ot been read (m	nust be cleared	in software)
bit 4	1 = Parallel S	Parallel Slave I lave Port mod Purpose I/O m	-	ect bit			
bit 3-0		ted: Read as					
FIGURE 12-	-4: PARA	LLEL SLAVI Q2 Q3	E PORT WRI	i i		1 Q2 Q	3 Q4



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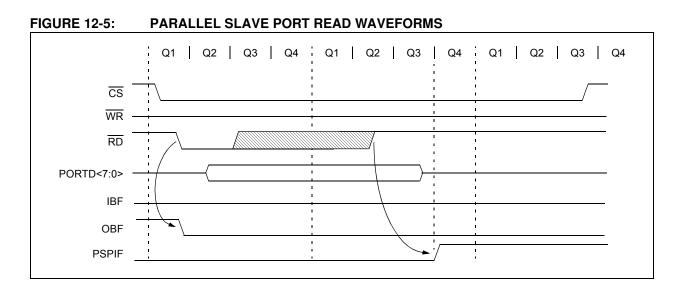


TABLE 12-19: REGISTERS ASSOCIATED WITH PARALLEL SLAVE PORT

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
PORTD	RD7	RD6	RD5	RD4	RD3	RD2	RD1	RD0
LATD	LATD7	LATD6	LATD5	LATD4	LATD3	LATD2	LATD1	LATD0
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0
PORTE	RE7	RE6	RE5	RE4	RE3	RE2	RE1	RE0
LATE	LATE7	LATE6	LATE5	LATE4	LATE3	LATE2	LATE1	LATE0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0
PSPCON	IBF	OBF	IBOV	PSPMODE		—	_	—
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBDM

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Parallel Slave Port.

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NOTES:

13.0 TIMER0 MODULE

The Timer0 module incorporates the following features:

- Software-selectable operation as a timer or counter in both 8-bit or 16-bit modes
- · Readable and writable registers
- Dedicated 8-bit, software programmable
 prescaler
- Selectable clock source (internal or external)
- · Edge select for external clock
- Interrupt-on-overflow

The T0CON register (Register 13-1) controls all aspects of the module's operation, including the prescale selection. It is both readable and writable.

Figure 13-1 provides a simplified block diagram of the Timer0 module in 8-bit mode. Figure 13-2 provides a simplified block diagram of the Timer0 module in 16-bit mode.

REGISTER 13-1: TOCON: TIMER0 CONTROL REGISTER

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
TMR00N	T08BIT	TOCS	T0SE	PSA	T0PS2	T0PS1	T0PS0
bit 7							bit 0

Legend:				
R = Reada	ble bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7	TMR0ON	: Timer0 On/Off Control bit		
	1 = Enabl	les Timer0		
	0 = Stops	Timer0		
bit 6	T08BIT : T	imer0 8-Bit/16-Bit Control bi	t	
	1 = Timer	0 is configured as an 8-bit tir	mer/counter	
	0 = Timer	0 is configured as a 16-bit tir	mer/counter	
bit 5	TOCS: Tir	ner0 Clock Source Select bit	:	
	1 = Trans	ition on T0CKI pin input edge	e	
	0 = Intern	al clock (Fosc/4)		
bit 4	TOSE: Tin	ner0 Source Edge Select bit		
	1 = Increr	ment on high-to-low transition	n on T0CKI pin	
	0 = Increr	ment on low-to-high transition	n on T0CKI pin	
bit 3	PSA: Tim	er0 Prescaler Assignment bi	t	
	1 = Timer	0 prescaler is not assigned;	Timer0 clock input bypasses p	prescaler
	0 = Timer	0 prescaler is assigned; Tim	er0 clock input comes from pr	escaler output
bit 2-0	T0PS<2:0)>: Timer0 Prescaler Select	bits	
	111 = 1 :2	256 Prescale value		
	110 = 1 :1	28 Prescale value		
		4 Prescale value		
		2 Prescale value		
		6 Prescale value		
		Prescale value Prescale value		
		Prescale value		

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13.1 Timer0 Operation

Timer0 can operate as either a timer or a counter. The mode is selected with the T0CS bit (T0CON<5>). In Timer mode (T0CS = 0), the module increments on every clock by default unless a different prescaler value is selected (see **Section 13.3 "Prescaler**"). If the TMR0 register is written to, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMR0 register.

The Counter mode is selected by setting the T0CS bit (= 1). In this mode, Timer0 increments either on every rising edge or falling edge of the T0CKI pin. The incrementing edge is determined by the Timer0 Source Edge Select bit, T0SE (T0CON<4>); clearing this bit selects the rising edge. Restrictions on the external clock input are discussed below.

An external clock source can be used to drive Timer0; however, it must meet certain requirements to ensure that the external clock can be synchronized with the internal phase clock (Tosc). There is a delay between synchronization and the onset of incrementing the timer/counter.

13.2 Timer0 Reads and Writes in 16-Bit Mode

TMR0H is not the actual high byte of Timer0 in 16-bit mode. It is actually a buffered version of the real high byte of Timer0, which is not directly readable nor writable (see Figure 13-2). TMR0H is updated with the contents of the high byte of Timer0 during a read of TMR0L. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte was valid, due to a rollover between successive reads of the high and low byte.

Similarly, a write to the high byte of Timer0 must also take place through the TMR0H Buffer register. The high byte is updated with the contents of TMR0H when a write occurs to TMR0L. This allows all 16 bits of Timer0 to be updated at once.

FIGURE 13-1: TIMER0 BLOCK DIAGRAM (8-BIT MODE)

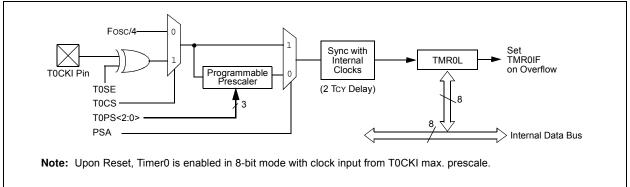
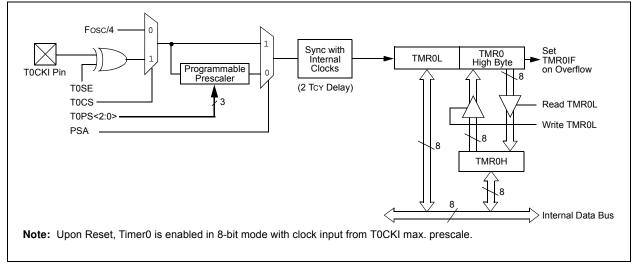


FIGURE 13-2: TIMER0 BLOCK DIAGRAM (16-BIT MODE)



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13.3 Prescaler

An 8-bit counter is available as a prescaler for the Timer0 module. The prescaler is not directly readable or writable. Its value is set by the PSA and T0PS<2:0> bits (T0CON<3:0>), which determine the prescaler assignment and prescale ratio.

Clearing the PSA bit assigns the prescaler to the Timer0 module. When it is assigned, prescale values from 1:2 through 1:256, in power-of-two increments, are selectable.

When assigned to the Timer0 module, all instructions writing to the TMR0 register (for example, CLRF TMR0, MOVWF TMR0, BSF TMR0) clear the prescaler count.

Note:	Writing to TMR0 when the prescaler is
	assigned to Timer0 will clear the prescaler
	count but will not change the prescaler
	assignment.

13.3.1 SWITCHING PRESCALER ASSIGNMENT

The prescaler assignment is fully under software control and can be changed "on-the-fly" during program execution.

13.4 Timer0 Interrupt

The TMR0 interrupt is generated when the TMR0 register overflows from FFh to 00h in 8-bit mode, or from FFFFh to 0000h in 16-bit mode. This overflow sets the TMR0IF flag bit. The interrupt can be masked by clearing the TMR0IE bit (INTCON<5>). Before reenabling the interrupt, the TMR0IF bit must be cleared in software by the Interrupt Service Routine (ISR).

Since Timer0 is shut down in Sleep mode, the TMR0 interrupt cannot awaken the processor from Sleep.

TABLE 13-1:	REGISTERS ASSOCIATED WITH TIMER0

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
TMR0L Timer0 Register Low Byte							
Timer0 Register High Byte							
GIE/GIEH	PEIE/GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	INT0IF	RBIF
TMR0ON	T08BIT	TOCS	TOSE	PSA	T0PS2	T0PS1	T0PS0
	Timer0 Regis Timer0 Regis GIE/GIEH	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL TMR0IE	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL TMR0IE INT0IE	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL TMR0IE INT0IE	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL TMR0IE INT0IE RBIE TMR0IF	Timer0 Register Low Byte Timer0 Register High Byte GIE/GIEH PEIE/GIEL TMR0IE INT0IE RBIE TMR0IF

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Timer0.

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NOTES:

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14.0 TIMER1 MODULE

The Timer1 timer/counter module incorporates these features:

- Software-selectable operation as a 16-bit timer or counter
- Readable and writable 8-bit registers (TMR1H and TMR1L)
- Selectable clock source (internal or external) with device clock or SOSC oscillator internal options
- Interrupt-on-overflow
- Reset on ECCP Special Event Trigger
- · Timer with gated control

Figure 14-1 displays a simplified block diagram of the Timer1 module.

The Timer1 oscillator can also be used as a low-power clock source for the microcontroller in power-managed operation. The Timer1 can also work on the SOSC oscillator.

Timer1 is controlled through the T1CON Control register (Register 14-1). It also contains the Secondary Oscillator Enable bit (SOSCEN). Timer1 can be enabled or disabled by setting or clearing control bit, TMR10N (T1CON<0>).

The Fosc clock source should not be used with the ECCP capture/compare features. If the timer will be used with the capture or compare features, always select one of the other timer clocking options.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N
bit 7							bit 0

REGISTER 14-1: T1CON: TIMER1 CONTROL REGISTER

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	TMR1CS<1:0>: Timer1 Clock Source Select bits
	10 = Timer1 clock source is either from a pin or oscillator, depending on the SOSCEN bit: SOSCEN = 0:
	External clock from the T1CKI pin (on the rising edge).
	SOSCEN = 1: Depending on the SOSCSEL Configuration bit, the clock source is either a crystal oscillator on the SOSCI/SOSCO pins or an internal clock from the SCLKI pin. 01 = Timer1 clock source is the system clock (Fosc) ⁽¹⁾ 00 = Timer1 clock source is the instruction clock (Fosc/4)
bit 5-4	T1CKPS<1:0>: Timer1 Input Clock Prescale Select bits
	11 = 1:8 Prescale value 10 = 1:4 Prescale value
	01 = 1:2 Prescale value
h :+ 0	00 = 1:1 Prescale value
bit 3	SOSCEN: SOSC Oscillator Enable bit 1 = SOSC is enabled and available for Timer1
	0 = SOSC is disabled for Timer1
	The oscillator inverter and feedback resistor are turned off to eliminate power drain.
bit 2	T1SYNC: Timer1 External Clock Input Synchronization Select bit
	<u>TMR1CS<1:0> = 10:</u> 1 = Do not synchronize external clock input
	0 = Synchronize external clock input
	$\underline{TMR1CS} = 0 \times :$
	This bit is ignored. Timer1 uses the internal clock when TMR1CS<1:0> = $1x$.
bit 1	RD16: 16-Bit Read/Write Mode Enable bit
	 Enables register read/write of Timer1 in one 16-bit operation Enables register read/write of Timer1 in two 8-bit operations
bit 0	TMR1ON: Timer1 On bit
	1 = Enables Timer1
	0 = Stops Timer1
Note 1.	The Face shall as the selected if the time will be used with the FCCD century (compare facture)

Note 1: The FOSC clock source should not be selected if the timer will be used with the ECCP capture/compare features.

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14.1 Timer1 Gate Control Register

The Timer1 Gate Control register (T1GCON), displayed in Register 14-2, is used to control the Timer1 gate.

REGISTER 14-2: T1GCON: TIMER1 GATE CONTROL REGISTER⁽¹⁾

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-x	R/W-0	R/W-0
TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/T1DONE	T1GVAL	T1GSS1	T1GSS0
bit 7							bit 0
-							
Legend:							
R = Readab	le bit	W = Writable	bit	U = Unimplemented	d bit, read as	ʻ0'	
-n = Value at	t POR	'1' = Bit is set		'0' = Bit is cleared		x = Bit is unkn	iown
bit 7	TMR1GE: Ti	mer1 Gate En	able bit				
	If TMR10N =						
	This bit is igr	nored.					
	If TMR1ON =						
		ounting is cont ounts regardle		Timer1 gate function gate function			
bit 6	T1GPOL: Tir	mer1 Gate Pol	arity bit				
				ounts when gate is hi unts when gate is low			
bit 5	-	er1 Gate Togg	-	Ū	,		
	1 = Timer1 (Gate Toggle m	ode is enable	d d and toggle flip-flop	is cleared		
	Timer1 gate	flip-flop toggle	s on every risi	ing edge.			
bit 4	TIGSPM: Ti	mer1 Gate Sin	gle Pulse Mo	de bit			
		•		nabled and is controll	ling Timer1 g	ate	
		Bate Single Pul					
bit 3			-	ulse Acquisition State			
				is ready, waiting for			
				has completed or ha GSPM is cleared.	is not been si	arted	
bit 2		ner1 Gate Cur					
~	Indicates the		of the Timer1	gate that could be p	provided to T	MR1H:TMR1L;	unaffected by
bit 1-0		: Timer1 Gate		ct bits			
		rator 2 output					
		rator 1 output					
		o match PR2 of	output				
	00 = Timer1	gate pin					
Note 1: P	rogramming th	e T1GCON pr	ior to T1CON	is recommended			

Note 1: Programming the T1GCON prior to T1CON is recommended.

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14.2 Timer1 Operation

The Timer1 module is an 8 or 16-bit incrementing counter that is accessed through the TMR1H:TMR1L register pair.

When used with an internal clock source, the module is a timer and increments on every instruction cycle. When used with an external clock source, the module can be used as either a timer or counter. It increments on every selected edge of the external source.

Timer1 is enabled by configuring the TMR1ON and TMR1GE bits in the T1CON and T1GCON registers, respectively.

When SOSC is selected as Crystal mode (by SOSCSEL), the RC1/SOSCI/ECCP2/P2A and RC0/ SOSCO/SCLKI pins become inputs. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.

14.3 Clock Source Selection

The TMR1CS<1:0> and SOSCEN bits of the T1CON register are used to select the clock source for Timer1. Register 14-1 displays the clock source selections.

14.3.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected, the TMR1H:TMR1L register pair will increment on multiples of Fosc, as determined by the Timer1 prescaler.

14.3.2 EXTERNAL CLOCK SOURCE

When the external clock source is selected, the Timer1 module may work as a timer or a counter.

When enabled to count, Timer1 is incremented on the rising edge of the external clock input, T1CKI. Either of these external clock sources can be synchronized to the microcontroller system clock or they can run asynchronously.

When used as a timer with a clock oscillator, an external, 32.768 kHz crystal can be used in conjunction with the dedicated internal oscillator circuit.

- **Note:** In Counter mode, a falling edge must be registered by the counter prior to the first incrementing rising edge after any one or more of the following conditions:
 - Timer1 enabled after POR Reset
 - Write to TMR1H or TMR1L
 - Timer1 is disabled
 - Timer1 is disabled (TMR1ON = 0)

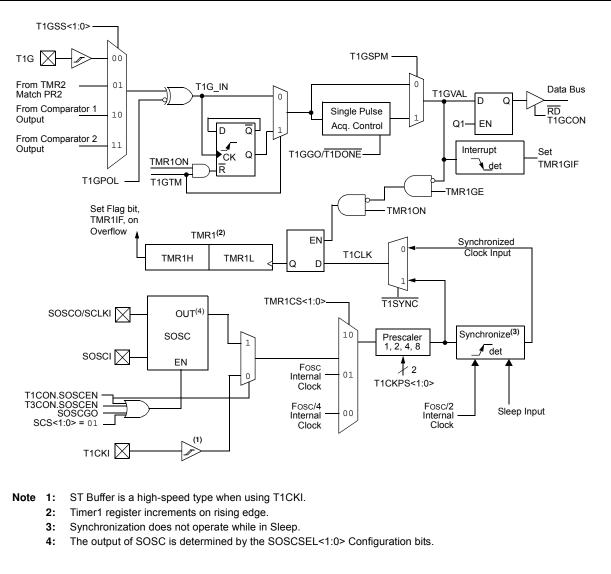
When T1CKI is high, Timer1 is enabled (TMR1ON = 1) when T1CKI is low.

|--|

TMR1CS1	TMR1CS0	SOSCEN	Clock Source
0	1	х	Clock Source (Fosc)
0	0	x	Instruction Clock (Fosc/4)
1	0	0	External Clock on T1CKI Pin
1	0	1	Oscillator Circuit on SOSCI/SOSCO Pins

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14.4 Timer1 16-Bit Read/Write Mode

Timer1 can be configured for 16-bit reads and writes. When the RD16 control bit (T1CON<1>) is set, the address for TMR1H is mapped to a buffer register for the high byte of Timer1. A read from TMR1L loads the contents of the high byte of Timer1 into the Timer1 High Byte Buffer register. This provides the user with the ability to accurately read all 16 bits of Timer1 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer1 must also take place through the TMR1H Buffer register. The Timer1 high byte is updated with the contents of TMR1H when a write occurs to TMR1L. This allows a user to write all 16 bits at once to both the high and low bytes of Timer1.

The high byte of Timer1 is not directly readable or writable in this mode. All reads and writes must take place through the Timer1 High Byte Buffer register. Writes to TMR1H do not clear the Timer1 prescaler; the prescaler is only cleared on writes to TMR1L.

14.5 SOSC Oscillator

An on-chip crystal oscillator circuit is incorporated between pins, SOSCI (input) and SOSCO (amplifier output). It is enabled by any peripheral that requests it. There are eight ways the SOSC can be enabled: if the SOSC is selected as the source by any of the odd timers, which is done by each respective SOSCEN bit (TxCON<3>), if the SOSC is selected as the RTCC source by the RTCOSC Configuration bit (CONFIG3L<1>), if the SOSC is selected as the CPU clock source by the SCS bits (OSCCON<1:0>) or if the SOSCGO bit is set (OSCCON2<3>). The SOSCGO bit is used to warm up the SOSC so that it is ready before any peripheral requests it. The oscillator is a low-power circuit, rated for 32 kHz crystals. It will continue to run during all power-managed modes. The circuit for a typical low-power oscillator is depicted in Figure 14-2. Table 14-2 provides the capacitor selection for the SOSC oscillator.

The user must provide a software time delay to ensure proper start-up of the SOSC oscillator.

FIGURE 14-2:

EXTERNAL COMPONENTS FOR THE SOSC LOW-POWER OSCILLATOR

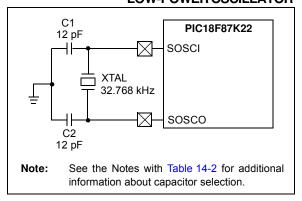


TABLE 14-2: CAPACITOR SELECTION FOR THE TIMER OSCILLATOR^(2,3,4,5)

Oscillator Type	Freq.	C1	C2
LP	32 kHz	12 pF ⁽¹⁾	12 pF ⁽¹⁾
Note 1:	Aicrochip sug	gests these	values as a

- starting point in validating the oscillator circuit.
 - **2:** Higher capacitance increases the stability of the oscillator, but also increases the start-up time.
 - 3: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - 4: Capacitor values are for design guidance only. Values listed would be typical of a CL = 10 pF rated crystal when SOSCSEL<1:0> = 11.
 - Incorrect capacitance value may result in a frequency not meeting the crystal manufacturer's tolerance specification.

The SOSC crystal oscillator drive level is determined based on the SOSCSEL<1:0> (CONFIG1L<4:3>) Configuration bits. The Higher Drive Level mode, SOSCSEL<1:0> = 11, is intended to drive a wide variety of 32.768 kHz crystals with a variety of Capacitance Load (CL) ratings.

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The Lower Drive Level mode is highly optimized for extremely low-power consumption. It is not intended to drive all types of 32.768 kHz crystals. In the Low Drive Level mode, the crystal oscillator circuit may not work correctly if excessively large discrete capacitors are placed on the SOSCO and SOSCI pins. This mode is designed to work only with discrete capacitances of approximately 3 pF-10 pF on each pin.

Crystal manufacturers usually specify a CL (Capacitance Load) rating for their crystals. This value is related to, but not necessarily the same as, the values that should be used for C1 and C2 in Figure 14-2.

For more details on selecting the optimum C1 and C2 for a given crystal, see the crystal manufacturer's applications information. The optimum value depends, in part, on the amount of parasitic capacitance in the circuit, which is often unknown. For that reason, it is highly recommended that thorough testing and validation of the oscillator be performed after values have been selected.

14.5.1 USING SOSC AS A CLOCK SOURCE

The SOSC oscillator is also available as a clock source in power-managed modes. By setting the System Clock Select bits, SCS<1:0> (OSCCON<1:0>), to '01', the device switches to SEC_RUN mode, and both the CPU and peripherals are clocked from the SOSC oscillator. If the IDLEN bit (OSCCON<7>) is cleared and a SLEEP instruction is executed, the device enters SEC_IDLE mode. Additional details are available in Section 4.0 "Power-Managed Modes".

Whenever the SOSC oscillator is providing the clock source, the SOSC System Clock Status flag, SOSCRUN (OSCCON2<6>), is set. This can be used to determine the controller's current clocking mode. It can also indicate the clock source currently being used by the Fail-Safe Clock Monitor (FSCM).

If the Clock Monitor is enabled and the SOSC oscillator fails while providing the clock, polling the SOCSRUN bit will indicate whether the clock is being provided by the SOSC oscillator or another source.

14.5.2 SOSC OSCILLATOR LAYOUT CONSIDERATIONS

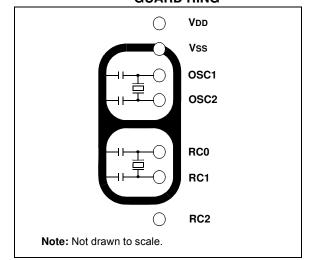
The SOSC oscillator circuit draws very little power during operation. Due to the low-power nature of the oscillator, it may also be sensitive to rapidly changing signals in close proximity. This is especially true when the oscillator is configured for extremely Low-Power mode, SOSCSEL<1:0> (CONFIG1L<4:3>) = 01.

The oscillator circuit, displayed in Figure 14-2, should be located as close as possible to the microcontroller. There should be no circuits passing within the oscillator circuit boundaries other than Vss or VDD.

If a high-speed circuit must be located near the oscillator, it may help to have a grounded guard ring around the oscillator circuit. The guard, as displayed in Figure 14-3, could be used on a single-sided PCB or in addition to a ground plane. (Examples of a high-speed circuit include the ECCP1 pin, in Output Compare or PWM mode, or the primary oscillator using the OSC2 pin.)



3: OSCILLATOR CIRCUIT WITH GROUNDED GUARD RING



In the Low Drive Level mode, SOSCSEL<1:0> = 01, it is critical that RC2 I/O pin signals be kept away from the oscillator circuit. Configuring RC2 as a digital output, and toggling it, can potentially disturb the oscillator circuit, even with a relatively good PCB layout. If possible, either leave RC2 unused or use it as an input pin with a slew rate limited signal source. If RC2 must be used as a digital output, it may be necessary to use the Higher Drive Level Oscillator mode (SOSCSEL<1:0> = 11) with many PCB layouts.

Even in the Higher Drive Level mode, careful layout procedures should still be followed when designing the oscillator circuit.

In addition to dV/dt induced noise considerations, it is important to ensure that the circuit board is clean. Even a very small amount of conductive soldering flux residue can cause PCB leakage currents that can overwhelm the oscillator circuit.

14.6 Timer1 Interrupt

The TMR1 register pair (TMR1H:TMR1L) increments from 0000h to FFFFh and rolls over to 0000h. The Timer1 interrupt, if enabled, is generated on overflow which is latched in interrupt flag bit, TMR1IF (PIR1<0>). This interrupt can be enabled or disabled by setting or clearing the Timer1 Interrupt Enable bit, TMR1IE (PIE1<0>).

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14.7 Resetting Timer1 Using the ECCP Special Event Trigger

If ECCP modules are configured to use Timer1 and to generate a Special Event Trigger in Compare mode (CCPxM<3:0> = 1011), this signal will reset Timer1. The trigger from ECCP2 will also start an A/D conversion, if the A/D module is enabled. (For more information, see **Section 20.3.4 "Special Event Trigger**".)

To take advantage of this feature, the module must be configured as either a timer or a synchronous counter. When used this way, the CCPRxH:CCPRxL register pair effectively becomes a period register for Timer1.

If Timer1 is running in Asynchronous Counter mode, this Reset operation may not work.

In the event that a write to Timer1 coincides with a Special Event Trigger, the write operation will take precedence.

Note:	The	Special	Event	Trigger	from	the	
	ECCPx module will only clear the T						
	register's content, but not set the TMR1IF						
	interru	upt flag b	oit (PIR1	<0>).			

14.8 Timer1 Gate

Timer1 can be configured to count freely or the count can be enabled and disabled using the Timer1 gate circuitry. This is also referred to as Timer1 gate count enable.

Timer1 gate can also be driven by multiple selectable sources.

14.8.1 TIMER1 GATE COUNT ENABLE

The Timer1 Gate Enable mode is enabled by setting the TMR1GE bit of the T1GCON register. The polarity of the Timer1 Gate Enable mode is configured using the T1GPOL bit (T1GCON<6>).

When Timer1 Gate Enable mode is enabled, Timer1 will increment on the rising edge of the Timer1 clock source. When Timer1 Gate Enable mode is disabled, no incrementing will occur and Timer1 will hold the current count. See Figure 14-4 for timing details.

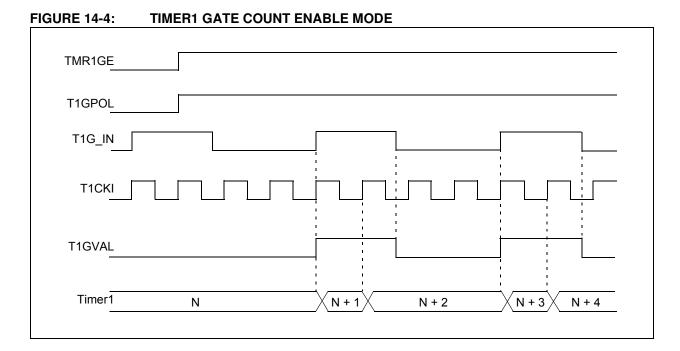
TABLE 14-3:	TIMER1 GATE ENABLE
	SELECTIONS

T1CLK ^(†)	T1GPOL (T1GCON<6>)	T1G Pin	Timer1 Operation	
1	0	0	Counts	
\uparrow	0	1	Holds Count	
\uparrow	1	0	Holds Count	
\uparrow	1	1	Counts	

† The clock on which TMR1 is running. For more information, see Figure 14-1.

Note: The CCP and ECCP modules use Timers, 1 through 8, for some modes. The assignment of a particular timer to a CCP/ECCP module is determined by the Timer to CCP enable bits in the CCPTMRSx registers. For more details, see Register 19-2, Register 19-3 and Register 20-2.

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14.8.2 TIMER1 GATE SOURCE SELECTION

The Timer1 gate source can be selected from one of four sources. Source selection is controlled by the T1GSSx (T1GCON<1:0>) bits (see Table 14-4).

TABLE 14-4:	TIMER1 GATE SOURCES
-------------	---------------------

T1GSS<1:0>	Timer1 Gate Source
00	Timer1 Gate Pin
01	TMR2 to Match PR2 (TMR2 increments to match PR2)
10	Comparator 1 Output (comparator logic high output)
11	Comparator 2 Output (comparator logic high output)

The polarity for each available source is also selectable, controlled by the T1GPOL bit (T1GCON<6>).

14.8.2.1 T1G Pin Gate Operation

The T1G pin is one source for Timer1 gate control. It can be used to supply an external source to the Timer1 gate circuitry.

14.8.2.2 Timer2 Match Gate Operation

The TMR2 register will increment until it matches the value in the PR2 register. On the very next increment cycle, TMR2 will be reset to 00h. When this Reset

occurs, a low-to-high pulse will automatically be generated and internally supplied to the Timer1 gate circuitry. The pulse will remain high for one instruction cycle and will return back to a low state until the next match.

The T1GPOL bit determines when the Timer1 counter increments based on this pulse. When T1GPOL = 1, Timer1 increments for a single instruction cycle following a TMR2 match with PR2. When T1GPOL = 0, Timer1 increments continuously, except for the cycle following the match, when the gate signal goes from low-to-high.

14.8.2.3 Comparator 1 Output Gate Operation

The output of Comparator 1 can be internally supplied to the Timer1 gate circuitry. After setting up Comparator 1 with the CM1CON register, Timer1 will increment depending on the transitions of the CMP1OUT (CMSTAT<5>) bit.

14.8.2.4 Comparator 2 Output Gate Operation

The output of Comparator 2 can be internally supplied to the Timer1 gate circuitry. After setting up Comparator 2 with the CM2CON register, Timer1 will increment depending on the transitions of the CMP2OUT (CMSTAT<6>) bit.

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14.8.3 TIMER1 GATE TOGGLE MODE

When Timer1 Gate Toggle mode is enabled, it is possible to measure the full cycle length of a Timer1 gate signal, as opposed to the duration of a single level pulse.

The Timer1 gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. (For timing details, see Figure 14-5.)

The T1GVAL bit (T1GCON<2>) indicates when the Toggled mode is active and the timer is counting.

The Timer1 Gate Toggle mode is enabled by setting the T1GTM bit (T1GCON<5>). When T1GTM is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

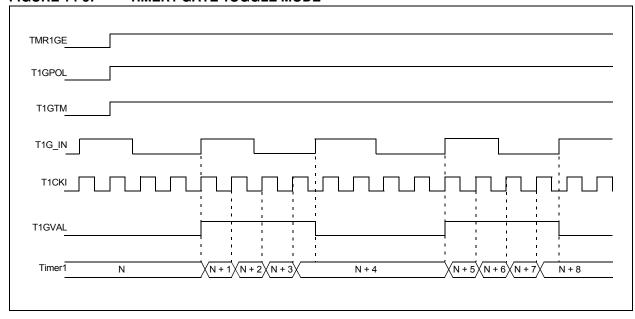


FIGURE 14-5: TIMER1 GATE TOGGLE MODE

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14.8.4 TIMER1 GATE SINGLE PULSE MODE

When Timer1 Gate Single Pulse mode is enabled, it is possible to capture a single pulse gate event. Timer1 Gate Single Pulse mode is enabled by setting the T1GSPM bit (T1GCON<4>) and the T1GGO/T1DONE bit (T1GCON<3>). The Timer1 will be fully enabled on the next incrementing edge.

On the next trailing edge of the pulse, the T1GGO/ T1DONE bit will automatically be cleared. No other gate events will be allowed to increment Timer1 until the T1GGO/T1DONE bit is once again set in software. Clearing the T1GSPM <u>bit of the</u> T1GCON register will also clear the T1GGO/T1DONE bit. (For timing details, see Figure 14-6.)

Simultaneously enabling the Toggle and Single Pulse modes will permit both sections to work together. This allows the cycle times on the Timer1 gate source to be measured. (For timing details, see Figure 14-7.)

14.8.5 TIMER1 GATE VALUE STATUS

When the Timer1 gate value status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the T1GVAL bit (T1GCON<2>). This bit is valid even when the Timer1 gate is not enabled (TMR1GE bit is cleared).

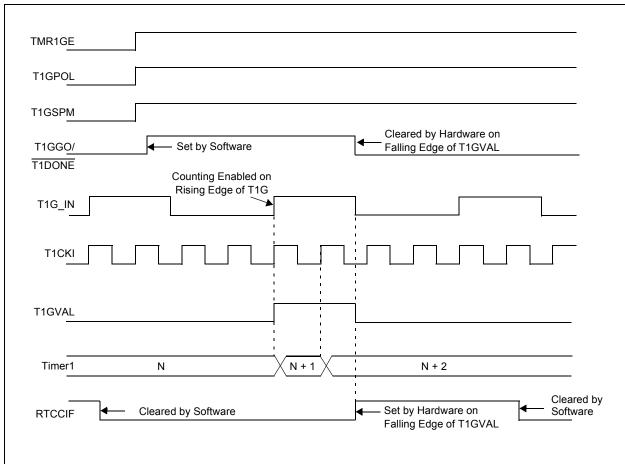


FIGURE 14-6: TIMER1 GATE SINGLE PULSE MODE

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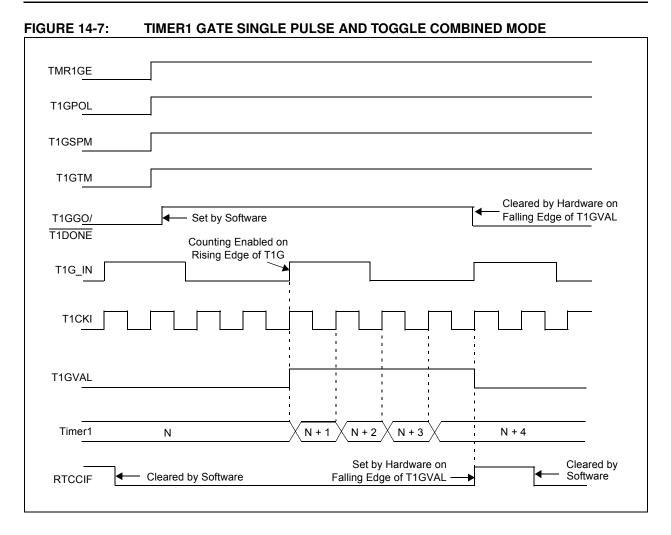


TABLE 14-5: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
TMR1L	Timer1 Regi	ster Low Byte	9					
TMR1H	Timer1 Regi	ster High Byte	е					
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/ T1DONE	T1GVAL	T1GSS1	T1GSS0
OSCCON2	_	SOSCRUN	_	—	SOSCGO		MFIOFS	MFIOSEL
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBDM

Legend: Shaded cells are not used by the Timer1 module.

Note 1: Unimplemented on 32-Kbyte devices (PIC18FX5K22).

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NOTES:

15.0 TIMER2 MODULE

The Timer2 module incorporates the following features:

- Eight-bit Timer and Period registers (TMR2 and PR2, respectively)
- · Both registers are readable and writable
- Software programmable prescaler (1:1, 1:4 and 1:16)
- Software programmable postscaler (1:1 through 1:16)
- Interrupt on TMR2 to PR2 match
- Optional use as the shift clock for the MSSP modules

This module is controlled through the T2CON register (Register 15-1) that enables or disables the timer, and configures the prescaler and postscaler. Timer2 can be shut off by clearing control bit, TMR2ON (T2CON<2>), to minimize power consumption.

A simplified block diagram of the module is shown in Figure 15-1.

15.1 Timer2 Operation

In normal operation, TMR2 is incremented from 00h on each clock (Fosc/4). A four-bit counter/prescaler on the clock input gives the prescale options of direct input, divide-by-4 or divide-by-16. These are selected by the prescaler control bits, T2CKPS<1:0> (T2CON<1:0>).

The value of TMR2 is compared to that of the Period register, PR2, on each clock cycle. When the two values match, the comparator generates a match signal as the timer output. This signal also resets the value of TMR2 to 00h on the next cycle and drives the output counter/ postscaler. (See Section 15.2 "Timer2 Interrupt".)

The TMR2 and PR2 registers are both directly readable and writable. The TMR2 register is cleared on any device Reset, while the PR2 register initializes at FFh. Both the prescaler and postscaler counters are cleared on the following events:

- A write to the TMR2 register
- · A write to the T2CON register
- Any device Reset Power-on Reset (POR), MCLR Reset, Watchdog Timer Reset (WDTR) or Brown-out Reset (BOR)

TMR2 is not cleared when T2CON is written.

Note: The CCP and ECCP modules use Timers, 1 through 8, for some modes. The assignment of a particular timer to a CCP/ECCP module is determined by the Timer to CCP enable bits in the CCPTMRSx registers. For more details, see Register 20-2, Register 19-2 and Register 19-3.

REGISTER 15-1: T2CON: TIMER2 CONTROL REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 bit 6-3	Unimplemented: Read as '0' T2OUTPS<3:0>: Timer2 Output Postscale Select bits 0000 = 1:1 Postscale 0001 = 1:2 Postscale •
	• 1111 = 1:16 Postscale
bit 2	TMR2ON: Timer2 On bit 1 = Timer2 is on 0 = Timer2 is off
bit 1-0	T2CKPS<1:0>: Timer2 Clock Prescale Select bits 00 = Prescaler is 1 01 = Prescaler is 4 1x = Prescaler is 16

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15.2 Timer2 Interrupt

Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2 to PR2 match) provides the input for the 4-bit output counter/postscaler. This counter generates the TMR2 match interrupt flag, which is latched in TMR2IF (PIR1<1>). The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE (PIE1<1>).

A range of 16 postscaler options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS<3:0> (T2CON<6:3>).

15.3 Timer2 Output

The unscaled output of TMR2 is available primarily to the ECCP modules, where it is used as a time base for operations in PWM mode.

Timer2 can optionally be used as the shift clock source for the MSSP modules operating in SPI mode. Additional information is provided in Section 21.0 "Master Synchronous Serial Port (MSSP) Module".

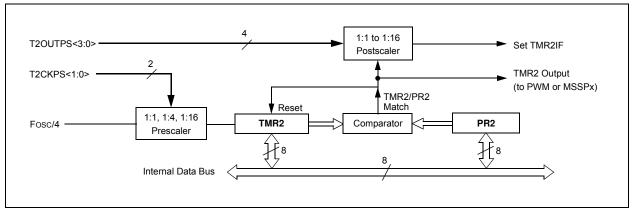


FIGURE 15-1: TIMER2 BLOCK DIAGRAM

TABLE 15-1: REGISTERS ASSOCIATED WITH TIMER2 AS A TIMER/COUNTER

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
Timer2 Reg	ister						
_	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
Timer2 Period Register							
PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBDM
	GIE/GIEH PSPIF PSPIE PSPIP Timer2 Reg — Timer2 Peri	GIE/GIEHPEIE/GIELPSPIFADIFPSPIEADIEPSPIPADIPTimer2 Resister	GIE/GIEHPEIE/GIELTMR0IEPSPIFADIFRC1IFPSPIEADIERC1IEPSPIPADIPRC1IPTimer2 RegisterT20UTPS3T20UTPS2Timer2 Period RegisterFF	GIE/GIEHPEIE/GIELTMR0IEINT0IEPSPIFADIFRC1IFTX1IFPSPIEADIERC1IETX1IEPSPIPADIPRC1IPTX1IPTimer2 RegisterT2OUTPS3T2OUTPS2T2OUTPS1Timer2 Period Register	GIE/GIEHPEIE/GIELTMR0IEINT0IERBIEPSPIFADIFRC1IFTX1IFSSP1IFPSPIEADIERC1IETX1IESSP1IEPSPIPADIPRC1IPTX1IPSSP1IPTimer2 RegisterT20UTPS3T20UTPS2T20UTPS1T20UTPS0Timer2 Period RegisterImage: Non-transferImage: Non-transferImage: Non-transfer	GIE/GIEHPEIE/GIELTMR0IEINT0IERBIETMR0IFPSPIFADIFRC1IFTX1IFSSP1IFTMR1GIFPSPIEADIERC1IETX1IESSP1IETMR1GIEPSPIPADIPRC1IPTX1IPSSP1IPTMR1GIPTimer2 RegisterT20UTPS3T20UTPS2T20UTPS1T20UTPS0TMR2ONTimer2 Period Register	GIE/GIEHPEIE/GIELTMR0IEINT0IERBIETMR0IFINT0IFPSPIFADIFRC1IFTX1IFSSP1IFTMR1GIFTMR2IFPSPIEADIERC1IETX1IESSP1IETMR1GIETMR2IEPSPIPADIPRC1IPTX1IPSSP1IPTMR1GIPTMR2IPTimer2 RegisterT20UTPS3T20UTPS2T20UTPS1T20UTPS0TMR2ONT2CKPS1Timer2 Period Register

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer2 module.

16.0 TIMER3/5/7 MODULES

The Timer3/5/7 timer/counter modules incorporate these features:

- Software-selectable operation as a 16-bit timer or counter
- Readable and writable eight-bit registers (TMRxH and TMRxL)
- Selectable clock source (internal or external) with device clock or SOSC oscillator internal options
- Interrupt-on-overflow
- · Module Reset on ECCP Special Event Trigger

Timer7 is unimplemented for devices with a program memory of 32 Kbytes (PIC18FX5K22).

Note: Throughout this section, generic references are used for register and bit names that are the same – except for an 'x' variable that indicates the item's association with the Timer3, Timer5 or Timer7 module. For example, the control register is named TxCON and refers to T3CON, T5CON and T7CON.

A simplified block diagram of the Timer3/5/7 module is shown in Figure 16-1.

The Timer3/5/7 module is controlled through the TxCON register (Register 16-1). It also selects the clock source options for the ECCP modules. (For more information, see Section 20.1.1 "ECCP Module and Timer Resources".)

The Fosc clock source should not be used with the ECCP capture/compare features. If the timer will be used with the capture or compare features, always select one of the other timer clocking options.

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REGISTER 1	6-1: TxCO	N: TIMERx C	ONTROL RE	GISTER	
R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
TMRxCS1	TMRxCS0	TxCKPS1	TxCKPS0	SOSCEN	TxSYNC	RD16	TMRxON			
bit 7							bit (
Legend:										
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	as '0'				
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown			
bit 7-6		>: Timerx Cloc								
		lock source de	pends on the S	SOSCEN bit:						
	<u>SOSCEN = 0</u> External clock	refrom the T1CI	(I nin (on the r	isina edae)						
	SOSCEN = 1			ising eage).						
			fuses, either a	a crystal oscillate	or on the SOSC	I/SOSCO pins	s or an externa			
	clock from the			(1)						
		lock source is								
		lock source is t		,						
bit 5-4	TxCKPS<1:0 >: Timerx Input Clock Prescale Select bits 11 = 1:8 Prescale value									
	11 = 1.6 Pres									
	01 = 1.2 Prescale value									
	00 = 1:1 Pres	0 = 1:1 Prescale value								
bit 3	SOSCEN: SC	SC Oscillator I	Enable bit							
	1 = SOSC/SCLKI are enabled for Timerx (based on the SOSCSEL fuses)									
1.11.0		DSC/SCLKI are disabled for Timerx and TxCKI is enabled 								
bit 2										
	(Not usable if the device clock comes from Timer1/Timer3.) When TMRxCS<1:0> = 10:									
		1 = Do not synchronize external clock input								
		ize external clo								
	When TMRxC									
	This bit is ignored; Timer3 uses the internal clock.									
bit 1	RD16: 16-Bit Read/Write Mode Enable bit									
	 Enables register read/write of Timerx in one 16-bit operation Enables register read/write of Timerx in two eight-bit operations 									
bit 0	TMRxON: Tin	•		0	•					
	1 = Enables T	Timerx								
	0 = Stops Tim	ierx								
Note 1. Th	a Foss alaak a	ource should n	ot be selected	if the timer will	bo used with th		iro/oomnoro			

Note 1: The Fosc clock source should not be selected if the timer will be used with the ECCP capture/compare features.

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16.1 Timer3/5/7 Gate Control Register

The Timer3/5/7 Gate Control register (TxGCON), provided in Register 14-2, is used to control the Timerx gate.

REGISTER 16-2: TxGCON: TIMERx GATE CONTROL REGISTER⁽¹⁾

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-x	R/W-0	R/W-0
TMRxGE	TxGPOL	TxGTM	TxGSPM	TxGGO/TxDONE	TxGVAL	TxGSS1	TxGSS0
bit 7							bit 0

Legend:					
R = Readable bit W = Writable bit U = Unimplement			ed bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7	TMRxGE: Timerx Gate Enable bit
	<u>If TMRxON = 0:</u>
	This bit is ignored.
	<u>If TMRxON = 1:</u>
	1 = Timerx counting is controlled by the Timerx gate function
	0 = Timerx counts regardless of Timerx gate function
bit 6	TxGPOL: Timerx Gate Polarity bit
	 1 = Timerx gate is active-high (Timerx counts when gate is high) 0 = Timerx gate is active-low (Timerx counts when gate is low)
bit 5	TxGTM: Timerx Gate Toggle Mode bit
	1 = Timerx Gate Toggle mode is enabled.
	0 = Timerx Gate Toggle mode is disabled and toggle flip-flop is cleared
	Timerx gate flip-flop toggles on every rising edge.
bit 4	TxGSPM: Timerx Gate Single Pulse Mode bit
	 1 = Timerx Gate Single Pulse mode is enabled and is controlling Timerx gate 0 = Timerx Gate Single Pulse mode is disabled
bit 3	TxGGO/TxDONE: Timerx Gate Single Pulse Acquisition Status bit
	1 = Timerx gate single pulse acquisition is ready, waiting for an edge
	0 = Timerx gate single pulse acquisition has completed or has not been started
	This bit is automatically cleared when TxGSPM is cleared.
bit 2	TxGVAL: Timerx Gate Current State bit
	Indicates the current state of the Timerx gate that could be provided to TMRxH:TMRxL. Unaffected by the Timerx Gate Enable (TMRxGE) bit.
bit 1-0	TxGSS<1:0>: Timerx Gate Source Select bits
	11 = Comparator 2 output
	10 = Comparator 1 output
	01 = TMR(x+1) to match PR(x+1) output ⁽²⁾
	00 = Timer1 gate pin The Watchdog Timer oscillator is turned on if TMRxGE = 1, regardless of the state of TMRxON.
Note 1:	Programming the TxGCON prior to TxCON is recommended.

2: Timer(x+1) will be Timer4/6/8 for Timerx (Timer3/5/7), respectively.

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REGISTER 16-3: OSCCON2: OSCILLATOR CONTROL REGISTER 2

U-0	R-0	U-0	U-0	R/W-0	U-0	R-x	R/W-0		
_	SOSCRUN	—	_	SOSCGO		MFIOFS	MFIOSEL		
bit 7							bit C		
Legend:									
R = Readable bit		W = Writable bit		U = Unimplemented bit, read as '0'					
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown			
bit 7	Unimplemented: Read as '0'								
bit 6	SOSCRUN: SOSC Run Status bit								
	1 = System clock comes from a secondary SOSC								
	0 = System clock comes from an oscillator other than SOSC								
bit 5-4	Unimplemen	Unimplemented: Read as '0'							
bit 3	SOSCGO: Oscillator Start Control bit								
	1 = Oscillator is running even if no other sources are requesting it								
	 0 = Oscillator is shut off if no other sources are requesting it (When the SOSC is selected to run fron a digital clock input, rather than an external crystal, this bit has no effect.) 								
bit 2	Unimplemen	Unimplemented: Read as '0'							
bit 1	. MFIOFS: MF-INTOSC Frequency Stable bit								
	1 = MF-INTOSC is stable								
	0 = MF-INTOSC is not stable								
bit 0	MFIOSEL: MF-INTOSC Select bit								
	1 = MF-INTOSC is used in place of HF-INTOSC frequencies of 500 kHz, 250 kHz and 31.25 kHz								
	0 = MF-INTO	SC is not used							

The operating mode is determined by the clock select

bits, TMRxCSx (TxCON<7:6>). When the TMRxCSx bits

are cleared (= 00), Timer3/5/7 increments on every inter-

nal instruction cycle (Fosc/4). When TMRxCSx = 01, the

Timer3/5/7 clock source is the system clock (Fosc), and when it is '10', Timer3/5/7 works as a counter from the

external clock from the TxCKI pin (on the rising edge after

the first falling edge) or the SOSC oscillator.

16.2 Timer3/5/7 Operation

Timer3, Timer5 and Timer7 can operate in these modes:

- Timer
- Synchronous Counter
- Asynchronous Counter
- · Timer with Gated Control

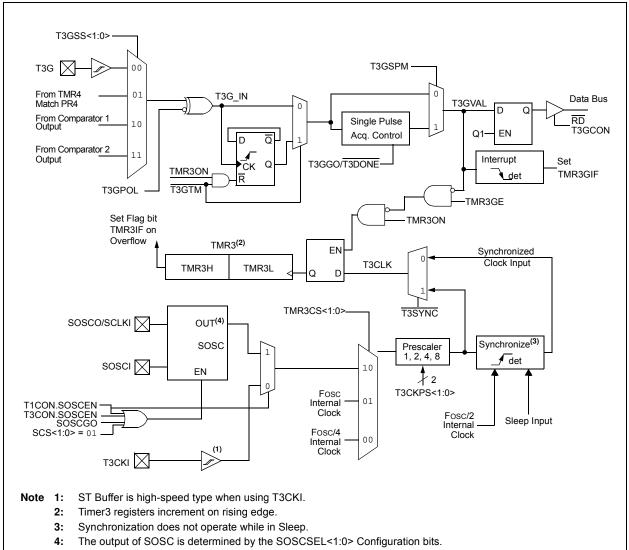


FIGURE 16-1: TIMER3/5/7 BLOCK DIAGRAM

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16.3 Timer3/5/7 16-Bit Read/Write Mode

Timer3/5/7 can be configured for 16-bit reads and writes (see Figure 16.3). When the RD16 control bit (TxCON<1>) is set, the address for TMRxH is mapped to a buffer register for the high byte of Timer3/5/7. A read from TMRxL will load the contents of the high byte of Timer3/5/7 into the Timerx High Byte Buffer register. This provides users with the ability to accurately read all 16 bits of Timer3/5/7 without having to determine whether a read of the high byte, followed by a read of the low byte, has become invalid due to a rollover between reads.

A write to the high byte of Timer3/5/7 must also take place through the TMRxH Buffer register. The Timer3/5/7 high byte is updated with the contents of TMRxH when a write occurs to TMRxL. This allows users to write all 16 bits to both the high and low bytes of Timer3/5/7 at once.

The high byte of Timer3/5/7 is not directly readable or writable in this mode. All reads and writes must take place through the Timerx High Byte Buffer register.

Writes to TMRxH do not clear the Timer3/5/7 prescaler. The prescaler is only cleared on writes to TMRxL.

16.4 Using the SOSC Oscillator as the Timer3/5/7 Clock Source

The SOSC internal oscillator may be used as the clock source for Timer3/5/7. The SOSC oscillator is enabled by any peripheral that requests it. There are eight ways the SOSC can be enabled: if the SOSC is selected as the source by any of the odd timers, which is done by each respective SOSCEN bit (TxCON<3>), if the SOSC is selected as the RTCC source by the RTCOSC Configuration bit (CONFIG3L<1>), if the SOSC is selected as the CPU clock source by the SCS bits (OSCCON<1:0>) or if the SOSCGO bit is set (OSCCON2<3>). The SOSCGO bit is used to warm up the SOSC so that it is ready before any peripheral requests it. To use it as the Timer3/5/7 clock source, the TMRxCS bit must also be set. As previously noted, this also configures Timer3/5/7 to increment on every rising edge of the oscillator source.

The SOSC oscillator is described in Section 14.5 "SOSC Oscillator".

16.5 Timer3/5/7 Gates

Timer3/5/7 can be configured to count freely or the count can be enabled and disabled using the Timer3/5/7 gate circuitry. This is also referred to as the Timer3/5/7 gate count enable.

The Timer3/5/7 gate can also be driven by multiple selectable sources.

16.5.1 TIMER3/5/7 GATE COUNT ENABLE

The Timerx Gate Enable mode is enabled by setting the TMRxGE bit (TxGCON<7>). The polarity of the Timerx Gate Enable mode is configured using the TxGPOL bit (TxGCON<6>).

When Timerx Gate Enable mode is enabled, Timer3/5/7 will increment on the rising edge of the Timer3/5/7 clock source. When Timerx Gate Enable mode is disabled, no incrementing will occur and Timer3/5/7 will hold the current count. See Figure 16-2 for timing details.

TABLE 16-1: TIMER3/5/7 GATE ENABLE SELECTIONS

TxCLK ^(†)	TxGPOL (TxGCON<6>)	TxG Pin	Timerx Operation
1	0	0	Counts
\uparrow	0	1	Holds Count
\uparrow	1	0	Holds Count
\uparrow	1	1	Counts

† The clock on which TMR3/5/7 is running. For more information, see TxCLK in Figure 16-1.

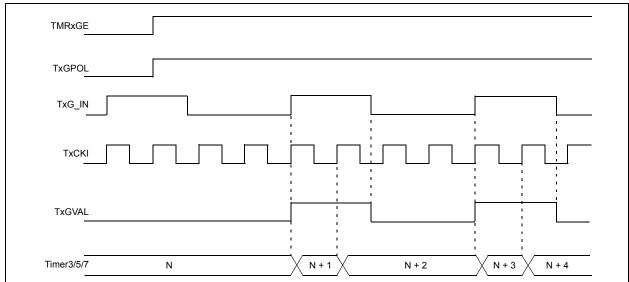


FIGURE 16-2: TIMER3/5/7 GATE COUNT ENABLE MODE

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16.5.2 TIMER3/5/7 GATE SOURCE SELECTION

The Timer3/5/7 gate source can be selected from one of four different sources. Source selection is controlled by the TxGSS<1:0> bits (TxGCON<1:0>). The polarity for each available source is also selectable and is controlled by the TxGPOL bit (TxGCON <6>).

TABLE 16-2:	TIMER3/5/7 GATE SOURCES
TADLE 10-2.	TIMENS/S/T GALE SOURCES

TxGSS<1:0>	Timerx Gate Source
00	Timerx Gate Pin
01	TMR(x+1) to Match PR(x+1) (TMR(x+1) increments to match PR(x+1))
10	Comparator 1 Output (comparator logic high output)
11	Comparator 2 Output (comparator logic high output)

16.5.2.1 TxG Pin Gate Operation

The TxG pin is one source for Timer3/5/7 gate control. It can be used to supply an external source to the Timerx gate circuitry.

16.5.2.2 Timer4/6/8 Match Gate Operation

The TMR(x+1) register will increment until it matches the value in the PR(x+1) register. On the very next increment cycle, TMR2 will be reset to 00h. When this Reset occurs, a low-to-high pulse will automatically be generated and internally supplied to the Timerx gate circuitry. The pulse will remain high for one instruction cycle and will return back to a low state until the next match.

Depending on TxGPOL, Timerx increments differently when TMR(x+1) matches PR(x+1). When TxGPOL = 1, Timerx increments for a single instruction

cycle following a TMR(x+1) match with PR(x+1). When TxGPOL = 0, Timerx increments continuously, except for the cycle following the match, when the gate signal goes from low-to-high.

16.5.2.3 Comparator 1 Output Gate Operation

The output of Comparator 1 can be internally supplied to the Timerx gate circuitry. After setting up Comparator 1 with the CM1CON register, Timerx will increment depending on the transitions of the CMP1OUT (CMSTAT<5>) bit.

16.5.2.4 Comparator 2 Output Gate Operation

The output of Comparator 2 can be internally supplied to the Timerx gate circuitry. After setting up Comparator 2 with the CM2CON register, Timerx will increment depending on the transitions of the CMP2OUT (CMSTAT<6>) bit.

16.5.3 TIMER3/5/7 GATE TOGGLE MODE

When Timer3/5/7 Gate Toggle mode is enabled, it is possible to measure the full cycle length of a Timer3/5/7 gate signal, as opposed to the duration of a single level pulse.

The Timerx gate source is routed through a flip-flop that changes state on every incrementing edge of the signal. (For timing details, see Figure 16-3.)

The TxGVAL bit will indicate when the Toggled mode is active and the timer is counting.

Timer3/5/7 Gate Toggle mode is enabled by setting the TxGTM bit (TxGCON<5>). When the TxGTM bit is cleared, the flip-flop is cleared and held clear. This is necessary in order to control which edge is measured.

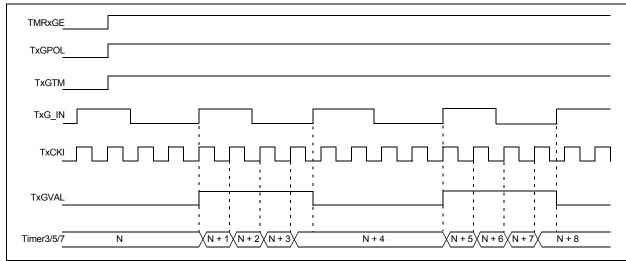


FIGURE 16-3: TIMER3/5/7 GATE TOGGLE MODE

16.5.4 TIMER3/5/7 GATE SINGLE PULSE MODE

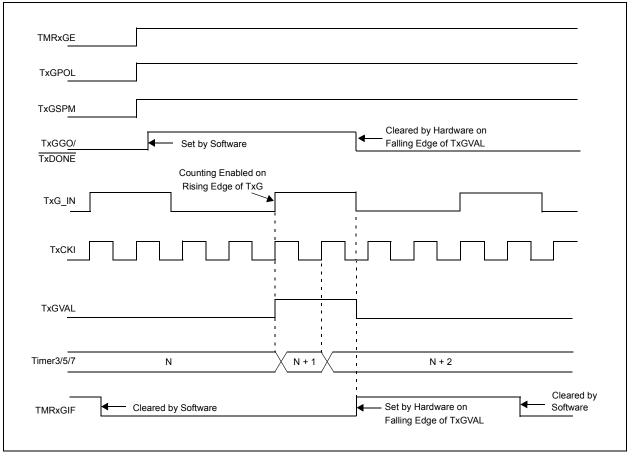
When Timer3/5/7 Gate Single Pulse mode is enabled, it is possible to capture a single pulse gate event. Timer3/5/7 Gate Single Pulse mode is first enabled by setting the TxGSPM bit (TxGCON<4>). Next, the TxGGO/TxDONE bit (TxGCON<3>) must be set.

The Timer3/5/7 will be fully enabled on the next incrementing edge. On the next trailing edge of the pulse, the TxGGO/TxDONE bit will automatically be cleared. No other gate events will <u>be allowed</u> to increment Timer3/5/7 until the TxGGO/TxDONE bit is once again set in software.

Clearing the TxGSPM bit also will clear the TxGGO/ TxDONE bit. (For timing details, see Figure 16-4.)

Simultaneously enabling the Toggle mode and the Single Pulse mode will permit both sections to work together. This allows the cycle times on the Timer3/5/7 gate source to be measured. (For timing details, see Figure 16-5.)

FIGURE 16-4: TIMER3/5/7 GATE SINGLE PULSE MODE



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TMRxGE						
TxGSPM						
TxGTM						
TxGG <u>O/</u> TxDONE	Set by Software				Cleared by Falling Edge	
TxG_IN	Rising Edge of Tx	G				
ТхСКІ						
TxGVAL						
ïmer3/5/7	Ν	N + 1	N + 2	N + 3	N + 4	
	 Cleared by Software 			Hardware on le of TxGVAL	¦ • • • •	Cleared by Software

FIGURE 16-5: TIMER3/5/7 GATE SINGLE PULSE AND TOGGLE COMBINED MODE

16.5.5 TIMER3/5/7 GATE VALUE STATUS

When Timer3/5/7 gate value status is utilized, it is possible to read the most current level of the gate control value. The value is stored in the TxGVAL bit (TxGCON<2>). The TxGVAL bit is valid even when the Timer3/5/7 gate is not enabled (TMRxGE bit is cleared).

16.5.6 TIMER3/5/7 GATE EVENT INTERRUPT

When the Timer3/5/7 gate event interrupt is enabled, it is possible to generate an interrupt upon the completion of a gate event. When the falling edge of TxGVAL occurs, the TMRxGIF flag bit in the PIRx register will be set. If the TMRxGIE bit in the PIEx register is set, then an interrupt will be recognized.

The TMRxGIF flag bit operates even when the Timer3/5/7 gate is not enabled (TMRxGE bit is cleared).

16.6 Timer3/5/7 Interrupt

The TMRx register pair (TMRxH:TMRxL) increments from 0000h to FFFFh and overflows to 0000h. The Timerx interrupt, if enabled, is generated on overflow and is latched in the interrupt flag bit, TMRxIF. Table 16-3 gives each module's flag bit.

TABLE 16-3:TIMER3/5/7 INTERRUPTFLAG BITS

Timer Module	Flag Bit
3	PIR2<1>
5	PIR5<1>
7	PIR5<3>

This interrupt can be enabled or disabled by setting or clearing the TMRxIE bit, respectively. Table 16-4 gives each module's enable bit.

TABLE 16-4:TIMER3/5/7 INTERRUPT
ENABLE BITS

Timer Module	Flag Bit
3	PIE2<1>
5	PIE5<1>
7	PIE5<3>

16.7 Resetting Timer3/5/7 Using the ECCP Special Event Trigger

If the ECCP modules are configured to use Timerx and to generate a Special Event Trigger in Compare mode (CCPxM<3:0> = 1011), this signal will reset Timerx. The trigger from ECCP2 will also start an A/D conversion if the A/D module is enabled. (For more information, see Section 20.3.4 "Special Event Trigger".)

The module must be configured as either a timer or synchronous counter to take advantage of this feature. When used this way, the CCPRxH:CCPRxL register pair effectively becomes a Period register for Timerx.

If Timerx is running in Asynchronous Counter mode, the Reset operation may not work.

In the event that a write to Timerx coincides with a Special Event Trigger from an ECCP module, the write will take precedence.

Note:	The Special Event Triggers from the				
	ECCPx module will only clear the TMR3				
	register's content, but not set the TMR3IF				
	interrupt flag bit (PIR1<0>).				

Note: The CCP and ECCP modules use Timers, 1 through 8, for some modes. The assignment of a particular timer to a CCP/ECCP module is determined by the Timer to CCP enable bits in the CCPTMRSx registers. For more details, see Register 19-2, Register 19-3 and Register 20-2

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TABLE 10-	ABLE 16-5: REGISTERS ASSOCIATED WITH TIMER3/5/7 AS A TIMER/COUNTER							
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR5	TMR7GIF ⁽¹⁾	TMR12IF ⁽¹⁾	TMR10IF ⁽¹⁾	TMR8IF	TMR7IF ⁽¹⁾	TMR6IF	TMR5IF	TMR4IF
IPR5	TMR7GIP ⁽¹⁾	TMR12IP ⁽¹⁾	TMR10IP ⁽¹⁾	TMR8IP	TMR7IP ⁽¹⁾	TMR6IP	TMR5IP	TMR4IP
PIE5	TMR7GIE ⁽¹⁾	TMR12IE ⁽¹⁾	TMR10IE ⁽¹⁾	TMR8IE	TMR7IE ⁽¹⁾	TMR6IE	TMR5IE	TMR4IE
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE2	OSCFIE	_	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE
PIR2	OSCFIF	_	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF
IPR2	OSCFIP	_	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP
TMR3H	Timer3 Regist	er High Byte						
TMR3L	Timer3 Regist	er Low Byte						
T3GCON	TMR3GE	T3GPOL	T3GTM	T3GSPM	T3GGO/ T3DONE	T3GVAL	T3GSS1	T3GSS0
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	SOSCEN	T3SYNC	RD16	TMR3ON
TMR5H	Timer5 Regist	er High Byte						
TMR5L	Timer5 Regist	er Low Byte						
T5GCON	TMR5GE	T5GPOL	T5GTM	T5GSPM	T5GGO/ T5DONE	T5GVAL	T5GSS1	T5GSS0
T5CON	TMR5CS1	TMR5CS0	T5CKPS1	T5CKPS0	SOSCEN	T5SYNC	RD16	TMR5ON
TMR7H ⁽¹⁾) Timer7 Register High Byte							
TMR7L ⁽¹⁾	Timer7 Regist	er Low Byte						
T7GCON ⁽¹⁾	TMR7GE	T7GPOL	T7GTM	T7GSPM	T7GGO/ T7DONE	T7GVAL	T7GSS1	T7GSS0
T7CON ⁽¹⁾	TMR7CS1	TMR7CS0	T7CKPS1	T7CKPS0	SOSCEN	T7SYNC	RD16	TMR7ON
OSCCON2	—	SOSCRUN	—	_	SOSCGO	—	MFIOFS	MFIOSEL
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD
PMD2	TMR10MD ⁽¹⁾	TMR8MD	TMR7MD ⁽¹⁾	TMR6MD	TMR5MD	CMP3MD	CMP2MD	CMP2MD

TABLE 16-5: REGISTERS ASSOCIATED WITH TIMER3/5/7 AS A TIMER/COUNTER

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer3/5/7 module.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

17.0 TIMER4/6/8/10/12 MODULES

The Timer4/6/8/10/12 timer modules have the following features:

- Eight-bit Timer register (TMRx)
- Eight-bit Period register (PRx)
- Readable and writable (all registers)
- Software programmable prescaler (1:1, 1:4, 1:16)
- Software programmable postscaler (1:1 to 1:16)
- Interrupt on TMRx match of PRx

Timer10 and Timer12 are unimplemented for devices with a program memory of 32 Kbytes (PIC18FX5K22).

Note:	Throughout this section, generic references
	are used for register and bit names that are the
	same, except for an 'x' variable that indicates
	the item's association with the Timer4, Timer6,
	Timer8, Timer10 or Timer12 module. For
	example, the control register is named TxCON
	and refers to T4CON, T6CON, T8CON,
	T10CON and T12CON.

The Timer4/6/8/10/12 modules have a control register, which is shown in Register 17-1. Timer4/6/8/10/12 can be shut off by clearing control bit, TMRxON (TxCON<2>), to minimize power consumption. The prescaler and post-scaler selection of Timer4/6/8/10/12 are also controlled by this register. Figure 17-1 is a simplified block diagram of the Timer4/6/8/10/12 modules.

17.1 Timer4/6/8/10/12 Operation

Timer4/6/8/10/12 can be used as the PWM time base for the PWM mode of the ECCP modules. The TMRx registers are readable and writable, and are cleared on any device Reset. The input clock (Fosc/4) has a prescale option of 1:1, 1:4 or 1:16, selected by control bits, TxCKPS<1:0> (TxCON<1:0>). The match output of TMRx goes through a four-bit postscaler (that gives a 1:1 to 1:16 inclusive scaling) to generate a TMRx interrupt, latched in the flag bit, TMRxIF. Table 17-1 shows each module's flag bit.

Timer Module	Flag Bit PIR5 <x></x>	Timer Module	Flag Bit PIR5 <x></x>
4	0	10	5
6	2	12	6
8	4		

TABLE 17-1: TIMER4/6/8/10/12 FLAG BITS

The interrupt can be enabled or disabled by setting or clearing the Timerx Interrupt Enable bit (TMRxIE), shown in Table 17-2.

TABLE 17-2:	TIMER4/6/8/10/12 INTERRUPT
	ENABLE BITS

Timer Module	Flag Bit PIE5 <x></x>	Timer Module	Flag Bit PIE5 <x></x>
4	0	10	5
6	2	12	6
8	4		

The prescaler and postscaler counters are cleared when any of the following occurs:

- A write to the TMRx register
- A write to the TxCON register
- <u>Any device Reset</u> Power-on Reset (POR), MCLR Reset, Watchdog Timer Reset (WDTR) or Brown-out Reset (BOR)

A TMRx is not cleared when a TxCON is written.

Note: The CCP and ECCP modules use Timers, 1 through 8, for some modes. The assignment of a particular timer to a CCP/ECCP module is determined by the Timer to CCP enable bits in the CCPTMRSx registers. For more details, see Register 19-2, Register 19-3 and Register 20-2.

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REGISTER 17-1: TxCON: TIMERx CONTROL REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	
_	TxOUTPS3	TxOUTPS2	TxOUTPS1	TxOUTPS0	TMRxON	TxCKPS1	TxCKPS0	
bit 7							bit (
Legend:								
R = Readal	ble bit	W = Writable bit		U = Unimplemented bit, read as '0'				
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown		
bit 7	Unimplemen	ted: Read as 'o	כ'					
bit 6-3	TxOUTPS<3:0>: Timerx Output Postscale Select bits							
	0000 = 1:1 P	ostscale						
	0001 = 1:2 P	ostscale						

	•
	•
	1111 = 1:16 Postscale
bit 2	TMRxON: Timerx On bit
	1 = Timerx is on
	0 = Timerx is off
bit 1-0	TxCKPS<1:0>: Timerx Clock Prescale Select bits
	00 = Prescaler is 1
	01 = Prescaler is 4
	1x = Prescaler is 16

17.2 Timer4/6/8/10/12 Interrupt

The Timer4/6/8/10/12 modules have eight-bit Period registers, PRx, that are both readable and writable. Timer4/6/8/10/12 increment from 00h until they match PR4/6/8/10/12 and then reset to 00h on the next increment cycle. The PRx registers are initialized to FFh upon Reset.

17.3 Output of TMRx

The outputs of TMRx (before the postscaler) are used only as a PWM time base for the ECCP modules. They are not used as baud rate clocks for the MSSP modules as is the Timer2 output.

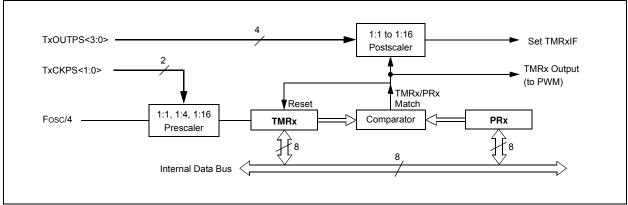


FIGURE 17-1: TIMER4 BLOCK DIAGRAM

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
IPR5	TMR7GIP ⁽¹⁾	TMR12IP ⁽¹⁾	TMR10IP ⁽¹⁾	TMR8IP	TMR7IP ⁽¹⁾	TMR6IP	TMR5IP	TMR4IP
PIR5	TMR7GIF ⁽¹⁾	TMR12IF ⁽¹⁾	TMR10IF ⁽¹⁾	TMR8IF	TMR7IF ⁽¹⁾	TMR6IF	TMR5IF	TMR4IF
PIE5	TMR7GIE ⁽¹⁾	TMR12IE ⁽¹⁾	TMR10IE ⁽¹⁾	TMR8IE	TMR7IE ⁽¹⁾	TMR6IE	TMR5IE	TMR4IE
TMR4	Timer4 Regis	ter						
T4CON	—	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR40N	T4CKPS1	T4CKPS0
PR4	Timer4 Period	d Register						
TMR6	Timer6 Regis	ter						
T6CON	_	T6OUTPS3	T6OUTPS2	T6OUTPS1	T6OUTPS0	TMR6ON	T6CKPS1	T6CKPS0
PR6	Timer6 Period	d Register						
TMR8	Timer8 Regis	ter						
T8CON	—	T8OUTPS3	T8OUTPS2	T8OUTPS1	T8OUTPS0	TMR8ON	T8CKPS1	T8CKPS0
PR8	Timer8 Period	d Register						
TMR10 ⁽¹⁾	Timer10 Regi	ster						
T10CON ⁽¹⁾	—	T10OUTPS3	T100UTPS2	T10OUTPS1	T10OUTPS0	TMR10ON	T10CKPS1	T10CKPS0
PR10 ⁽¹⁾	Timer10 Perio	od Register						
TMR12 ⁽¹⁾	Timer12 Regi	ster						
T12CON ⁽¹⁾	_	T12OUTPS3	T12OUTPS2	T12OUTPS1	T12OUTPS0	TMR12ON	T12CKPS1	T12CKPS0
PR12 ⁽¹⁾	Timer12 Perio	od Register						
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBMD
PMD2	TMR10MD ⁽¹⁾		TMR7MD ⁽¹⁾	TMR6MD	TMR5MD	CMP3MD	CMP2MD	CMP2MD
PMD3	CCP10MD ⁽¹⁾	CCP9MD ⁽¹⁾	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD ⁽¹⁾
					a a al las sitta a Tiu			

TABLE 17-3:	REGISTERS ASSOCIATED WITH TIMER4/6/8/10/12 AS A TIMER/COUNTER
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Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer4/6/8/10/12 module.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18FX5K22).

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NOTES:

18.0 REAL-TIME CLOCK AND CALENDAR (RTCC)

The key features of the Real-Time Clock and Calendar (RTCC) module are:

- · Time: hours, minutes and seconds
- Twenty-four hour format (military time)
- Calendar: weekday, date, month and year
- · Alarm configurable
- Year range: 2000 to 2099
- · Leap year correction
- · BCD format for compact firmware
- · Optimized for low-power operation
- · User calibration with auto-adjust
- Calibration range: ±2.64 seconds error per month
- Requirements: external 32.768 kHz clock crystal
- · Alarm pulse or seconds clock output on RTCC pin

The RTCC module is intended for applications where accurate time must be maintained for extended period with minimum to no intervention from the CPU. The module is optimized for low-power usage in order to provide extended battery life while keeping track of time.

The module is a 100-year clock and calendar with automatic leap year detection. The range of the clock is from 00:00:00 (midnight) on January 1, 2000 to 23:59:59 on December 31, 2099.

Hours are measured in 24-hour (military time) format. The clock provides a granularity of one second with half-second visibility to the user.

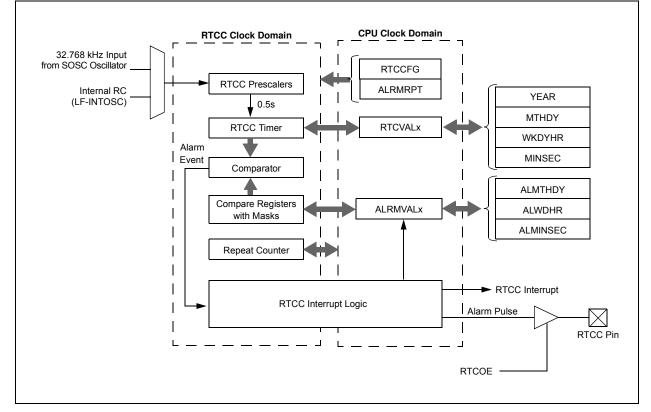


FIGURE 18-1: RTCC BLOCK DIAGRAM

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18.1 RTCC MODULE REGISTERS

The RTCC module registers are divided into the following categories:

RTCC Control Registers

- RTCCFG
- RTCCAL
- PADCFG1
- ALRMCFG
- ALRMRPT

RTCC Value Registers

- RTCVALH
- RTCVALL

Both registers access the following registers:

- YEAR
- MONTH
- DAY
- WEEKDAY
- HOUR
- MINUTE
- SECOND

Alarm Value Registers

- ALRMVALH
- ALRMVALL
 Both registers access the following registers:
 - ALRMMNTH
 - ALRMDAY
 - ALRMWD
 - ALRMHR
 - ALRMMIN
 - ALRMSEC
- Note: The RTCVALH and RTCVALL registers can be accessed through RTCRPT<1:0> (RTCCFG<1:0>). ALRMVALH and ALRMVALL can be accessed through ALRMPTR<1:0> (ALRMCFG<1:0>).

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18.1.1 RTCC CONTROL REGISTERS

R/W-0	U-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0
RTCEN ⁽²⁾	_	RTCWREN ⁽⁴⁾	RTCSYNC	HALFSEC ⁽³⁾	RTCOE	RTCPTR1	RTCPTR0
bit 7							bit C
Legend:							
R = Readable	e bit	W = Writable I	oit	U = Unimplem	ented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7		FCC Enable bit ⁽²⁾					
		module is enable	h				
		module is disable					
bit 6	Unimpleme	ented: Read as 'o)'				
bit 5	RTCWREN	: RTCC Value Re	gisters Write	Enable bit ⁽⁴⁾			
		ALH and RTCVAL					
		ALH and RTCVAL	•		•	en to by the use	r
bit 4		RTCC Value Reg	-	-			
		ALH, RTCVALL a in an invalid data					
		assumed to be v					
	0 = RTCVA	ALH, RTCVALL a	nd ALCFGR	PT registers ca	n be read wit	hout concern c	over a rollove
	ripple		(0)				
bit 3		Half-Second Stat					
		d half period of a sec					
bit 2		TCC Output Enab					
SIL 2		clock output is en					
		clock output is dis					
bit 1-0	RTCPTR<1	:0>: RTCC Value	Register Wir	ndow Pointer bit	S		
		he corresponding he RTCPTR<1:0					
	reaches '00				y roud or min		
	RTCVALH:						
	00 = Minute						
	01 = Week 10 = Month	,					
	11 = Reser						
	RTCVALL:						
	00 = Secon						
	01 = Hours						
	10 = Day 11 = Year						
Note 1: Th		gister is only affe	rted by a PO	P			
		CEN bit is only ane	-				
		only: it is cleared t				SEC register	

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- 3: This bit is read-only; it is cleared to '0' on a write to the lower half of the MINSEC register.
- 4: RTCWREN can only be written with the unlock sequence (see Example 18-1).

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R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0
bit 7							bit 0
Legend:							

REGISTER 18-2: RTCCAL: RTCC CALIBRATION REGISTER

Legena:					
R = Readable bit	W = Writable bit	U = Unimplemented bit,	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-0 CAL<7:0>: RTC Drift Calibration bits

.

01111111 = Maximum positive adjustment. Adds 508 RTC clock pulses every minute.

• 00000001 = Minimum positive adjustment. Adds four RTC clock pulses every minute. 00000000 = No adjustment
11111111 = Minimum negative adjustment. Subtracts four RTC clock pulses every minute.
10000000 = Maximum negative adjustment. Subtracts 512 RTC clock pulses every minute.

REGISTER 18-3: PADCFG1: PAD CONFIGURATION REGISTER

					-			
R/W-0	R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	U-0	
RDPU	REPU	RJPU ⁽²⁾	_		RTSECSEL1 ⁽¹	RTSECSEL0 ⁽¹⁾		
bit 7		1				1	bit 0	
Legend:								
R = Reada	ble bit	W = Writable b	bit	U = Unimplen	nented bit, read	as '0'		
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unknow	n	
bit 7	RDPU: POR	TD Pull-up Enal	ole bit					
	1 = PORTD	pull-up resistors	s are enabled	by individual p	oort latch values	6		
	0 = All POR	TD pull-up resis	tors are disat	oled				
bit 6	REPU: POR	TE Pull-up Enat	ole bit					
	1 = PORTE	pull-up resistors	are enabled	by individual p	oort latch values			
		TE pull-up resis		led				
bit 5	RJPU: POR	TJ Pull-up Enab	le bit ⁽²⁾					
		pull-up resistors			ort latch values			
		TJ pull-up resist						
bit 2-1	RTSECSEL	<1:0>: RTCC Se	econds Clock	Output Select	bits ⁽¹⁾			
		ed; do not use						
					oin can be LF-IN	ITOSC or SOSC, d	epending on	
	the RTCOSC (CONFIG3L<1>) bit setting) 01 = RTCC seconds clock is selected for the RTCC pin							
		alarm pulse is s						
bit 0		nted: Read as '						
	Gimpleme		5					
Note 1:	To enable the a	ctual RTCC outp	out, the RTCC	E (RTCCFG<	2>) bit must be	set.		
2:	Unimplemented	on 64-pin devic	es (PIC18F6)	XK22) read as	• 'O'			

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
ALRMEN	CHIME	AMASK3	AMASK2	AMASK1	AMASK0	ALRMPTR1	ALRMPTR0			
bit 7							bit (
Legend:										
R = Readable	e bit	W = Writable	bit	U = Unimpler	nented bit, rea	d as '0'				
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown			
bit 7		larm Enable bit								
		enabled (cleare	ed automatica	lly after an alar	m event when	ever ARPT<7:0	> = 00			
	and CHI 0 = Alarm is	,								
bit 6		ne Enable bit								
DILO		s enabled; ALRI	UDTD-1.05 h	ite are allowed	to roll over free	m 00h ta EEh				
		s disabled; ALR								
bit 5-2		>: Alarm Mask		•						
511 0 2		ry half second	comgaration	510						
	0001 = Eve									
		ry 10 seconds								
	0011 = Eve	•								
	0100 = Eve	ry 10 minutes								
	0101 = Eve									
	0110 = Onc	•								
	0111 = Onc									
	1000 = Onc		turbon config	urad for Cabru	and 20 th analy	over four veer				
		e a year (excep			ary 29 th , once o	every four years	5)			
	101x = Reserved – Do not use 11xx = Reserved – Do not use									
hit 1 0				lindow Dointor	hito					
bit 1-0	ALRMPTR<1:0>: Alarm Value Register Window Pointer bits									
	Points to the corresponding Alarm Value registers when reading the ALRMVALH and ALRMVALL registers. The ALRMVALH until it reaches									
	registers. The ALRMPTR<1:0> value decrements on every read or write of ALRMVALH until it reaches '00'.									
	ALRMVALH:									
	00 = ALRMM									
	01 = ALRMV									
	10 = ALRMN									
	11 = Unimple	emented								
	ALRMVALL:									
	00 = ALRMS	EC								
	01 = ALRMH									
	10 = ALRMD									
	11 = Unimple									

REGISTER 18-4: ALRMCFG: ALARM CONFIGURATION REGISTER

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REGISTER 18-5: ALRMRPT: ALARM REPEAT REGISTER

| R/W-0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| ARPT7 | ARPT6 | ARPT5 | ARPT4 | ARPT3 | ARPT2 | ARPT1 | ARPT0 |
| bit 7 | | | | | | | bit 0 |

Legend:				
R = Readable bit	dable bit W = Writable bit U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7-0 ARPT<7:0>: Alarm Repeat Counter Value bits

11111111 = Alarm will repeat 255 more times

. . 00000000 = Alarm will not repeat

The counter decrements on any alarm event. The counter is prevented from rolling over from 00h to FFh unless CHIME = 1.

18.1.2 RTCVALH AND RTCVALL REGISTER MAPPINGS

REGISTER 18-6: RESERVED REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	U-0	U-0
—	—	—	_	—	—	—	—
bit 7							bit 0

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7-0 Unimplemented: Read as '0'

REGISTER 18-7: YEAR: YEAR VALUE REGISTER⁽¹⁾

Lananda							
							bit 0
bit 7							bit 0
YRTEN3	YRTEN2	YRTEN1	YRTEN0	YRONE3	YRONE2	YRONE1	YRONE0
R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4	YRTEN<3:0>: Binary Coded Decimal Value of Year's Tens Digit bits
	Contains a value from 0 to 9.

bit 3-0 **YRONE<3:0>:** Binary Coded Decimal Value of Year's Ones Digit bits Contains a value from 0 to 9.

Note 1: A write to the YEAR register is only allowed when RTCWREN = 1.

REGISTER 18-8: MONTH: MONTH VALUE REGISTER⁽¹⁾

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	—	MTHTEN0	MTHONE3	MTHONE2	MTHONE1	MTHONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	d as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

Unimplemented: Read as '0'
MTHTEN0: Binary Coded Decimal Value of Month's Tens Digit bit
Contains a value of 0 or 1.
MTHONE<3:0>: Binary Coded Decimal Value of Month's Ones Digit bits
Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 18-9: DAY: DAY VALUE REGISTER⁽¹⁾

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	DAYTEN1	DAYTEN0	DAYONE3	DAYONE2	DAYONE1	DAYONE0
bit 7							bit 0

Legend:			
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	Unimplemented: Read as '0'
bit 5-4	DAYTEN<1:0>: Binary Coded Decimal value of Day's Tens Digit bits
	Contains a value from 0 to 3.
bit 3-0	DAYONE<3:0>: Binary Coded Decimal Value of Day's Ones Digit bits
	Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 18-10: WEEKDAY: WEEKDAY VALUE REGISTER⁽¹⁾

U-0	U-0	U-0	U-0	U-0	R/W-x	R/W-x	R/W-x
—	—			—	WDAY2	WDAY1	WDAY0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-3 Unimplemented: Read as '0'

bit 2-0 **WDAY<2:0>:** Binary Coded Decimal Value of Weekday Digit bits Contains a value from 0 to 6.

Note 1: A write to this register is only allowed when RTCWREN = 1.

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REGISTER 18-11: HOUR: HOUR VALUE REGISTER⁽¹⁾

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	HRTEN1	HRTEN0	HRONE3	HRONE2	HRONE1	HRONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	Unimplemented: Read as '0'
---------	----------------------------

bit 5-4	HRTEN<1:0>: Binary Coded Decimal Value of Hour's Tens Digit bits
	Contains a value from 0 to 2.
bit 3-0	HRONE<3:0>: Binary Coded Decimal Value of Hour's Ones Digit bits
	Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 18-12: MINUTE: MINUTE VALUE REGISTER

U-0	R/W-x						
_	MINTEN2	MINTEN1	MINTEN0	MINONE3	MINONE2	MINONE1	MINONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-4	MINTEN<2:0>: Binary Coded Decimal Value of Minute's Tens Digit bits
	Contains a value from 0 to 5.
bit 3-0	MINONE<3:0>: Binary Coded Decimal Value of Minute's Ones Digit bits
	Contains a value from 0 to 9.

REGISTER 18-13: SECOND: SECOND VALUE REGISTER

U-0	R/W-x						
—	SECTEN2	SECTEN1	SECTEN0	SECONE3	SECONE2	SECONE1	SECONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-4	SECTEN<2:0>: Binary Coded Decimal Value of Second's Tens Digit bits
	Contains a value from 0 to 5.
bit 3-0	SECONE<3:0>: Binary Coded Decimal Value of Second's Ones Digit bits Contains a value from 0 to 9.

18.1.3 ALRMVALH AND ALRMVALL REGISTER MAPPINGS

REGISTER 18-14: ALRMMNTH: ALARM MONTH VALUE REGISTER⁽¹⁾

U-0	U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
_	_	_	MTHTEN0	MTHONE3	MTHONE2	MTHONE1	MTHONE0
bit 7 bit 0							
Legend:							
R = Readable b	oit	W = Writable	bit	U = Unimpler	nented bit, read	as '0'	
-n = Value at POR '1' = Bit is set			'0' = Bit is cleared x = Bit is unknow			iown	
bit 7-5	Unimplemer	ted: Read as '	0'				

bit 4	MTHTEN0: Binary Coded Decimal Value of Month's Tens Digit bits
	Contains a value of 0 or 1.
bit 3-0	MTHONE<3:0>: Binary Coded Decimal Value of Month's Ones Digit bits
	Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 18-15: ALRMDAY: ALARM DAY VALUE REGISTER⁽¹⁾

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	—	DAYTEN1	DAYTEN0	DAYONE3	DAYONE2	DAYONE1	DAYONE0
bit 7							bit 0

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7-6	Unimplemented: Read as '0'
bit 5-4	DAYTEN<1:0>: Binary Coded Decimal Value of Day's Tens Digit bits
	Contains a value from 0 to 3.
bit 3-0	DAYONE<3:0>: Binary Coded Decimal Value of Day's Ones Digit bits
	Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 18-16: ALRMWD: ALARM WEEKDAY VALUE REGISTER⁽¹⁾

U-0	U-0	U-0	U-0	U-0	R/W-x	R/W-x	R/W-x
—	—	—	—	—	WDAY2	WDAY1	WDAY0
bit 7							bit 0
Legend:							
R = Readable b	oit	W = Writable	bit	U = Unimplemented bit, read as '0'			
-n = Value at POR (1' = Bit is set			'0' = Bit is cle	eared	x = Bit is unkr	nown	

bit 7-3 Unimplemented: Read as '0'

bit 2-0 **WDAY<2:0>:** Binary Coded Decimal Value of Weekday Digit bits Contains a value from 0 to 6.

Note 1: A write to this register is only allowed when RTCWREN = 1.

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REGISTER 18-17: ALRMHR: ALARM HOURS VALUE REGISTER⁽¹⁾

U-0	U-0	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
—	_	HRTEN1	HRTEN0	HRONE3	HRONE2	HRONE1	HRONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

Unimplemented: Read as '0'
HRTEN<1:0>: Binary Coded Decimal Value of Hour's Tens Digit bits
Contains a value from 0 to 2.
HRONE<3:0>: Binary Coded Decimal Value of Hour's Ones Digit bits

Contains a value from 0 to 9.

Note 1: A write to this register is only allowed when RTCWREN = 1.

REGISTER 18-18: ALRMMIN: ALARM MINUTES VALUE REGISTER

U-0	R/W-x						
—	MINTEN2	MINTEN1	MINTEN0	MINONE3	MINONE2	MINONE1	MINONE0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
-------	----------------------------

- bit 6-4 MINTEN<2:0>: Binary Coded Decimal Value of Minute's Tens Digit bits Contains a value from 0 to 5.
- bit 3-0 MINONE<3:0>: Binary Coded Decimal Value of Minute's Ones Digit bits Contains a value from 0 to 9.

REGISTER 18-19: ALRMSEC: ALARM SECONDS VALUE REGISTER

U-0	R/W-x						
—	SECTEN2	SECTEN1	SECTEN0	SECONE3	SECONE2	SECONE1	SECONE0
bit 7							bit 0

Legend:				
R = Reada	able bit	W = Writable bit	U = Unimplemented bit,	read as '0'
-n = Value at POR		'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown
bit 7 bit 6-4	SECTEN	mented: Read as '0' < 2:0>: Binary Coded Decin a value from 0 to 5.	nal Value of Second's Tens Dig	it bits
bit 3-0	SECONE	3:0>: Binary Coded Decir a value from 0 to 9.	nal Value of Second's Ones Di	git bits

18.1.4 RTCEN BIT WRITE

RTCWREN (RTCCFG<5>) must be set before a write to RTCEN can take place. Any write to the RTCEN bit, while RTCWREN = 0, will be ignored.

Like the RTCEN bit, the RTCVALH and RTCVALL registers can only be written to when RTCWREN = 1. A write to these registers, while RTCWREN = 0, will be ignored.

FIGURE 18-2: TIMER DIGIT FORMAT

18.2 Operation

18.2.1 REGISTER INTERFACE

The register interface for the RTCC and alarm values is implemented using the Binary Coded Decimal (BCD) format. This simplifies the firmware when using the module, as each of the digits is contained within its own 4-bit value (see Figure 18-2 and Figure 18-3).

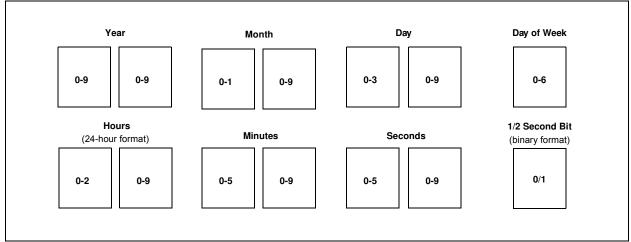
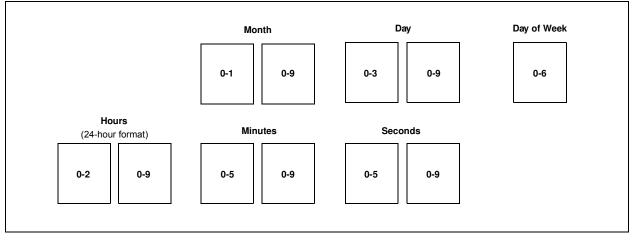


FIGURE 18-3: ALARM DIGIT FORMAT



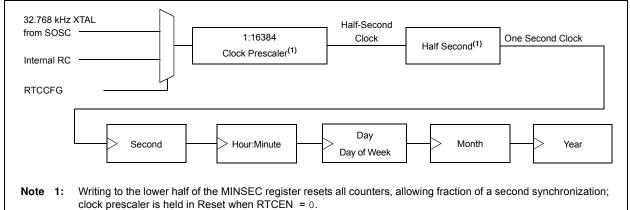
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18.2.2 CLOCK SOURCE

As previously mentioned, the RTCC module is intended to be clocked by an external Real-Time Clock (RTC) crystal, oscillating at 32.768 kHz, but an internal oscillator can be used. The RTCC clock selection is decided by the RTCOSC bit (CONFIG3L<0>).

Calibration of the crystal can be done through this module to yield an error of 3 seconds or less per month. (For further details, see Section 18.2.9 "Calibration".)





18.2.2.1 Real-Time Clock Enable

The RTCC module can be clocked by an external, 32.768 kHz crystal (SOSC oscillator) or the LF-INTOSC oscillator, which can be selected in CONFIG3L<0>.

If the external clock is used, the SOSC oscillator should be enabled. If LF-INTOSC is providing the clock, the INTOSC clock can be brought out to the RTCC pin by the RTSECSEL<1:0> bits (PADCFG<2:1>).

18.2.3 DIGIT CARRY RULES

This section explains which timer values are affected when there is a rollover:

- Time of Day: From 23:59:59 to 00:00:00 with a carry to the Day field
- Month: From 12/31 to 01/01 with a carry to the Year field
- Day of Week: From 6 to 0 with no carry (see Table 18-1)
- Year Carry: From 99 to 00; this also surpasses the use of the RTCC

For the day-to-month rollover schedule, see Table 18-2.

Because the following values are in BCD format, the carry to the upper BCD digit occurs at the count of 10, not 16 (SECONDS, MINUTES, HOURS, WEEKDAY, DAYS and MONTHS).

TABLE 18-1: DAY OF WEEK SCHEDULE

Day of Week							
Sunday	0						
Monday	1						
Tuesday	2						
Wednesday	3						
Thursday	4						
Friday	5						
Saturday	6						

TABLE 18-2:DAY TO MONTH ROLLOVER
SCHEDULE

Maximum Day Field
31
28 or 29 ⁽¹⁾
31
30
31
30
31
31
30
31
30
31

Note 1: See Section 18.2.4 "Leap Year".

18.2.4 LEAP YEAR

Since the year range on the RTCC module is 2000 to 2099, the leap year calculation is determined by any year divisible by four in the above range. Only February is affected in a leap year.

February will have 29 days in a leap year and 28 days in any other year.

18.2.5 GENERAL FUNCTIONALITY

All Timer registers containing a time value of seconds or greater are writable. The user configures the time by writing the required year, month, day, hour, minutes and seconds to the Timer registers, via register pointers. (See Section 18.2.8 "Register Mapping".)

The timer uses the newly written values and proceeds with the count from the required starting point.

The RTCC is enabled by setting the RTCEN bit (RTCCFG<7>). If enabled, while adjusting these registers, the timer still continues to increment. However, any time the MINSEC register is written to, both of the timer prescalers are reset to '0'. This allows fraction of a second synchronization.

The Timer registers are updated in the same cycle as the write instruction's execution by the CPU. The user must ensure that when RTCEN = 1, the updated registers will not be incremented at the same time. This can be accomplished in several ways:

- By checking the RTCSYNC bit (RTCCFG<4>)
- By checking the preceding digits from which a carry can occur
- By updating the registers immediately following the seconds pulse (or an alarm interrupt)

The user has visibility to the half-second field of the counter. This value is read-only and can be reset only by writing to the lower half of the SECONDS register.

18.2.6 SAFETY WINDOW FOR REGISTER READS AND WRITES

The RTCSYNC bit indicates a time window during which the RTCC Clock Domain registers can be safely read and written without concern about a rollover. When RTCSYNC = 0, the registers can be safely accessed by the CPU.

Whether RTCSYNC = 1 or 0, the user should employ a firmware solution to ensure that the data read did not fall on a rollover boundary, resulting in an invalid or partial read. This firmware solution would consist of reading each register twice and then comparing the two values. If the two values matched, then a rollover did not occur.

18.2.7 WRITE LOCK

In order to perform a write to any of the RTCC Timer registers, the RTCWREN bit (RTCCFG<5>) must be set.

To avoid accidental writes to the RTCC Timer register, it is recommended that the RTCWREN bit (RTCCFG<5>) be kept clear when not writing to the register. For the RTCWREN bit to be set, there is only one instruction cycle time window allowed between the 55h/AA sequence and the setting of RTCWREN. For that reason, it is recommended that users follow the code example in Example 18-1.

EXAMPLE 18-1: SETTING THE RTCWREN BIT

movlw	0x55
movwf	EECON2
movlw	0xAA
movwf	EECON2
bsf	RTCCFG, RTCWREN

18.2.8 REGISTER MAPPING

To limit the register interface, the RTCC Timer and Alarm Timer registers are accessed through corresponding register pointers. The RTCC Value register window (RTCVALH and RTCVALL) uses the RTCPTRx bits (RTCCFG<1:0>) to select the required Timer register pair.

By reading or writing to the RTCVALH register, the RTCC Pointer value (RTCPTR<1:0>) decrements by '1' until it reaches '00'. When '00' is reached, the MINUTES and SECONDS value is accessible through RTCVALH and RTCVALL until the pointer value is manually changed.

TABLE 18-3: RTCVALH AND RTCVALL REGISTER MAPPING

RTCPTR<1:0>	RTCC Value Register Window				
nicrin<1.0>	RTCVALH	RTCVALL			
00	MINUTES	SECONDS			
01	WEEKDAY	HOURS			
10	MONTH	DAY			
11	—	YEAR			

The Alarm Value register windows (ALRMVALH and ALRMVALL) use the ALRMPTR bits (ALRMCFG<1:0>) to select the desired alarm register pair.

By reading or writing to the ALRMVALH register, the Alarm Pointer value, ALRMPTR<1:0>, decrements by one until it reaches '00'. When it reaches '00', the ALRMMIN and ALRMSEC values are accessible through ALRMVALH and ALRMVALL until the pointer value is manually changed.

TABLE 18-4:ALRMVAL REGISTER
MAPPING

ALRMPTR<1:0>	Alarm Value Register Window				
ALNMP IN<1.0>	ALRMVALH	ALRMVALL			
00	ALRMMIN	ALRMSEC			
01	ALRMWD	ALRMHR			
10	ALRMMNTH	ALRMDAY			
11	—	—			

18.2.9 CALIBRATION

The real-time crystal input can be calibrated using the periodic auto-adjust feature. When properly calibrated, the RTCC can provide an error of less than three seconds per month.

To perform this calibration, find the number of error clock pulses and store the value into the lower half of the RTCCAL register. The eight-bit, signed value, loaded into RTCCAL, is multiplied by four and will be either added or subtracted from the RTCC timer, once every minute.

To calibrate the RTCC module:

- 1. Use another timer resource on the device to find the error of the 32.768 kHz crystal.
- 2. Convert the number of error clock pulses per minute (see Equation 18-1).

EQUATION 18-1: CONVERTING ERROR CLOCK PULSES

(Ideal Frequency (32,758) – Measured Frequency) * 60 = Error Clocks per Minute

- If the oscillator is *faster* than ideal (negative result from Step 2), the RCFGCALL register value needs to be negative. This causes the specified number of clock pulses to be subtracted from the timer counter, once every minute.
- If the oscillator is *slower* than ideal (positive result from Step 2), the RCFGCALL register value needs to be positive. This causes the specified number of clock pulses to be added to the timer counter, once every minute.
- 3. Load the RTCCAL register with the correct value.

Writes to the RTCCAL register should occur only when the timer is turned off or immediately after the rising edge of the seconds pulse.

Note:	In determining the crystal's error value, it								
	is the user's responsibility to include the								
	crystal's initial error from drift due to								
	temperature or crystal aging.								

18.3 Alarm

The Alarm features and characteristics are:

- Configurable from half a second to one year
- Enabled using the ALRMEN bit (ALRMCFG<7>, Register 18-4)
- · Offers one-time and repeat alarm options

18.3.1 CONFIGURING THE ALARM

The alarm feature is enabled using the ALRMEN bit. This bit is cleared when an alarm is issued. The bit will not be cleared if the CHIME bit = $1 \text{ or if ALRMRPT} \neq 0$.

The interval selection of the alarm is configured through the ALRMCFG bits (AMASK<3:0>); see Figure 18-5. These bits determine which, and how many, digits of the alarm must match the clock value for the alarm to occur.

The alarm can also be configured to repeat based on a preconfigured interval. The number of times this occurs, after the alarm is enabled, is stored in the ALRMRPT register.

Note: While the alarm is enabled (ALRMEN = 1), changing any of the registers, other than the RTCCAL, ALRMCFG and ALRMRPT registers and the CHIME bit, can result in a false alarm event leading to a false alarm interrupt. To avoid this, only change the timer and alarm values while the alarm is disabled (ALRMEN = 0). It is recommended that the ALRMCFG and ALRMRPT registers and CHIME bit be changed when RTCSYNC = 0.



Alarm Mask Setting AMASK<3:0>	Day of the Week	e Month Day	Hours	Minutes Seconds
0000 – Every half second 0001 – Every second				:
0010 – Every 10 seconds				:
0011 – Every minute				: : : : :
0100 – Every 10 minutes				: m : s s
0101 – Every hour				: m m : s s
0110 – Every day			h h	• m m • s s
0111 – Every week	d		h h	: m m : s s
1000 – Every month		/d	d h h	: m m : s s
1001 – Every year ⁽¹⁾		m m / d	d h h	: m m : s s
Note 1: Annually, except when c	onfigured for	February 29.		

When ALRMCFG = 00 and the CHIME bit = 0 (ALRMCFG<6>), the repeat function is disabled and only a single alarm will occur. The alarm can be repeated up to 255 times by loading the ALRMRPT register with FFh.

After each alarm is issued, the ALRMRPT register is decremented by one. Once the register has reached '00', the alarm will be issued one last time.

After the alarm is issued a last time, the ALRMEN bit is cleared automatically and the alarm is turned off. Indefinite repetition of the alarm can occur if the CHIME bit = 1.

When CHIME = 1, the alarm is not disabled when the ALRMRPT register reaches '00', but it rolls over to FF and continues counting indefinitely.

18.3.2 ALARM INTERRUPT

At every alarm event, an interrupt is generated. Additionally, an alarm pulse output is provided that operates at half the frequency of the alarm.

The alarm pulse output is completely synchronous with the RTCC clock and can be used as a trigger clock to other peripherals. This output is available on the RTCC pin. The output pulse is a clock with a 50% duty cycle and a frequency half that of the alarm event (see Figure 18-6).

The RTCC pin also can output the seconds clock. The user can select between the alarm pulse, generated by the RTCC module, or the seconds clock output.

The RTSECSEL<1:0> bits (PADCFG1<2:1>) select between these two outputs:

- Alarm pulse RTSECSEL<1:0> = 00
- Seconds clock RTSECSEL<1:0> = 01

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FIGURE 18-6: TIMER PULSE GENERATION

RTCEN bit -	
ALRMEN bit	
RTCC Alarm Event	
RTCC Pin	

18.4 Sleep Mode

The timer and alarm continue to operate while in Sleep mode. The operation of the alarm is not affected by Sleep, as an alarm event can always wake up the CPU.

The Idle mode does not affect the operation of the timer or alarm.

18.5 Reset

18.5.1 DEVICE RESET

When a device Reset occurs, the ALRMRPT register is forced to its Reset state, causing the alarm to be disabled (if enabled prior to the Reset). If the RTCC was enabled, it will continue to operate when a basic device Reset occurs.

18.5.2 POWER-ON RESET (POR)

The RTCCFG and ALRMRPT registers are reset only on a POR. Once the device exits the POR state, the clock registers should be reloaded with the desired values.

The timer prescaler values can be reset only by writing to the SECONDS register. No device Reset can affect the prescalers.

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18.6 Register Maps

Table 18-5, Table 18-6 and Table 18-7 summarize theregisters associated with the RTCC module.

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RTCCFG	RTCEN	_	RTCWREN	RTCSYNC	HALFSEC	RTCOE	RTCPTR1	RTCPTR0
RTCCAL	CAL7	CAL6	CAL5	CAL4	CAL3	CAL2	CAL1	CAL0
PADCFG1	RDPU	REPU	RJPU ⁽¹⁾	—		RTSECSEL1	RTSECSEL0	_
ALRMCFG	ALRMEN	CHIME	AMASK3	AMASK2	AMASK1	AMASK0	ALRMPTR1	ALRMPTR0
ALRMRPT	ARPT7	ARPT6	ARPT5	ARPT4	ARPT3	ARPT2	ARPT1	ARPT0
PMD1	PSPMD	CTMUMD	RTCCMD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	EMBDM

TABLE 18-5: RTCC CONTROL REGISTERS

Legend: — = unimplemented, read as '0'. Reset values are shown in hexadecimal for 80-pin devices.

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

TABLE 18-6: RTCC VALUE REGISTERS

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
RTCVALH	RTCC Value	RTCC Value High Register Window Based on RTCPTR<1:0>							
RTCVALL	RTCC Value	RTCC Value Low Register Window Based on RTCPTR<1:0>							

TABLE 18-7: ALARM VALUE REGISTERS

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
ALRMVALH Alarm Value High Register Window Based on ALRMPTR<1:0>								
ALRMVALL	Alarm Value Low Register Window Based on ALRMPTR<1:0>							

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NOTES:

19.0 CAPTURE/COMPARE/PWM (CCP) MODULES

PIC18F87K22 family devices have seven CCP (Capture/Compare/PWM) modules, designated CCP4 through CCP10. All the modules implement standard Capture, Compare and Pulse-Width Modulation (PWM) modes.

Note: Throughout this section, generic references are used for register and bit names that are the same, except for an 'x' variable that indicates the item's association with the specific CCP module. For example, the control register is named CCPxCON and refers to CCP4CON through CCP10CON.

Each CCP module contains a 16-bit register that can operate as a 16-bit Capture register, a 16-bit Compare register or a PWM Master/Slave Duty Cycle register. For the sake of clarity, all CCP module operation in the following sections is described with respect to CCP4, but is equally applicable to CCP5 through CCP10.

Note: The CCP9 and CCP10 modules are disabled on the devices with 32 Kbytes of program memory (PIC18FX5K22).

REGISTER 19-1: CCPxCON: CCPx CONTROL REGISTER (CCP4-CCP10 MODULES)⁽¹⁾

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PxM1	PxM0	DCxB1	DCxB0	CCPxM3 ⁽²⁾	CCPxM2 ⁽²⁾	CCPxM1 ⁽²⁾	CCPxM0 ⁽²⁾
bit 7 bit 0							

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-6	PxM<1:0>: PWM Output Configuration bits
	If CCPxM<3:2> = 00, 01, 10:
	xx = PxA is assigned as a capture/compare input/output; PxB, PxC and PxD are assigned as port pins
	<u>If CCPxM<3:2> = 11:</u>
	00 = Single output: PxA, PxB, PxC and PxD are controlled by steering
	01 = Full-bridge output forward: PxD is modulated; PxA is active; PxB, PxC are inactive
	10 = Half-bridge output: PxA, PxB are modulated with dead-band control; PxC and PxD are assigned as port pins
	11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive
bit 5-4	DCxB<1:0>: PWM Duty Cycle bit 1 and bit 0 for CCPx Module
	Capture mode:
	Unused.
	Compare mode:
	Unused.
	PWM mode:
	These bits are the two Least Significant bits (bit 1 and bit 0) of the 10-bit PWM duty cycle. The eight
	Most Significant bits (DCx<9:2>) of the duty cycle are found in CCPRxL.
Note 1:	The CCP9 and CCP10 modules are not available on devices with 32 Kbytes of program memory

- (PIC18FX5K22).
- 2: CCPxM<3:0> = 1011 will only reset the timer and not start AN A/D conversion on CCPx match.

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REGISTER 19-1: CCPxCON: CCPx CONTROL REGISTER (CCP4-CCP10 MODULES)⁽¹⁾

bit 3-0 CCPxM<3:0>: CCPx Module Mode Select bits⁽²⁾

- 0000 = Capture/Compare/PWM disabled (resets CCPx module)
- 0001 = Reserved
- 0010 = Compare mode, toggle output on match (CCPxIF bit is set)
- 0011 = Reserved
- 0100 = Capture mode: every falling edge
- 0101 = Capture mode: every rising edge
- 0110 = Capture mode: every 4th rising edge
- 0111 = Capture mode: every 16th rising edge
- 1000 = Compare mode: initialize CCPx pin low; on compare match, force CCPx pin high (CCPxIF bit is set)
- 1001 = Compare mode: initialize CCPx pin high; on compare match, force CCPx pin low (CCPxIF bit is set)
- 1010 = Compare mode: generate software interrupt on compare match (CCPxIF bit is set, CCPx pin reflects I/O state)
- 1011 = Compare mode: Special Event Trigger; reset timer on CCPx match (CCPxIF bit is set)
- 11xx = PWM mode

2: CCPxM<3:0> = 1011 will only reset the timer and not start AN A/D conversion on CCPx match.

REGISTER 19-2: CCPTMRS1: CCP TIMER SELECT REGISTER 1

R/W-0	R/W-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
C7TSEL1	C7TSEL0	—	C6TSEL0	—	C5TSEL0	C4TSEL1	C4TSEL0
bit 7 bit 0							

Legend:						
R = Readable bit V		W = Writable bit	U = Unimplemented bit,	U = Unimplemented bit, read as '0'		
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		
bit 7-6	00 = CC 01 = CC	<1:0>: CCP7 Timer Selectior P7 is based off of TMR1/TMF P7 is based off of TMR5/TMF P7 is based off of TMR5/TMF	82 84			
		P7 is based off of TMR5/TMF	88			
bit 5	Unimple	emented: Read as '0'				
bit 4	0 = CCF	0: CCP6 Timer Selection bit P6 is based off of TMR1/TMR P6 is based off of TMR5/TMR				
bit 3	Unimple	mented: Read as '0'				
bit 2	C5TSEL0: CCP5 Timer Selection bit 0 = CCP5 is based off of TMR1/TMR2 1 = CCP5 is based off of TMR5/TMR4					
bit 1-0	00 = CC 01 = CC 10 = CC	<1:0>: CCP4 Timer Selection P4 is based off of TMR1/TMF P4 is based off of TMR3/TMF P4 is based off of TMR3/TMF served; do not use	82 84			

Note 1: The CCP9 and CCP10 modules are not available on devices with 32 Kbytes of program memory (PIC18FX5K22).

REGISTER 19-3: CCPTMRS2: CCP TIMER SELECT REGISTER 2

U-0	U-0	U-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0
—	—		C10TSEL0 ⁽¹⁾	—	C9TSEL0 ⁽¹⁾	C8TSEL1	C8TSEL0
bit 7							bit 0

W = Writable bit	U = Unimplemented bit, read as '0'				
'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown			
plemented: Read as '0'					
•		'1' = Bit is set '0' = Bit is cleared			

bit 4	C10TSEL0: CCP10 Timer Selection bit ⁽¹⁾
	0 = CCP10 is based off of TMR1/TMR2
	1 = CCP10 is based off of TMR7/TMR2
bit 3	Unimplemented: Read as '0'
bit 2	C9TSEL0: CCP9 Timer Selection bit ⁽¹⁾
	0 = CCP9 is based off of TMR1/TMR2
	1 = CCP9 is based off of TMR7/TMR4
bit 1-0	C8TSEL<1:0>: CCP8 Timer Selection bits
	On Non 32-Byte Device Variants:
	00 = CCP8 is based off of TMR1/TMR2
	01 = CCP8 is based off of TMR7/TMR4
	10 = CCP8 is based off of TMR7/TMR6
	11 = Reserved; do not use
	On 32-Byte Device Variants (PIC18F65K22 and PIC18F85K22):
	00 = CCP8 is based off of TMR1/TMR2
	01 = CCP8 is based off of TMR1/TMR4
	10 = CCP8 is based off of TMR1/TMR6
	11 = Reserved; do not use

Note 1: This bit is unimplemented and reads as '0' on devices with 32 Kbytes of program memory (PIC18FX5K22).

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REGISTER 19-4: CCPRxL: CCPx PERIOD LOW BYTE REGISTER

| R/W-x |
|---------|---------|---------|---------|---------|---------|---------|---------|
| CCPRxL7 | CCPRxL6 | CCPRxL5 | CCPRxL4 | CCPRxL3 | CCPRxL2 | CCPRxL1 | CCPRxL0 |
| bit 7 | | | | | | | bit 0 |

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 CCPRxL<7:0>: CCPx Period Register Low Byte bits Capture Mode: Capture Register Low Byte Compare Mode: Compare Register Low Byte PWM Mode: Duty Cycle Register

REGISTER 19-5: CCPRxH: CCPx PERIOD HIGH BYTE REGISTER

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
CCPRxH7	CCPRxH6	CCPRxH5	CCPRxH4	CCPRxH3	CCPRxH2	CCPRxH1	CCPRxH0
bit 7 bit 0							

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 CCPRxH<7:0>: CCPx Period Register High Byte bits Capture Mode: Capture Register High Byte Compare Mode: Compare Register High Byte PWM Mode: Duty Cycle Buffer Register

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19.1 CCP Module Configuration

Each Capture/Compare/PWM module is associated with a control register (generically, CCPxCON) and a data register (CCPRx). The data register, in turn, is comprised of two 8-bit registers: CCPRxL (low byte) and CCPRxH (high byte). All registers are both readable and writable.

19.1.1 CCP MODULES AND TIMER RESOURCES

The CCP modules utilize Timers, 1 through 8, which vary with the selected mode. Various timers are available to the CCP modules in Capture, Compare or PWM modes, as shown in Table 19-1.

TABLE 19-1: CCP MODE – TIMER RESOURCE

CCP Mode	Timer Resource				
Capture	Timeri Timeri Timeri 5 en Timeri				
Compare	Timer1, Timer3, Timer 5 or Timer7				
PWM	Timer2, Timer4, Timer 6 or Timer8				

The assignment of a particular timer to a module is determined by the timer to CCP enable bits in the CCPTMRSx registers (see Register 19-2 and Register 19-3). All of the modules may be active at once and may share the same timer resource if they are configured to operate in the same mode (Capture/Compare or PWM) at the same time.

The CCPTMRS1 register selects the timers for CCP modules, 7, 6, 5 and 4, and the CCPTMRS2 register selects the timers for CCP modules, 10, 9 and 8. The possible configurations are shown in Table 19-2 and Table 19-3.

TABLE 19-2: TIMER ASSIGNMENTS FOR CCP MODULES 4, 5, 6 AND 7

	CCPTMRS1 Register										
	CCP4 CCP5			CCP6			CCP7				
C4TSEL <1:0>	Capture/ Compare Mode	PWM Mode	C5TSEL0	Capture/ Compare Mode	PWM Mode	C6TSEL0	Capture/ Compare Mode	PWM Mode	C7TSEL <1:0>	Capture/ Compare Mode	PWM Mode
0 0	TMR1	TMR2	0	TMR1	TMR2	0	TMR1	TMR2	0 0	TMR1	TMR2
0 1	TMR3	TMR4	1	TMR5	TMR4	1	TMR5	TMR2	0 1	TMR5	TMR4
1 0	TMR3	TMR6							1 0	TMR5	TMR6
1 1 Reserved ⁽¹⁾									1 1	TMR5	TMR8

Note 1: Do not use the reserved bits.

TABLE 19-3: TIMER ASSIGNMENTS FOR CCP MODULES 8, 9 AND 10

	CCPTMRS2 Register										
CCP8 CCP8 Devices with 32 Kbytes				CCP9 ⁽¹⁾		c	CP10 ⁽¹⁾				
C8TSEL <1:0>	Capture/ Compare Mode	PWM Mode	C8TSEL <1:0>	Capture/ Compare Mode	PWM Mode	C9TSEL0	Capture/ Compare Mode	PWM Mode	C10TSEL0	Capture/ Compare Mode	PWM Mode
0 0	TMR1	TMR2	0 0	TMR1	TMR2	0	TMR1	TMR2	0	TMR1	TMR2
0 1	TMR7	TMR4	0 1	TMR1	TMR4	1	TMR7	TMR4	1	TMR7	TMR2
1 0	TMR7	TMR6	1 0	TMR1	TMR6						
1 1	Reserv	ed ⁽²⁾	1 1	Reserv	ed ⁽²⁾						

Note 1: The module is not available for devices with 32 Kbytes of program memory (PIC18F65K22 and PIC18F85K22).

2: Do not use the reserved setting.

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19.1.2 OPEN-DRAIN OUTPUT OPTION

When operating in Output mode (the Compare or PWM modes), the drivers for the CCPx pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor and allows the output to communicate with external circuits without the need for additional level shifters.

The open-drain output option is controlled by the CCPxOD bits (ODCON2<7:0>). Setting the appropriate bit configures the pin for the corresponding module for open-drain operation.

19.1.3 PIN ASSIGNMENT FOR CCP6, CCP7, CCP8 AND CCP9

The pin assignment for CCP6/7/8/9 (Capture input, Compare and PWM output) can change, based on the device configuration.

The ECCPMX Configuration bit (CONFIG3H<1>) determines the pin to which CCP6/7/8/9 is multiplexed. The pin assignments for these CCP modules are given in Table 19-4.

TABLE 19-4: CCP PIN ASSIGNMENT

ECCPMX	Pin Mapped to							
Value	CCP6	CCP7	CCP8	CC9				
1 (Default)	RE6	RE5	RE4	RE3				
0	RH7	RH6	RH5	RH4				

19.2 Capture Mode

In Capture mode, the CCPR4H:CCPR4L register pair captures the 16-bit value of the Timer register selected in the CCPTMRS1 when an event occurs on the CCP4 pin. An event is defined as one of the following:

- · Every falling edge
- Every rising edge
- · Every 4th rising edge
- · Every 16th rising edge

The event is selected by the mode select bits, CCP4M<3:0> (CCP4CON<3:0>). When a capture is made, the interrupt request flag bit, CCP4IF (PIR4<1>), is set. (It must be cleared in software.) If another capture occurs before the value in CCPR4 is read, the old captured value is overwritten by the new captured value.

Figure 19-1 shows the Capture mode block diagram.

19.2.1 CCP PIN CONFIGURATION

In Capture mode, the appropriate CCPx pin should be configured as an input by setting the corresponding TRIS direction bit.

Note:	If RC1	or	RE7 is	s co	onfigu	ured a	as a CCF	₽4
	output,	а	write	to	the	port	causes	а
	capture condition.							

19.2.2 TIMER1/3/5/7 MODE SELECTION

For the available timers (1/3/5/7) to be used for the capture feature, the used timers must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the capture operation may not work.

The timer to be used with each CCP module is selected in the CCPTMRSx registers. (See Section 19.1.1 "CCP Modules and Timer Resources".)

Details of the timer assignments for the CCP modules are given in Table 19-2 and Table 19-3.

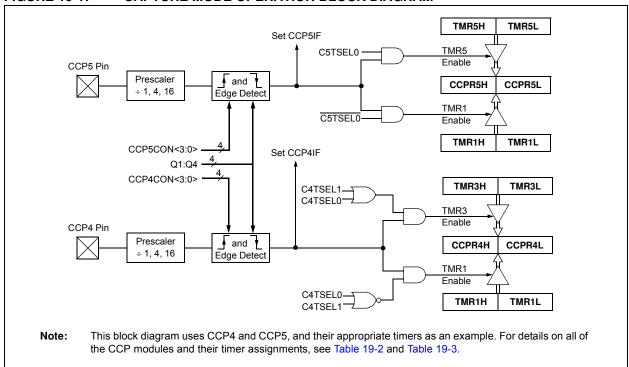


FIGURE 19-1: CAPTURE MODE OPERATION BLOCK DIAGRAM

19.2.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCP4IE bit (PIE4<1>) clear to avoid false interrupts and should clear the flag bit, CCP4IF, following any such change in operating mode.

19.2.4 CCP PRESCALER

There are four prescaler settings in Capture mode. They are specified as part of the operating mode selected by the mode select bits (CCP4M<3:0>). Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. This means that any Reset will clear the prescaler counter. Switching from one capture prescaler to another may generate an interrupt. Doing that will also not clear the prescaler counter – meaning the first capture may be from a non-zero prescaler.

Example 19-1 shows the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

EXAM	PLE 19-1:	-	HANGING BETWEEN APTURE PRESCALERS
CLRF	CCP4CON	;	Turn CCP module off
MOVLW	NEW_CAPT_PS	;	Load WREG with the
		;	new prescaler mode
		;	value and CCP ON
MOVWF	CCP4CON	;	Load CCP4CON with
		;	this value

19.3 Compare Mode

In Compare mode, the 16-bit CCPR4 register value is constantly compared against the Timer register pair value selected in the CCPTMR1 register. When a match occurs, the CCP4 pin can be:

- Driven high
- Driven low
- Toggled (high-to-low or low-to-high)
- Unchanged (that is, reflecting the state of the I/O latch)

The action on the pin is based on the value of the mode select bits (CCP4M<3:0>). At the same time, the interrupt flag bit, CCP4IF, is set.

Figure 19-2 gives the Compare mode block diagram

19.3.1 CCP PIN CONFIGURATION

The user must configure the CCPx pin as an output by clearing the appropriate TRIS bit.

Note: Clearing the CCP4CON register will force the RC1 or RE7 compare output latch (depending on device configuration) to the default low level. This is not the PORTC or PORTE I/O data latch.

19.3.2 TIMER1/3/5/7 MODE SELECTION

If the CCP module is using the compare feature in conjunction with any of the Timer1/3/5/7 timers, the timers must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the compare operation may not work.

Note: Details of the timer assignments for the CCP modules are given in Table 19-2 and Table 19-3.

19.3.3 SOFTWARE INTERRUPT MODE

When the Generate Software Interrupt mode is chosen (CCP4M<3:0> = 1010), the CCP4 pin is not affected. Only a CCP interrupt is generated, if enabled, and the CCP4IE bit is set.

19.3.4 SPECIAL EVENT TRIGGER

Both CCP modules are equipped with a Special Event Trigger. This is an internal hardware signal generated in Compare mode to trigger actions by other modules. The Special Event Trigger is enabled by selecting the Compare Special Event Trigger mode (CCP4M<3:0> = 1011).

For either CCP module, the Special Event Trigger resets the Timer register pair for whichever timer resource is currently assigned as the module's time base. This allows the CCPRx registers to serve as a programmable Period register for either timer.

The Special Event Trigger for CCP4 cannot start an A/D conversion.

Note:	The Special Event Trigger of ECCP2 can start
	an A/D conversion, but the A/D Converter
	must be enabled. For more information, see
	Section 19.0 "Capture/Compare/PWM
	(CCP) Modules".

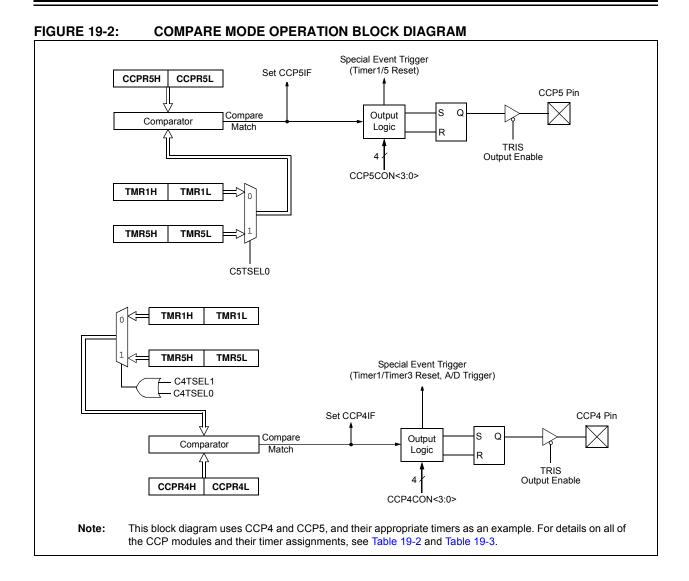


 TABLE 19-5:
 REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, TIMER1/3/5/7

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
PIR4	CCP10IF ⁽¹⁾	CCP9IF ⁽¹⁾	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
PIE4	CCP10IE ⁽¹⁾	CCP9IE ⁽¹⁾	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
IPR4	CCP10IP ⁽¹⁾	CCP9IP ⁽¹⁾	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0
TRISH ⁽²⁾	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
TMR1L	Timer1 Register Low Byte							

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Capture/Compare or Timer1/3/5/7.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

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TABLE 19-5:	REGISTERS ASSOCIATED WITH CAPTURE, COMPARE, TIMER1/3/5/7 (CONTINUED)								
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
TMR1H	Timer1 Regis	Timer1 Register High Byte							
TMR3L	Timer3 Regis	ter Low Byte							
TMR3H	Timer3 Regis	ter High Byte							
TMR5L	Timer5 Regis	ter Low Byte							
TMR5H	Timer5 Regis	ter High Byte							
TMR7L ⁽¹⁾	Timer7 Regis	ter Low Byte							
TMR7H ⁽¹⁾	Timer7 Regis	ter High Byte							
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N	
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	SOSCEN	T3SYNC	RD16	TMR3ON	
T5CON	TMR5CS1	TMR5CS0	T5CKPS1	T5CKPS0	SOSCEN	T5SYNC	RD16	TMR5ON	
T7CON ⁽¹⁾	TMR7CS1	TMR7CS0	T7CKPS1	T7CKPS0	SOSCEN	T7SYNC	RD16	TMR7ON	
CCPR4L	Capture/Com	pare/PWM R	egister 4 Lo	ow Byte					
CCPR4H	Capture/Com	pare/PWM R	egister 4 Hi	gh Byte					
CCPR5L	Capture/Com	pare/PWM R	egister 5 Lo	ow Byte					
CCPR5H	Capture/Com	pare/PWM R	egister 5 Hi	gh Byte					
CCPR6L	Capture/Com	pare/PWM R	egister 6 Lo	ow Byte					
CCPR6H	Capture/Com	pare/PWM R	egister 6 Hi	gh Byte					
CCPR7L	Capture/Com	pare/PWM R	egister 7 Lo	ow Byte					
CCPR7H	Capture/Com	pare/PWM R	egister 7 Hi	gh Byte					
CCPR8L	Capture/Com	pare/PWM R	egister 8 Lo	ow Byte					
CCPR8H	Capture/Com	pare/PWM R	egister 8 Hi	gh Byte					
CCPR9L ⁽¹⁾	Capture/Com	pare/PWM R	egister 9 Lo	ow Byte					
CCPR9H ⁽¹⁾	Capture/Com	pare/PWM R	egister 9 Hi	gh Byte					
CCPR10L ⁽¹⁾	Capture/Com	pare/PWM R	egister 10 L	ow Byte					
CCPR10H ⁽¹⁾	Capture/Com	pare/PWM R	egister 10 F	• •					
CCP4CON			DC4B1	DC4B0	CCP4M3	CCP4M2	CCP4M1	CCP4M0	
CCP5CON			DC5B1	DC5B0	CCP5M3	CCP5M2	CCP5M1	CCP5M0	
CCP6CON			DC6B1	DC6B0	CCP6M3	CCP6M2	CCP6M1	CCP6M0	
CCP7CON			DC7B1	DC7B0	CCP7M3	CCP7M2	CCP7M1	CCP7M0	
CCP8CON			DC8B1	DC8B0	CCP8M3	CCP8M2	CCP8M1	CCP8M0	
CCP9CON ⁽¹⁾	_	_	DC9B1	DC9B0	CCP9M3	CCP9M2	CCP9M1	CCP9M0	
CCP10CON ⁽¹⁾	_	_	DC10B1	DC10B0	CCP10M3	CCP10M2	CCP10M1	CCP10M0	
CCPTMRS1	C7TSEL1	C7TSEL0		C6TSEL0		C5TSEL0	C4TSEL1	C4TSEL0	
CCPTMRS2	—	—	_	C10TSEL0 ⁽¹⁾		C9TSEL0 ⁽¹⁾	C8TSEL1	C8TSEL0	
PMD3	CCP10MD ⁽¹⁾	CCP9MD ⁽¹⁾	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD ⁽¹⁾	

Legend: — = unimplemented, read as '0'. Shaded cells are not used by Capture/Compare or Timer1/3/5/7.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

19.4 PWM Mode

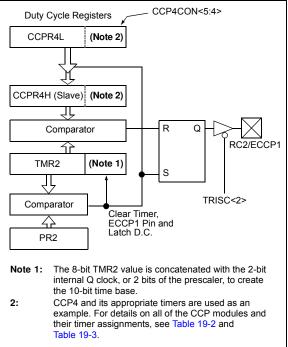
In Pulse-Width Modulation (PWM) mode, the CCP4 pin produces up to a 10-bit resolution PWM output. Since the CCP4 pin is multiplexed with a PORTC or PORTE data latch, the appropriate TRIS bit must be cleared to make the CCP4 pin an output.

Note:	Clearing the CCP4CON register will force the RC1 or RE7 output latch (depending on device configuration) to the default low level. This is not the PORTC or PORTE
	level. This is not the PORTC of PORTE
	I/O data latch.

Figure 19-3 shows a simplified block diagram of the ECCP1 module in PWM mode.

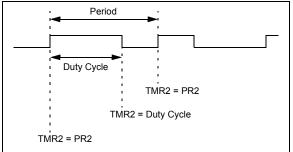
For a step-by-step procedure on how to set up the CCP module for PWM operation, see Section 19.4.3 "Setup for PWM Operation".

FIGURE 19-3: SIMPLIFIED PWM BLOCK DIAGRAM



A PWM output (Figure 19-4) has a time base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).

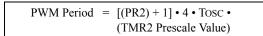
FIGURE 19-4: PWM OUTPUT



19.4.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

EQUATION 19-1:



PWM frequency is defined as 1/[PWM period].

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- · The CCP4 pin is set

(An exception: If PWM duty cycle = 0%, the CCP4 pin will not be set)

 The PWM duty cycle is latched from CCPR4L into CCPR4H

Note:	The			(see
	Sectio	n 15.0 "Tin	n <mark>er2 Module</mark> ") a	are not
	used i	n the detei	mination of the	PWM
	frequer	ncy. The po	stscaler could be	e used
	to have	e a servo up	odate rate at a di	ifferent
	frequer	ncy than the	e PWM output.	

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19.4.2 PWM DUTY CYCLE

The PWM duty cycle is specified to use CCP4, as an example, by writing to the CCPR4L register and to the CCP4CON<5:4> bits. Up to 10-bit resolution is available. The CCPR4L contains the eight MSbs and the CCP4CON<5:4> bits contain the two LSbs. This 10-bit value is represented by CCPR4L:CCP4CON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

EQUATION 19-2:

PWM Duty Cycle = (CCPR4L:CCP4CON<5:4>) • Tosc • (TMR2 Prescale Value)

CCPR4L and CCP4CON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPR4H until after a match between PR2 and TMR2 occurs (that is, the period is complete). In PWM mode, CCPR4H is a read-only register.

The CCPR4H register and a two-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitchless PWM operation.

When the CCPR4H and two-bit latch match TMR2, concatenated with an internal two-bit Q clock or two bits of the TMR2 prescaler, the CCP4 pin is cleared.

The maximum PWM resolution (bits) for a given PWM frequency is shown in Equation 19-3:

EQUATION 19-3:

PWM Resolution (max) =
$$\frac{\log(\frac{FOSC}{FPWM})}{\log(2)}$$
 bits

Note: If the PWM duty cycle value is longer than the PWM period, the CCP4 pin will not be cleared.

TABLE 19-6: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 40 MHz

PWM Frequency	2.44 kHz	9.77 kHz	39.06 kHz	156.25 kHz	312.50 kHz	416.67 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	FFh	FFh	FFh	3Fh	1Fh	17h
Maximum Resolution (bits)	10	10	10	8	7	6.58

19.4.3 SETUP FOR PWM OPERATION

To configure the CCP module for PWM operation, using CCP4 as an example:

- 1. Set the PWM period by writing to the PR2 register.
- 2. Set the PWM duty cycle by writing to the CCPR4L register and CCP4CON<5:4> bits.
- 3. Make the CCP4 pin an output by clearing the appropriate TRIS bit.
- 4. Set the TMR2 prescale value, then enable Timer2 by writing to T2CON.
- 5. Configure the CCP4 module for PWM operation.

TABLE 19-7: REGISTERS ASSOCIATED WITH PWM AND TIMERS

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
PIR4	CCP10IF ⁽¹⁾	CCP9IF ⁽¹⁾	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
PIE4	CCP10IE ⁽¹⁾	CCP9IE ⁽¹⁾	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
IPR4	CCP10IP ⁽¹⁾	CCP9IP ⁽¹⁾	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0

Legend: — = unimplemented, read as '0'. Shaded cells are not used by PWM or Timer2/4/6/8.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

IABLE 19-7:								
Name	Bit 7	Bit 6	Bit 5		Bit 3	Bit 2	Bit 1	Bit 0
TRISH ⁽²⁾	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
TMR2	Timer2 Regis							
TMR4	Timer4 Regis	ter						
TMR6	Timer6 Regis	ter						
TMR8	Timer8 Regis	ter						
PR2	Timer2 Perio	d Register						
PR4	Timer4 Perio	d Register						
PR6	Timer6 Perio	d Register						
PR8	Timer8 Perio	d Register						
T2CON	—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
T4CON	—	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR4ON	T4CKPS1	T4CKPS0
T6CON	_	T6OUTPS3	T6OUTPS2	T6OUTPS1	T6OUTPS0	TMR6ON	T6CKPS1	T6CKPS0
T8CON	—	T8OUTPS3	T8OUTPS2	T8OUTPS1	T8OUTPS0	TMR8ON	T8CKPS1	T8CKPS0
CCPR4L	Capture/Com	pare/PWM R	Register 4 Lo	w Byte				
CCPR4H	Capture/Com	pare/PWM R	Register 4 Hi	gh Byte				
CCPR5L	Capture/Com	pare/PWM R	Register 5 Lo	w Byte				
CCPR5H	Capture/Com	pare/PWM R	Register 5 Hig	gh Byte				
CCPR6L	Capture/Com	pare/PWM R	Register 6 Lo	w Byte				
CCPR6H	Capture/Com	pare/PWM R	Register 6 Hig	gh Byte				
CCPR7L	Capture/Com	pare/PWM R	Register 7 Lo	w Byte				
CCPR7H	Capture/Com	pare/PWM R	Register 7 Hig	gh Byte				
CCPR8L	Capture/Com	pare/PWM R	Register 8 Lo	w Byte				
CCPR8H	Capture/Com	pare/PWM R	Register 8 Hig	gh Byte				
CCPR9L ⁽¹⁾	Capture/Com	pare/PWM R	Register 9 Lo	w Byte				
CCPR9H ⁽¹⁾	Capture/Com	pare/PWM R	Register 9 Hi	gh Byte				
CCPR10L ⁽¹⁾	Capture/Com	pare/PWM R	Register 10 L	ow Byte				
CCPR10H ⁽¹⁾	Capture/Com	pare/PWM R	Register 10 H	ligh Byte				
CCP4CON	—	_	DC4B1	DC4B0	CCP4M3	CCP4M2	CCP4M1	CCP4M0
CCP5CON	—	_	DC5B1	DC5B0	CCP5M3	CCP5M2	CCP5M1	CCP5M0
CCP6CON	—	—	DC6B1	DC6B0	CCP6M3	CCP6M2	CCP6M1	CCP6M0
CCP7CON	—	_	DC7B1	DC7B0	CCP7M3	CCP7M2	CCP7M1	CCP7M0
CCP8CON	—	—	DC8B1	DC8B0	CCP8M3	CCP8M2	CCP8M1	CCP8M0
CCP9CON ⁽¹⁾	—	—	DC9B1	DC9B0	CCP9M3	CCP9M2	CCP9M1	CCP9M0
CCP10CON ⁽¹⁾	_	_	DC10B1	DC10B0	CCP10M3	CCP10M2	CCP10M1	CCP10M0
CCPTMRS1	C7TSEL1	C7TSEL0		C6TSEL0		C5TSEL0	C4TSEL1	C4TSEL0
CCPTMRS2	_		_	C10TSEL0 ⁽¹⁾		C9TSEL0 ⁽¹⁾	C8TSEL1	C8TSEL0
PMD3	CCP10MD ⁽¹⁾	CCP9MD ⁽¹⁾	CCP8MD	CCP7MD	CCP6MD	CCP5MD	CCP4MD	TMR12MD ⁽¹⁾

TABLE 19-7:	REGISTERS ASSOCIATED WITH PWM AND TIMERS (CONTINUED)
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Legend: — = unimplemented, read as '0'. Shaded cells are not used by PWM or Timer2/4/6/8.

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

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NOTES:

20.0 ENHANCED CAPTURE/COMPARE/PWM (ECCP) MODULE

PIC18F87K22 family devices have three Enhanced Capture/Compare/PWM (ECCP) modules: ECCP1, ECCP2 and ECCP3. These modules contain a 16-bit register, which can operate as a 16-bit Capture register, a 16-bit Compare register or a PWM Master/Slave Duty Cycle register. These ECCP modules are upward compatible with CCP.

Note: Throughout this section, generic references are used for register and bit names that are the same, except for an 'x' variable that indicates the item's association with the ECCP1, ECCP2 or ECCP3 module. For example, the control register is named CCPxCON and refers to CCP1CON, CCP2CON and CCP3CON.

ECCP1, ECCP2 and ECCP3 are implemented as standard CCP modules with Enhanced PWM capabilities. These include:

- Provision for two or four output channels
- · Output Steering modes
- · Programmable polarity
- Programmable dead-band control
- · Automatic shutdown and restart

The enhanced features are discussed in detail in Section 20.4 "PWM (Enhanced Mode)".

The ECCP1, ECCP2 and ECCP3 modules use the control registers: CCP1CON, CCP2CON and CCP3CON. The control registers, CCP4CON through CCP10CON, are for the modules, CCP4 through CCP10.

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Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7-6 PxM<1:0>: Enhanced PWM Output Configuration bits If CCPxM<3:2> = 00, 01, 10: xx = PxA is assigned as capture/compare input/output; PxB, PxC and PxD are assigned as port pin If CCPxM<3:2> = 11: 00 = Single output: PxA, PxB, PxC and PxD are controlled by steering (see Section 20.4.7 "Puls Steering Mode") 01 = Full-bridge output: PxA, PxB are modulated; PxA is active; PxB, PxC are inactive 10 = Half-bridge output: PxA, PxB are modulated with dead-band control; PxC and PxD are assigned as port pins 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive bit 5-4 DCxB<1:0>: PWM Duty Cycle Bit 1 and Bit 0 Capture mode: Unused. Unused. Compare mode: Unused. PWM mode: These bits are the two LSbs of the 10-bit PWM duty cycle. The eight MSbs of the duty cycle are foun in CCPRxL.	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0				
Legend: R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' :n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7-6 PxM<1:0>: Enhanced PWM Output Configuration bits If CCPxM<3:2> = 0.0.01.10: xx = PxA is assigned as capture/compare input/output; PxB, PxC and PxD are assigned as port pin If CCPxM<3:2> = 11: 00 = Single output: PxA, PxB, PxC and PxD are controlled by steering (see Section 20.4.7 "Pute Steering Mode") 01 = Full-bridge output forward: PxD is modulated; PxA is active; PxB, PxC are inactive 10 = Half-bridge output forward: PxD is modulated; PxC is active; PxA and PxD are inactive 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are active in assigned as port pins 11 = Full-bridge output reverse: FxB is modulated; PxC is active; PxA and PxD are active intervers 0 DcxB-1:0:: PVM Mouty Cycle Bit 1 and Bit 0 Capt	PxM1	PxM0	DCxB1	DCxB0	CCPxM3	CCPxM2	CCPxM1	CCPxM0				
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7-6 PxM<1:0>: Enhanced PWM Output Configuration bits If CCPxM<3:2> = 0.0, 01, 10: xx = PxA is assigned as capture/compare input/output; PxB, PxC and PxD are assigned as port pin If CCPxM<3:2> = 11: 00 = Single output: PxA, PxB, PxC and PxD are controlled by steering (see Section 20.4.7 "Pute Steering Mode") 01 = Full-bridge output forward: PxD is modulated; PxA is active; PxB, PxC are inactive 10 = Hail-bridge output forward: PxD is modulated; PxG is active; PxA and PxD are inactive 10 = Hail-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive bit 5-4 DCxB-1:0>: PVMM Duty Cycle Bit 1 and Bit 0 Capture mode: Unused. Compare mode: Unused. Unused. EVM. ECPx Mode Select bits 0000 = Capture/Compare/PVM off (resets ECCPx module) 001 = Capture mode: toggle output on match 0011 = Capture mode: every falling edge 010 = Capture mode: every falling edge 0110 = Capture mode: every falling edge 010 = Capture mode: every falling edge 0110 = Capture mode: every falling edge 010 = Capture mode: every falling edge 0100 = Compare mode: initialize ECCPx pin hish, et and thy or compare match (set CC	bit 7	1						bit (
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0' -n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7-6 PxM<1:0>: Enhanced PWM Output Configuration bits If CCPxM<3:2> = 0.0, 01, 10: xx = PxA is assigned as capture/compare input/output; PxB, PxC and PxD are assigned as port pin If CCPxM<3:2> = 11: 00 = Single output: PxA, PxB, PxC and PxD are controlled by steering (see Section 20.4.7 "Pute Steering Mode") 01 = Full-bridge output forward: PxD is modulated; PxA is active; PxB, PxC are inactive 10 = Hail-bridge output forward: PxD is modulated; PxG is active; PxA and PxD are inactive 10 = Hail-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive bit 5-4 DCxB-1:0>: PVMM Duty Cycle Bit 1 and Bit 0 Capture mode: Unused. Compare mode: Unused. Unused. EVM. ECPx Mode Select bits 0000 = Capture/Compare/PVM off (resets ECCPx module) 001 = Capture mode: toggle output on match 0011 = Capture mode: every falling edge 010 = Capture mode: every falling edge 0110 = Capture mode: every falling edge 010 = Capture mode: every falling edge 0110 = Capture mode: every falling edge 010 = Capture mode: every falling edge 0100 = Compare mode: initialize ECCPx pin hish, et and thy or compare match (set CC												
n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown bit 7-6 PxM<1:0>: Enhanced PWM Output Configuration bits If CCPxM<3:2> = 0.0.01.10: xx = PxA is assigned as capture/compare input/output; PxB, PxC and PxD are assigned as port pins 0 = Single output: PxA, PxB, PxC and PxD are controlled by steering (see Section 20.4.7 "Pute Steering Mode") 01 = Full-bridge output: PxA, PxB are modulated; PxA is active; PxB, PxC are inactive 10 = Half-bridge output forward: PxD is modulated; PxA is active; PxB, PxC and PxD are inactive 10 = Half-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive Unused. Compare mode; Unused. Compare mode; Unused. 0000 = Capture mode; Unused. 00001 = Reserved 0000 = Capture/Compare/PWM off (resets ECCPx module) 0011 = Capture mode: every falling edge 0100 = Capture mode: every falling edge 0100 = Capture mode: every falling edge 010 = Capture mode: every falling edge 0110 = Capture mode: every falling ECCPx pin low, set output on compare match (set CCPxIF) 0101 = Compare mode: initialize ECCPx pin low, set output on compare match (set CCPxIF) 0101 = Compare mode: initialize ECCPx pin low, set output on compare match (set CCPxIF)	Legend:											
 PxM<-1:0>: Enhanced PWM Output Configuration bits If CCPxM<3:2> = 0.0.01.10: xx = PXA is assigned as capture/compare input/output; PxB, PxC and PxD are assigned as port pin If CCPxM<3:2> = 11: 01 = Single output: PxA, PxB, PxC and PxD are controlled by steering (see Section 20.4.7 "Puts Steering Mode") 01 = Full-bridge output forward: PxD is modulated; PxA is active; PxB, PxC are inactive 10 = Half-bridge output: PxA, PxB are modulated; with dead-band control; PxC and PxD are assigned as port pins 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive bit 5-4 DCxB<1:0>: PWM Duty Cycle Bit 1 and Bit 0 Capture mode: Unused. Compare mode: Unused. 2000 = Ccpture/Compare/PWM off (resets ECCPx module) 0000 = Capture/Compare/PWM off (resets ECCPx module) 0010 = Capture mode: 0100 = Capture mode: 0111 = Capture mode: every falling edge 0111 = Capture mode: every falling edge 0112 = Capture mode: every falling edge 0114 = Capture mode: initialize ECCPx pin low, set output on compare match (set CCPxIF) 0101 = Compare mode: initialize ECCPx pin low, set output on compare match (set CCPxIF) 0101 = Compare mode: initialize ECCPx pin low, set output on compare match (set CCPxIF) 0101 = Compare mode: initialize ECCPx pin low, set output on compare match (set CCPxIF) 0101 = Compare mode: initialize ECCPx pin low, set output on compare match (set CCPxIF) 0101 = Compare mode: initialize ECCPx pin low, set output on compare match (set CCPxIF) 0103 = Compare mode: initialize ECCPx pin low, set output on compare match (set CCPxIF) 0104 = Compare mode: initialize ECCPx pin low, set output on co	R = Readabl	le bit	W = Writable	bit	U = Unimplei	mented bit, read	d as '0'					
If CCPxM<3:2> = 00, 01, 10: xx = PxA is assigned as capture/compare input/output; PxB, PxC and PxD are assigned as port pin If CCPxM<3:2> = 11: 00 = Single output: PxA, PxB, PxC and PxD are controlled by steering (see Section 20.4.7 "Pute Steering Mode") 01 = Full-bridge output forward: PxD is modulated; PxA is active; PxB, PxC are inactive 10 = Half-bridge output: PxA, PxB are modulated with dead-band control; PxC and PxD at assigned as port pins 11 = Full-bridge output reverse: PxB is modulated; PxC is active; PxA and PxD are inactive DvxB<1:0-: PVM Duty Cycle Bit 1 and Bit 0	-n = Value at	t POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unk	nown				
If CCPXM<3:2> = 00.01.10: xx = PxA is assigned as capture/compare input/output; PxB, PxC and PxD are assigned as port pin If CCPXM<3:2> = 11: 00 = Single output: PxA, PxB, PxC and PxD are controlled by steering (see Section 20.4.7 "Pute Steering Mode") 01 = Full-bridge output forward: PxD is modulated; PxA is active; PxB, PxC are inactive 10 = Half-bridge output: PxA, PxB are modulated with dead-band control; PxC and PxD at assigned as port pins 11 = Full-bridge output: reverse: PxB is modulated; PxC is active; PxA and PxD are inactive DCxB<1:0:: PVM Duty Cycle Bit 1 and Bit 0	bit 7-6	PxM<1:0>: E	Enhanced PWM	Output Confi	guration bits							
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1110 = PWM mode: PxA and PxC are active-low; PxB and PxD are active-high		1100 = PWN	/I mode: PxA ar									
					•		0					

REGISTER 20-1: CCPxCON: ENHANCED CAPTURE/COMPARE/PWMx CONTROL

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REGISTER 20-2: CCPTMRS0: CCP TIMER SELECT 0 REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
C3TSEL1	C3TSEL0	C2TSEL2	C2TSEL1	C2TSEL0	C1TSEL2	C1TSEL1	C1TSEL0		
bit 7	•	·	•	•	•	•	bit (
Legend:									
R = Readabl	e bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'			
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cleared			x = Bit is unknown		
bit 7-6	00 = ECCP3 01 = ECCP3 10 = ECCP3 11 = ECCP3	>: ECCP3 Tim is based off of is based off of is based off of is based off of	TMR1/TMR2 TMR3/TMR4 TMR3/TMR6 TMR3/TMR8						
bit 5-3	000 = ECCP 001 = ECCP 010 = ECCP	: ECCP2 Tim 2 is based off 2 is based off 2 is based off 2 is based off	of TMR1/TMR of TMR3/TMR of TMR3/TMR	2 4 6					

100 = ECCP2 is based off of TMR3/TMR10: option reserved on the 32-Kbyte device variant; do not use

101 = Reserved; do not use

- 110 = Reserved; do not use
- 111 = Reserved; do not use

bit 2-0 C1TSEL<2:0>: ECCP1 Timer Selection bits

- 000 = ECCP1 is based off of TMR1/TMR2
- 001 = ECCP1 is based off of TMR3/TMR4
- 010 = ECCP1 is based off of TMR3/TMR6
- 011 = ECCP1 is based off of TMR3/TMR8
- 100 = ECCP1 is based off of TMR3/TMR10: option reserved on the 32-Kbyte device variant; do not use
- 101 = ECCP1 is based off of TMR3/TMR12: option reserved on the 32-Kbyte device variant; do not use
- 110 = Reserved; do not use
- 111 = Reserved; do not use

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In addition to the expanded range of modes available through the CCPxCON and ECCPxAS registers, the ECCP modules have two additional registers associated with Enhanced PWM operation and auto-shutdown features. They are:

- ECCPxDEL Enhanced PWM Control
- PSTRxCON Pulse Steering Control

20.1 ECCP Outputs and Configuration

The Enhanced CCP module may have up to four PWM outputs, depending on the selected operating mode. The CCPxCON register is modified to allow control over four PWM outputs: ECCPx/PxA, PxB, PxC and PxD. Applications can use one, two or four of these outputs.

The outputs that are active depend on the ECCP selected operating mode. The pin assignments are summarized in Table 20-3.

To configure the I/O pins as PWM outputs, the proper PWM mode must be selected by setting the PxM<1:0> and CCPxM<3:0> bits. The appropriate TRIS direction bits for the port pins must also be set as outputs.

20.1.1 ECCP MODULE AND TIMER RESOURCES

The ECCP modules use Timers, 1, 2, 3, 4, 6, 8, 10 or 12, depending on the mode selected. These timers are available to CCP modules in Capture, Compare or PWM modes, as shown in Table 20-1.

TABLE 20-1: ECCP MODE – TIMER RESOURCE

ECCP Mode Timer Resource			
Capture	Timer1 or Timer3		
Compare	Timer1 or Timer3		
PWM	Timer2, Timer4, Timer6, Timer8, Timer10 or Timer12		

The assignment of a particular timer to a module is determined by the timer to ECCP enable bits in the CCPTMRSx register (Register 20-2). The interactions between the two modules are depicted in Figure 20-1. Capture operations are designed to be used when the timer is configured for Synchronous Counter mode. Capture operations may not work as expected if the associated timer is configured for Asynchronous Counter mode.

20.1.2 ECCP PIN ASSIGNMENT

The pin assignment for ECCPx (Capture input, Compare and PWM output) can change, based on device configuration. The ECCPMX (CONFIG3H<1>) Configuration bit determines which pins ECCP1 and ECCP3 are multiplexed to.

- Default/ECCPMX = 1:
 - ECCP1 (P1B/P1C) is multiplexed onto RE6 and RE5
 - ECCP3 (P3B/P3C) is multiplexed onto RE4 and RE3
- ECCPMX = 0:
 - ECCP1 (P1B/P1C) is multiplexed onto RH7 and RH6
 - ECCP3 (P3B/P3C) is multiplexed onto RH5 and RH4.

The pin assignment for ECCP2 (Capture input, Compare and PWM output) can change, based on the device configuration.

The CCP2MX Configuration bit (CONFIG3H<0>) determines which pin ECCP2 is multiplexed to.

- If CCP2MX = 1 (default) ECCP2 is multiplexed to RC1
- If CCP2MX = 0 ECCP2 is multiplexed to RE7

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20.2 Capture Mode

In Capture mode, the CCPRxH:CCPRxL register pair captures the 16-bit value of the TMR1 or TMR3 registers when an event occurs on the corresponding ECCPx pin. An event is defined as one of the following:

- Every falling edge
- Every rising edge
- Every fourth rising edge
- Every 16th rising edge

The event is selected by the mode select bits, CCPxM<3:0> (CCPxCON<3:0>). When a capture is made, the interrupt request flag bit, CCPxIF, is set (see Table 20-2). The flag must be cleared by software. If another capture occurs before the value in the CCPRxH/L register is read, the old captured value is overwritten by the new captured value.

TABLE 20-2: ECCP1/2/3 INTERRUPT FLAG BITS

ECCP Module	Flag Bit
1	PIR3<1>
2	PIR3<2>
3	PIR4<0>

20.2.1 ECCP PIN CONFIGURATION

In Capture mode, the appropriate ECCPx pin should be configured as an input by setting the corresponding TRIS direction bit.

Note:	If the ECCPx pin is configured as an out-
	put, a write to the port can cause a capture
	condition.

20.2.2 TIMER1/2/3/4/6/8/10/12 MODE SELECTION

The timers that are to be used with the capture feature (Timer1 2, 3, 4, 6, 8, 10 or 12) must be running in Timer mode or Synchronized Counter mode. In Asynchronous Counter mode, the capture operation may not work. The timer to be used with each ECCP module is selected in the CCPTMRS0 register (Register 20-2).

20.2.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep the CCPxIE interrupt enable bit clear to avoid false interrupts. The interrupt flag bit, CCPxIF, should also be cleared following any such change in operating mode.

20.2.4 ECCP PRESCALER

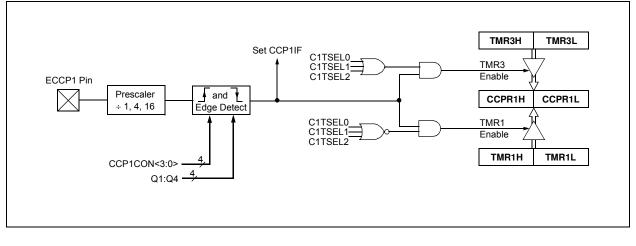
There are four prescaler settings in Capture mode; they are specified as part of the operating mode selected by the mode select bits (CCPxM<3:0>). Whenever the ECCP module is turned off, or Capture mode is disabled, the prescaler counter is cleared. This means that any Reset will clear the prescaler counter.

Switching from one capture prescaler to another may generate an interrupt. Also, the prescaler counter will not be cleared; therefore, the first capture may be from a non-zero prescaler. Example 20-1 provides the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

EXAMPLE 20-1: CHANGING BETWEEN CAPTURE PRESCALERS

CLRF	CCP1CON	;	Turn ECCP module off
MOVLW	NEW_CAPT_PS	;	Load WREG with the
		;	new prescaler mode
		;	value and ECCP ON
MOVWF	CCP1CON	;	Load CCP1CON with
		;	this value

FIGURE 20-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



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20.3 Compare Mode

In Compare mode, the 16-bit CCPRx register value is constantly compared against the Timer register pair value selected in the CCPTMR1 register. When a match occurs, the ECCPx pin can be:

- Driven high
- Driven low
- Toggled (high-to-low or low-to-high)
- Unchanged (that is, reflecting the state of the I/O latch)

The action on the pin is based on the value of the mode select bits (CCPxM<3:0>). At the same time, the interrupt flag bit, CCPxIF, is set.

20.3.1 ECCP PIN CONFIGURATION

Users must configure the ECCPx pin as an output by clearing the appropriate TRIS bit.

Note:	Clearing the CCPxCON register will force							
	the ECCPx compare output latch							
	(depending on device configuration) to the							
	default low level. This is not the PORTx							
	I/O data latch.							

20.3.2 TIMER1/2/3/4/6/8/10/12 MODE SELECTION

Timer1 2, 3, 4, 6, 8, 10 or 12 must be running in Timer mode or Synchronized Counter mode if the ECCP module is using the compare feature. In Asynchronous Counter mode, the compare operation will not work reliably.

20.3.3 SOFTWARE INTERRUPT MODE

When the Generate Software Interrupt mode is chosen (CCPxM<3:0> = 1010), the ECCPx pin is not affected; only the CCPxIF interrupt flag is affected.

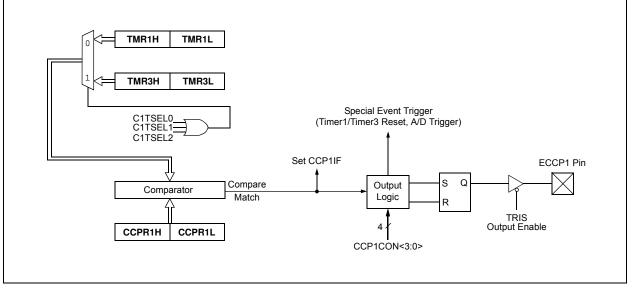
20.3.4 SPECIAL EVENT TRIGGER

The ECCP module is equipped with a Special Event Trigger. This is an internal hardware signal generated in Compare mode to trigger actions by other modules. The Special Event Trigger is enabled by selecting the Compare Special Event Trigger mode (CCPxM<3:0> = 1011).

The Special Event Trigger resets the Timer register pair for whichever timer resource is currently assigned as the module's time base. This allows the CCPRx registers to serve as a programmable Period register for either timer.

The Special Event Trigger can also start an A/D conversion. In order to do this, the A/D Converter must already be enabled.

FIGURE 20-2: COMPARE MODE OPERATION BLOCK DIAGRAM



20.4 PWM (Enhanced Mode)

The Enhanced PWM mode can generate a PWM signal on up to four different output pins with up to 10 bits of resolution. It can do this through four different PWM Output modes:

- Single PWM
- Half-Bridge PWM
- Full-Bridge PWM, Forward mode
- · Full-Bridge PWM, Reverse mode

To select an Enhanced PWM mode, the PxM bits of the CCPxCON register must be set appropriately.

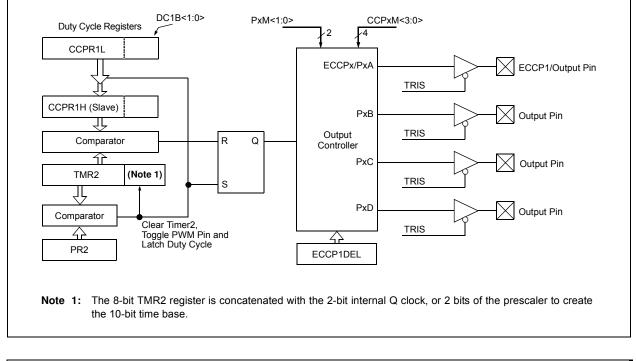
The PWM outputs are multiplexed with I/O pins and are designated: PxA, PxB, PxC and PxD. The polarity of the PWM pins is configurable and is selected by setting the CCPxM bits in the CCPxCON register appropriately.

Table 20-1provides the pin assignments for eachEnhanced PWM mode.

Figure 20-3 provides an example of a simplified block diagram of the Enhanced PWM module.

Note: To prevent the generation of an incomplete waveform when the PWM is first enabled, the ECCP module waits until the start of a new PWM period before generating a PWM signal.

FIGURE 20-3: EXAMPLE SIMPLIFIED BLOCK DIAGRAM OF THE ENHANCED PWM MODE



Note 1: The TRIS register value for each PWM output must be configured appropriately.2: Any pin not used by an Enhanced PWM mode is available for alternate pin functions.

ECCP Mode	PxM<1:0>	PxA	PxB	PxC	PxD
Single	00	Yes ⁽¹⁾	Yes ⁽¹⁾	Yes ⁽¹⁾	Yes ⁽¹⁾
Half-Bridge	10	Yes	Yes	No	No
Full-Bridge, Forward	01	Yes	Yes	Yes	Yes
Full-Bridge, Reverse	11	Yes	Yes	Yes	Yes

TABLE 20-3: EXAMPLE PIN ASSIGNMENTS FOR VARIOUS PWM ENHANCED MODES

Note 1: Outputs are enabled by pulse steering in Single mode (see Register 20-5).

FIGURE 20-4: EXAMPLE PWM (ENHANCED MODE) OUTPUT RELATIONSHIPS (ACTIVE-HIGH STATE)

00	(Single Output)	PxA Modulated			İ
		PxA Modulated	Delay ⁽¹⁾	Delay ⁽¹⁾	
10	(Half-Bridge)	PxB Modulated	- '		i
		PxA Active	- <u>'</u>		
	(Full-Bridge,	PxB Inactive	_ ;		- - -
01	Forward)	PxC Inactive	_ ' _ '		
		PxD Modulated		—_į́	
		PxA Inactive	- ;		I I
11	(Full-Bridge,	PxB Modulated			
	Reverse)	PxC Active			 i i
		PxD Inactive	_ ! _ / _ !		
Relat	• Pulse Width = To	c * (PR2 + 1) * (TMR2 Pre osc * (CCPRxL<7:0>:CCP; c * (ECCPxDEL<6:0>)		scale Value)	, , , , , , , , , , , , , , , , , , ,

00	(Single Output)	PxA Modulated	<u> </u>		<u> </u>	1 1 1
		PxA Modulated	- Vel	► ay ⁽¹⁾	Delay ⁽¹⁾	
10	(Half-Bridge)	PxB Modulated		.,		
		PxA Active	— ; — ;			
01	(Full-Bridge, Forward)	PxB Inactive	— <u> </u>		 	; ; ;
	r orward)	PxC Inactive				
		PxD Modulated				
		PxA Inactive			1 1 1	1 1 1
11	(Full-Bridge, Reverse)	PxB Modulated				1
		PxC Active	:_		1 	
		PxD Inactive	'		 I I	I I
	 Pulse Width = Tos Delay = 4 * Tosc 	* (PR2 + 1) * (TMR2 Pre sc * (CCPRxL<7:0>:CCP; * (ECCPxDEL<6:0>) I delay is programmed usin	«CON<5:4>) *	-		ammable Dead-Ban

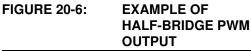
FIGURE 20-5: EXAMPLE ENHANCED PWM OUTPUT RELATIONSHIPS (ACTIVE-LOW STATE)

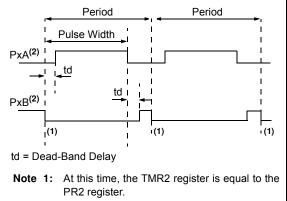
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20.4.1 HALF-BRIDGE MODE

In Half-Bridge mode, two pins are used as outputs to drive push-pull loads. The PWM output signal is output on the PxA pin, while the complementary PWM output signal is output on the PxB pin (see Figure 20-6). This mode can be used for half-bridge applications, as shown in Figure 20-7, or for full-bridge applications, where four power switches are being modulated with two PWM signals.

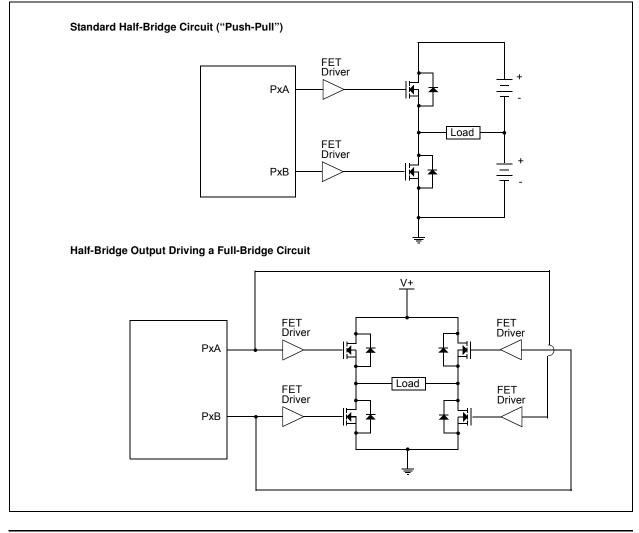
In Half-Bridge mode, the programmable dead-band delay can be used to prevent shoot-through current in half-bridge power devices. The value of the PxDC<6:0> bits of the ECCPxDEL register sets the number of instruction cycles before the output is driven active. If the value is greater than the duty cycle, the corresponding output remains inactive during the entire cycle. For more details on the dead-band delay operations, see Section 20.4.6 "Programmable Dead-Band Delay Mode". Since the PxA and PxB outputs are multiplexed with the port data latches, the associated TRIS bits must be cleared to configure PxA and PxB as outputs.





^{2:} Output signals are shown as active-high.

FIGURE 20-7: EXAMPLE OF HALF-BRIDGE APPLICATIONS



In the Reverse mode, the PxC pin is driven to its active state and the PxB pin is modulated, while the PxA and

PxD pins are driven to their inactive state, as provided

The PxA, PxB, PxC and PxD outputs are multiplexed

with the port data latches. The associated TRIS bits

must be cleared to configure the PxA, PxB, PxC and

20.4.2 FULL-BRIDGE MODE

In Full-Bridge mode, all four pins are used as outputs. An example of a full-bridge application is provided in Figure 20-8.

In the Forward mode, the PxA pin is driven to its active state and the PxD pin is modulated, while the PxB and PxC pins are driven to their inactive state, as provided in Figure 20-9.

FIGURE 20-8: EXAMPLE OF FULL-BRIDGE APPLICATION

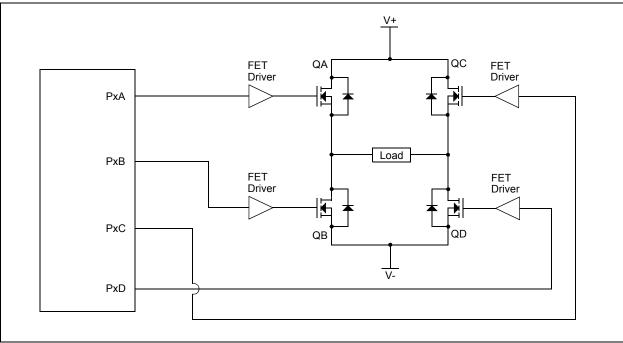
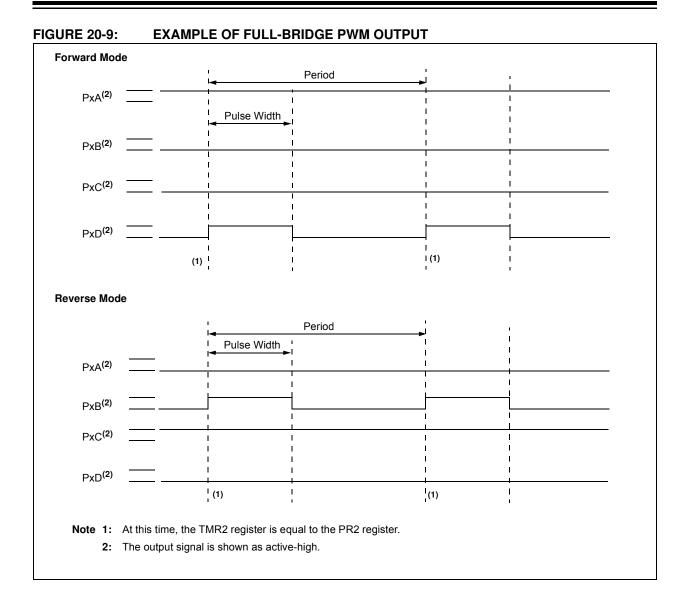


Figure 20-9.

PxD pins as outputs.

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20.4.2.1 Direction Change in Full-Bridge Mode

In Full-Bridge mode, the PxM1 bit in the CCPxCON register allows users to control the forward/reverse direction. When the application firmware changes this direction control bit, the module will change to the new direction on the next PWM cycle.

A direction change is initiated in software by changing the PxM1 bit of the CCPxCON register. The following sequence occurs prior to the end of the current PWM period:

- The modulated outputs (PxB and PxD) are placed in their inactive state.
- The associated unmodulated outputs (PxA and PxC) are switched to drive in the opposite direction.
- PWM modulation resumes at the beginning of the next period.

For an illustration of this sequence, see Figure 20-10.

The Full-Bridge mode does not provide a dead-band delay. As one output is modulated at a time, a dead-band delay is generally not required. There is a situation where a dead-band delay is required. This situation occurs when both of the following conditions are true:

- The direction of the PWM output changes when the duty cycle of the output is at or near 100%.
- The turn-off time of the power switch, including the power device and driver circuit, is greater than the turn-on time.

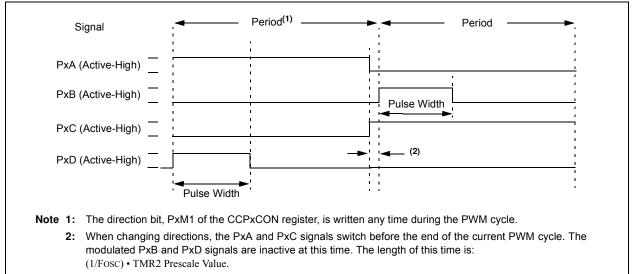
Figure 20-11 shows an example of the PWM direction changing from forward to reverse, at a near 100% duty cycle. In this example, at time, t1, the PxA and PxD outputs become inactive, while the PxC output becomes active. Since the turn-off time of the power devices is longer than the turn-on time, a shoot-through current will flow through power devices, QC and QD (see Figure 20-8), for the duration of 't'. The same phenomenon will occur to power devices, QA and QB, for PWM direction change from reverse to forward.

If changing PWM direction at high duty cycle is required for an application, two possible solutions for eliminating the shoot-through current are:

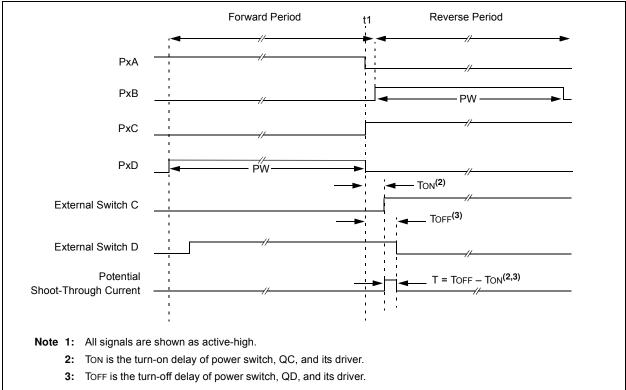
- Reduce PWM duty cycle for one PWM period before changing directions.
- Use switch drivers that can drive the switches off faster than they can drive them on.

Other options to prevent shoot-through current may exist.

FIGURE 20-10: EXAMPLE OF PWM DIRECTION CHANGE







20.4.3 START-UP CONSIDERATIONS

When any PWM mode is used, the application hardware must use the proper external pull-up and/or pull-down resistors on the PWM output pins.

Note:	When the microcontroller is released from
	Reset, all of the I/O pins are in the
	high-impedance state. The external
	circuits must keep the power switch
	devices in the OFF state until the micro-
	controller drives the I/O pins with the
	proper signal levels or activates the PWM
	output(s).

The CCPxM<1:0> bits of the CCPxCON register allow the user to choose whether the PWM output signals are active-high or active-low for each pair of PWM output pins (PxA/PxC and PxB/PxD). The PWM output polarities must be selected before the PWM pin output drivers are enabled. Changing the polarity configuration while the PWM pin output drivers are enabled is not recommended since it may result in damage to the application circuits.

The PxA, PxB, PxC and PxD output latches may not be in the proper states when the PWM module is initialized. Enabling the PWM pin output drivers at the same time as the Enhanced PWM modes may cause damage to the application circuit. The Enhanced PWM modes must be enabled in the proper Output mode and complete a full PWM cycle before enabling the PWM pin output drivers. The completion of a full PWM cycle is indicated by the TMR2IF or TMR4IF bit of the PIR1 or PIR5 register being set as the second PWM period begins.

20.4.4 ENHANCED PWM AUTO-SHUTDOWN MODE

The PWM mode supports an Auto-Shutdown mode that will disable the PWM outputs when an external shutdown event occurs. Auto-Shutdown mode places the PWM output pins into a predetermined state. This mode is used to help prevent the PWM from damaging the application.

The auto-shutdown sources are selected using the ECCPxAS<2:0> bits (ECCPxAS<6:4>). A shutdown event may be generated by:

- A logic '0' on the pin that is assigned the FLT0 input function
- · C1 Comparator
- · C2 Comparator
- · Setting the ECCPxASE bit in firmware

A shutdown condition is indicated by the ECCPxASE (Auto-Shutdown Event Status) bit (ECCPxAS<7>). If the bit is a '0', the PWM pins are operating normally. If the bit is a '1', the PWM outputs are in the shutdown state.

When a shutdown event occurs, two things happen:

- The ECCPxASE bit is set to '1'. The ECCPxASE will remain set until cleared in firmware or an auto-restart occurs. (See Section 20.4.5 "Auto-Restart Mode".)
- The enabled PWM pins are asynchronously placed in their shutdown states. The PWM output pins are grouped into pairs (PxA/PxC) and (PxB/PxD). The state of each pin pair is determined by the PSSxAC and PSSxBD bits (ECCPxAS<3:2> and <1:0>, respectively).

Each pin pair may be placed into one of three states:

- Drive logic '1'
- Drive logic '0'
- Tri-state (high-impedance)

REGISTER 20-3: ECCPxAS: ECCPx AUTO-SHUTDOWN CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
ECCPxASE	ECCPxAS2	ECCPxAS1	ECCPxAS0	PSSxAC1	PSSxAC0	PSSxBD1	PSSxBD0
bit 7							bit 0

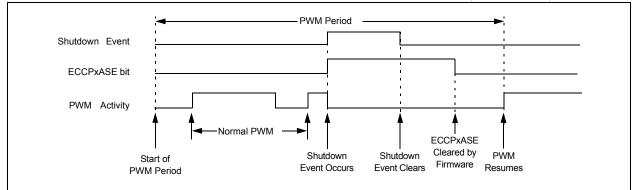
Legend:							
R = Reada	R = Readable bit W = Writable bit		U = Unimplemented bit	, read as '0'			
-n = Value	at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown			
bit 7	ECCPxA	SE: ECCP Auto-Shutdown I	Event Status bit				
		utdown event has occurred; P outputs are operating	ECCP outputs are in a shutdo	wn state			
bit 6-4	ECCPxA	S<2:0>: ECCP Auto-Shutdo	own Source Select bits				
	001 = C 010 = C 011 = Ei 100 = VI 101 = VI 110 = VI	uto-shutdown is disabled omparator C1OUT output is l omparator C2OUT output is l ther Comparator C1OUT or (L on FLT0 pin L on FLT0 pin or Comparator L on FLT0 pin or Comparator L on FLT0 pin or Comparator	high C2OUT is high r C1OUT output is high	JT is high			
bit 3-2	00 = Driv 01 = Driv	C <1:0>: PxA and PxC Pins S ve pins, PxA and PxC, to '0' ve pins, PxA and PxC, to '1' A and PxC pins tri-state	hutdown State Control bits				
bit 1-0	PSSxBD<1:0>: Pins PxB and PxD Shutdown State Control bits						
	00 = Driv 01 = Driv	ve pins, PxB and PxD, to '0' ve pins, PxB and PxD, to '1' and PxD pins tri-state					

Note 1: The auto-shutdown condition is a level-based signal, not an edge-based signal. As long as the level is present, the auto-shutdown will persist.

- 2: Writing to the ECCPxASE bit is disabled while an auto-shutdown condition persists.
- 3: Once the auto-shutdown condition has been removed and the PWM restarted (either through firmware or auto-restart), the PWM signal will always restart at the beginning of the next PWM period.

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FIGURE 20-12: PWM AUTO-SHUTDOWN WITH FIRMWARE RESTART (PxRSEN = 0)

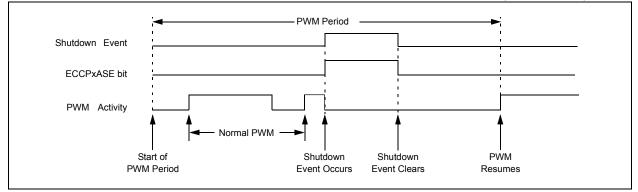


20.4.5 AUTO-RESTART MODE

The Enhanced PWM can be configured to automatically restart the PWM signal once the auto-shutdown condition has been removed. Auto-restart is enabled by setting the PxRSEN bit (ECCPxDEL<7>).

If auto-restart is enabled, the ECCPxASE bit will remain set as long as the auto-shutdown condition is active. When the auto-shutdown condition is removed, the ECCPxASE bit will be cleared via hardware and normal operation will resume. The module will wait until the next PWM period begins, however, before re-enabling the output pin. This behavior allows the auto-shutdown with auto-restart features to be used in applications based on current mode of PWM control.

FIGURE 20-13: PWM AUTO-SHUTDOWN WITH AUTO-RESTART ENABLED (PxRSEN = 1)



20.4.6 PROGRAMMABLE DEAD-BAND DELAY MODE

In half-bridge applications, where all power switches are modulated at the PWM frequency, the power switches normally require more time to turn off than to turn on. If both the upper and lower power switches are switched at the same time (one turned on and the other turned off), both switches may be on for a short period until one switch completely turns off. During this brief interval, a very high current (shoot-through current) will flow through both power switches, shorting the bridge supply. To avoid this potentially destructive shoot-through current from flowing during switching, turning on either of the power switches is normally delayed to allow the other switch to completely turn off.

In Half-Bridge mode, a digitally programmable dead-band delay is available to avoid shoot-through current from destroying the bridge power switches. The delay occurs at the signal transition from the non-active state to the active state. For an illustration, see Figure 20-14. The lower seven bits of the associated ECCPxDEL register (Register 20-4) set the delay period in terms of microcontroller instruction cycles (TcY or 4 Tosc).

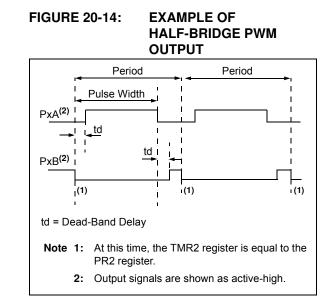
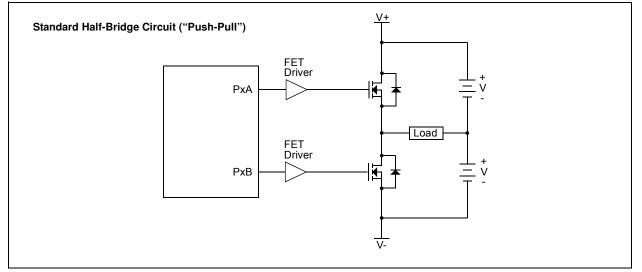


FIGURE 20-15: EXAMPLE OF HALF-BRIDGE APPLICATIONS



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REGISTER 20-4: ECCPxDEL: ENHANCED PWM CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PxRSEN	PxDC6	PxDC5	PxDC4	PxDC3	PxDC2	PxDC1	PxDC0
bit 7							bit 0

Legend:						
R = Readable bit	W = Writable bit	U = Unimplemented bit	U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown			

bit 7 PxRSEN: PWM Restart Enable bit

- 1 = Upon auto-shutdown, the ECCPxASE bit clears automatically once the shutdown event goes away; the PWM restarts automatically
- 0 = Upon auto-shutdown, ECCPxASE must be cleared by software to restart the PWM

bit 6-0 Px

PxDC<6:0>: PWM Delay Count bits

PxDCn = Number of Fosc/4 (4 * Tosc) cycles between the scheduled time when a PWM signal **should** transition active and the **actual** time it does transition active.

20.4.7 PULSE STEERING MODE

In Single Output mode, pulse steering allows any of the PWM pins to be the modulated signal. Additionally, the same PWM signal can simultaneously be available on multiple pins.

Once the Single Output mode is selected (CCPxM<3:2> = 11 and PxM<1:0> = 00 of the CCPxCON register), the user firmware can bring out the same PWM signal to one, two, three or four output pins by setting the appropriate STR<D:A> bits (PSTRxCON<3:0>), as provided in Table 20-3.

Note: The associated TRIS bits must be set to output ('0') to enable the pin output driver in order to see the PWM signal on the pin.

While the PWM Steering mode is active, the CCPxM<1:0> bits (CCPxCON<1:0>) select the PWM output polarity for the Px<D:A> pins.

The PWM auto-shutdown operation also applies to the PWM Steering mode, as described in Section 20.4.4 "Enhanced PWM Auto-shutdown mode". An auto-shutdown event will only affect pins that have PWM outputs enabled.

REGISTER 20-5: PSTRxCON: PULSE STEERING CONTROL⁽¹⁾

R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-1
CMPL1	CMPL0	—	STRSYNC	STRD	STRC	STRB	STRA
bit 7							bit 0
Lanandi							

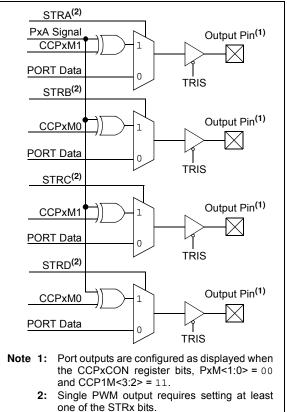
Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

> = 11 and
, ,

PxM<1:0> = 00.

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FIGURE 20-16: SIMPLIFIED STEERING BLOCK DIAGRAM



20.4.7.1 Steering Synchronization

The STRSYNC bit of the PSTRxCON register gives the user two choices for when the steering event will happen. When the STRSYNC bit is '0', the steering event will happen at the end of the instruction that writes to the PSTRxCON register. In this case, the output signal at the Px<D:A> pins may be an incomplete PWM waveform. This operation is useful when the user firmware needs to immediately remove a PWM signal from the pin.

When the STRSYNC bit is '1', the effective steering update will happen at the beginning of the next PWM period. In this case, steering on/off the PWM output will always produce a complete PWM waveform.

Figures 20-17 and 20-18 illustrate the timing diagrams of the PWM steering depending on the STRSYNC setting.

FIGURE 20-17: EXAMPLE OF STEERING EVENT AT END OF INSTRUCTION (STRSYNC = 0)

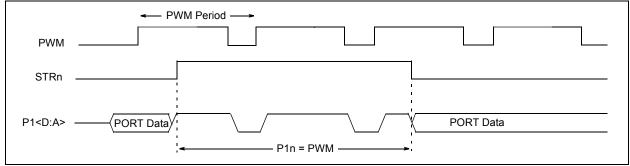
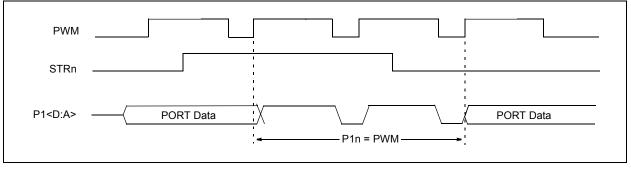


FIGURE 20-18: EXAMPLE OF STEERING EVENT AT BEGINNING OF INSTRUCTION (STRSYNC = 1)



20.4.8 OPERATION IN POWER-MANAGED MODES

In Sleep mode, all clock sources are disabled. Timer2/4/6/8 will not increment and the state of the module will not change. If the ECCPx pin is driving a value, it will continue to drive that value. When the device wakes up, it will continue from this state. If Two-Speed Start-ups are enabled, the initial start-up frequency from HF-INTOSC and the postscaler may not be immediately stable.

In PRI_IDLE mode, the primary clock will continue to clock the ECCPx module without change.

20.4.8.1 Operation with Fail-Safe Clock Monitor (FSCM)

If the Fail-Safe Clock Monitor (FSCM) is enabled, a clock failure will force the device into the power-managed RC_RUN mode and the OSCFIF bit of the PIR2 register will be set. The ECCPx will then be clocked from the internal oscillator clock source, which may have a different clock frequency than the primary clock.

20.4.9 EFFECTS OF A RESET

Both Power-on Reset and subsequent Resets will force all ports to Input mode and the ECCP registers to their Reset states. This forces the ECCP module to reset to a state compatible with previous, non-Enhanced CCP modules used on other PIC18 and PIC16 devices.

	111VIER1/2/3/4/0/0/10/12							
File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIR4	CCP10IF ⁽¹⁾	CCP9IF ⁽¹⁾	CCP8IF	CCP7IF	CCP6IF	CCP5IF	CCP4IF	CCP3IF
PIE3	TMR5GIE	—	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
PIE4	CCP10IE ⁽¹⁾	CCP9IE ⁽¹⁾	CCP8IE	CCP7IE	CCP6IE	CCP5IE	CCP4IE	CCP3IE
IPR3	TMR5GIP	—	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
IPR4	CCP10IP ⁽¹⁾	CCP9IP ⁽¹⁾	CCP8IP	CCP7IP	CCP6IP	CCP5IP	CCP4IP	CCP3IP
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0
TRISE	TRISE7	TRISE6	TRISE5	TRISE4	TRISE3	TRISE2	TRISE1	TRISE0
TRISH ⁽²⁾	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
TMR1H	Timer1 Register	High Byte						
TMR1L	Timer1 Register	Low Byte						
TMR2	Timer2 Register							
TMR3H	Timer3 Register High Byte							
TMR3L	Timer3 Register	Low Byte						
TMR4	Timer4 Register							
TMR6	Timer6 Register							
TMR8	Timer8 Register							
TMR10 ⁽¹⁾	TMR10 Register							
TMR12 ⁽¹⁾	TMR10 Register	TMR10 Register						
PR2	Timer2 Period Register							
PR4	Timer4 Period Register							
PR6	Timer6 Period Register							
PR8	Timer8 Period Register							
PR10 ⁽¹⁾	Timer10 Period	Register						
PR12 ⁽¹⁾	Timer12 Period	Register						

TABLE 20-4: REGISTERS ASSOCIATED WITH ECCP1/2/3 MODULE AND TIMER1/2/3/4/6/8/10/12

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

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TABLE 20-4: REGISTERS ASSOCIATED WITH ECCP1/2/3 MODULE AND TIMER1/2/3/4/6/8/10/12 (CONTINUED)

File Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
T1CON	TMR1CS1	TMR1CS0	T1CKPS1	T1CKPS0	SOSCEN	T1SYNC	RD16	TMR10N
T2CON	—	T2OUTPS3	T2OUTPS2	T2OUTPS1	T2OUTPS0	TMR2ON	T2CKPS1	T2CKPS0
T3CON	TMR3CS1	TMR3CS0	T3CKPS1	T3CKPS0	SOSCEN	T3SYNC	RD16	TMR3ON
T4CON	—	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR4ON	T4CKPS1	T4CKPS0
T6CON	—	T6OUTPS3	T6OUTPS2	T6OUTPS1	T6OUTPS0	TMR6ON	T6CKPS1	T6CKPS0
T8CON		T8OUTPS3	T8OUTPS2	T8OUTPS1	T8OUTPS0	TMR8ON	T8CKPS1	T8CKPS0
T10CON ⁽¹⁾	—	T10OUTPS3	T10OUTPS2	T10OUTPS1	T10OUTPS0	TMR100N	T10CKPS1	T10CKPS0
T12CON ⁽¹⁾	—	T12OUTPS3	T12OUTPS2	T12OUTPS1	T12OUTPS0	TMR12ON	T12CKPS1	T12CKPS0
CCPR1H	Capture/Compare/PWM Register 1 High Byte							
CCPR1L	Capture/Compare/PWM Register 1 Low Byte							
CCPR2H	Capture/Compare/PWM Register 2 High Byte							
CCPR2L	Capture/Compa	Capture/Compare/PWM Register 2 Low Byte						
CCPR3H	Capture/Compa	re/PWM Regis	ter 3 High Byt	e				
CCPR3L	Capture/Compa	re/PWM Regis	ter 3 Low Byte	e				
CCP1CON	P1M1	P1M0	DC1B1	DC1B0	CCP1M3	CCP1M2	CCP1M1	CCP1M0
CCP2CON	P2M1	P2M0	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0
CCP3CON	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD

Note 1: Unimplemented on devices with a program memory of 32 Kbytes (PIC18F65K22 and PIC18F85K22).

2: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

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21.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

21.1 Master SSP (MSSP) Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface, useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D Converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit[™] (I²C[™])
 - Full Master mode
 - Slave mode (with general address call)

The I^2C interface supports the following modes in hardware:

- Master mode
- · Multi-Master mode
- Slave mode with 5-bit and 7-bit address masking (with address masking for both 10-bit and 7-bit addressing)

All members of the PIC18F87K22 family have two MSSP modules, designated as MSSP1 and MSSP2. Each module operates independently of the other.

Note: Throughout this section, generic references to an MSSP module in any of its operating modes may be interpreted as being equally applicable to MSSP1 or MSSP2. Register names and module I/O signals use the generic designator 'x' to indicate the use of a numeral to distinguish a particular module when required. Control bit names are not individuated.

21.2 Control Registers

Each MSSP module has three associated control registers. These include a status register (SSPxSTAT) and two control registers (SSPxCON1 and SSPxCON2). The use of these registers and their individual configuration bits differ significantly depending on whether the MSSP module is operated in SPI or I²C mode.

Additional details are provided under the individual sections.

Note:	In devices with more than one MSSP
	module, it is very important to pay close
	attention to SSPxCON register names.
	SSP1CON1 and SSP1CON2 control
	different operational aspects of the same
	module, while SSP1CON1 and
	SSP2CON1 control the same features for
	two different modules.

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Note: The SSPxBUF register cannot be used with read-modify-write instructions, such as BCF, COMF, etc.

To avoid lost data in Master mode, a read of the SSPxBUF must be performed to clear the Buffer Full (BF) detect bit (SSPSTAT<0>) between each transmission.

21.3 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received simultaneously. All four modes of SPI are supported. To accomplish communication, typically three pins are used:

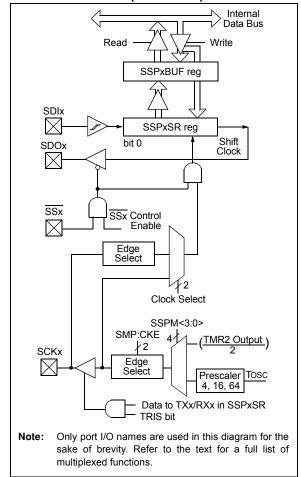
- Serial Data Out (SDOx) RC5/SDO1 or RD4/PSP4/SDO2
- Serial Data In (SDIx) RC4/SDI1/SDA1 or RD5/PSP5/SDI2/SDA2
- Serial Clock (SCKx) RC3/SCK1/SCL1 or RD6/PSP6/SCK2/SCL2

Additionally, a fourth pin may be used when in a Slave mode of operation:

• Slave Select (SSx) – RF7/AN5/SS1 or RD7/SS2

Figure 21-1 shows the block diagram of the MSSP module when operating in SPI mode.

FIGURE 21-1: MSSP BLOCK DIAGRAM (SPI MODE)



21.3.1 REGISTERS

Each MSSP module has four registers for SPI mode operation. These are:

- MSSPx Control Register 1 (SSPxCON1)
- MSSPx Status Register (SSPxSTAT)
- Serial Receive/Transmit Buffer Register (SSPxBUF)
- MSSPx Shift Register (SSPxSR) Not directly accessible

SSPxCON1 and SSPxSTAT are the control and status registers in SPI mode operation. The SSPxCON1 register is readable and writable. The lower 6 bits of the SSPxSTAT are read-only. The upper two bits of the SSPxSTAT are read/write.

SSPxSR is the shift register used for shifting data in or out. SSPxBUF is the buffer register to which data bytes are written to or read from.

In receive operations, SSPxSR and SSPxBUF together create a double-buffered receiver. When SSPxSR receives a complete byte, it is transferred to SSPxBUF and the SSPxIF interrupt is set.

During transmission, the SSPxBUF is not double-buffered. A write to SSPxBUF will write to both SSPxBUF and SSPxSR.

REGISTER 21-1: SSPxSTAT: MSSPx STATUS REGISTER (SPI MODE)

				```	,		
R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0
SMP	CKE ⁽¹⁾	D/A	Р	S	R/W	UA	BF
bit 7							bit C
Legend:							
R = Readabl		W = Writable	bit	•	mented bit, rea	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	0 = Input data <u>SPI Slave mo</u>	<u>iode:</u> a is sampled at a is sampled at	the middle of	data output tim	e		
bit 6		ock Select bit ⁽¹⁾					
bit 5	<ul> <li>1 = Transmit occurs on the transition from active to Idle clock state</li> <li>0 = Transmit occurs on the transition from Idle to active clock state</li> <li>D/A: Data/Address bit</li> </ul>						
bit 0	Used in I ² C™						
bit 4	<ul> <li>P: Stop bit</li> <li>Used in I²C mode only. This bit is cleared when the MSSPx module is disabled; SSPEN is cleared.</li> </ul>						
bit 3	<b>S:</b> Start bit Used in I ² C mode only.						
bit 2		<b>R/W:</b> Read/Write Information bit Used in I ² C mode only.					
bit 1	<b>UA:</b> Update Address bit Used in I ² C mode only.						
bit 0	1 = Receive i	Ill Status bit (Re s complete, SS s not complete,	PxBUF is full				
Note 1. P	olarity of clock s	state is set by th	e CKP hit (S				

Note 1: Polarity of clock state is set by the CKP bit (SSPxCON1<4>).

REGISTER 21-2:	SSPxCON1: MSSPx CONTROL REGISTER 1 (SPI MODE)
----------------	-----------------------------------------------

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
WCOL	SSPOV ⁽¹⁾	SSPEN ⁽²⁾	CKP	SSPM3 ⁽³⁾	SSPM2 ⁽³⁾	SSPM1 ⁽³⁾	SSPM0 ⁽³⁾		
bit 7							bit C		
Legend:									
R = Readat	ole bit	W = Writable I	oit	U = Unimplen	nented bit, rea	d as '0'			
-n = Value at POR		'1' = Bit is set		'0' = Bit is cleared		x = Bit is unknown			
bit 7		-		e it is still transm	nitting the prev	ious word (mus	t be cleared in		
	0 = No collisi								
bit 6	SSPOV: Rece	eive Overflow Ir	ndicator bit ⁽¹⁾						
	SPI Slave mode:								
	the SSP software) 0 = No overfl	xBUF, even if ). ow	only transmi	Overflow can on tting data, to a					
bit 5		Master Synchronous Serial Port Enable bit ⁽²⁾							
				Kx, SDOx, SDIx ese pins as I/O p		erial port pins			
bit 4	CKP: Clock F	olarity Select b	it						
		for the clock is for the clock is							
bit 3-0	SSPM<3:0>:	Master Synchro	onous Serial I	Port Mode Selec	ct bits ⁽³⁾				
	0101 = SPI S 0100 = SPI S 0011 = SPI N 0010 = SPI N 0001 = SPI N		ck = SCKx pir ck = SCKx pir ock = TMR2 c ock = FOSC/6 ock = FOSC/10	4		5x can be used	as I/O pin		
	n Master mode, t vriting to the SSF		is not set sind	ce each new rec	ception (and tra	ansmission) is ii	nitiated by		
0. 1		-	ha proporty of	opfigured op inn	uto or outputo				

- 2: When enabled, these pins must be properly configured as inputs or outputs.
- **3:** Bit combinations not specifically listed here are either reserved or implemented in  $I^2C^{TM}$  mode only.

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### 21.3.2 OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSPxCON1<5:0> and SSPxSTAT<7:6>). These control bits allow the following to be specified:

- · Master mode (SCKx is the clock output)
- Slave mode (SCKx is the clock input)
- Clock Polarity (Idle state of SCKx)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCKx)
- Clock Rate (Master mode only)
- · Slave Select mode (Slave mode only)

Each MSSP module consists of a Transmit/Receive Shift register (SSPxSR) and a Serial Input Buffer register (SSPxBUF). The SSPxSR shifts the data in and out of the device, MSb first. The SSPxBUF holds the data that was written to the SSPxSR until the received data is ready. Once the 8 bits of data have been received, that byte is moved to the SSPxBUF register. Then, the Buffer Full detect bit, BF (SSPxSTAT<0>), and the interrupt flag bit, SSPxIF, are set. This double-buffering of the received data (SSPxBUF) allows the next byte to start reception before reading the data that was just received. Any write to the SSPxBUF register during transmission/reception of data will be ignored and the Write Collision Detect bit, WCOL (SSPxCON1<7>), will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPxBUF register completed successfully.

When the application software is expecting to receive valid data, the SSPxBUF should be read before the next byte of data to transfer is written to the SSPxBUF. The Buffer Full bit, BF (SSPxSTAT<0>), indicates when SSPxBUF has been loaded with the received data (transmission is complete). When the SSPxBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP interrupt is used to determine when the transmission/reception has completed. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 21-1 shows the loading of the SSPxBUF (SSPxSR) for data transmission.

The SSPxSR is not directly readable or writable and can only be accessed by addressing the SSPxBUF register. Additionally, the SSPxSTAT register indicates the various status conditions.

### 21.3.3 OPEN-DRAIN OUTPUT OPTION

The drivers for the SDOx output and SCKx clock pins can be optionally configured as open-drain outputs. This feature allows the voltage level on the pin to be pulled to a higher level through an external pull-up resistor, and allows the output to communicate with external circuits without the need for additional level shifters. For more information, see Section 12.1.3 "Open-Drain Outputs".

The open-drain output option is controlled by the SSP2OD and SSP1OD bits (ODCON3<1:0>). Setting an SSPxOD bit configures the SDOx and SCKx pins for the corresponding module for open-drain operation.

**Note:** To avoid lost data in Master mode, a read of the SSPxBUF must be performed to clear the Buffer Full (BF) detect bit (SSPxSTAT<0>) between each transmission.

### EXAMPLE 21-1: LOADING THE SSP1BUF (SSP1SR) REGISTER

LOOP	BTFSS BRA MOVF	LOOP	<pre>;Has data been received (transmit complete)? ;No ;WREG reg = contents of SSP1BUF</pre>
	MOVWF	RXDATA	;Save in user RAM, if data is meaningful
	MOVF MOVWF	TXDATA, W SSP1BUF	;W reg = contents of TXDATA ;New data to xmit

### 21.3.4 ENABLING SPI I/O

To enable the serial port, MSSP Enable bit, SSPEN (SSPxCON1<5>), must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, re-initialize the SSPxCON registers and then set the SSPEN bit. This configures the SDIx, SDOx, SCKx and SSx pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed as follows:

- SDIx must have the TRISC<4> or TRISD<5> bit set
- SDOx must have the TRISC<5> or TRISD<4> bit cleared
- SCKx (Master mode) must have the TRISC<3> or TRISD<6>bit cleared
- SCKx (Slave mode) must have the TRISC<3> or TRISD<6> bit set
- SSx must have the TRISF<7> or TRISD<7> bit set

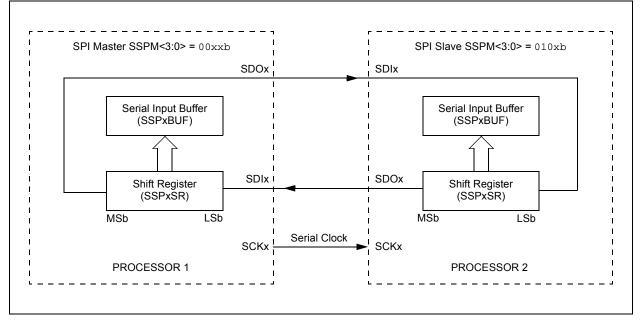
Any serial port function that is not desired may be overridden by programming the corresponding Data Direction (TRIS) register to the opposite value.

### 21.3.5 TYPICAL CONNECTION

Figure 21-2 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCKx signal. Data is shifted out of both shift registers on their programmed clock edge and latched on the opposite edge of the clock. Both processors should be programmed to the same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- · Master sends data Slave sends dummy data
- Master sends data Slave sends data
- Master sends dummy data Slave sends data

### FIGURE 21-2: SPI MASTER/SLAVE CONNECTION



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### 21.3.6 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCKx. The master determines when the slave (Processor 1, Figure 21-2) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPxBUF register is written to. If the SPI is only going to receive, the SDOx output could be disabled (programmed as an input). The SSPxSR register will continue to shift in the signal present on the SDIx pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPxBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode.

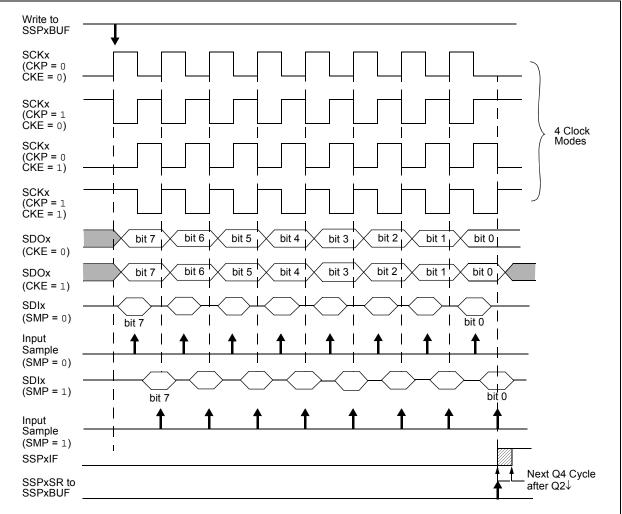
The clock polarity is selected by appropriately programming the CKP bit (SSPxCON1<4>). This, then, would give waveforms for SPI communication, as

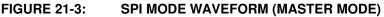
shown in Figure 21-3, Figure 21-5 and Figure 21-6, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user-programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum data rate (at 40 MHz) of 10.00 Mbps.

Figure 21-3 shows the waveforms for Master mode. When the CKE bit is set, the SDOx data is valid before there is a clock edge on SCKx. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPxBUF is loaded with the received data is shown.





### 21.3.7 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCKx. When the last bit is latched, the SSPxIF interrupt flag bit is set.

While in Slave mode, the external clock is supplied by the external clock source on the SCKx pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. When a byte is received, the device can be configured to wake up from Sleep.

### 21.3.8 SLAVE SELECT SYNCHRONIZATION

The  $\overline{SSx}$  pin allows a Synchronous Slave mode. The SPI must be in Slave mode with the  $\overline{SSx}$  pin control enabled (SSPxCON1<3:0> = 04h). When the  $\overline{SSx}$  pin is low, transmission and reception are enabled and the SDOx pin is driven. When the  $\overline{SSx}$  pin goes high, the SDOx pin is no longer driven, even if in the middle of a

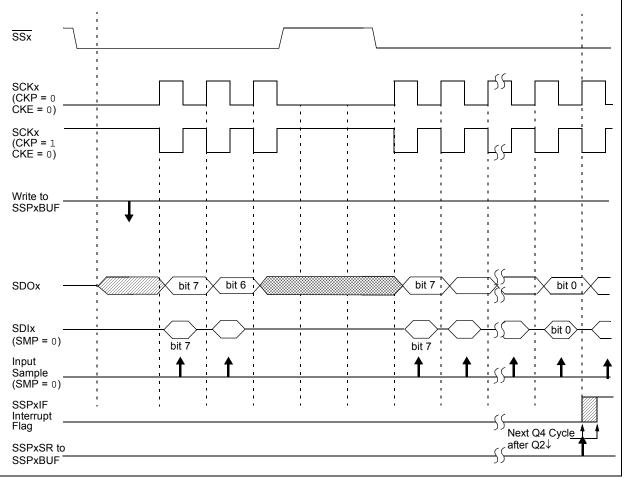
transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable depending on the application.

Note: When the SPI is in Slave mode with the SSx pin control enabled (SSPxCON1<3:0> = 0100), the SPI module will reset if the SSx pin is set to VDD.
 If the SPI is used in Slave mode with CKE set, then the SSx pin control must be enabled.

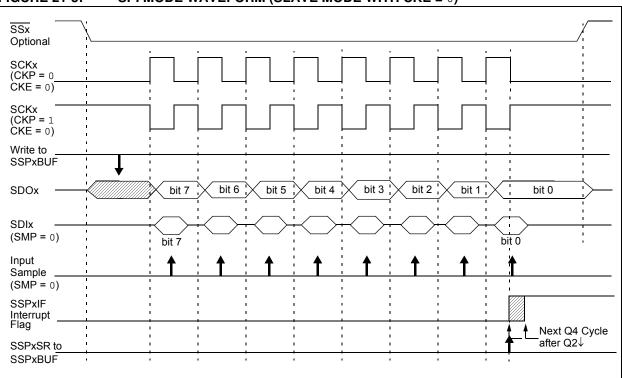
When the SPI module resets, the bit counter is forced to '0'. This can be done by either forcing the SSx pin to a high level or clearing the SSPEN bit.

To emulate two-wire communication, the SDOx pin can be connected to the SDIx pin. When the SPI needs to operate as a receiver, the SDOx pin can be configured as an input. This disables transmissions from the SDOx. The SDIx can always be left as an input (SDIx function) since it cannot create a bus conflict.



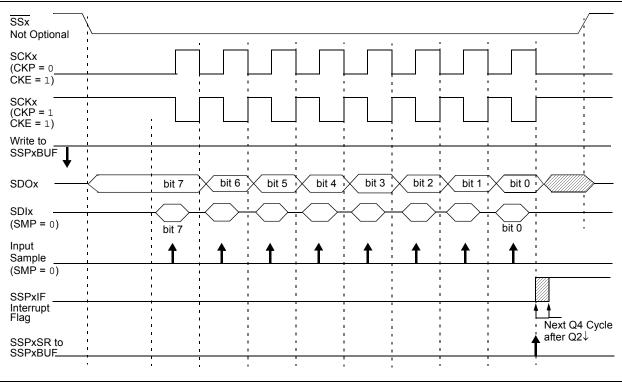


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### FIGURE 21-5: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 0)





### 21.3.9 OPERATION IN POWER-MANAGED MODES

In SPI Master mode, module clocks may be operating at a different speed than when in Full-Power mode. In the case of Sleep mode, all clocks are halted.

In Idle modes, a clock is provided to the peripherals. That clock can be from the primary clock source, the secondary clock (SOSC oscillator) or the INTOSC source. See Section 3.3 "Clock Sources and Oscillator Switching" for additional information.

In most cases, the speed that the master clocks SPI data is not important; however, this should be evaluated for each system.

If MSSP interrupts are enabled, they can wake the controller from Sleep mode, or one of the Idle modes, when the master completes sending data. If an exit from Sleep or Idle mode is not desired, MSSP interrupts should be disabled.

If the Sleep mode is selected, all module clocks are halted and the transmission/reception will remain in that state until the device wakes. After the device returns to Run mode, the module will resume transmitting and receiving data.

In SPI Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in any power-managed mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP interrupt flag bit will be set, and if enabled, will wake the device.

### 21.3.10 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

### 21.3.11 BUS MODE COMPATIBILITY

Table 21-1shows the compatibility between thestandard SPI modes and the states of the CKP andCKE control bits.

	TABLE	21-1:	SPI B	US M	ODES
--	-------	-------	-------	------	------

Standard SPI Mode	Control Bits State			
Terminology	СКР	CKE		
0, 0	0	1		
0, 1	0	0		
1, 0	1	1		
1, 1	1	0		

There is also an SMP bit which controls when the data is sampled.

## 21.3.12 SPI CLOCK SPEED AND MODULE INTERACTIONS

Because MSSP1 and MSSP2 are independent modules, they can operate simultaneously at different data rates. Setting the SSPM<3:0> bits of the SSPxCON1 register determines the rate for the corresponding module.

An exception is when both modules use Timer2 as a time base in Master mode. In this instance, any changes to the Timer2 module's operation will affect both MSSP modules equally. If different bit rates are required for each module, the user should select one of the other three time base options for one of the modules.

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TADLE 21-2	nedistens associated with set openation								
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	
PIR2	OSCFIF	—	SSP2IF	BCL2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF	
PIE2	OSCFIE	—	SSP2IE	BCL2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE	
IPR2	OSCFIP	—	SSP2IP	BCL2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP	
PIR3	TMR5GIF	—	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	
PIE3	TMR5GIE	—	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	
IPR3	TMR5GIP	—	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	
TRISD	TRISD7	TRISD6	TRISD5	TRISD4	TRISD3	TRISD2	TRISD1	TRISD0	
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—	
SSP1BUF	MSSP1 Red	MSSP1 Receive Buffer/Transmit Register							
SSP1CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	
SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	
SSP1STAT	SMP	CKE	D/Ā	Р	S	R/W	UA	BF	
SSP2CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	
SSP2CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	
SSP2STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	
SSP2BUF	MSSP2 Receive Buffer/Transmit Register								
ODCON1	SSP10D	CCP2OD	CCP10D	—	—	—	—	SSP2OD	
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD	
Logond: Sh	adad aalla ar	o not used by	the MSSD	modulo in SD	Imada				

### TABLE 21-2: REGISTERS ASSOCIATED WITH SPI OPERATION

**Legend:** Shaded cells are not used by the MSSP module in SPI mode.

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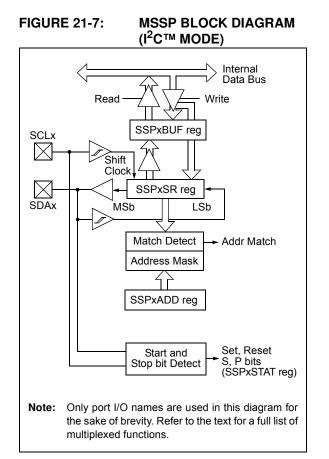
### 21.4 I²C Mode

The MSSP module in I²C mode fully implements all master and slave functions (including general call support), and provides interrupts on Start and Stop bits in hardware to determine a free bus (multi-master function). The MSSP module implements the standard mode specifications, as well as 7-bit and 10-bit addressing.

Two pins are used for data transfer:

- Serial Clock (SCLx) RC3/SCK1/SCL1 or RD6/SCK2/SCL2
- Serial Data (SDAx) RC4/SDI1/SDA1 or RD5/SDI2/SDA2

The user must configure these pins as inputs by setting the associated TRIS bits.



### 21.4.1 REGISTERS

The MSSP module has seven registers for  ${\rm I}^2{\rm C}$  operation. These are:

- MSSPx Control Register 1 (SSPxCON1)
- MSSPx Control Register 2 (SSPxCON2)
- MSSPx Status Register (SSPxSTAT)
- Serial Receive/Transmit Buffer Register (SSPxBUF)
- MSSPx Shift Register (SSPxSR) Not directly accessible
- MSSPx Address Register (SSPxADD)
- I²C Slave Address Mask Register (SSPxMSK)

SSPxCON1, SSPxCON2 and SSPxSTAT are the control and status registers in  $I^2C$  mode operation. The SSPxCON1 and SSPxCON2 registers are readable and writable. The lower 6 bits of the SSPxSTAT are read-only. The upper two bits of the SSPxSTAT are read/write.

SSPxSR is the shift register used for shifting data in or out. SSPxBUF is the buffer register to which data bytes are written to or read from.

SSPxADD contains the slave device address when the MSSP is configured in  $I^2C$  Slave mode. When the MSSP is configured in Master mode, the lower seven bits of SSPxADD act as the Baud Rate Generator reload value.

SSPxMSK holds the slave address mask value when the module is configured for 7-Bit Address Masking mode. While it is a separate register, it shares the same SFR address as SSPxADD; it is only accessible when the SSPM<3:0> bits are specifically set to permit access. Additional details are provided in Section 21.4.3.4 "7-Bit Address Masking Mode".

In receive operations, SSPxSR and SSPxBUF together, create a double-buffered receiver. When SSPxSR receives a complete byte, it is transferred to SSPxBUF and the SSPxIF interrupt is set.

During transmission, the SSPxBUF is not double-buffered. A write to SSPxBUF will write to both SSPxBUF and SSPxSR.

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R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0				
SMP	CKE	D/A	P ⁽¹⁾	S ⁽¹⁾	R/W ^(2,3)	UA	BF				
bit 7		I					bit				
Legend:											
R = Readal	ble bit	W = Writable	bit	U = Unimple	emented bit, read	d as '0'					
-n = Value a	at POR	'1' = Bit is se	t	'0' = Bit is cl		x = Bit is unkr	nown				
bit 7	SMP: Slew I	Rate Control bit									
		Slave mode:									
					ode (100 kHz and	d 1 MHz)					
L:1 0		te control is ena	abled for High-	Speea moae (	400 KHZ)						
bit 6	CKE: SMBu										
		<u>Slave mode:</u> SMBus-specific	innuts								
		SMBus-specific									
bit 5	D/A: Data/A	ddress bit									
	In Master me										
	Reserved.										
	In Slave mo										
		s that the last by s that the last by									
bit 4	P: Stop bit ⁽¹⁾										
	•	1 = Indicates that a Stop bit has been detected last									
		was not detecte									
bit 3	S: Start bit ⁽¹	)									
		s that a Start bit was not detecte		ected last							
bit 2	R/W: Read/	Write Informatio	n bit ^(2,3)								
	In Slave mo	<u>de:</u>									
	1 = Read										
	0 = Write In Master mode:										
		t is in progress									
		t is not in progre	ess								
bit 1	UA: Update	Address bit (10	-Bit Slave mod	le only)							
		s that the user r does not need	•	e the address	in the SSPxADE	) register					
bit 0	BF: Buffer F	ull Status bit									
	In Transmit r	<u>mode:</u>									
	1 = SSPxBL										
	0 = SSPxBL										
	<u>In Receive n</u> 1 = SSPxBL	<u>noue.</u> IF is full (does r	not include the	ACK and Ston	bits)						
		IF is empty (doe									
	This bit is cleare										
2:	This bit holds the address match t	e R/W bit inforn o the next Start	nation following bit, Stop bit or	the last addre not ACK bit.	ess match. This	bit is only valid	from the				

### REGISTER 21-3: SSPxSTAT: MSSPx STATUS REGISTER (I²C[™] MODE)

3: ORing this bit with SEN, RSEN, PEN, RCEN or ACKEN will indicate if the MSSPx is in Active mode.

### REGISTER 21-4: SSPxCON1: MSSPx CONTROL REGISTER 1 (I²C[™] MODE)

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
WCOL	SSPOV	SSPEN ⁽¹⁾	CKP	SSPM3 ⁽²⁾	SSPM2 ⁽²⁾	SSPM1 ⁽²⁾	SSPM0 ⁽²⁾			
bit 7						1	bit			
Legend:										
R = Reada	able bit	W = Writable I	bit	U = Unimpler	nented bit, read	d as '0'				
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle		x = Bit is unkr	nown			
bit 7	WCOL: Write	e Collision Deteo	ct bit							
		ansmit mode:			0					
		to the SSPxBU				nditions were	not valid for			
	0 = No collis	sion to be starte	ed (must be cl	eared in softwa	re)					
	In Slave Tra									
		PxBUF register i	s written while	e it is still transn	nitting the previ	ous word (mus	t be cleared i			
	software	e)			<b>C</b> .	·				
	0 = No collis									
	In Receive m This is a "doi	node (Master or	Slave modes)	<u>.</u>						
bit 6		ceive Overflow Ir	adiaatar hit							
	In Receive m									
		received while	the SSPxBUF	reaister is still I	nolding the prev	/ious byte (mus	st be cleared i			
	•	software)								
	0 = No over	flow								
	In Transmit r									
		n't care" bit in Tr								
bit 5		ster Synchronou								
		the serial port and serial port and o				ie senai port pi	ns			
bit 4		Release Control	-							
	In Slave mod									
	1 = Releases									
		ock low (clock st	retch), used to	o ensure data s	etup time					
	In Master mo Unused in th									
bit 3-0		: Master Synchro	onous Serial F	Port Mode Sele	rt hits(2)					
		Slave mode: 10-				enabled				
	1110 <b>= I²C S</b>	Slave mode: 7-bi	it address with	Start and Stop	bit interrupts e					
		Firmware Contro								
		I SSPMSK regis /laster mode: clo								
		Blave mode: 10-		(SSFXADD +	1))					
		Slave mode: 7-bi								
Note 1:	When enabled, t	he SDAx and S	CLx pins must	be configured	as inputs.					
2:	Bit combinations	not specifically	listed here are	e either reserve	d or implement	ed in SPI mod	e only.			
3:	When SSPM<3:0 SSPxMSK regist				=		-			
		er.								
4:	This mode is onl					/ISSPMSK Cor	figuration bit			

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bit 7	OR GCEN: Gener	ACKDT ⁽¹⁾ W = Writable '1' = Bit is set	ACKEN ⁽²⁾	U = Unimplen '0' = Bit is clea		RSEN ⁽²⁾	SEN ⁽²⁾ bit (
L <b>egend:</b> R = Readable n = Value at F pit 7	OR GCEN: Gener	'1' = Bit is set	bit	-		d as '0'	bit
R = Readable n = Value at F bit 7	OR GCEN: Gener	'1' = Bit is set	bit	-		d as '0'	
n = Value at F bit 7	OR GCEN: Gener	'1' = Bit is set	bit	-		d as '0'	
	GCEN: Gene			'0' = Bit is clea			
bit 7					ared	x = Bit is unkn	own
- :		al Call Enable	bit				
h 11 O	Unused in Ma	ster mode.					
bit 6	ACKSTAT: Ad	knowledge Sta	itus bit (Mastei	r Transmit mod	e only)		
		dge was not re dge was receiv		ave			
bit 5	ACKDT: Ackr	owledge Data	bit (Master Re	ceive mode onl	y) ⁽¹⁾		
	1 = Not Ackno 0 = Acknowle						
bit 4	ACKEN: Ackr	nowledge Sequ	ence Enable b	pit ⁽²⁾			
	Automati	Acknowledge cally cleared by edge sequence	hardware.	SDAx and SO	CLx pins and	transmits ACk	(DT data bi
bit 3				e mode only) ⁽²⁾			
		Receive mode f					
bit 2	PEN: Stop Co	ndition Enable	bit ⁽²⁾				
		top condition o		CLx pins. Autor	matically clear	ed by hardware.	
bit 1	RSEN: Repea	ated Start Cond	lition Enable bi	t ⁽²⁾			
		Repeated Start Start condition		DAx and SCLx	pins. Automa	tically cleared by	/ hardware.
bit 0	SEN: Start Co	ndition Enable	bit ⁽²⁾				
	1 = Initiates S 0 = Start conc		n SDAx and S	CLx pins. Auto	matically clear	ed by hardware.	

### **REGISTER 21-5:** SSPxCON2: MSSPx CONTROL REGISTER 2 (I²C[™] MASTER MODE)

2: If the l²C module is active, these bits may not be set (no spooling) and the SSPxBUF may not be written (or writes to the SSPxBUF are disabled).

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
GCEN	ACKSTAT	ACKDT ⁽¹⁾	ACKEN ⁽¹⁾	RCEN ⁽¹⁾	PEN ⁽¹⁾	RSEN ⁽¹⁾	SEN ⁽¹⁾
bit 7							bit
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimplen	nented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	own
bit 7	1 = Enables ir	ral Call Enable nterrupt when a all address is c	general call a	ddress (0000h	) is received in	the SSPxSR	
bit 6	ACKSTAT: Ac Unused in Sla	cknowledge Sta ave mode.	itus bit				
bit 5	ACKDT: Ackn 1 = Not Ackno 0 = Acknowle		bit (Master Red	ceive mode onl	y) ⁽¹⁾		
bit 4	1 = Initiates Automati	nowledge Sequ Acknowledge cally cleared by edge sequence	sequence on / hardware.		CLx pins and	transmits ACk	(DT data bi
bit 3	RCEN: Recein 1 = Enables F 0 = Receive is	ve Enable bit (I Receive mode f s Idle	Master Receive or I ² C	e mode only) ⁽¹⁾			
bit 2	<b>PEN:</b> Stop Co 1 = Initiates S 0 = Stop cond		bit ⁽¹⁾ n SDAx and S0	CLx pins. Auto	matically cleare	ed by hardware.	
bit 1	1 = Initiates F	ated Start Cond Repeated Start d Start condition	condition on SI		pins. Automat	ically cleared by	/ hardware.
bit 0	SEN: Stretch 1 = Clock stre 0 = Clock stre	tching is enabl		ve transmit and	l slave receive	(stretch enable	d)

**Note 1:** If the I²C module is active, this bit may not be set (no spooling) and the SSPxBUF may not be written (or writes to the SSPxBUF are disabled).

### REGISTER 21-7: SSPxMSK: I²C[™] SLAVE ADDRESS MASK REGISTER (7-BIT MASKING MODE)⁽¹⁾

					•		,
R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0 ⁽²⁾
bit 7				·	•		bit 0
Legend:							
R = Readable bit W = Writable bit			U = Unimplemented bit, read as '0'				
-n = Value at F	n = Value at POR '1' = Bit is set			'0' = Bit is clea	ared	x = Bit is unk	nown

bit 7-0 MSK<7:0>: Slave Address Mask Select bit

1 = Masking of corresponding bit of SSPxADD is enabled

0 = Masking of corresponding bit of SSPxADD is disabled

Note 1: This register shares the same SFR address as SSPxADD and is only addressable in select MSSPx operating modes. See Section 21.4.3.4 "7-Bit Address Masking Mode" for more details.

2: MSK0 is not used as a mask bit in 7-bit addressing.

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### 21.4.2 OPERATION

The MSSP module functions are enabled by setting the MSSP Enable bit, SSPEN (SSPxCON1<5>).

The SSPxCON1 register allows control of the  $I^2C$  operation. Four mode selection bits (SSPxCON1<3:0>) allow one of the following  $I^2C$  modes to be selected:

- I²C Master mode, clock
- I²C Slave mode (7-bit address)
- I²C Slave mode (10-bit address)
- I²C Slave mode (7-bit address) with Start and Stop bit interrupts enabled
- I²C Slave mode (10-bit address) with Start and Stop bit interrupts enabled
- I²C Firmware Controlled Master mode, slave is Idle

Selection of any I²C mode with the SSPEN bit set forces the SCLx and SDAx pins to be open-drain, provided these pins are programmed as inputs by setting the appropriate TRISC or TRISD bits. To ensure proper operation of the module, pull-up resistors must be provided externally to the SCLx and SDAx pins.

### 21.4.3 SLAVE MODE

In Slave mode, the SCLx and SDAx pins must be configured as inputs (TRISC<4:3> set). The MSSP module will override the input state with the output data when required (slave-transmitter).

The  $I^2C$  Slave mode hardware will always generate an interrupt on an address match. Address masking will allow the hardware to generate an interrupt for more than one address (up to 31 in 7-bit addressing and up to 63 in 10-bit addressing). Through the mode select bits, the user can also choose to interrupt on Start and Stop bits.

When an address is matched, or the data transfer after an address match is received, the hardware automatically will generate the Acknowledge (ACK) pulse and load the SSPxBUF register with the received value currently in the SSPxSR register.

Any combination of the following conditions will cause the MSSP module not to give this ACK pulse:

- The Buffer Full bit, BF (SSPxSTAT<0>), was set before the transfer was received.
- The overflow bit, SSPOV (SSPxCON1<6>), was set before the transfer was received.

In this case, the SSPxSR register value is not loaded into the SSPxBUF, but bit, SSPxIF, is set. The BF bit is cleared by reading the SSPxBUF register, while bit, SSPOV, is cleared through software.

The SCLx clock input must have a minimum high and low for proper operation. The high and low times of the  $I^2C$  specification, as well as the requirement of the MSSP module, are shown in timing Parameter 100 and Parameter 101.

### 21.4.3.1 Addressing

Once the MSSP module has been enabled, it waits for a Start condition to occur. Following the Start condition, the 8 bits are shifted into the SSPxSR register. All incoming bits are sampled with the rising edge of the clock (SCLx) line. The value of register, SSPxSR<7:1>, is compared to the value of the SSPxADD register. The address is compared on the falling edge of the eighth clock (SCLx) pulse. If the addresses match and the BF and SSPOV bits are clear, the following events occur:

- 1. The SSPxSR register value is loaded into the SSPxBUF register.
- 2. The Buffer Full bit, BF, is set.
- 3. An ACK pulse is generated.
- 4. The MSSP Interrupt Flag bit, SSPxIF, is set (and an interrupt is generated, if enabled) on the falling edge of the ninth SCLx pulse.

In 10-Bit Addressing mode, two address bytes need to be received by the slave. The five Most Significant bits (MSbs) of the first address byte specify if this is a 10-bit address. The  $R/\overline{W}$  (SSPxSTAT<2>) bit must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '11110 A9 A8 0', where 'A9' and 'A8' are the two MSbs of the address. The sequence of events for 10-bit addressing is as follows, with Steps 7 through 9 for the slave-transmitter:

- 1. Receive first (high) byte of address (bits, SSPxIF, BF and UA, are set on address match).
- 2. Update the SSPxADD register with second (low) byte of address (clears bit, UA, and releases the SCLx line).
- 3. Read the SSPxBUF register (clears bit, BF) and clear flag bit, SSPxIF.
- 4. Receive second (low) byte of address (bits, SSPxIF, BF and UA, are set).
- 5. Update the SSPxADD register with the first (high) byte of address. If match releases SCLx line, this will clear bit, UA.
- 6. Read the SSPxBUF register (clears bit, BF) and clear flag bit, SSPxIF.
- 7. Receive Repeated Start condition.
- 8. Receive first (high) byte of address (bits, SSPxIF and BF, are set).
- 9. Read the SSPxBUF register (clears bit, BF) and clear flag bit, SSPxIF.

### 21.4.3.2 Address Masking Modes

Masking an address bit causes that bit to become a "don't care". When one address bit is masked, two addresses will be Acknowledged and cause an interrupt. It is possible to mask more than one address bit at a time, which greatly expands the number of addresses Acknowledged.

The  $l^2C$  slave behaves the same way whether address masking is used or not. However, when address masking is used, the  $l^2C$  slave can Acknowledge multiple addresses and cause interrupts. When this occurs, it is necessary to determine which address caused the interrupt by checking the SSPxBUF.

The PIC18F87K22 family of devices is capable of using two different Address Masking modes in I²C slave operation: 5-Bit Address Masking and 7-Bit Address Masking. The Masking mode is selected at device configuration using the MSSPMSK Configuration bit. The default device configuration is 7-Bit Address Masking.

Both Masking modes, in turn, support address masking of 7-bit and 10-bit addresses. The combination of Masking modes and addresses provides different ranges of Acknowledgable addresses for each combination.

While both Masking modes function in roughly the same manner, the way they use address masks are different.

### 21.4.3.3 5-Bit Address Masking Mode

As the name implies, 5-Bit Address Masking mode uses an address mask of up to 5 bits to create a range of addresses to be Acknowledged, using bits, 5 through 1, of the incoming address. This allows the module to Acknowledge up to 31 addresses when using 7-bit addressing, or 63 addresses with 10-bit addressing (see Example 21-2). This Masking mode is selected when the MSSPMSK Configuration bit is programmed ('0').

The address mask in this mode is stored in the SSPxCON2 register, which stops functioning as a control register in I²C Slave mode (Register 21-6). In 7-Bit Address Masking mode, address mask bits, ADMSK<5:1> (SSPxCON2<5:1>), mask the corresponding address bits in the SSPxADD register. For any ADMSK bits that are set (ADMSK<n> = 1), the corresponding address bit is ignored (SSPxADD<n> = x). For the module to issue an address Acknowledge, it is sufficient to match only on addresses that do not have an active address mask.

In 10-Bit Address Masking mode, bits, ADMSK<5:2>, mask the corresponding address bits in the SSPxADD register. In addition, ADMSK1 simultaneously masks the two LSbs of the address (SSPxADD<1:0>). For any ADMSK bits that are active (ADMSK<n> = 1), the corresponding address bit is ignored (SPxADD<n> = x). Also note, that although in 10-Bit Address Masking mode, the upper address bits reuse part of the SSPxADD register bits. The address mask bits do not interact with those bits; they only affect the lower address bits.

- Note 1: ADMSK1 masks the two Least Significant bits of the address.
  - The two Most Significant bits of the address are not affected by address masking.

### EXAMPLE 21-2: ADDRESS MASKING EXAMPLES IN 5-BIT MASKING MODE

### 7-Bit Addressing:

SSPxADD<7:1>= A0h (1010000) (SSPxADD<0> is assumed to be '0')

### ADMSK<5:1> = 00111

Addresses Acknowledged: A0h, A2h, A4h, A6h, A8h, AAh, ACh, AEh

### 10-Bit Addressing:

SSPxADD<7:0> = A0h (10100000) (The two MSb of the address are ignored in this example, since they are not affected by masking)

ADMSK<5:1> = 00111

Addresses Acknowledged: A0h, A1h, A2h, A3h, A4h, A5h, A6h, A7h, A8h, A9h, AAh, ABh, ACh, ADh, AEh, AFh

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### 21.4.3.4 7-Bit Address Masking Mode

Unlike 5-bit masking, 7-Bit Address Masking mode uses a mask of up to 8 bits (in 10-bit addressing) to define a range of addresses that can be Acknowledged, using the lowest bits of the incoming address. This allows the module to Acknowledge up to 127 different addresses with 7-bit addressing, or 255 with 10-bit addressing (see Example 21-3). This mode is the default configuration of the module, which is selected when MSSPMSK is unprogrammed ('1').

The address mask for 7-Bit Address Masking mode is stored in the SSPxMSK register, instead of the SSPxCON2 register. SSPxMSK is a separate hardware register within the module, but it is not directly addressable. Instead, it shares an address in the SFR space with the SSPxADD register. To access the SSPxMSK register, it is necessary to select MSSP mode, '1001' (SSPxCON1<3:0> = 1001) and then read or write to the location of SSPxADD.

To use 7-Bit Address Masking mode, it is necessary to initialize SSPxMSK with a value before selecting the  $I^2C$  Slave Addressing mode. Thus, the required sequence of events is:

- 1. Select SSPxMSK Access mode (SSPxCON2<3:0> = 1001).
- 2. Write the mask value to the appropriate SSPADD register address (FC8h for MSSP1, F6Eh for MSSP2).
- 3. Set the appropriate I²C Slave mode (SSPxCON2<3:0> = 0111 for 10-bit addressing, '0110' for 7-bit addressing).

Setting or clearing mask bits in SSPxMSK behaves in the opposite manner of the ADMSK bits in 5-Bit Address Masking mode. That is, clearing a bit in SSPxMSK causes the corresponding address bit to be masked; setting the bit requires a match in that position. SSPxMSK resets to all '1's upon any Reset condition and, therefore, has no effect on the standard MSSP operation until written with a mask value.

With 7-bit addressing, SSPxMSK<7:1> bits mask the corresponding address bits in the SSPxADD register. For any SSPxMSK bits that are active (SSPxMSK<n> = 0), the corresponding SSPxADD address bit is ignored (SSPxADD<n> = x). For the module to issue an address Acknowledge, it is sufficient to match only on addresses that do not have an active address mask.

With 10-bit addressing, SSPxMSK<7:0> bits mask the corresponding address bits in the SSPxADD register. For any SSPxMSK bits that are active (= 0), the corresponding SSPxADD address bit is ignored (SSPxADD

Note: The two Most Significant bits of the address are not affected by address masking.

### EXAMPLE 21-3: ADDRESS MASKING EXAMPLES IN 7-BIT MASKING MODE

### 7-Bit Addressing:

SSPxADD<7:1> = 1010 000

SSPxMSK<7:1> = 1111 001

Addresses Acknowledged = ACh, A8h, A4h, A0h

### 10-Bit Addressing:

SSPxADD<7:0> = 1010 0000 (The two MSb are ignored in this example since they are not affected)

SSPxMSK<5:1> = 1111 0011

Addresses Acknowledged = ACh, A8h, A4h, A0h

### 21.4.3.5 Reception

When the  $R/\overline{W}$  bit of the address byte is clear and an address match occurs, the  $R/\overline{W}$  bit of the SSPxSTAT register is cleared. The received address is loaded into the SSPxBUF register and the SDAx line is held low (ACK).

When the address byte overflow condition exists, then the no Acknowledge (ACK) pulse is given. An overflow condition is defined as either bit, BF (SSPxSTAT<0>), is set or bit, SSPOV (SSPxCON1<6>), is set.

An MSSP interrupt is generated for each data transfer byte. The interrupt flag bit, SSPxIF, must be cleared in software. The SSPxSTAT register is used to determine the status of the byte.

If SEN is enabled (SSPxCON2<0> = 1), SCLx will be held low (clock stretch) following each data transfer. The clock must be released by setting bit, CKP (SSPxCON1<4>). See Section 21.4.4 "Clock Stretching" for more details.

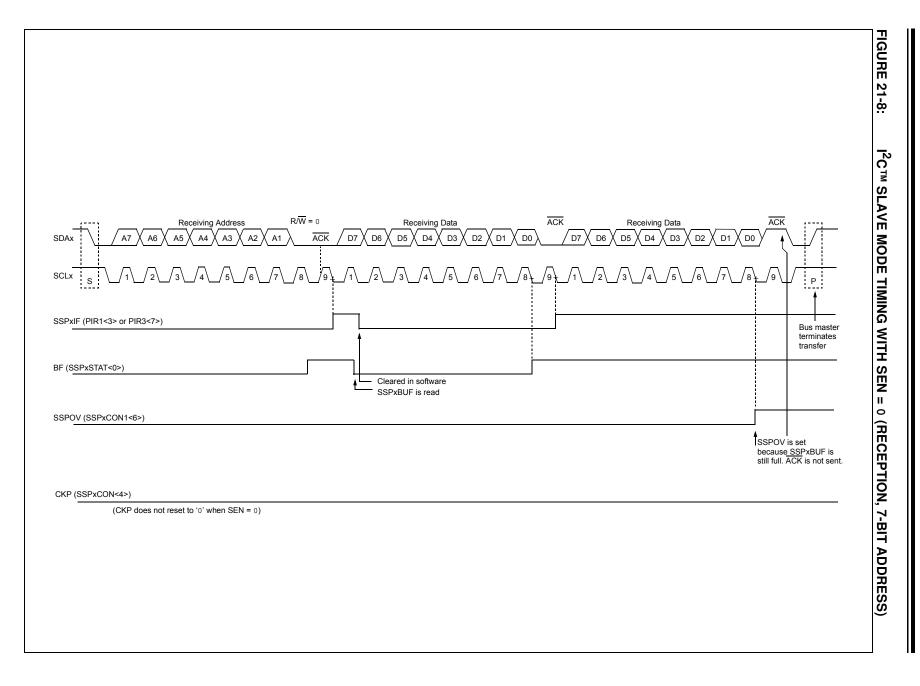
### 21.4.3.6 Transmission

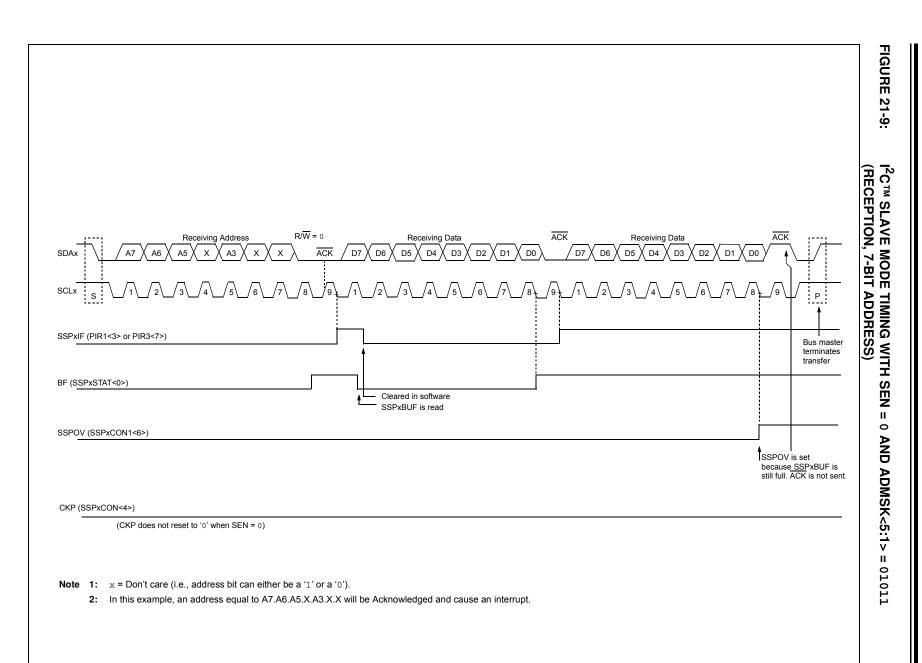
When the  $R/\overline{W}$  bit of the incoming address byte is set and an address match occurs, the  $R/\overline{W}$  bit of the SSPxSTAT register is set. The received address is loaded into the SSPxBUF register. The ACK pulse will be sent on the ninth bit and pin SCLx is held low regardless of SEN (see Section 21.4.4 "Clock Stretching" for more details). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data. The transmit data must be loaded into the SSPxBUF register which also loads the SSPxSR register. Then, pin SCLx should be enabled by setting bit, CKP (SSPxCON1<4>). The eight data bits are shifted out on the falling edge of the SCLx input. This ensures that the SDAx signal is valid during the SCLx high time (Figure 21-10).

The  $\overline{ACK}$  pulse from the master-receiver is latched on the rising edge of the ninth SCLx input pulse. If the SDAx line is high (not  $\overline{ACK}$ ), then the data transfer is complete. In this case, when the  $\overline{ACK}$  is latched by the slave, the slave logic is reset and the slave monitors for another occurrence of the Start bit. If the SDAx line was low ( $\overline{ACK}$ ), the next transmit data must be loaded into the SSPxBUF register. Again, pin SCLx must be enabled by setting bit, CKP.

An MSSP interrupt is generated for each data transfer byte. The SSPxIF bit must be cleared in software and the SSPxSTAT register is used to determine the status of the byte. The SSPxIF bit is set on the falling edge of the ninth clock pulse.

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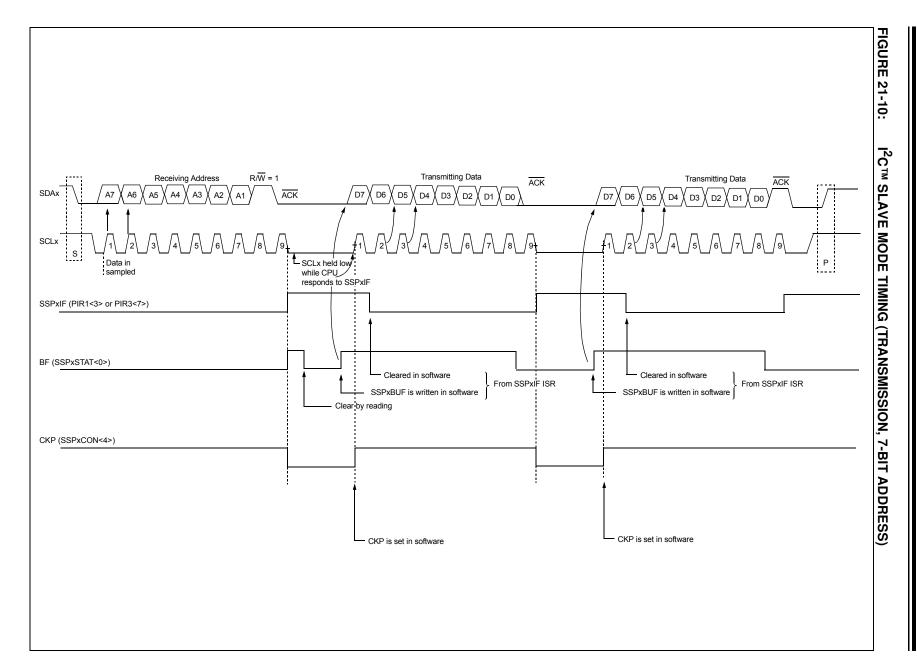


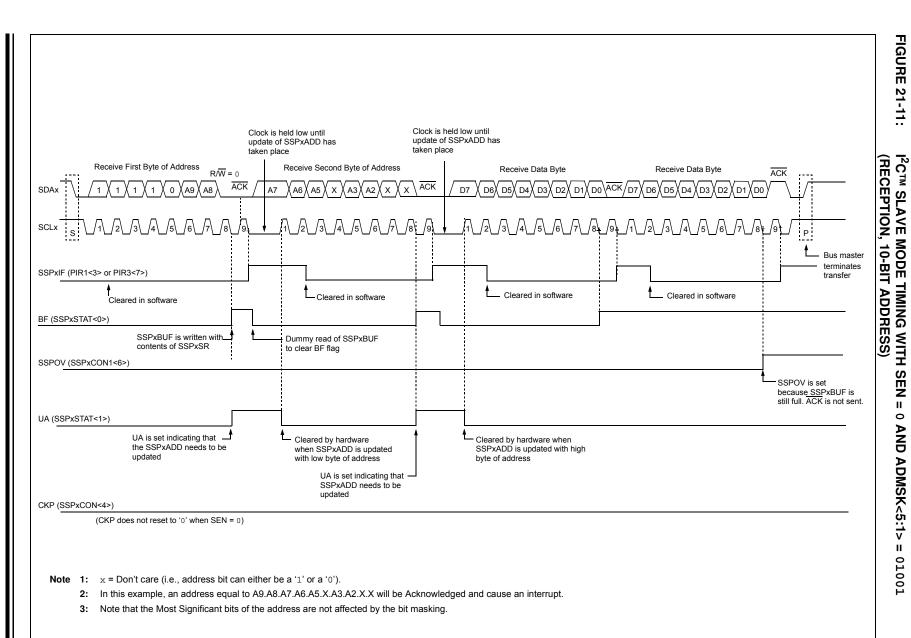


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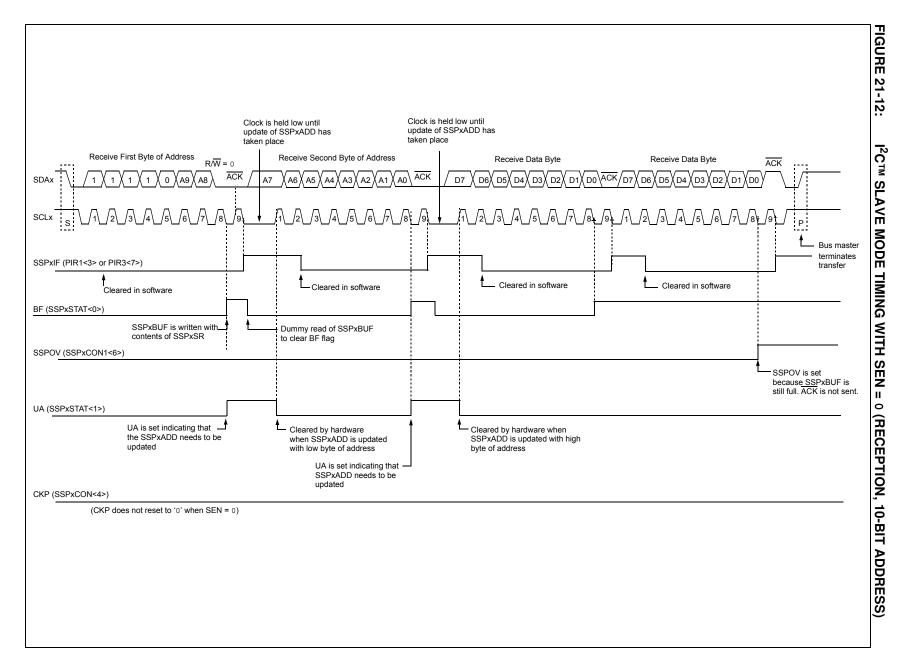






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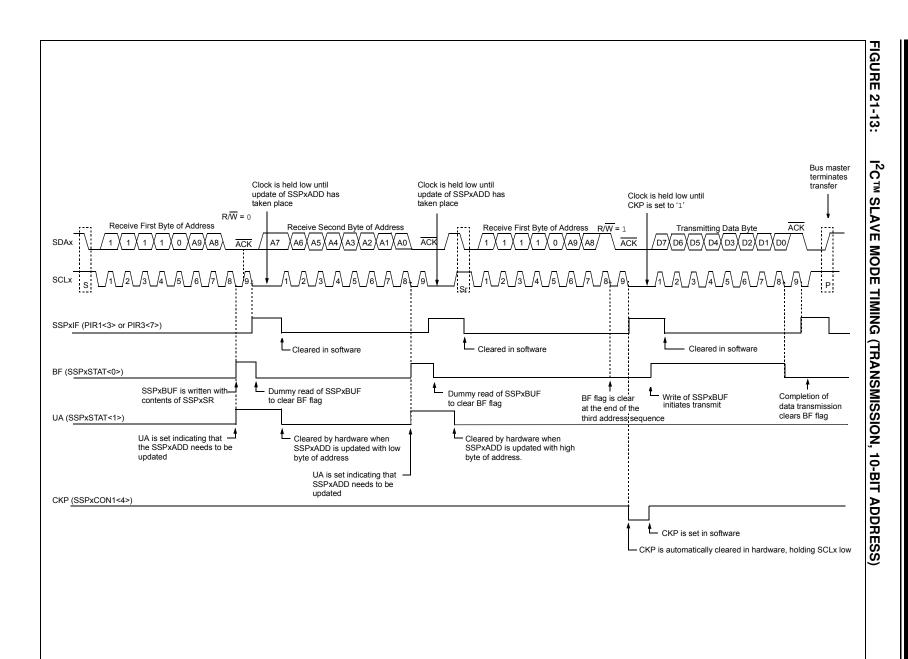


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### 21.4.4 CLOCK STRETCHING

Both 7-Bit and 10-Bit Slave modes implement automatic clock stretching during a transmit sequence.

The SEN bit (SSPxCON2<0>) allows clock stretching to be enabled during receives. Setting SEN will cause the SCLx pin to be held low at the end of each data receive sequence.

### 21.4.4.1 Clock Stretching for 7-Bit Slave Receive Mode (SEN = 1)

In 7-Bit Slave Receive mode, <u>on the falling edge of the</u> ninth clock at the end of the ACK sequence, if the BF bit is set, the CKP bit in the SSPxCON1 register is automatically cleared, forcing the SCLx output to be held low. The CKP bit being cleared to '0' will assert the SCLx line low. The CKP bit must be set in the user's ISR before reception is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and read the contents of the SSPxBUF before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring (see Figure 21-15).

- Note 1: If the user reads the contents of the SSPxBUF before the falling edge of the ninth clock, thus clearing the BF bit, the CKP bit will not be cleared and clock stretching will not occur.
  - 2: The CKP bit can be set in software regardless of the state of the BF bit. The user should be careful to clear the BF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

### 21.4.4.2 Clock Stretching for 10-Bit Slave Receive Mode (SEN = 1)

In 10-Bit Slave Receive mode, during the address sequence, clock stretching automatically takes place but CKP is not cleared. During this time, if the UA bit is set after the ninth clock, clock stretching is initiated. The UA bit is set after receiving the upper byte of the 10-bit address and following the receive of the second byte of the 10-bit address with the R/W bit cleared to '0'. The release of the clock line occurs upon updating SSPxADD. Clock stretching will occur on each data receive sequence as described in 7-bit mode.

Note: If the user polls the UA bit and clears it by updating the SSPxADD register before the falling edge of the ninth clock occurs, and if the user hasn't cleared the BF bit by reading the SSPxBUF register before that time, then the CKP bit will still NOT be asserted low. Clock stretching, on the basis of the state of the BF bit, only occurs during a data sequence, not an address sequence.

### 21.4.4.3 Clock Stretching for 7-Bit Slave Transmit Mode

The 7-Bit Slave Transmit mode implements clock stretching by clearing the CKP bit after the falling edge of the ninth clock if the BF bit is clear. This occurs regardless of the state of the SEN bit.

The user's ISR must set the CKP bit before transmission is allowed to continue. By holding the SCLx line low, the user has time to service the ISR and load the contents of the SSPxBUF before the master device can initiate another transmit sequence (see Figure 21-10).

- Note 1: If the user loads the contents of SSPxBUF, setting the BF bit before the falling edge of the ninth clock, the CKP bit will not be cleared and clock stretching will not occur.
  - **2:** The CKP bit can be set in software regardless of the state of the BF bit.

### 21.4.4.4 Clock Stretching for 10-Bit Slave Transmit Mode

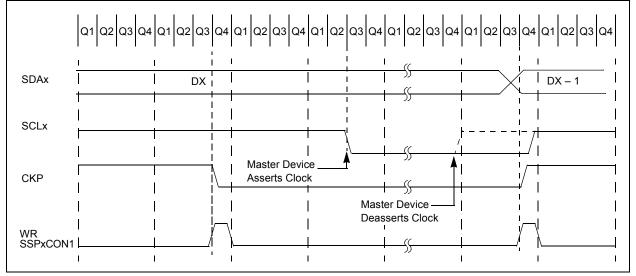
In 10-Bit Slave Transmit mode, clock stretching is controlled during the first two address sequences by the state of the UA bit, just as it is in 10-Bit Slave Receive mode. The first two addresses are followed by a third address sequence, which contains the high-order bits of the 10-bit address and the R/W bit set to '1'. After the third address sequence is performed, the UA bit is not set, the module is now configured in Transmit mode and clock stretching is controlled by the BF flag as in 7-Bit Slave Transmit mode (see Figure 21-13).

## 21.4.4.5 Clock Synchronization and the CKP bit

When the CKP bit is cleared, the SCLx output is forced to '0'. However, clearing the CKP bit will not assert the SCLx output low until the SCLx output is already sampled low. Therefore, the CKP bit will not assert the SCLx line until an external  $I^2C$  master device has

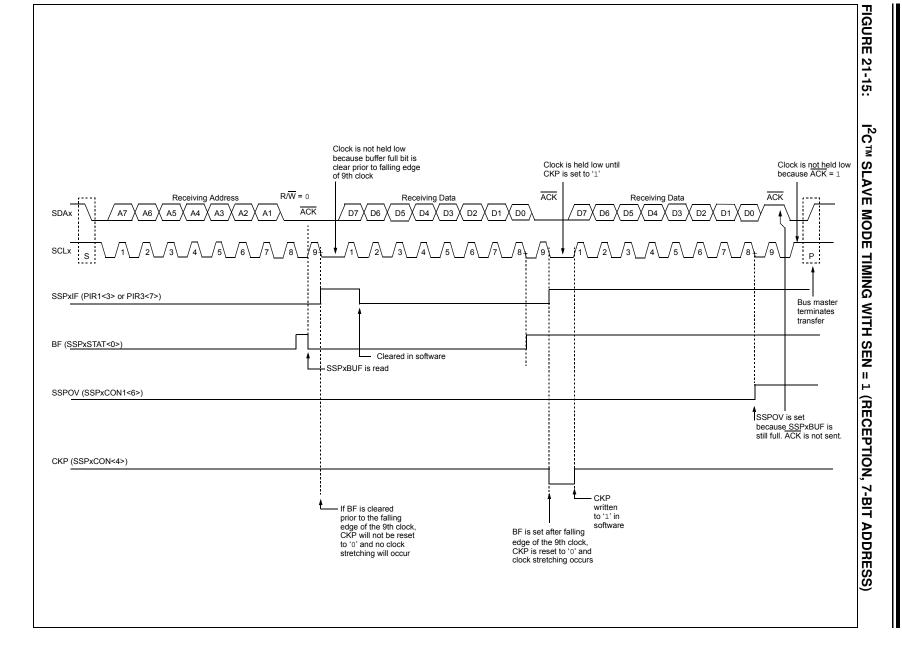
already asserted the SCLx line. The SCLx output will remain low until the CKP bit is set and all other devices on the  $I^2C$  bus have deasserted SCLx. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCLx (see Figure 21-14).





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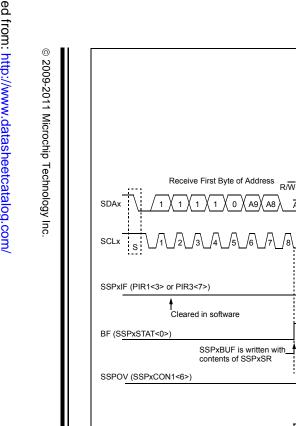


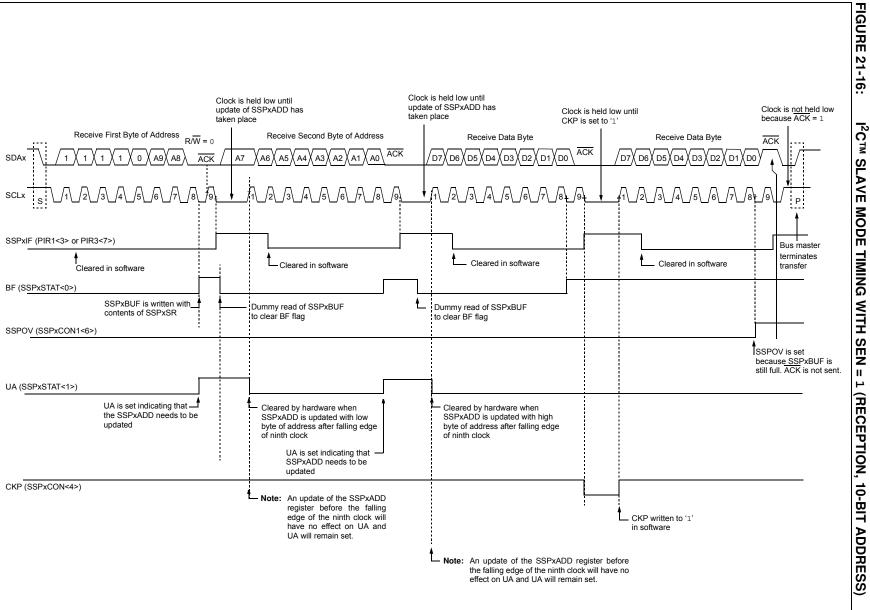
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### 21.4.5 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I²C bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an Acknowledge.

The general call address is one of eight addresses reserved for specific purposes by the  $I^2C$  protocol. It consists of all '0's with R/W = 0.

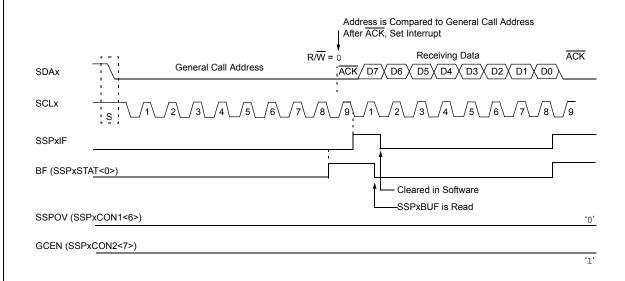
The general call address is recognized when the General Call Enable bit, GCEN, is enabled (SSPxCON2<7> set). Following a Start bit detect, eight bits are shifted into the SSPxSR and the address is compared against the SSPxADD. It is also compared to the general call address and fixed in hardware.

If the general call address matches, the SSPxSR is transferred to the SSPxBUF, the BF flag bit is set (eighth bit), and on the falling edge of the ninth bit (ACK bit), the SSPxIF interrupt flag bit is set.

When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPxBUF. The value can be used to determine if the address was device-specific or a general call address.

In 10-Bit Addressing mode, the SSPxADD is required to be updated for the second half of the address to match and the UA bit is set (SSPxSTAT<1>). If the general call address is sampled when the GCEN bit is set, while the slave is configured in 10-Bit Addressing mode, then the second half of the address is not necessary, the UA bit will not be set and the slave will begin receiving data after the Acknowledge (Figure 21-17).





### 21.4.6 MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPxCON1, and by setting the SSPEN bit. In Master mode, the SCLx and SDAx lines are manipulated by the MSSP hardware if the TRIS bits are set.

The Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the  $I^2C$  bus may be taken when the P bit is set, or the bus is Idle, with both the S and P bits clear.

In Firmware Controlled Master mode, user code conducts all  $I^2C$  bus operations based on Start and Stop bit conditions.

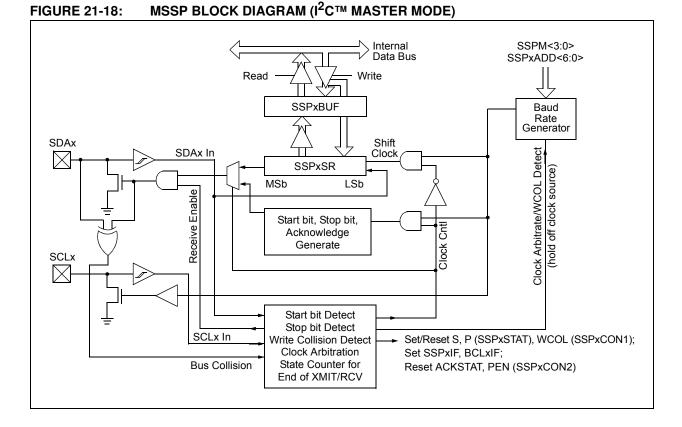
Once Master mode is enabled, the user has six options.

- 1. Assert a Start condition on SDAx and SCLx.
- 2. Assert a Repeated Start condition on SDAx and SCLx.
- 3. Write to the SSPxBUF register, initiating transmission of data/address.
- 4. Configure the I²C port to receive data.
- 5. Generate an Acknowledge condition at the end of a received byte of data.
- 6. Generate a Stop condition on SDAx and SCLx.

Note: The MSSP module, when configured in I²C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPxBUF register to initiate transmission before the Start condition is complete. In this case, the SSPxBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPxBUF did not occur.

The following events will cause the MSSP Interrupt Flag bit, SSPxIF, to be set (and MSSP interrupt, if enabled):

- Start condition
- Stop condition
- Data transfer byte transmitted/received
- Acknowledge transmitted
- Repeated Start



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### 21.4.6.1 I²C[™] Master Mode Operation

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDAx while SCLx outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted, 8 bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the  $R/\overline{W}$  bit. In this case, the  $R/\overline{W}$  bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address, followed by a '1' to indicate the receive bit. Serial data is received via SDAx, while SCLx outputs the serial clock. Serial data is received, 8 bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

The Baud Rate Generator, used for the SPI mode operation, is used to set the SCLx clock frequency for either 100 kHz, 400 kHz or 1 MHz I²C operation. See **Section 21.4.7 "Baud Rate**" for more details.

A typical transmit sequence would go as follows:

- 1. The user generates a Start condition by setting the Start Enable bit, SEN (SSPxCON2<0>).
- SSPxIF is set. The MSSP module will wait the required start time before any other operation takes place.
- 3. The user loads the SSPxBUF with the slave address to transmit.
- 4. Address is shifted out the SDAx pin until all 8 bits are transmitted.
- The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPxCON2 register (SSPxCON2<6>).
- The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
- 7. The user loads the SSPxBUF with eight bits of data.
- 8. Data is shifted out the SDAx pin until all 8 bits are transmitted.
- The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPxCON2 register (SSPxCON2<6>).
- 10. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPxIF bit.
- 11. The user generates a Stop condition by setting the Stop Enable bit, PEN (SSPxCON2<2>).
- 12. Interrupt is generated once the Stop condition is complete.

### 21.4.7 BAUD RATE

In I²C Master mode, the Baud Rate Generator (BRG) reload value is placed in the lower 7 bits of the SSPxADD register (Figure 21-19). When a write occurs to SSPxBUF, the Baud Rate Generator will automatically begin counting. The BRG counts down to 0 and stops until another reload has taken place. The BRG count is decremented twice per instruction cycle (TcY) on the Q2 and Q4 clocks. In I²C Master mode, the BRG is reloaded automatically.

Once the given operation is complete (i.e., transmission of the last data bit is followed by ACK), the internal clock will automatically stop counting and the SCLx pin will remain in its last state.

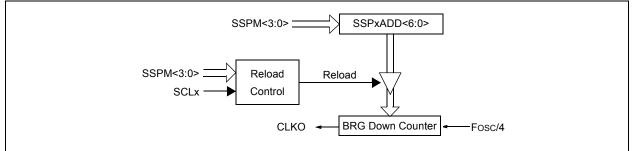
Table 21-3 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSPxADD. The SSPxADD BRG value of  $0 \ge 00$  is not supported.

### 21.4.7.1 Baud Rate and Module Interdependence

Because MSSP1 and MSSP2 are independent, they can operate simultaneously in I²C Master mode at different baud rates. This is done by using different BRG reload values for each module.

Because this mode derives its basic clock source from the system clock, any changes to the clock will affect both modules in the same proportion. It may be possible to change one or both baud rates back to a previous value by changing the BRG reload value.

### FIGURE 21-19: BAUD RATE GENERATOR BLOCK DIAGRAM



### TABLE 21-3: I²C[™] CLOCK RATE w/BRG

Fosc	Fcy	Fcy * 2	BRG Value	FscL (2 Rollovers of BRG)
40 MHz	10 MHz	20 MHz	18h	400 kHz
40 MHz	10 MHz	20 MHz	1Fh	312.5 kHz
40 MHz	10 MHz	20 MHz	63h	100 kHz
16 MHz	4 MHz	8 MHz	09h	400 kHz
16 MHz	4 MHz	8 MHz	0Ch	308 kHz
16 MHz	4 MHz	8 MHz	27h	100 kHz
4 MHz	1 MHz	2 MHz	02h	333 kHz
4 MHz	1 MHz	2 MHz	09h	100 kHz
16 MHz	4 MHz	8 MHz	03h	1 MHz ⁽¹⁾

**Note 1:** A minimum of 16 MHz Fosc is required to get 1 MHz  $l^2C$ .

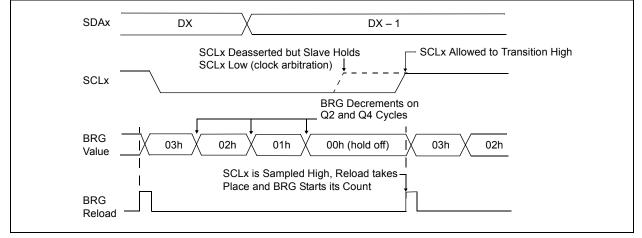
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### 21.4.7.2 Clock Arbitration

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, deasserts the SCLx pin (SCLx allowed to float high). When the SCLx pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCLx pin is actually sampled high. When the

SCLx pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and begins counting. This ensures that the SCLx high time will always be at least one BRG rollover count in the event that the clock is held low by an external device (Figure 21-20).





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### 21.4.8 I²C[™] MASTER MODE START CONDITION TIMING

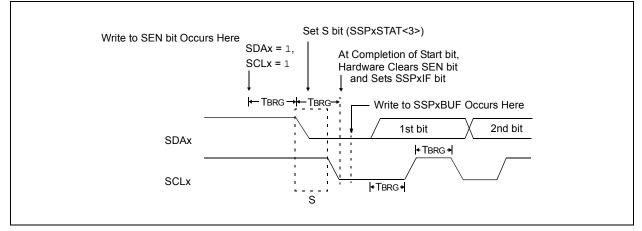
To initiate a Start condition, the user sets the Start Enable bit, SEN (SSPxCON2<0>). If the SDAx and SCLx pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and starts its count. If SCLx and SDAx are both sampled high when the Baud Rate Generator times out (TBRG), the SDAx pin is driven low. The action of the SDAx being driven low while SCLx is high is the Start condition and causes the S bit (SSPxSTAT<3>) to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit (SSPxCON2<0>) will be automatically cleared by hardware. The Baud Rate Generator is suspended, leaving the SDAx line held low and the Start condition is complete.

Note: If, at the beginning of the Start condition, the SDAx and SCLx pins are already sampled low, or if during the Start condition, the SCLx line is sampled low before the SDAx line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag, BCLxIF, is set, the Start condition is aborted and the I²C module is reset into its Idle state.

### 21.4.8.1 WCOL Status Flag

If the user writes the SSPxBUF when a Start sequence is in progress, the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSPxCON2 is disabled until the Start condition is complete.



### FIGURE 21-21: FIRST START BIT TIMING

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### 21.4.9 I²C[™] MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit (SSPxCON2<1>) is programmed high and the I²C logic module is in the Idle state. When the RSEN bit is set, the SCLx pin is asserted low. When the SCLx pin is sampled low, the Baud Rate Generator is loaded with the contents of SSPxADD<5:0> and begins counting. The SDAx pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, and if SDAx is sampled high, the SCLx pin will be deasserted (brought high). When SCLx is sampled high, the Baud Rate Generator is reloaded with the contents of SSPxADD<6:0> and begins counting. SDAx and SCLx must be sampled high for one TBRG. This action is then followed by assertion of the SDAx pin (SDAx = 0) for one TBRG while SCLx is high. Following this, the RSEN bit (SSPxCON2<1>) will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDAx pin held low. As soon as a Start condition is detected on the SDAx and SCLx pins, the S bit (SSPxSTAT<3>) will be set. The SSPxIF bit will not be set until the Baud Rate Generator has timed out.

- **Note 1:** If RSEN is programmed while any other event is in progress, it will not take effect.
  - **2:** A bus collision during the Repeated Start condition occurs if:
    - SDAx is sampled low when SCLx goes from low-to-high.
    - SCLx goes low before SDAx is asserted low. This may indicate that another master is attempting to transmit a data '1'.

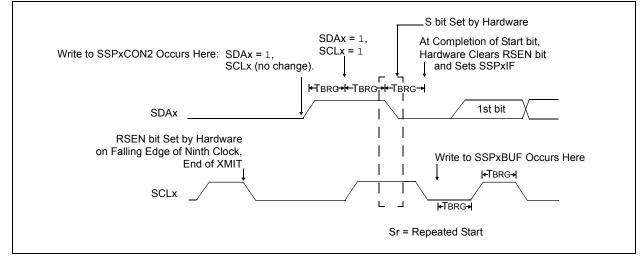
Immediately following the SSPxIF bit getting set, the user may write the SSPxBUF with the 7-bit address in 7-bit mode or the default first address in 10-bit mode. After the first eight bits are transmitted and an ACK is received, the user may then transmit an additional eight bits of address (10-bit mode) or eight bits of data (7-bit mode).

### 21.4.9.1 WCOL Status Flag

If the user writes the SSPxBUF when a Repeated Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing of the lower 5 bits of SSPxCON2 is disabled until the Repeated Start condition is complete.

### FIGURE 21-22: REPEATED START CONDITION WAVEFORM



### 21.4.10 I²C[™] MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address, is accomplished by simply writing a value to the SSPxBUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDAx pin after the falling edge of SCLx is asserted (see data hold time specification Parameter 106). SCLx is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCLx is released high (see data setup time specification Parameter 107). When the SCLx pin is released high, it is held that way for TBRG. The data on the SDAx pin must remain stable for that duration and some hold time after the next falling edge of SCLx. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDAx. This allows the slave device being addressed to respond with an ACK bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKDT bit on the falling edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared; if not, the bit is set. After the ninth clock, the SSPxIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPxBUF, leaving SCLx low and SDAx unchanged (Figure 21-23).

After the write to the SSPxBUF, each bit of the address will be shifted out on the falling edge of SCLx until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will deassert the SDAx pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDAx pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT status bit (SSPxCON2<6>). Following the falling edge of the ninth clock transmission of the address, the SSPxIF flag is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSPxBUF takes place, holding SCLx low and allowing SDAx to float.

### 21.4.10.1 BF Status Flag

In Transmit mode, the BF bit (SSPxSTAT<0>) is set when the CPU writes to SSPxBUF and is cleared when all 8 bits are shifted out.

### 21.4.10.2 WCOL Status Flag

If the user writes the SSPxBUF when a transmit is already in progress (i.e., SSPxSR is still shifting out a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur) after 2 TCY after the SSPxBUF write. If SSPxBUF is rewritten within 2 TCY, the WCOL bit is set and SSPxBUF is updated. This may result in a corrupted transfer.

The user should verify that the WCOL bit is clear after each write to SSPxBUF to ensure the transfer is correct. In all cases, WCOL must be cleared in software.

### 21.4.10.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit (SSPxCON2<6>) is cleared when the slave has sent an Acknowledge  $(\overline{ACK} = 0)$  and is set when the slave does not Acknowledge  $(\overline{ACK} = 1)$ . A slave sends an Acknowledge when it has recognized its address (including a general call), or when the slave has properly received its data.

### 21.4.11 I²C[™] MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN (SSPxCON2<3>).

Note: The MSSP module must be in an inactive state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting, and on each rollover, the state of the SCLx pin changes (high-to-low/low-to-high) and data is shifted into the SSPxSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPxSR are loaded into the SSPxBUF, the BF flag bit is set, the SSPxIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCLx low. The MSSP is now in Idle state awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable bit, ACKEN (SSPxCON2<4>).

### 21.4.11.1 BF Status Flag

In receive operation, the BF bit is set when an address or data byte is loaded into SSPxBUF from SSPxSR. It is cleared when the SSPxBUF register is read.

### 21.4.11.2 SSPOV Status Flag

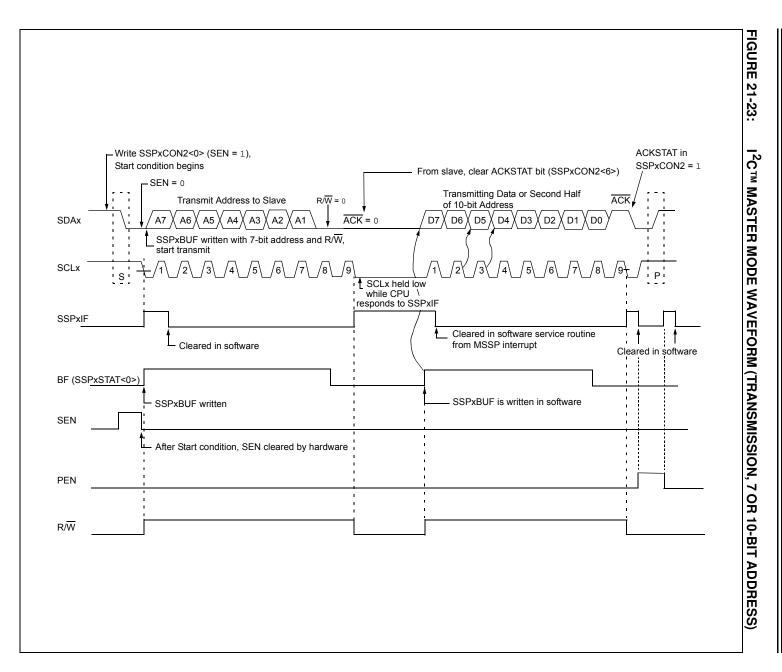
In receive operation, the SSPOV bit is set when 8 bits are received into the SSPxSR and the BF flag bit is already set from a previous reception.

### 21.4.11.3 WCOL Status Flag

If the user writes the SSPxBUF when a receive is already in progress (i.e., SSPxSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

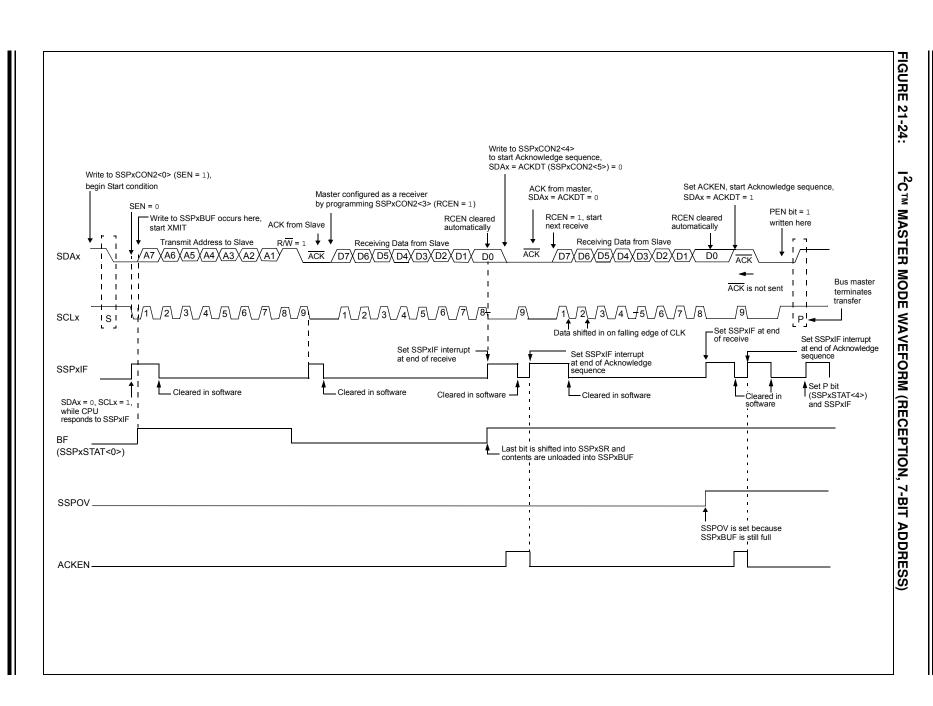
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### 21.4.12 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit. ACKEN (SSPxCON2<4>). When this bit is set, the SCLx pin is pulled low and the contents of the Acknowledge data bit are presented on the SDAx pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCLx pin is deasserted (pulled high). When the SCLx pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG; the SCLx pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the Baud Rate Generator is turned off and the MSSP module then goes into an inactive state (Figure 21-25).

### 21.4.12.1 WCOL Status Flag

If the user writes the SSPxBUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

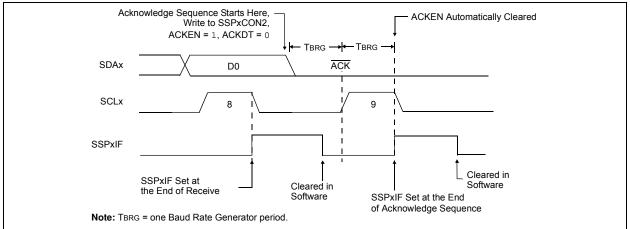
### 21.4.13 STOP CONDITION TIMING

A Stop bit is asserted on the SDAx pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN (SSPxCON2<2>). At the end of a receive/transmit, the SCLx line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDAx line low. When the SDAx line is sampled low, the Baud Rate Generator is reloaded and counts down to 0. When the Baud Rate Generator times out, the SCLx pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDAx pin will be deasserted. When the SDAx pin is sampled high while SCLx is high, the P bit (SSPxSTAT<4>) is set. A TBRG later, the PEN bit is cleared and the SSPxIF bit is set (see Figure 21-26).

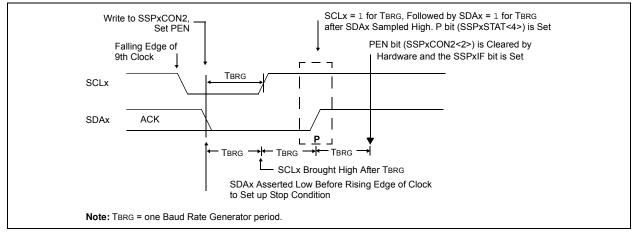
### 21.4.13.1 WCOL Status Flag

If the user writes the SSPxBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

### FIGURE 21-25: ACKNOWLEDGE SEQUENCE WAVEFORM



### FIGURE 21-26: STOP CONDITION RECEIVE OR TRANSMIT MODE



### 21.4.14 SLEEP OPERATION

While in Sleep mode, the  $I^2C$  module can receive addresses or data, and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

### 21.4.15 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

### 21.4.16 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I²C bus may be taken when the P bit (SSPxSTAT<4>) is set, or the bus is Idle, with both the S and P bits clear. When the bus is busy, enabling the MSSP interrupt will generate the interrupt when the Stop condition occurs.

In multi-master operation, the SDAx line must be monitored for arbitration to see if the signal level is the expected output level. This check is performed in hardware with the result placed in the BCLxIF bit.

The states where arbitration can be lost are:

- · Address Transfer
- Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition

### 21.4.17 MULTI -MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDAx pin, arbitration takes place when the master outputs a '1' on SDAx, by letting SDAx float high, and another master asserts a '0'. When the SCLx pin floats high, data should be stable. If the expected data on SDAx is a '1' and the data sampled on the SDAx pin = 0, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLxIF, and reset the  $l^2C$  port to its Idle state (Figure 21-27).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDAx and SCLx lines are deasserted and the SSPxBUF can be written to. When the user services the bus collision Interrupt Service Routine and if the  $I^2C$  bus is free, the user can resume communication by asserting a Start condition.

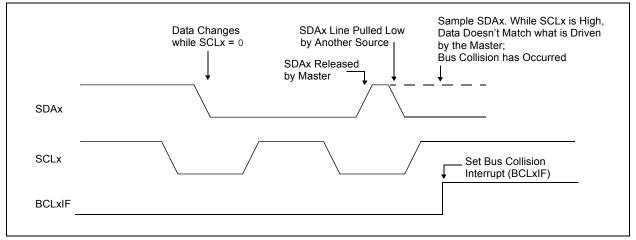
If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDAx and SCLx lines are deasserted and the respective control bits in the SSPxCON2 register are cleared. When the user services the bus collision Interrupt Service Routine (ISR), and if the  $I^2C$  bus is free, the user can resume communication by asserting a Start condition.

The master will continue to monitor the SDAx and SCLx pins. If a Stop condition occurs, the SSPxIF bit will be set.

A write to the SSPxBUF will start the transmission of data at the first data bit regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the I²C bus can be taken when the P bit is set in the SSPxSTAT register, or the bus is Idle and the S and P bits are cleared.

### FIGURE 21-27: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE



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### 21.4.17.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDAx or SCLx is sampled low at the beginning of the Start condition (Figure 21-28).
- b) SCLx is sampled low before SDAx is asserted low (Figure 21-29).

During a Start condition, both the SDAx and the SCLx pins are monitored.

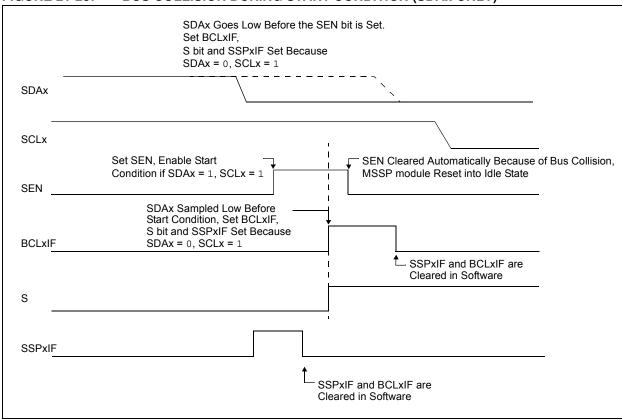
If the SDAx pin is already low, or the SCLx pin is already low, then all of the following occur:

- · The Start condition is aborted
- · The BCLxIF flag is set
- The MSSP module is reset to its inactive state (Figure 21-28)

The Start condition begins with the SDAx and SCLx pins deasserted. When the SDAx pin is sampled high, the Baud Rate Generator is loaded from SSPxADD<6:0> and counts down to 0. If the SCLx pin is sampled low while SDAx is high, a bus collision occurs because it is assumed that another master is attempting to drive a data '1' during the Start condition.

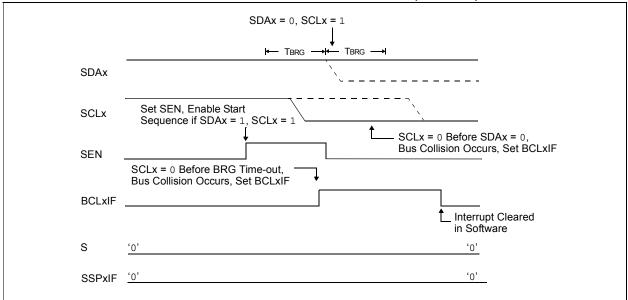
If the SDAx pin is sampled low during this count, the BRG is reset and the SDAx line is asserted early (Figure 21-30). If, however, a '1' is sampled on the SDAx pin, the SDAx pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to 0. If the SCLx pin is sampled as '0' during this time, a bus collision does not occur. At the end of the BRG count, the SCLx pin is asserted low.

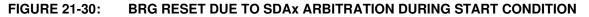
Note: The reason that a bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDAx before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

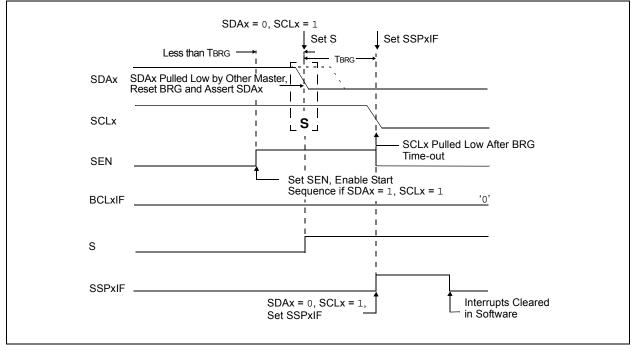


### FIGURE 21-28: BUS COLLISION DURING START CONDITION (SDAx ONLY)









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## 21.4.17.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

- a) A low level is sampled on SDAx when SCLx goes from a low level to a high level.
- SCLx goes low before SDAx is asserted low, indicating that another master is attempting to transmit a data '1'.

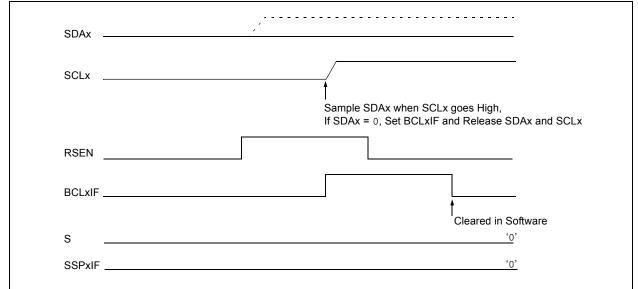
When the user deasserts SDAx and the pin is allowed to float high, the BRG is loaded with SSPxADD<6:0> and counts down to 0. The SCLx pin is then deasserted and when sampled high, the SDAx pin is sampled.

If SDAx is low, a bus collision has occurred (i.e., another master is attempting to transmit a data '0', Figure 21-31). If SDAx is sampled high, the BRG is reloaded and begins counting. If SDAx goes from high-to-low before the BRG times out, no bus collision occurs because no two masters can assert SDAx at exactly the same time.

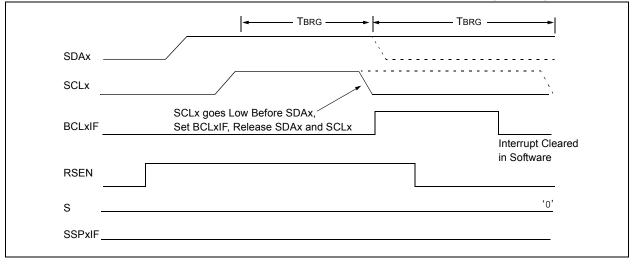
If SCLx goes from high-to-low before the BRG times out and SDAx has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data '1' during the Repeated Start condition (see Figure 21-32).

If, at the end of the BRG time-out, both SCLx and SDAx are still high, the SDAx pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCLx pin, the SCLx pin is driven low and the Repeated Start condition is complete.

### FIGURE 21-31: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)



### FIGURE 21-32: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)



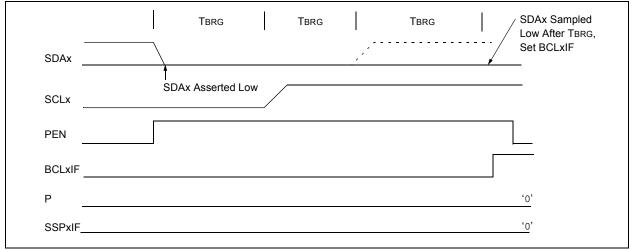
#### 21.4.17.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

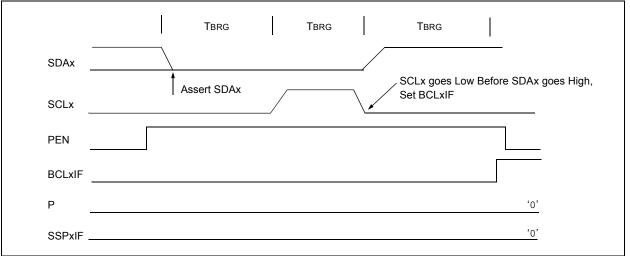
- a) After the SDAx pin has been deasserted and allowed to float high, SDAx is sampled low after the BRG has timed out.
- b) After the SCLx pin is deasserted, SCLx is sampled low before SDAx goes high.

The Stop condition begins with SDAx asserted low. When SDAx is sampled low, the SCLx pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSPxADD<6:0> and counts down to 0. After the BRG times out, SDAx is sampled. If SDAx is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0' (Figure 21-33). If the SCLx pin is sampled low before SDAx is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 21-34).

#### FIGURE 21-33: BUS COLLISION DURING A STOP CONDITION (CASE 1)



#### FIGURE 21-34: BUS COLLISION DURING A STOP CONDITION (CASE 2)



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<b>IABLE 21-4:</b>	REGISTERS ASSOCIATED WITH I'C M OPERATION														
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0							
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF							
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF							
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE							
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP							
PIR2	OSCFIF		SSP2IF	BLC2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF							
PIE2	OSCFIE		SSP2IE	BLC2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE							
IPR2	OSCFIP		SSP2IP	BLC2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP							
PIR3	TMR5GIF		RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF							
PIE3	TMR5GIE	—	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE							
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP							
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0							
TRISD	TRISD7	TRISD7 TRISD6 TRISD5 TRISD4 TRISD3 TRISD2 TRISD1 TRISD0													
SSP1BUF	MSSP1 Receive Buffer/Transmit Register														
SSP1ADD			r (I ² C Slave r ad Register (I		ode)										
SSP1MSK ⁽¹⁾	MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0							
SSP1CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0							
SSP1CON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN							
	GCEN	ACKSTAT	ADMSK5 ⁽²⁾	ADMSK4 ⁽²⁾	ADMSK3 ⁽²⁾	ADMSK2 ⁽²⁾	ADMSK1 ⁽²⁾	SEN							
SSP1STAT	SMP	CKE	D/A	Р	S	R/W	UA	BF							
SSP2BUF			ransmit Regi												
SSP2ADD	MSSP2 Add MSSP2 Bau	lress Registe Id Rate Reloa	r (I ² C Slave r ad Register (I	node), ² C Master mo	ode)										
SSP2MSK ⁽¹⁾	MSK7	MSK6	MSK5	MSK4	MSK3	MSK2	MSK1	MSK0							
SSP2CON1	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0							
SSP2CON2	GCEN	ACKSTAT	STAT ACKDT ACKEN		RCEN	PEN	RSEN	SEN							
	GCEN	ACKSTAT	ADMSK5 ⁽²⁾	ADMSK5 ⁽²⁾ ADMSK4 ⁽²⁾ ADMSK3 ⁽²⁾ ADMSK2 ⁽²⁾		ADMSK2 ⁽²⁾	ADMSK1 ⁽²⁾	SEN							
SSP2STAT	SMP	CKE	D/A	Р	S	S R/W UA		BF							
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD							

# TABLE 21-4: REGISTERS ASSOCIATED WITH I²C[™] OPERATION

**Legend:** -= unimplemented, read as '0'. Shaded cells are not used by the MSSP module in  $I^2C^{TM}$  mode.

Note 1: SSPxMSK shares the same address in SFR space as SSPxADD, but is only accessible in certain I²C Slave operating modes in 7-Bit Masking mode. See Section 21.4.3.4 "7-Bit Address Masking Mode" for more details.

2: Alternate bit definitions for use in I²C Slave mode operations only.

# 22.0 ENHANCED UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (EUSART)

The Enhanced Universal Synchronous Asynchronous Receiver Transmitter (EUSART) module is one of two serial I/O modules. (Generically, the EUSART is also known as a Serial Communications Interface or SCI.) The EUSART can be configured as a full-duplex, asynchronous system that can communicate with peripheral devices, such as CRT terminals and personal computers. It can also be configured as a half-duplex synchronous system that can communicate with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs, etc.

The Enhanced USART module implements additional features, including automatic baud rate detection and calibration, automatic wake-up on Sync Break reception and 12-bit Break character transmit. These make it ideally suited for use in Local Interconnect Network bus (LIN/J2602 bus) systems.

All members of the PIC18F87K22 family are equipped with two independent EUSART modules, referred to as EUSART1 and EUSART2. They can be configured in the following modes:

- Asynchronous (full duplex) with:
  - Auto-wake-up on character reception
  - Auto-baud calibration
  - 12-bit Break character transmission
- Synchronous Master (half duplex) with selectable clock polarity
- Synchronous Slave (half duplex) with selectable clock polarity

The pins of EUSART1 and EUSART2 are multiplexed with the functions of PORTC (RC6/TX1/CK1 and RC7/RX1/DT1) and PORTG (RG1/TX2/CK2/AN19/C3OUT and RG2/RX2/DT2/AN18/C3INA), respectively. In order to configure these pins as an EUSART:

- For EUSART1:
  - Bit, SPEN (RCSTA1<7>), must be set (= 1)
  - Bit, TRISC<7>, must be set (= 1)
  - Bit, TRISC<6>, must be cleared (= 0) for Asynchronous and Synchronous Master modes
  - Bit, TRISC<6>, must be set (= 1) for Synchronous Slave mode
- For EUSART2:
  - Bit, SPEN (RCSTA2<7>), must be set (= 1)
  - Bit, TRISG<2>, must be set (= 1)
  - Bit TRISG<1> must be cleared (= 0) for Asynchronous and Synchronous Master modes
  - Bit, TRISC<6>, must be set (= 1) for Synchronous Slave mode

Note: The EUSART control will automatically reconfigure the pin from input to output as needed.

The operation of each Enhanced USART module is controlled through three registers:

- Transmit Status and Control (TXSTAx)
- Receive Status and Control (RCSTAx)
- Baud Rate Control (BAUDCONx)

These are detailed on the following pages in Register 22-1, Register 22-2 and Register 22-3, respectively.

Note: Throughout this section, references to register and bit names that may be associated with a specific EUSART module are referred to generically by the use of 'x' in place of the specific module number. Thus, "RCSTAx" might refer to the Receive Status register for either EUSART1 or EUSART2.

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R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-1	R/W-0
CSRC	TX9	TXEN ⁽¹⁾	SYNC	SENDB	BRGH	TRMT	TX9D
bit 7		·				·	bit
Legend:							
R = Reada	ble bit	W = Writable	bit	U = Unimplen	nented bit, rea	d as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7		ck Source Select	bit				
	<u>Asynchrono</u> Don't care.	us mode:					
	<u>Synchronou</u>						
		mode (clock gen Iode (clock from					
bit 6		ransmit Enable		.,			
		9-bit transmissio					
	0 = Selects	8-bit transmissio	n				
bit 5	TXEN: Tran	smit Enable bit ⁽¹	)				
		it is enabled					
		it is disabled					
bit 4		SART Mode Sele	ct bit				
	1 = Synchro 0 = Asynchr	ronous mode					
bit 3	-	end Break Chara	cter bit				
	Asynchrono						
		ync Break on ne			rdware upon c	ompletion)	
	-	eak transmissio	n has complete	d			
	<u>Synchronou</u> Don't care.	<u>is mode:</u>					
bit 2		h Baud Rate Sel	ect bit				
	Asynchrono						
	1 = High sp	eed					
	0 = Low spe						
	<u>Synchronou</u> Unused in th						
bit 1		smit Shift Regist	er Status bit				
	1 = TSR is 6	-					
	0 = TSR is f	full					
L:1 0	TX9D· 9th h						
bit 0		oit of Transmit Da ress/data bit or a					

### REGISTER 22-1: TXSTAX: TRANSMIT STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
oit 7							bit
_egend:							
R = Readab	le bit	W = Writable b	bit	U = Unimplem	ented bit, read	as '0'	
n = Value a	t POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	iown
oit 7	SPEN: Seria	al Port Enable bit					
		ort is enabled					
	•	ort is disabled (h	eld in Reset)				
oit 6	<b>RX9:</b> 9-Bit R	Receive Enable bi	it				
		9-bit reception					
sit E		8-bit reception	a hit				
bit 5	Asynchrono	le Receive Enabl	ebit				
	Don't care.	<u>uo mouo</u> .					
		<u>s mode – Master</u>	<u>.</u>				
		s single receive					
		s single receive eared after recep	tion is comple	ete.			
	<u>Synchronou</u>	<u>s mode – Slave:</u>	·				
	Don't care.						
oit 4		tinuous Receive I	Enable bit				
	Asynchronou 1 = Enables						
	0 = Disables						
	Synchronou	<u>s mode:</u>					
		s continuous rece		le bit, CREN, is	cleared (CREN	N overrides SR	EN)
		s continuous rece					
bit 3		dress Detect Ena					
		us mode 9-Bit (R address detection		nterrupt and load	ls the receive h	ouffer when RS	R<8> is set
		s address detecti					
		<u>us mode 8-Bit (R</u>	X9 = 0):				
	Don't care.						
oit 2	FERR: Fram	-					
	1 = Framing 0 = No fram	g error (can be clo ning error	eared by read	ling the RCREG	x register and i	receiving the n	ext valid byte
oit 1	OERR: Ove	rrun Error bit					
	1 = Overrur 0 = No over	n error (can be cle rrun error	eared by clea	ring bit, CREN)			
oit 0		it of Received Da	ata				

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R/W-0	R-1	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0
ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN
bit 7			•				bit
Legend:							
R = Readabl	e bit	W = Writable	bit	U = Unimplem	ented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	red	x = Bit is unki	nown
bit 7	1 = A BRG r	ito-Baud Acqui bllover has occ rollover has oc	urred during A	Status bit uto-Baud Rate D	Detect mode (I	must be cleare	d in software)
bit 6	RCIDL: Rece	ive Operation I	dle Status bit				
		operation is Idle					
bit 5	RXDTP: Data	/Receive Polar	ity Select bit				
		<u>s mode:</u> lata (RXx) is in lata (RXx) is no					
	· ·	<u>mode:</u> x) is inverted (a x) is not inverte	,	)			
bit 4	TXCKP: Syno	chronous Clock	Polarity Selec	t bit			
		<u>s mode:</u> for transmit (T) for transmit (T)					
		<u>mode:</u> for clock (CKx) for clock (CKx)		I			
bit 3	BRG16: 16-B	it Baud Rate R	egister Enable	e bit			
				Hx and SPBRGx only (Compatible		RGHx value ig	nored
bit 2	Unimplemen	ted: Read as '	0'				
bit 1	WUE: Wake-	up Enable bit					
	hardware		sing edge	RXx pin – interru detected	ipt generated	on falling edge	e; bit cleared i
	Synchronous Unused in thi						
bit 0	ABDEN: Auto	Baud Detect	Enable bit				
	cleared in 0 = Baud rate	aud rate meas n hardware upo e measuremen	on completion.	e next character has completed	r. Requires re	ception of a Sy	ync field (55h
	Synchronous Unused in thi						

# 22.1 Baud Rate Generator (BRG)

The BRG is a dedicated, 8-bit or 16-bit generator that supports both the Asynchronous and Synchronous modes of the EUSART. By default, the BRG operates in 8-bit mode; setting the BRG16 bit (BAUDCONx<3>) selects 16-bit mode.

The SPBRGHx:SPBRGx register pair controls the period of a free-running timer. In Asynchronous mode, bits, BRGH (TXSTAx<2>) and BRG16 (BAUDCONx<3>), also control the baud rate. In Synchronous mode, BRGH is ignored. Table 22-1 shows the formula for computation of the baud rate for different EUSART modes which only apply in Master mode (internally generated clock).

Given the desired baud rate and Fosc, the nearest integer value for the SPBRGHx:SPBRGx registers can be calculated using the formulas in Table 22-1. From this, the error in baud rate can be determined. An example calculation is shown in Example 22-1. Typical baud rates and error values for the various Asynchronous modes are shown in Table 22-2. It may be advantageous to use the high baud rate (BRGH = 1) or the 16-bit BRG to reduce the baud rate error, or achieve a slow baud rate for a fast oscillator frequency.

Writing a new value to the SPBRGHx:SPBRGx registers causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate. When operated in the Synchronous mode, SPBRGH:SPBRG values of 0000h and 0001h are not supported. In the Asynchronous mode, all BRG values may be used.

# 22.1.1 OPERATION IN POWER-MANAGED MODES

The device clock is used to generate the desired baud rate. When one of the power-managed modes is entered, the new clock source may be operating at a different frequency. This may require an adjustment to the value in the SPBRGx register pair.

### 22.1.2 SAMPLING

The data on the RXx pin (either RC7/RX1/DT1 or RG2/RX2/DT2/AN18/C3INA) is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RXx pin.

C	onfiguration B	its	BRG/EUSART Mode	Baud Rate Formula
SYNC	BRG16	BRGH	BRG/EUSART Mode	Baud Rate Formula
0	0	0	8-bit/Asynchronous	Fosc/[64 (n + 1)]
0	0	1	8-bit/Asynchronous	$E_{0000}/[16 (n + 1)]$
0	1	0	16-bit/Asynchronous	Fosc/[16 (n + 1)]
0	1	1	16-bit/Asynchronous	
1	0	x	8-bit/Synchronous	Fosc/[4 (n + 1)]
1	1	x	16-bit/Synchronous	

#### TABLE 22-1: BAUD RATE FORMULAS

**Legend:** x = Don't care, n = value of SPBRGHx:SPBRGx register pair

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## EXAMPLE 22-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 10	6 MH	z, desired baud rate of 9600, Asynchronous mode, and 8-bit BRG:
Desired Baud Rate	=	Fosc/(64 ([SPBRGHx:SPBRGx] + 1))
Solving for SPBRGHx:	SPBI	RGx:
Х	=	((FOSC/Desired Baud Rate)/64) – 1
	=	((16000000/9600)/64) - 1
	=	[25.042] = 25
Calculated Baud Rate	=	16000000/(64 (25 + 1))
	=	9615
Error	=	(Calculated Baud Rate - Desired Baud Rate)/Desired Baud Rate
	=	(9615 - 9600)/9600 = 0.16%

### TABLE 22-2: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

	TX9 RX9	TXEN	SYNC	SENDB	BRGH					
	RX9			OLINDD	вкоп	TRMT	TX9D			
		SREN	CREN	ADDEN	FERR	OERR	RX9D			
OVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN			
EUSART1 Baud Rate Generator Register High Byte										
EUSART1 Baud Rate Generator Register										
RC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D			
EN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D			
OVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN			
RT2 Bauc	d Rate Ger	nerator Regis	ter High Byte	;						
EUSART2 Baud Rate Generator Register										
3MD C	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD			
	RT1 Bau RT1 Bau RC EN OVF RT2 Bau RT2 Bau 3MD (	RT1 Baud Rate GerRT1 Baud Rate GerRCTX9ENRX9OVFRCIDLRT2 Baud Rate GerRT2 Baud Rate Ger3MDCCP2MD	RT1 Baud Rate Generator RegisRT1 Baud Rate Generator RegisRCTX9TXENENRX9SRENOVFRCIDLRT2 Baud Rate Generator RegisRT2 Baud Rate Generator RegisSMDCCP2MDCCP1MD	RT1 Baud Rate Generator Register High ByteRT1 Baud Rate Generator RegisterRCTX9TXENSYNCENRX9SRENCRENOVFRCIDLRXDTPTXCKPRT2 Baud Rate Generator RegisterHigh ByteRT2 Baud Rate Generator RegisterSMDCCP2MDCCP1MDUART2MD	RT1 Baud Rate Generator Register High Byte         RT1 Baud Rate Generator Register         RC       TX9         TX9       TXEN         SYNC       SENDB         EN       RX9       SREN         OVF       RCIDL       RXDTP         RT2 Baud Rate Generator Register       High Byte         RT2 Baud Rate Generator Register       High Byte	RT1 Baud Rate Generator Register High ByteRT1 Baud Rate Generator RegisterRCTX9TXENSYNCSENDBBRGHENRX9SRENCRENADDENFERROVFRCIDLRXDTPTXCKPBRG16—RT2 Baud Rate Generator RegisterHigh ByteRT2 Baud Rate Generator RegisterSRENSREN3MDCCP2MDCCP1MDUART2MDUART1MDSSP2MD	RT1 Baud Rate Generator Register High Byte         RT1 Baud Rate Generator Register         RC       TX9       TXEN       SYNC       SENDB       BRGH       TRMT         EN       RX9       SREN       CREN       ADDEN       FERR       OERR         OVF       RCIDL       RXDTP       TXCKP       BRG16       —       WUE         RT2 Baud Rate Generator Register High Byte       RT2 Baud Rate Generator Register       SSP1MD       SSP1MD         3MD       CCP2MD       CCP1MD       UART2MD       UART1MD       SSP2MD       SSP1MD			

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the BRG.

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					SYNC	= 0, BRGH	l = 0, BRC	<b>G16 =</b> 0					
BAUD	Fosc	= 40.000	) MHz	Fosc	Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	_						_		_	_	_		
1.2	—	—	—	1.221	1.73	255	1.202	0.16	129	1.201	-0.16	103	
2.4	2.441	1.73	255	2.404	0.16	129	2.404	0.16	64	2.403	-0.16	51	
9.6	9.615	0.16	64	9.766	1.73	31	9.766	1.73	15	9.615	-0.16	12	
19.2	19.531	1.73	31	19.531	1.73	15	19.531	1.73	7	_	_	_	
57.6	56.818	-1.36	10	62.500	8.51	4	52.083	-9.58	2	_	_	_	
115.2	125.000	8.51	4	104.167	-9.58	2	78.125	-32.18	1	_	_	_	

#### TABLE 22-3: BAUD RATES FOR ASYNCHRONOUS MODES

			S	YNC = 0, E	BRGH = 0	, BRG16 =	0			
BAUD	Fos	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz			
RATE (K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.16	207	0.300	-0.16	103	0.300	-0.16	51	
1.2	1.202	0.16	51	1.201	-0.16	25	1.201	-0.16	12	
2.4	2.404	0.16	25	2.403	-0.16	12	_	_	_	
9.6	8.929	-6.99	6	—	_	_	_	_	_	
19.2	20.833	8.51	2	—	_	_	_	_	_	
57.6	62.500	8.51	0	—	_	_	—	_	_	
115.2	62.500	-45.75	0	_	—	_	_	—	_	

					SYNC	= 0, BRGH	l = 1, BRG	i <b>16 =</b> 0				
BAUD RATE	Fosc	= 40.000	) MHz	Fosc	= 20.000	) MHz	Fosc = 10.000 MHz			Fosc = 8.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	—	_	_	_	_	_	_	_	_	_	_	_
1.2	—	—	—	—	—	—	—	—	—	—	—	—
2.4	—	_	_	—	_	_	2.441	1.73	255	2.403	-0.16	207
9.6	9.766	1.73	255	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	—	_	_

			S	YNC = 0, E	SYNC = 0, BRGH = 1, BRG16 = 0												
BAUD RATE	Foso	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz										
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)								
0.3	_	_	_		_	_	0.300	-0.16	207								
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51								
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25								
9.6	9.615	0.16	25	9.615	-0.16	12	_	_	_								
19.2	19.231	0.16	12	_	_	_	_	_	_								
57.6	62.500	8.51	3	—	_	_	—	_	_								
115.2	125.000	8.51	1	_	—	—	_	_	—								

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					SYNC	= 0, BRGH	<b>1</b> = 0, BRG	16 = 1				
BAUD	Fosc	= 40.000	) MHz	Fosc	= 20.000	) MHz	Fosc = 10.000 MHz Fosc = 8.0			c = 8.000	8.000 MHz	
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)
0.3	0.300	0.00	8332	0.300	0.02	4165	0.300	0.02	2082	0.300	-0.04	1665
1.2	1.200	0.02	2082	1.200	-0.03	1041	1.200	-0.03	520	1.201	-0.16	415
2.4	2.402	0.06	1040	2.399	-0.03	520	2.404	0.16	259	2.403	-0.16	207
9.6	9.615	0.16	259	9.615	0.16	129	9.615	0.16	64	9.615	-0.16	51
19.2	19.231	0.16	129	19.231	0.16	64	19.531	1.73	31	19.230	-0.16	25
57.6	58.140	0.94	42	56.818	-1.36	21	56.818	-1.36	10	55.555	3.55	8
115.2	113.636	-1.36	21	113.636	-1.36	10	125.000	8.51	4	_	_	—

#### TABLE 22-3: BAUD RATES FOR ASYNCHRONOUS MODES (CONTINUED)

			S	YNC = 0, E	BRGH = 0	, BRG16 =	1			
BAUD	Foso	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz			
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.04	832	0.300	-0.16	415	0.300	-0.16	207	
1.2	1.202	0.16	207	1.201	-0.16	103	1.201	-0.16	51	
2.4	2.404	0.16	103	2.403	-0.16	51	2.403	-0.16	25	
9.6	9.615	0.16	25	9.615	-0.16	12	—	_	_	
19.2	19.231	0.16	12	_	_	_	_	_	_	
57.6	62.500	8.51	3	—	_	_	—	_	_	
115.2	125.000	8.51	1	_	_	—	_	_	_	

				SYNC = 0,	BRGH :	= 1, BRG16	= 1 or SY	NC = 1,	BRG16 = 1			
BAUD RATE	Fosc	= 40.000	) MHz	Fosc = 20.000 MHz			Fosc = 10.000 MHz			Fosc = 8.000 MHz		
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)									
0.3	0.300	0.00	33332	0.300	0.00	16665	0.300	0.00	8332	0.300	-0.01	6665
1.2	1.200	0.00	8332	1.200	0.02	4165	1.200	0.02	2082	1.200	-0.04	1665
2.4	2.400	0.02	4165	2.400	0.02	2082	2.402	0.06	1040	2.400	-0.04	832
9.6	9.606	0.06	1040	9.596	-0.03	520	9.615	0.16	259	9.615	-0.16	207
19.2	19.193	-0.03	520	19.231	0.16	259	19.231	0.16	129	19.230	-0.16	103
57.6	57.803	0.35	172	57.471	-0.22	86	58.140	0.94	42	57.142	0.79	34
115.2	114.943	-0.22	86	116.279	0.94	42	113.636	-1.36	21	117.647	-2.12	16

		SYN	IC = 0, BR(	GH = 1, BI	<b>RG16 =</b> 1	or SYNC =	: 1, BRG1	6 = 1		
BAUD RATE	Fost	c = 4.000	MHz	Fos	c = 2.000	MHz	Fosc = 1.000 MHz			
(K)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	Actual Rate (K)	% Error	SPBRG value (decimal)	
0.3	0.300	0.01	3332	0.300	-0.04	1665	0.300	-0.04	832	
1.2	1.200	0.04	832	1.201	-0.16	415	1.201	-0.16	207	
2.4	2.404	0.16	415	2.403	-0.16	207	2.403	-0.16	103	
9.6	9.615	0.16	103	9.615	-0.16	51	9.615	-0.16	25	
19.2	19.231	0.16	51	19.230	-0.16	25	19.230	-0.16	12	
57.6	58.824	2.12	16	55.555	3.55	8	—	—	—	
115.2	111.111	-3.55	8	—	_	—	—	_	—	

### 22.1.3 AUTO-BAUD RATE DETECT

The Enhanced USART module supports the automatic detection and calibration of baud rate. This feature is active only in Asynchronous mode and while the WUE bit is clear.

The automatic baud rate measurement sequence (Figure 22-1) begins whenever a Start bit is received and the ABDEN bit is set. The calculation is self-averaging.

In the Auto-Baud Rate Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RXx signal, the RXx signal is timing the BRG. In ABD mode, the internal Baud Rate Generator is used as a counter to time the bit period of the incoming serial byte stream.

Once the ABDEN bit is set, the state machine will clear the BRG and look for a Start bit. The Auto-Baud Rate Detect must receive a byte with the value, 55h (ASCII "U", which is also the LIN/J2602 bus Sync character), in order to calculate the proper bit rate. The measurement is taken over both a low and a high bit time in order to minimize any effects caused by asymmetry of the incoming signal. After a Start bit, the SPBRGx begins counting up, using the preselected clock source on the first rising edge of RXx. After eight bits on the RXx pin or the fifth rising edge, an accumulated value totalling the proper BRG period is left in the SPBRGHx:SPBRGx register pair. Once the 5th edge is seen (this should correspond to the Stop bit), the ABDEN bit is automatically cleared.

If a rollover of the BRG occurs (an overflow from FFFFh to 0000h), the event is trapped by the ABDOVF status bit (BAUDCONx<7>). It is set in hardware by BRG rollovers and can be set or cleared by the user in software. ABD mode remains active after rollover events and the ABDEN bit remains set (Figure 22-2).

While calibrating the baud rate period, the BRG registers are clocked at 1/8th the preconfigured clock rate. The BRG clock will be configured by the BRG16 and BRGH bits. The BRG16 bit must be set to use both SPBRG1 and SPBRGH1 as a 16-bit counter. This allows the user to verify that no carry occurred for 8-bit modes by checking for 00h in the SPBRGHx register. Refer to Table 22-4 for counter clock rates to the BRG.

While the ABD sequence takes place, the EUSART state machine is held in Idle. The RCxIF interrupt is set once the fifth rising edge on RXx is detected. The value in the RCREGx needs to be read to clear the RCxIF interrupt. The contents of RCREGx should be discarded.

- Note 1: If the WUE bit is set with the ABDEN bit, Auto-Baud Rate Detection will occur on the byte *following* the Break character.
  - 2: It is up to the user to determine that the incoming character baud rate is within the range of the selected BRG clock source. Some combinations of oscillator frequency and EUSART baud rates are not possible due to bit error rates. Overall system timing and communication baud rates must be taken into consideration when using the Auto-Baud Rate Detection feature.
  - **3:** To maximize baud rate range, if that feature is used, it is recommended that the BRG16 bit (BAUDCONx<3>) be set.

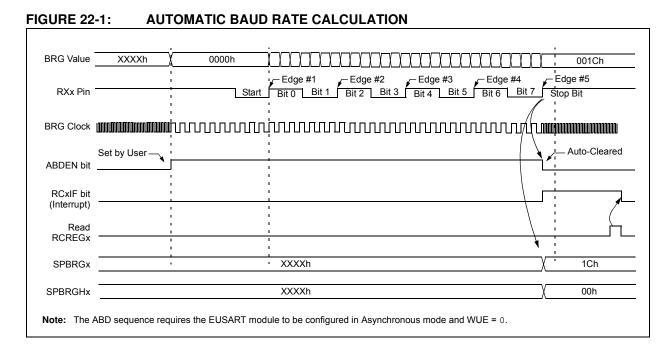
# TABLE 22-4:BRG COUNTER<br/>CLOCK RATES

BRG16	BRGH	BRG Counter Clock
0	0	Fosc/512
0	1	Fosc/128
1	0	Fosc/128
1	1	Fosc/32

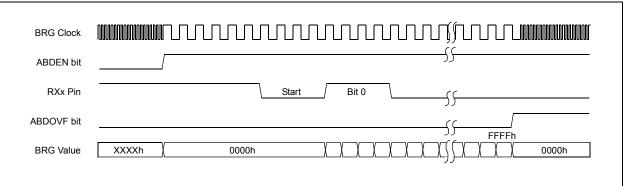
### 22.1.3.1 ABD and EUSART Transmission

Since the BRG clock is reversed during ABD acquisition, the EUSART transmitter cannot be used during ABD. This means that whenever the ABDEN bit is set, TXREGx cannot be written to. Users should also ensure that ABDEN does not become set during a transmit sequence. Failing to do this may result in unpredictable EUSART operation.

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#### FIGURE 22-2: BRG OVERFLOW SEQUENCE



# 22.2 EUSART Asynchronous Mode

The Asynchronous mode of operation is selected by clearing the SYNC bit (TXSTAx<4>). In this mode, the EUSART uses standard Non-Return-to-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip, dedicated 8-bit/16-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator.

The EUSART transmits and receives the LSb first. The EUSART's transmitter and receiver are functionally independent but use the same data format and baud rate. The Baud Rate Generator produces a clock, either x16 or x64 of the bit shift rate, depending on the BRGH and BRG16 bits (TXSTAx<2> and BAUDCONx<3>). Parity is not supported by the hardware but can be implemented in software and stored as the 9th data bit.

When operating in Asynchronous mode, the EUSART module consists of the following important elements:

- · Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver
- Auto-Wake-up on Sync Break Character
- 12-Bit Break Character Transmit
- Auto-Baud Rate Detection

#### 22.2.1 EUSART ASYNCHRONOUS TRANSMITTER

The EUSART transmitter block diagram is shown in Figure 22-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREGx. The TXREGx register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREGx register (if available).

Once the TXREGx register transfers the data to the TSR register (occurs in one TcY), the TXREGx register is empty and the TXxIF flag bit is set. This interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXxIE. TXxIF will be set regardless of the state of TXxIE; it cannot be cleared in software. TXxIF is also not cleared immediately upon loading TXREGx, but becomes valid in the second instruction cycle following the load instruction. Polling TXxIF immediately following a load of TXREGx will return invalid results.

While TXxIF indicates the status of the TXREGx register; another bit, TRMT (TXSTAx<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

Note 1: The TSR register is not mapped in data memory, so it is not available to the user.

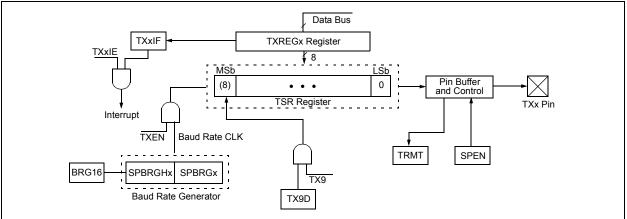
2: Flag bit, TXxIF, is set when enable bit, TXEN, is set.

To set up an Asynchronous Transmission:

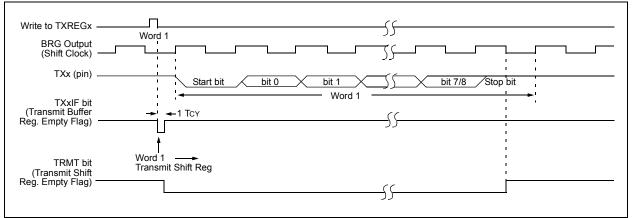
- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, TXxIE.
- 4. If 9-bit transmission is desired, set transmit bit, TX9; can be used as an address/data bit.
- 5. Enable the transmission by setting bit, TXEN, which will also set bit, TXxIF.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Load data to the TXREGx register (starts transmission).
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

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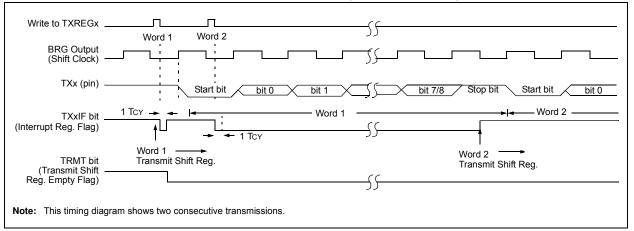
### FIGURE 22-3: EUSART TRANSMIT BLOCK DIAGRAM







# FIGURE 22-5: ASYNCHRONOUS TRANSMISSION (BACK-TO-BACK)



TADLE 22-J.											
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF			
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF			
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE			
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP			
PIR3	TMR5GIF	—	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF			
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE			
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP			
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D			
TXREG1	EUSART1 T	ransmit Regi	ster								
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D			
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN			
SPBRGH1	EUSART1 E	Baud Rate Ge	enerator Regi	ister High By	te						
SPBRG1	EUSART1 E	Baud Rate Ge	enerator Regi	ister							
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D			
TXREG2	EUSART2 T	ransmit Regi	ster								
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D			
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN			
SPBRGH2	EUSART2 E	aud Rate Ge	enerator Regi	ister High By	te						
SPBRG2	EUSART2 E	Baud Rate Ge	enerator Regi	ister							
ODCON3	U2OD	U10D	—	_	_	_	—	CTMUDS			
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD			
Legend: _ =	unimplement	ed locations	read as '0' S	shaded cells a	are not used	for asynchro	nous transm	ission			

# TABLE 22-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Legend: - = unimplemented locations read as '0'. Shaded cells are not used for asynchronous transmission.

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#### 22.2.2 EUSART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 22-6. The data is received on the RXx pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at x16 times the baud rate, whereas the main receive serial shifter operates at the bit rate or at Fosc. This mode would typically be used in RS-232 systems.

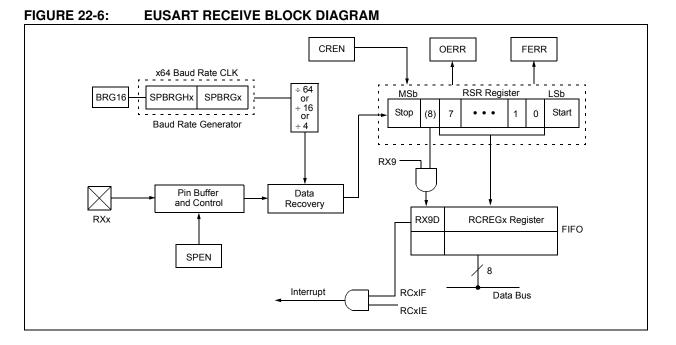
To set up an Asynchronous Reception:

- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
- 3. If interrupts are desired, set enable bit, RCxIE.
- 4. If 9-bit reception is desired, set bit, RX9.
- 5. Enable the reception by setting bit, CREN.
- Flag bit, RCxIF, will be set when reception is complete and an interrupt will be generated if enable bit, RCxIE, was set.
- 7. Read the RCSTAx register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 8. Read the 8-bit received data by reading the RCREGx register.
- 9. If any error occurred, clear the error by clearing enable bit, CREN.
- 10. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

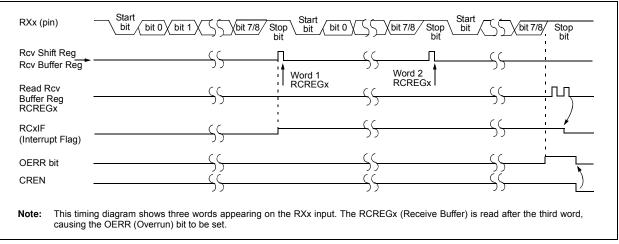
# 22.2.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

This mode would typically be used in RS-485 systems. To set up an Asynchronous Reception with Address Detect Enable:

- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
- 2. Enable the asynchronous serial port by clearing the SYNC bit and setting the SPEN bit.
- 3. If interrupts are required, set the RCEN bit and select the desired priority level with the RCxIP bit.
- 4. Set the RX9 bit to enable 9-bit reception.
- 5. Set the ADDEN bit to enable address detect.
- 6. Enable reception by setting the CREN bit.
- The RCxIF bit will be set when reception is complete. The interrupt will be Acknowledged if the RCxIE and GIE bits are set.
- 8. Read the RCSTAx register to determine if any error occurred during reception, as well as read bit 9 of data (if applicable).
- 9. Read RCREGx to determine if the device is being addressed.
- 10. If any error occurred, clear the CREN bit.
- 11. If the device has been addressed, clear the ADDEN bit to allow all received data into the receive buffer and interrupt the CPU.



### FIGURE 22-7: ASYNCHRONOUS RECEPTION



#### TABLE 22-6: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF	
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE	
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP	
PIR3	TMR5GIF	—	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	
PIE3	TMR5GIE		RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	
IPR3	TMR5GIP	—	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	
RCREG1	EUSART1 F	Receive Regist	er						
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	—	WUE	ABDEN	
SPBRGH1	EUSART1 E	aud Rate Ger	nerator Regis	ster High Byte	9				
SPBRG1	EUSART1 E	aud Rate Ger	nerator Regis	ster					
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	
RCREG2	EUSART2 F	Receive Regist	er						
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	
SPBRGH2	EUSART2 E	aud Rate Ger	nerator Regis	ster High Byte	9				
SPBRG2	EUSART2 Baud Rate Generator Register								
ODCON3	U2OD	U10D	_	—	_	_	—	CTMUDS	
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD	

**Legend:** — = unimplemented locations read as '0'. Shaded cells are not used for asynchronous reception.

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#### 22.2.4 AUTO-WAKE-UP ON SYNC BREAK CHARACTER

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper byte reception cannot be performed. The auto-wake-up feature allows the controller to wake up due to activity on the RXx/DTx line while the EUSART is operating in Asynchronous mode.

The auto-wake-up feature is enabled by setting the WUE bit (BAUDCONx<1>). Once set, the typical receive sequence on RXx/DTx is disabled and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RXx/DTx line. (This coincides with the start of a Sync Break or a Wake-up Signal character for the LIN/J2602 protocol.)

Following a wake-up event, the module generates an RCxIF interrupt. The interrupt is generated synchronously to the Q clocks in normal operating modes (Figure 22-8) and asynchronously if the device is in Sleep mode (Figure 22-9). The interrupt condition is cleared by reading the RCREGx register.

The WUE bit is automatically cleared once a low-to-high transition is observed on the RXx line following the wake-up event. At this point, the EUSART module is in Idle mode and returns to normal operation. This signals to the user that the Sync Break event is over.

### 22.2.4.1 Special Considerations Using Auto-Wake-up

Since auto-wake-up functions by sensing rising edge transitions on RXx/DTx, information with any state changes before the Stop bit may signal a false End-of-Character (EOC) and cause data or framing errors. To work properly, therefore, the initial character in the transmission must be all '0's. This can be 00h (8 bits) for standard RS-232 devices or 000h (12 bits) for the LIN/J2602 bus.

Oscillator start-up time must also be considered, especially in applications using oscillators with longer start-up intervals (i.e., HS or HSPLL mode). The Sync Break (or Wake-up Signal) character must be of sufficient length and be followed by a sufficient interval to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

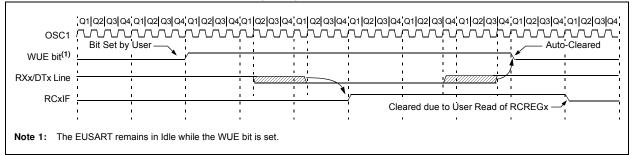
# 22.2.4.2 Special Considerations Using the WUE Bit

The timing of WUE and RCxIF events may cause some confusion when it comes to determining the validity of received data. As noted, setting the WUE bit places the EUSART in an Idle mode. The wake-up event causes a receive interrupt by setting the RCxIF bit. The WUE bit is cleared after this when a rising edge is seen on RXx/DTx. The interrupt condition is then cleared by reading the RCREGx register. Ordinarily, the data in RCREGx will be dummy data and should be discarded.

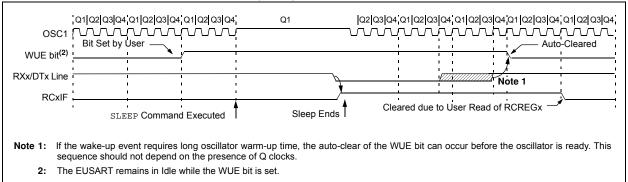
The fact that the WUE bit has been cleared (or is still set), and the RCxIF flag is set, should not be used as an indicator of the integrity of the data in RCREGx. Users should consider implementing a parallel method in firmware to verify received data integrity.

To assure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process. If a receive operation is not occurring, the WUE bit may then be set just prior to entering Sleep mode.

### FIGURE 22-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING NORMAL OPERATION



#### FIGURE 22-9: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP



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# 22.2.5 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN/J2602 bus standard. The Break character transmit consists of a Start bit, followed by twelve '0' bits and a Stop bit. The Frame Break character is sent whenever the SENDB and TXEN bits (TXSTAx<3> and TXSTAx<5>, respectively) are set while the Transmit Shift Register is loaded with data. Note that the value of data written to TXREGx will be ignored and all '0's will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN/J2602 specification).

Note that the data value written to the TXREGx for the Break character is ignored. The write simply serves the purpose of initiating the proper sequence.

The TRMT bit indicates when the transmit operation is active or Idle, just as it does during normal transmission. See Figure 22-10 for the timing of the Break character sequence.

#### 22.2.5.1 Break and Sync Transmit Sequence

The following sequence will send a message frame header made up of a Break, followed by an Auto-Baud Sync byte. This sequence is typical of a LIN/J2602 bus master.

- 1. Configure the EUSART for the desired mode.
- 2. Set the TXEN and SENDB bits to set up the Break character.
- 3. Load the TXREGx with a dummy character to initiate transmission (the value is ignored).
- 4. Write '55h' to TXREGx to load the Sync character into the transmit FIFO buffer.
- 5. After the Break has been sent, the SENDB bit is reset by hardware. The Sync character now transmits in the preconfigured mode.

When the TXREGx becomes empty, as indicated by the TXxIF, the next data byte can be written to TXREGx.

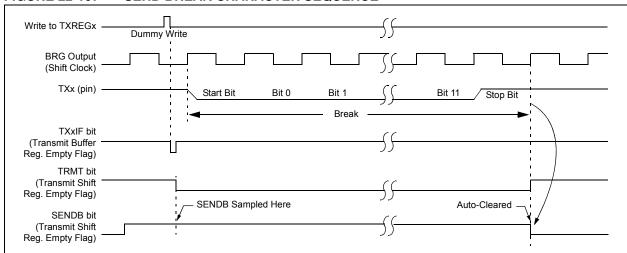
#### 22.2.6 RECEIVING A BREAK CHARACTER

The Enhanced USART module can receive a Break character in two ways.

The first method forces configuration of the baud rate at a frequency of 9/13 the typical speed. This allows for the Stop bit transition to be at the correct sampling location (13 bits for Break versus Start bit and 8 data bits for typical data).

The second method uses the auto-wake-up feature described in Section 22.2.4 "Auto-Wake-up on Sync Break Character". By enabling this feature, the EUSART will sample the next two transitions on RXx/DTx, cause an RCxIF interrupt and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Rate Detect feature. For both methods, the user can set the ABDEN bit once the TXxIF interrupt is observed.



# FIGURE 22-10: SEND BREAK CHARACTER SEQUENCE

# 22.3 EUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTAx<7>). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit, SYNC (TXSTAx<4>). In addition, enable bit, SPEN (RCSTAx<7>), is set in order to configure the TXx and RXx pins to CKx (clock) and DTx (data) lines, respectively.

The Master mode indicates that the processor transmits the master clock on the CKx line. Clock polarity is selected with the TXCKP bit (BAUDCONx<4>). Setting TXCKP sets the Idle state on CKx as high, while clearing the bit sets the Idle state as low. This option is provided to support Microwire devices with this module.

#### 22.3.1 EUSART SYNCHRONOUS MASTER TRANSMISSION

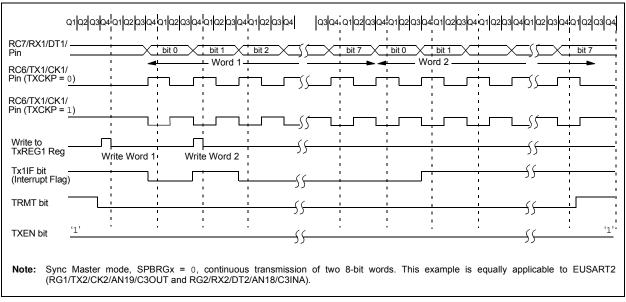
The EUSART transmitter block diagram is shown in Figure 22-3. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREGx. The TXREGx register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREGx (if available).

Once the TXREGx register transfers the data to the TSR register (occurs in one TcY), the TXREGx is empty and the TXxIF flag bit is set. The interrupt can be enabled or disabled by setting or clearing the interrupt enable bit, TXxIE. TXxIF is set regardless of the state of enable bit, TXxIE; it cannot be cleared in software. It will reset only when new data is loaded into the TXREGx register.

While flag bit, TXxIF, indicates the status of the TXREGx register, another bit, TRMT (TXSTAx<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user must poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

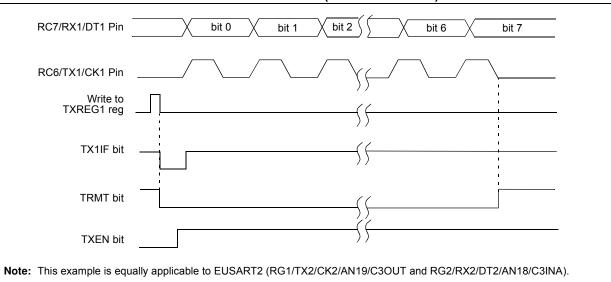
To set up a Synchronous Master Transmission:

- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
- 3. If interrupts are desired, set enable bit, TXxIE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Start transmission by loading data to the TXREGx register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.



# FIGURE 22-11: SYNCHRONOUS TRANSMISSION

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#### FIGURE 22-12: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP		RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
TXREG1	EUSART1 T	ransmit Regis	ster					
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN
SPBRGH1	EUSART1 E	aud Rate Ge	nerator Regis	ster High Byt	е			
SPBRG1	EUSART1 E	aud Rate Ge	nerator Regis	ster				
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
TXREG2	EUSART2 T	ransmit Regis	ster					
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN
SPBRGH2	EUSART2 E	aud Rate Ge	nerator Regis	ster High Byt	e			
SPBRG2	EUSART2 E	aud Rate Ge	nerator Regis	ster				
ODCON3	U2OD	U10D						CTMUDS
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD

TABLE 22-7:	REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER TRANSMISSION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

#### 22.3.2 EUSART SYNCHRONOUS MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either the Single Receive Enable bit, SREN (RCSTAx<5>), or the Continuous Receive Enable bit, CREN (RCSTAx<4>). Data is sampled on the RXx pin on the falling edge of the clock.

If enable bit, SREN, is set, only a single word is received. If enable bit, CREN, is set, the reception is continuous until CREN is cleared. If both bits are set, then CREN takes precedence.

To set up a Synchronous Master Reception:

- 1. Initialize the SPBRGHx:SPBRGx registers for the appropriate baud rate. Set or clear the BRG16 bit, as required, to achieve the desired baud rate.
- 2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.

- 3. Ensure bits, CREN and SREN, are clear.
- 4. If interrupts are desired, set enable bit, RCxIE.
- 5. If 9-bit reception is desired, set bit, RX9.
- 6. If a single reception is required, set bit, SREN. For continuous reception, set bit, CREN.
- 7. Interrupt flag bit, RCxIF, will be set when reception is complete and an interrupt will be generated if the enable bit, RCxIE, was set.
- 8. Read the RCSTAx register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 9. Read the 8-bit received data by reading the RCREGx register.
- 10. If any error occurred, clear the error by clearing bit CREN.
- 11. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

# FIGURE 22-13: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)

RC7/RX1/D	Pin	bit	0 bi	t 1 bit 2	2 bit 3	bit 4	bit 5	bit 6	bit 7	
RC6/TX1/C n (TXCKP =		'						r <u>÷</u> rr	<u>.</u>	1 1 1
RC6/TX1/C n (TXCKP =							·:		; 	, , ,
Write bit, SRI			1 1 1 1		1 1 1	1 1 1		1 1 1	1 1 1	1 1 1
SREN	bit					1 1 1		1 1 1	<u>.</u>	1 1 1
CREN	bót	1	1	1	1	1	1	1	1 1	"(
RC1IF (Interru		1 1 1	1 1 1	1 1 1	1 1 1	1	1 1 1	, , ,		<u> </u>
Re RCRE		;		<u> </u>		1 1 1				<u> </u> ́

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IADLE 22-0:	REGISTE	REGISTERS ASSOCIATED WITH STNCHRONOUS MASTER RECEPTION										
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0				
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF				
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF				
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE				
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP				
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF				
PIE3	TMR5GIE	—	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE				
IPR3	TMR5GIP	—	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP				
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D				
RCREG1	EUSART1 R	eceive Regist	er									
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D				
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN				
SPBRGH1	EUSART1 B	aud Rate Ger	erator Regi	ster High Byt	e							
SPBRG1	EUSART1 Ba	aud Rate Ger	erator Regi	ster								
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D				
RCREG2	EUSART2 R	eceive Regist	er									
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D				
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN				
SPBRGH2	EUSART2 Ba	aud Rate Ger	erator Regi	ster High Byt	e							
SPBRG2	EUSART2 B	aud Rate Ger	erator Regi	ster								
ODCON3	U2OD	U10D	_					CTMUDS				
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD				
Legend: -=	unimplement	ed, read as '0	. Shaded c	ells are not u	sed for sync	hronous mas	ter reception					

## TABLE 22-8: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used for synchronous master reception.

### 22.4 EUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit, CSRC (TXSTAx<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the CKx pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any low-power mode.

#### 22.4.1 EUSART SYNCHRONOUS SLAVE TRANSMISSION

The operation of the Synchronous Master and Slave modes is identical, except in the case of Sleep mode.

If two words are written to the TXREGx and then the SLEEP instruction is executed, the following will occur:

- a) The first word will immediately transfer to the TSR register and transmit.
- b) The second word will remain in the TXREGx register.
- c) Flag bit, TXxIF, will not be set.
- d) When the first word has been shifted out of TSR, the TXREGx register will transfer the second word to the TSR and flag bit, TXxIF, will now be set.

e) If enable bit, TXxIE, is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

- 1. Enable the synchronous slave serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. Clear bits, CREN and SREN.
- 3. If interrupts are desired, set enable bit, TXxIE.
- 4. If 9-bit transmission is desired, set bit, TX9.
- 5. Enable the transmission by setting enable bit, TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
- 7. Start transmission by loading data to the TXREGx register.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF	
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE	
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP	
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF	
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE	
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP	
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	
TXREG1	EUSART1 T	ransmit Regis	ster						
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	
SPBRGH1	EUSART1 B	aud Rate Gei	nerator Regi	ster High Byt	e				
SPBRG1	EUSART1 B	aud Rate Ger	nerator Regi	ster					
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	
TXREG2	EUSART2 T	ransmit Regis	ster						
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN	
SPBRGH2	EUSART2 B	aud Rate Gei	nerator Regi	ster High Byt	e				
SPBRG2	EUSART2 Baud Rate Generator Register								
ODCON3	U2OD	U10D	_	_			_	CTMUDS	
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD	

# TABLE 22-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

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#### 22.4.2 EUSART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical except in the case of Sleep, or any Idle mode, and bit, SREN, which is a "don't care" in Slave mode.

If receive is enabled by setting the CREN bit prior to entering Sleep or any Idle mode, then a word may be received while in this low-power mode. Once the word is received, the RSR register will transfer the data to the RCREGx register. If the RCxIE enable bit is set, the interrupt generated will wake the chip from the low-power mode. If the global interrupt is enabled, the program will branch to the interrupt vector. To set up a Synchronous Slave Reception:

- 1. Enable the synchronous master serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
- 2. If interrupts are desired, set enable bit, RCxIE.
- 3. If 9-bit reception is desired, set bit, RX9.
- 4. To enable reception, set enable bit, CREN.
- 5. Flag bit, RCxIF, will be set when reception is complete. An interrupt will be generated if enable bit, RCxIE, was set.
- Read the RCSTAx register to get the 9th bit (if enabled) and determine if any error occurred during reception.
- 7. Read the 8-bit received data by reading the RCREGx register.
- 8. If any error occurred, clear the error by clearing bit, CREN.
- If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
RCSTA1	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
RCREG1	EUSART1 F	Receive Regis	ster					
TXSTA1	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D
BAUDCON1	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16	_	WUE	ABDEN
SPBRGH1	EUSART1 E	aud Rate Ge	nerator Regi	ster High Byt	e			
SPBRG1	EUSART1 E	Baud Rate Ge	nerator Regi	ster				
RCSTA2	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
RCREG2	EUSART2 F	Receive Regis	ster					
TXSTA2	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D
BAUDCON2	ABDOVF	RCIDL	RXDTP	TXCKP	BRG16		WUE	ABDEN
SPBRGH2	EUSART2 E	aud Rate Ge	nerator Regi	ster High Byt	e			
SPBRG2	EUSART2 E	aud Rate Ge	nerator Regi	ster				
ODCON3	U2OD	U10D	—	_	_		_	CTMUDS
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD

TABLE 22-10: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave reception.

# 23.0 12-BIT ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The Analog-to-Digital (A/D) Converter module in the PIC18F87K22 family of devices has 16 inputs for the 64-pin devices and 24 inputs for the 80-pin devices. This module allows conversion of an analog input signal to a corresponding 12-bit digital number.

The module has these registers:

- A/D Control Register 0 (ADCON0)
- A/D Control Register 1 (ADCON1)
- A/D Control Register 2 (ADCON2)
- A/D Port Configuration Register 0 (ANCON0)
- A/D Port Configuration Register 1 (ANCON1)
- A/D Port Configuration Register 2 (ANCON2)
- ADRESH (the upper, A/D Results register)
- · ADRESL (the lower, A/D Results register)

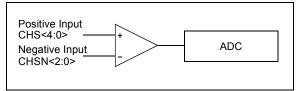
The ADCON0 register, shown in Register 23-1, controls the operation of the A/D module. The ADCON1 register, shown in Register 23-2, configures the voltage reference and special trigger selection. The ADCON2 register, shown in Register 23-3, configures the A/D clock source and programmed acquisition time and justification.

# 23.1 Differential A/D Converter

The converter in PIC18F87K22 family devices is implemented as a differential A/D where the differential voltage between two channels is measured and converted to digital values (see Figure 23-1).

The converter can also be configured to measure a voltage from a single input by clearing the CHSN bits (ADCON1<2:0>). With this configuration, the negative channel input is connected internally to AVss (see Figure 23-2).

# FIGURE 23-1: DIFFERENTIAL CHANNEL MEASUREMENT

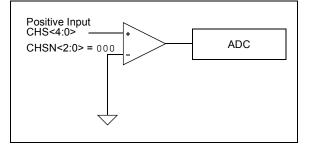


Differential conversion feeds the two input channels to a unity gain differential amplifier. The positive channel input is selected using the CHS bits (ADCON0<6:2>) and the negative channel input is selected using the CHSN bits (ADCON1<2:0>).

The output from the amplifier is fed to the A/D Converter, as shown in Figure 23-1. The 12-bit result is available on the ADRESH and ADRESL registers. An additional bit indicates if the 12-bit result is a positive or negative value.

FIGURE 23-2:

#### SINGLE CHANNEL MEASUREMENT



In the Single Channel Measurement mode, the negative input is connected to Avss by clearing the CHSN bits (ADCON1<2:0>).

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# 23.2 A/D Registers

23.2.1 A/D CONTROL REGISTERS

### REGISTER 23-1: ADCON0: A/D CONTROL REGISTER 0

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
_	CHS4	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON		
bit 7 bit 0									
Legend:									
R = Readable bit W = Writable bit			U = Unimplem	nented bit, read	d as '0'				
-n = Value at I	POR	'1' = Bit is set		'0' = Bit is cleared x = Bit is unknow		own			

#### bit 7 Unimplemented: Read as '0'

DIL 7	Unimplemented: Read as 0
bit 6-2	CHS<4:0>: Analog Channel Select bits
	00000 = Channel 00 (AN0) 10000 = Channel 16 (AN16)
	00001 = Channel 01 (AN1) 10001 = Channel 17 (AN17)
	00010 = Channel 02 (AN2) 10010 = Channel 18 (AN18)
	00011 = Channel 03 (AN3) 10011 = Channel 19 (AN19)
	00100 = Channel 04 (AN4) 10100 = Channel 20 (AN20) ^(1,2)
	00101 = Channel 05 (AN5) 10101 = Channel 21 (AN21) ^(1,2)
	00110 = Channel 06 (AN6) 10110 = Channel 22 (AN22)(1,2)
	00111 = Channel 07 (AN7) 10111 = Channel 23 (AN23) ^(1,2)
	01000 = Channel 08 (AN8) 11000 = (Reserved) ⁽²⁾
	01001 = Channel 09 (AN9) 11001 = (Reserved) ⁽²⁾
	01010 = Channel 10 (AN10 11010 = (Reserved) ⁽²⁾
	01011 = Channel 11 (AN11) 11011 = (Reserved) ⁽²⁾
	01100 = Channel 12 (AN12) ^(1,2) 11100 = Channel 28 (Reserved CTMU)
	01101 = Channel 13 (AN13) ^(1,20) 11101 = Channel 29 (Internal temperature diode)
	01110 = Channel 14 (AN14) ^(1,2) 11110 = Channel 30 (VDDCORE)
	01111 = Channel 15 (AN15) ^(1,2) 11111 = Channel 31 (v1.024V band gap)
bit 1	GO/DONE: A/D Conversion Status bit
	1 = A/D (or calibration) cycle in progress. Setting this bit starts an A/D conversion cycle. The bit is
	cleared automatically by hardware when the A/D conversion is completed.
	0 = A/D conversion has completed or is not in progress
bit 0	ADON: A/D On bit
	1 = A/D Converter is operating
	0 = A/D conversion module is shut off and consuming no operating current
	- · ·
Note 1:	These channels are not implemented on 64-pin devices.
<b>o</b> .	Derforming a conversion on unimplemented channels will return rendem values

**2:** Performing a conversion on unimplemented channels will return random values.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
TRIGSEL1	TRIGSEL0	VCFG1	VCFG0	VNCFG	CHSN2	CHSN1	CHSN0
bit 7							bit 0
Legend:							
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	d as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7-6	11 = Selects t 10 = Selects t 01 = Selects t	D>: Special Trig the special trigg the special trigg the special trigg the special trigg	ger from the R ⁻ ger from the Ti ger from the C ⁻	TCC mer1 TMU			
bit 5-4	11 = Internal	A/D VREF+ Cor VREF+ (4.096V VREF+ (2.048V VREF+	)				
bit 3	<b>VNCFG:</b> A/D 1 = External 0 = AVss	VREF- Configur VREF	ation bit				
bit 2-0	CHSN<2:0>: 111 = Channe 110 = Channe 101 = Channe 011 = Channe 011 = Channe 010 = Channe 001 = Channe 000 = Channe	el 06 (AN5) el 05 (AN4) el 04 (AN3) el 03 (AN2) el 02 (AN1) el 01 (AN0)	ve Channel Se	lect bits			

#### REGISTER 23-2: ADCON1: A/D CONTROL REGISTER 1

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R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0			
ADFM	0-0	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0			
bit 7		AUQIZ	ACQTT	ACQTO	ADCOZ	ADCOT	bit			
Legend:										
R = Reada	ble bit	W = Writable	bit	U = Unimpler	nented bit, read	d as '0'				
-n = Value	at POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown			
bit 7	<b>ADFM:</b> A/D 1 = Right jus 0 = Left justi		select bit							
bit 6	Unimpleme	nted: Read as '	0'							
bit 5-3	ACQT<2:0>: A/D Acquisition Time Select bits									
	111 = 20 TAI 110 = 16 TAI 101 = 12 TAI 100 = 8 TAD 011 = 6 TAD 010 = 4 TAD 001 = 2 TAD 000 = 0 TAD	D								
bit 2-0	111 = FRC (0 110 = Fosc/ 101 = Fosc/ 100 = Fosc/	16 4 clock derived fro 32 8	om A/D RC oso	cillator) ⁽¹⁾						

#### REGISTER 23-3: ADCON2: A/D CONTROL REGISTER 2

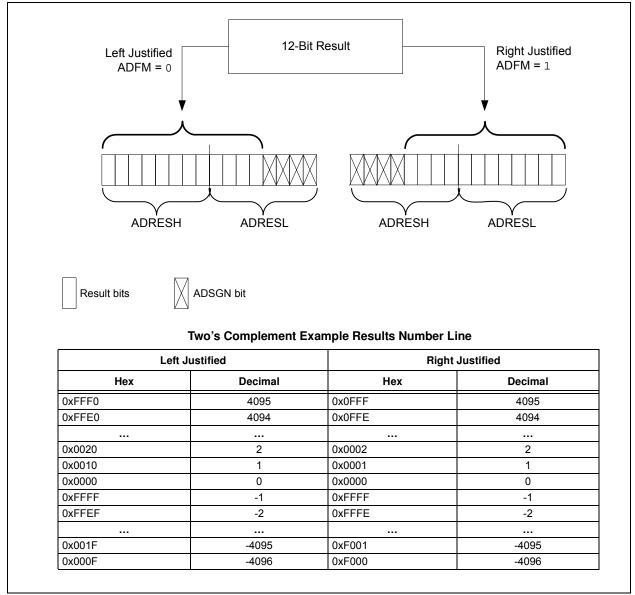
**Note 1:** If the A/D FRC clock source is selected, a delay of one TCY (instruction cycle) is added before the A/D clock starts. This allows the SLEEP instruction to be executed before starting a conversion.

### 23.2.2 A/D RESULT REGISTERS

The ADRESH:ADRESL register pair is where the 12-bit A/D result and extended sign bits (ADSGN) are loaded at the completion of a conversion. This register pair is 16 bits wide. The A/D module gives the flexibility of left or right justifying the 12-bit result in the 16-Bit Result register. The A/D Format Select bit (ADFM) controls this justification. Figure 23-3 shows the operation of the A/D result justification and location of the extended sign bits (ADSGN). The extended sign bits allow for easier 16-bit math to be performed on the result. The results are represented as a two's compliment binary value. This means that when sign bits and magnitude bits are considered together in right justification, the ADRESH and ADRESL registers can be read as a single signed integer value.

When the A/D Converter is disabled, these 8-bit registers can be used as two general purpose registers.

FIGURE 23-3: A/D RESULT JUSTIFICATION



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#### **REGISTER 23-4:** ADRESH: A/D RESULT HIGH BYTE REGISTER, LEFT JUSTIFIED (ADFM = 0)

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADRES11	ADRES10	ADRES9	ADRES8	ADRES7	ADRES6	ADRES5	ADRES4
bit 7							bit 0
Legend:							

Legena.				
R = Readable bit	W = Writable bit U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7-0 ADRES<11:4>: A/D Result High Byte bits

#### REGISTER 23-5: ADRESL: A/D RESULT HIGH BYTE REGISTER, LEFT JUSTIFIED (ADFM = 0)

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADRES3	ADRES2	ADRES1	ADRES0	ADSGN	ADSGN	ADSGN	ADSGN
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-4 ADRES<3:0>: A/D Result Low Byte bits

bit 3-0 ADSGN: A/D Result Sign bit

.

1 = A/D result is negative

0 = A/D result is positive

# REGISTER 23-6: ADRESH: A/D RESULT HIGH BYTE REGISTER, RIGHT JUSTIFIED (ADFM = 1)

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADSGN	ADSGN	ADSGN	ADSGN	ADRES11	ADRES10	ADRES9	ADRES8
bit 7							bit 0

Legend:					
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

bit 7-4	ADSGN: A/D Result Sign bit
	1 = A/D result is negative
	0 = A/D result is positive
bit 3-0	ADRESH<11:8>: A/D Result High Byte bits

# REGISTER 23-7: ADRESL: A/D RESULT LOW BYTE REGISTER, RIGHT JUSTIFIED (ADFM = 1)

R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x	R/W-x
ADRES7	ADRES6	ADRES5	ADRES4	ADRES3	ADRES2	ADRES1	ADRES0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit,	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 ADRES<7:0>: A/D Result Low Byte bits

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The ANCONx registers are used to configure the operation of the I/O pin associated with each analog channel. Clearing an ANSELx bit configures the corresponding pin (ANx) to operate as a digital only I/O. Setting a bit configures the pin to operate as an analog input for either the A/D Converter or the comparator module, with all digital peripherals disabled and digital inputs read as '0'.

As a rule, I/O pins that are multiplexed with analog inputs default to analog operation on any device Reset.

### REGISTER 23-8: ANCON0: A/D PORT CONFIGURATION REGISTER 0

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	1 as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0

#### **ANSEL<7:0>:** Analog Port Configuration bits (AN7 and AN0)

1 = Pin is configured as an analog channel; digital input is disabled and any inputs read as '0'

0 = Pin is configured as a digital port

#### REGISTER 23-9: ANCON1: A/D PORT CONFIGURATION REGISTER 1

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSEL15 ⁽¹⁾	ANSEL14 ⁽¹⁾	ANSEL13 ⁽¹⁾	ANSEL12 ⁽¹⁾	ANSEL11	ANSEL10	ANSEL9	ANSEL8
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7-0 ANSEL<15:8>: Analog Port Configuration bits (AN15 through AN8)⁽¹⁾

1 = Pin is configured as an analog channel; digital input is disabled and any inputs read as '0'
 0 = Pin is configured as a digital port

**Note 1:** AN15 through AN12 and AN23 to AN20 are implemented only on 80-pin devices. For 64-pin devices, the corresponding ANSELx bits are still implemented for these channels, but have no effect.

#### REGISTER 23-10: ANCON2: A/D PORT CONFIGURATION REGISTER 2

R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
ANSEL23 ⁽¹⁾	ANSEL22 ⁽¹⁾	ANSEL21 ⁽¹⁾	ANSEL20 ⁽¹⁾	ANSEL19	ANSEL18	ANSEL17	ANSEL16
bit 7							bit 0
Leaend:							

Legenu.			
R = Readable bit	= Readable bit W = Writable bit U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

#### bit 7-0 ANSEL<23:16>: Analog Port Configuration bits (AN23 through AN16)⁽¹⁾

- 1 = Pin is configured as an analog channel; digital input is disabled and any inputs read as '0'
   0 = Pin is configured as a digital port
- **Note 1:** AN15 through AN12 and AN23 through AN20 are implemented only on 80-pin devices. For 64-pin devices, the corresponding ANSELx bits are still implemented for these channels, but have no effect.

The analog reference voltage is software-selectable to either the device's positive and negative supply voltage (AVDD and AVSS) or the voltage level on the RA3/AN3/VREF+ and RA2/AN2/VREF- pins. VREF+ has two additional Internal Reference Voltage selections: 2.048V and 4.096V.

The A/D Converter can uniquely operate while the device is in Sleep mode. To operate in Sleep, the A/D conversion clock must be derived from the A/D Converter's internal RC oscillator.

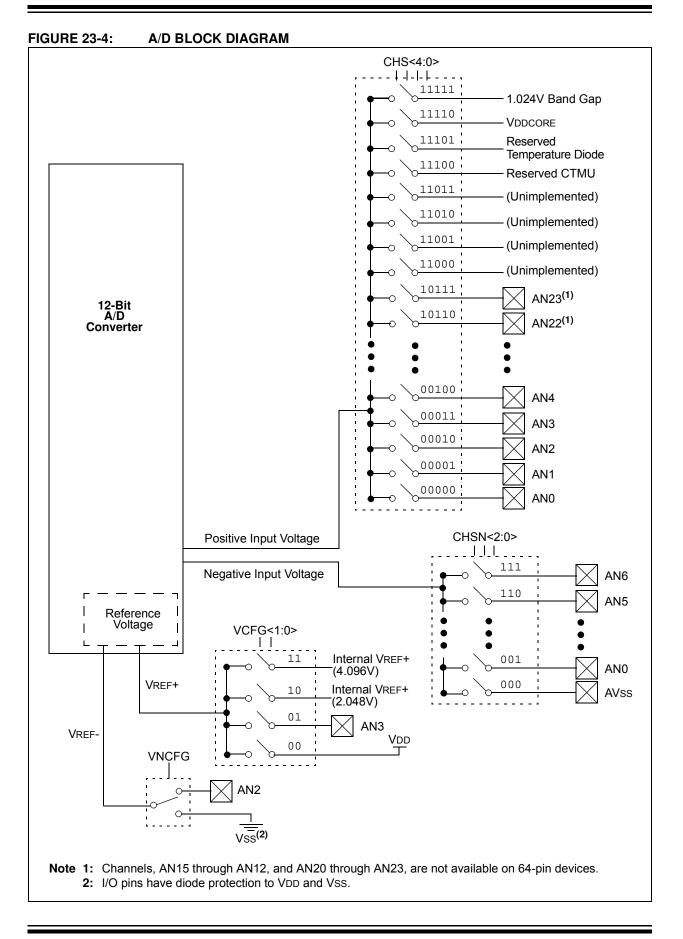
The output of the Sample-and-Hold (S/H) is the input into the converter, which generates the result via successive approximation.

Each port pin associated with the A/D Converter can be configured as an analog input or a digital I/O. The ADRESH and ADRESL registers contain the result of the A/D conversion. When the A/D conversion is complete, the result is loaded into the ADRESH:ADRESL register pair, the GO/DONE bit (ADCON0<1>) is cleared and the A/D Interrupt Flag bit, ADIF (PIR1<6>), is set.

A device Reset forces all registers to their Reset state. This forces the A/D module to be turned off and any conversion in progress is aborted. The value in the ADRESH:ADRESL register pair is not modified for a Power-on Reset. These registers will contain unknown data after a Power-on Reset.

The block diagram of the A/D module is shown in Figure 23-4.

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After the A/D module has been configured as desired, the selected channel must be acquired before the conversion can start. The analog input channels must have their corresponding TRIS bits selected as inputs. To determine acquisition time, see Section 23.3 "A/D Acquisition Requirements". After this acquisition time has elapsed, the A/D conversion can be started. An acquisition time can be programmed to occur between setting the GO/DONE bit and the actual start of the conversion.

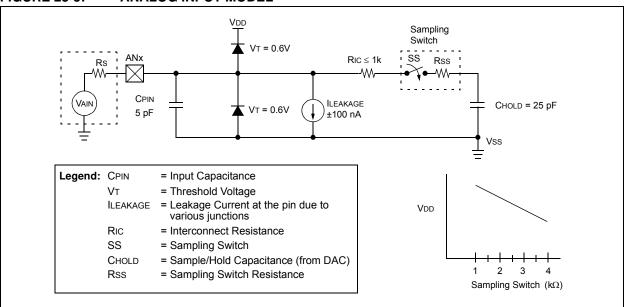
To do an A/D conversion, follow these steps:

- 1. Configure the A/D module:
  - Configure the required ADC pins as analog pins (ANCON0, ANCON1 and ANCON2)
  - Set the voltage reference (ADCON1)
  - Select the A/D positive and negative input channels (ADCON0 and ADCON1)
  - Select the A/D acquisition time (ADCON2)
  - Select the A/D conversion clock (ADCON2)
  - Turn on the A/D module (ADCON0)



- 2. Configure the A/D interrupt (if desired):
  - Clear the ADIF bit (PIR1<6>)
  - Set the ADIE bit (PIE1<6>)
  - Set the GIE bit (INTCON<7>)
- 3. Wait the required acquisition time (if required).
- 4. Start the conversion:
  - Set the GO/DONE bit (ADCON0<1>)
- 5. Wait for A/D conversion to complete, by either:
  Polling for the GO/DONE bit to be cleared OR
  - Waiting for the A/D interrupt
- 6. Read A/D Result registers (ADRESH:ADRESL) and, if required, clear bit, ADIF.
- 7. For the next conversion, begin with Step 1 or 2, as required.

The A/D conversion time per bit is defined as TAD. Before the next acquisition starts, a minimum Wait of 2 TAD is required.



## 23.3 A/D Acquisition Requirements

For the A/D Converter to meet its specified accuracy, the Charge Holding (CHOLD) capacitor must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 23-5. The source impedance (Rs) and the internal sampling switch (Rss) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD).

The source impedance affects the offset voltage at the analog input (due to pin leakage current). The maximum recommended impedance for analog sources is 2.5 k $\Omega$ . After the analog input channel is selected or changed, the channel must be sampled for at least the minimum acquisition time before starting a conversion.

Note:	When	the	conversion	is	started,	the
	holding	, capa	acitor is disco	nne	ected from	the
	input p	in.				

To calculate the minimum acquisition time, Equation 23-1 can be used. This equation assumes that 1/2 LSb error is used (1,024 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified resolution.

Equation 23-3 shows the calculation of the minimum required acquisition time, TACQ. This calculation is based on the following application system assumptions:

• CHOLD =	25 pF
• Rs =	2.5 kΩ
• Conversion Error $\leq$	1/2 LSb
• Vdd =	$3V \rightarrow Rss = 2 \ k\Omega$
Temperature =	85°C

#### EQUATION 23-1: ACQUISITION TIME

TACQ	=	Amplifier Settling Time + Holding Capacitor Charging Time + Temperature Coefficient
	=	TAMP + TC + TCOFF

#### EQUATION 23-2: A/D MINIMUM CHARGING TIME

VHOLD	=	$(\text{VREF} - (\text{VREF}/2048)) \bullet (1 - e^{(-\text{TC/CHOLD}(\text{Ric} + \text{Rss} + \text{Rs}))})$
or		
TC	=	-(CHOLD)(RIC + RSS + RS) $\ln(1/2048)$

### EQUATION 23-3: CALCULATING THE MINIMUM REQUIRED ACQUISITION TIME

TACQ	=	TAMP + TC + TCOFF
TAMP	=	0.2 μs
TCOFF	=	(Temp – 25°C)(0.02 μs/°C) (85°C – 25°C)(0.02 μs/°C) 1.2 μs
Tempera	ature c	oefficient is only required for temperatures $> 25^{\circ}$ C. Below 25°C, TCOFF = 0 ms.
ТС	=	-(ChOLD)(RIC + RSS + RS) $\ln(1/2048) \ \mu s$ -(25 pF) (1 k $\Omega$ + 2 k $\Omega$ + 2.5 k $\Omega$ ) ln(0.0004883) $\mu s$ 1.05 $\mu s$
TACQ	=	0.2 μs + 1.05 μs + 1.2 μs 2.45 μs

## 23.4 Selecting and Configuring Automatic Acquisition Time

The ADCON2 register allows the user to select an acquisition time that occurs each time the  $GO/\overline{DONE}$  bit is set.

When the GO/DONE bit is set, sampling is stopped and a conversion begins. The user is responsible for ensuring the required acquisition time has passed between selecting the desired input channel and setting the GO/DONE bit.

This occurs when the ACQT<2:0> bits (ADCON2<5:3>) remain in their Reset state ('000'), which is compatible with devices that do not offer programmable acquisition times.

If desired, the ACQTx bits can be set to select a programmable acquisition time for the A/D module. When the GO/DONE bit is set, the A/D module continues to sample the input for the selected acquisition time, then automatically begins a conversion. Since the acquisition time is programmed, there may be no need to wait for an acquisition time between selecting a channel and setting the GO/DONE bit.

In either case, when the conversion is completed, the GO/DONE bit is cleared, the ADIF flag is set and the A/D begins sampling the currently selected channel again. If an acquisition time is programmed, there is nothing to indicate if the acquisition time has ended or if the conversion has begun.

## 23.5 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires 14 TAD per 12-bit conversion. The source of the A/D conversion clock is software-selectable.

The possible options for TAD are:

- 2 Tosc
- 4 Tosc
- 8 Tosc
- 16 Tosc
- 32 Tosc
- 64 Tosc
- Using the internal RC Oscillator

For correct A/D conversions, the A/D conversion clock (TAD) must be as short as possible, but greater than the minimum TAD. (For more information, see Parameter 130 in Table 31-28.)

Table 23-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

### TABLE 23-1: TAD vs. DEVICE OPERATING FREQUENCIES

AD Clock S	Maximum	
Operation	ADCS<2:0>	Device Frequency
2 Tosc	000	2.50 MHz
4 Tosc	100	5.00 MHz
8 Tosc	001	10.00 MHz
16 Tosc	101	20.00 MHz
32 Tosc	010	40.00 MHz
64 Tosc	110	64.00 MHz
RC ⁽²⁾	x11	1.00 MHz ⁽¹⁾

Note 1: The RC source has a typical TAD time of  $4 \ \mu$ s.

2: For device frequencies above 1 MHz, the device must be in Sleep mode for the entire conversion or the A/D accuracy may be out of specification.

## 23.6 Configuring Analog Port Pins

The ANCON0, ANCON1, ANCON2, TRISA, TRISF, TRISG and TRISH registers control the operation of the A/D port pins. The port pins needed as analog inputs must have their corresponding TRISx bits set (input). If the TRISx bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS<3:0> bits and the TRISx bits.

- Note 1: When reading the PORT register, all pins configured as analog input channels will read as cleared (a low level). Pins configured as digital inputs will convert an analog input. Analog levels on a digitally configured input will be accurately converted.
  - 2: Analog levels on any pin defined as a digital input may cause the digital input buffer to consume current out of the device's specification limits.

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## 23.7 A/D Conversions

Figure 23-6 shows the operation of the A/D Converter after the GO/DONE bit has been set and the ACQT<2:0> bits are cleared. A conversion is started after the following instruction to allow entry into Sleep mode before the conversion begins.

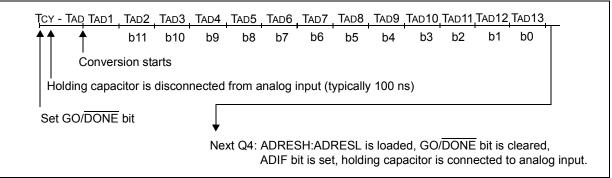
Figure 23-7 shows the operation of the A/D Converter after the GO/DONE bit has been set, the ACQT<2:0> bits set to '010' and a 4 TAD acquisition time selected.

Clearing the GO/DONE bit during a conversion will abort the current conversion. The A/D Result register pair will NOT be updated with the partially completed A/D conversion sample. This means the ADRESH:ADRESL registers will continue to contain the value of the last completed conversion (or the last value written to the ADRESH:ADRESL registers).

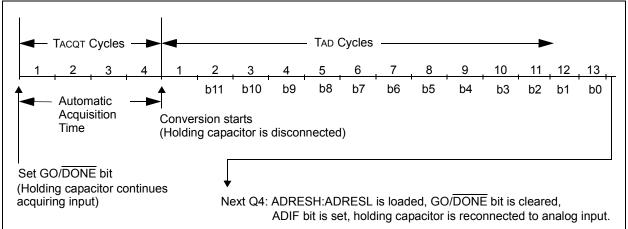
After the A/D conversion is completed or aborted, a 2 TAD Wait is required before the next acquisition can be started. After this Wait, acquisition on the selected channel is automatically started.

Note: The GO/DONE bit should NOT be set in the same instruction that turns on the A/D.

## FIGURE 23-6: A/D CONVERSION TAD CYCLES (ACQT<2:0> = 000, TACQ = 0)



### FIGURE 23-7: A/D CONVERSION TAD CYCLES (ACQT<2:0> = 010, TACQ = 4 TAD)



## 23.8 Use of the Special Event Triggers

A/D conversion can be started by the Special Event Trigger of any of these modules:

- ECCP2 Requires CCP2M<3:0> bits (CCP2CON<3:0>) set at '1011'
- CTMU Requires the setting of the CTTRIG bit (CTMUCONH<0>)
- Timer1
- RTCC

To start an A/D conversion:

- The A/D module must be enabled (ADON = 1)
- The appropriate analog input channel is selected
- The minimum acquisition period is set one of these ways:
- Timing provided by the user
- Selection made of an appropriate TACQ time

With these conditions met, the trigger sets the GO/DONE bit and the A/D acquisition starts.

If the A/D module is not enabled (ADON = 0), the module ignores the Special Event Trigger.

Note: With an ECCP2 trigger, Timer1 or Timer 3 is cleared. The timers reset to automatically repeat the A/D acquisition period with minimal software overhead (moving ADRESH:ADRESL to the desired location). If the A/D module is not enabled, the Special Event Trigger is ignored by the module, but the timer's counter resets.

### 23.9 Operation in Power-Managed Modes

The selection of the automatic acquisition time and A/D conversion clock is determined, in part, by the clock source and frequency while in a power-managed mode.

If the A/D is expected to operate while the device is in a power-managed mode, the ACQT<2:0> and ADCS<2:0> bits in ADCON2 should be updated in accordance with the power-managed mode clock that will be used.

After the power-managed mode is entered (either of the power-managed Run modes), an A/D acquisition or conversion may be started. Once an acquisition or conversion is started, the device should continue to be clocked by the same power-managed mode clock source until the conversion has been completed. If desired, the device may be placed into the corresponding power-managed Idle mode during the conversion.

If the power-managed mode clock frequency is less than 1 MHz, the A/D RC clock source should be selected.

Operation in Sleep mode requires that the A/D RC clock be selected. If bits, ACQT<2:0>, are set to '000' and a conversion is started, the conversion will be delayed one instruction cycle to allow execution of the SLEEP instruction and entry into Sleep mode. The IDLEN and SCS<1:0> bits in the OSCCON register must have already been cleared prior to starting the conversion.

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TADLE 23-2:	REGISTERS ASSOCIATED WITH A/D MODULE							
Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR1	PSPIF	ADIF	RC1IF	TX1IF	SSP1IF	TMR1GIF	TMR2IF	TMR1IF
PIE1	PSPIE	ADIE	RC1IE	TX1IE	SSP1IE	TMR1GIE	TMR2IE	TMR1IE
IPR1	PSPIP	ADIP	RC1IP	TX1IP	SSP1IP	TMR1GIP	TMR2IP	TMR1IP
PIR3	TMR5GIF	_	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE3	TMR5GIE	_	RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP
ADRESH	A/D Result	Register Higl	n Byte					
ADRESL	A/D Result	Register Low	[,] Byte					
ADCON0	_	CHS4	CHS3	CHS2	CHS1	CHS0	GO/DONE	ADON
ADCON1	TRIGSEL1	TRIGSEL0	VCFG1	VCFG0	VNCFG	CHSN2	CHSN1	CHSN0
ADCON2	ADFM	_	ACQT2	ACQT1	ACQT0	ADCS2	ADCS1	ADCS0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8
ANCON2	ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16
CCP2CON	P2M1	P2M0	DC2B1	DC2B0	CCP2M3	CCP2M2	CCP2M1	CCP2M0
PORTA	RA7 ⁽²⁾	RA6 ⁽²⁾	RA5	RA4	RA3	RA2	RA1	RA0
TRISA	TRISA7 ⁽²⁾	TRISA6 ⁽²⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
PORTF	RF7	RF6	RF5	RF4	RF3	RF2	RF1	—
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—
PORTG	_	_	RG5 ⁽³⁾	RG4	RG3	RG2	RG1	RG0
TRISG	_	_	_	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0
PORTH ⁽¹⁾	RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0
TRISH ⁽¹⁾	TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
PMD0	CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD
		whad waad aa				-		

#### TABLE 23-2: REGISTERS ASSOCIATED WITH A/D MODULE

Legend: — = unimplemented, read as '0'. Shaded cells are not used for A/D conversion.

Note 1: This register is not implemented on 64-pin devices.

**2:** These bits are available only in certain oscillator modes, when the FOSC2 Configuration bit = 0. If that Configuration bit is cleared, this signal is not implemented.

**3:** This bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented.

## 24.0 COMPARATOR MODULE

The analog comparator module contains three comparators that can be independently configured in a variety of ways. The inputs can be selected from the analog inputs and two Internal Reference Voltages. The digital outputs are available at the pin level and can also be read through the control register. Multiple output and interrupt event generation are also available. A generic single comparator from the module is shown in Figure 24-1.

Key features of the module includes:

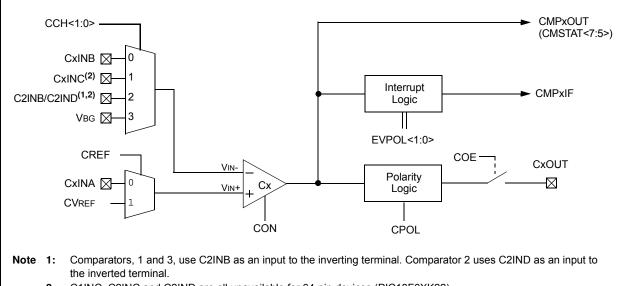
- · Independent comparator control
- · Programmable input configuration
- · Output to both pin and register levels
- · Programmable output polarity
- Independent interrupt generation for each comparator with configurable interrupt-on-change

### 24.1 Registers

The CMxCON registers (CM1CON, CM2CON and CM3CON) select the input and output configuration for each comparator, as well as the settings for interrupt generation (see Register 24-1).

The CMSTAT register (Register 24-2) provides the output results of the comparators. The bits in this register are read-only.





2: C1INC, C2INC and C2IND are all unavailable for 64-pin devices (PIC18F6XK22).

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R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
bit 7							bit 0
Legend:							
R = Reada	ble bit	W = Writable	bit	U = Unimplem	nented bit, rea	d as '0'	
-n = Value	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	nown
bit 7	CON: Compa	arator Enable b	t				
	1 = Compara	tor is enabled					
	0 = Compara	tor is disabled					
bit 6	COE: Compa	rator Output E	nable bit				
		tor output is pre tor output is int		OUT pin			
bit 5		•	-	h:+			
DIL D	•	parator Output I	•	DIL			
		tor output is inv tor output is no					
bit 4-3	EVPOL<1:0>	: Interrupt Pola	rity Select bits				
		t generation on					
				w transition of t			
		t generation on		h transition of t	ine output		
bit 2		•		on-inverting inp	out)		
	•		•	CVREF voltage	,		
	0 = Non-invei	rting input conr	ects to CxINA	pin			
bit 1-0	CCH<1:0>: C	Comparator Cha	annel Select bit	S			
		g input of comp			(0.0)		
				s to C2INB or C	21ND pin ^(2,3)		
		g input of comp g input of comp					
	The CMPxIF bit is	•	set any time th	is mode is seled	cted and must	be cleared by t	he application
	after the initial co	0		o the invertice	torminal Car	parator 2 uses	
2:	Comparators, 1 a	inu 3, use CZIN	id as an input i	o the inverting	terminal. Com	iparator 2 uses (	UZIND.

### REGISTER 24-1: CMxCON: COMPARATOR CONTROL x REGISTER

**3:** C1INC, C2INC and C2IND are all unavailable for 64-pin devices (PIC18F6XK22).

## REGISTER 24-2: CMSTAT: COMPARATOR STATUS REGISTER

R-x	R-x	R-x	U-0	U-0	U-0	U-0	U-0
CMP3OUT	CMP2OUT	CMP1OUT	—	—	—	—	—
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable b	bit	U = Unimplem	nented bit, read	l as '0'	
-n = Value at POR '1' = Bit is set			'0' = Bit is clea	ared	x = Bit is unkr	nown	
bit 7-5	CMP<3:1>OL	JT: Comparator	x Status bits				
<u>If CPOL (CMxCON&lt;5&gt;)= </u> <u>ο</u> (non-inverted polaritγ):							
1 = Comparator x's VIN+ > VIN-							

0 = Comparator x's VIN+ < VIN-

If CPOL = 1 (inverted polarity):

1 = Comparator x's VIN+ < VIN-

0 = Comparator x's VIN+ > VIN-

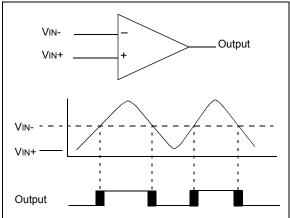
bit 4-0 Unimplemented: Read as '0'

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## 24.2 Comparator Operation

A single comparator is shown in Figure 24-2, along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input, VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input, VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator, in Figure 24-2, represent the uncertainty due to input offsets and response time.

FIGURE 24-2: SINGLE COMPARATOR



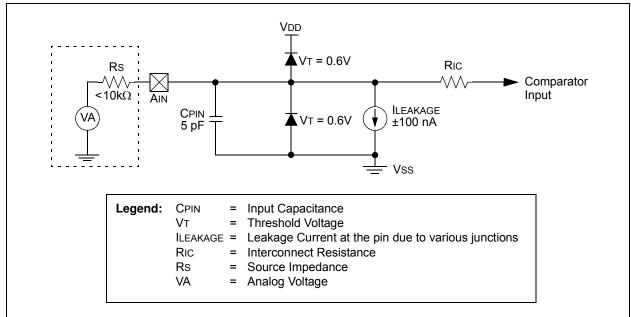
### 24.3 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output has a valid level. The response time of the comparator differs from the settling time of the voltage reference. Therefore, both of these times must be considered when determining the total response to a comparator input change; otherwise, the maximum delay of the comparators should be used (see **Section 31.0 "Electrical Characteristics"**).

## 24.4 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 24-3. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and VSs. The analog input, therefore, must be between VSs and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up condition may occur.

A maximum source impedance of  $10 \text{ k}\Omega$  is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.



### FIGURE 24-3: COMPARATOR ANALOG INPUT MODEL

## 24.5 Comparator Control and Configuration

Each comparator has up to eight possible combinations of inputs: up to four external analog inputs and one of two Internal Reference Voltages.

All of the comparators allow a selection of the signal from pin, CxINA, or the voltage from the comparator reference (CVREF) on the non-inverting channel. This is compared to either CxINB, CxINC, C2INB/C2IND or the microcontroller's fixed Internal Reference Voltage (VBG, 1.024V nominal) on the inverting channel. The comparator inputs and outputs are tied to fixed I/O pins, defined in Table 24-1. The available comparator configurations and their corresponding bit settings are shown in Figure 24-4.

TABLE 24-1:	COMPARATOR INPUTS AND
	OUTPUTS

0011 010					
Comparator	Input or Output	I/O Pin			
	C1INA (VIN+)	RF6			
	C1INB (VIN-)	RF5			
1	C1INC ⁽¹⁾ (VIN-)	RH6			
	C2INB (VIN-)	RF3			
	C1OUT	RF2			
	C2INA (VIN+)	RF4			
	C2INB (VIN-)	RF3			
2	C2INC ⁽¹⁾ (VIN-)	RH4			
	C2IND ⁽¹⁾ (VIN-)	RH5			
	C2OUT	RF1			
	C3INA (VIN+)	RG2			
	C3INB (VIN-)	RG3			
3	C3INC (VIN-)	RG4			
	C2INB (VIN-)	RF3			
	C3OUT	RG1			

Note 1: C1INC, C2INC and C2IND are all unavailable for 64-pin devices (PIC18F6XK22).

## 24.5.1 COMPARATOR ENABLE AND INPUT SELECTION

Setting the CON bit of the CMxCON register (CMxCON<7>) enables the comparator for operation. Clearing the CON bit disables the comparator, resulting in minimum current consumption.

The CCH<1:0> bits in the CMxCON register (CMxCON<1:0>) direct either one of three analog input pins, or the Internal Reference Voltage (VBG), to the comparator, VIN-. Depending on the Comparator

operating mode, either an external or Internal Reference Voltage may be used. For external analog pins that are unavailable in 64-pin devices (C1INC, C2INC and C2IND), the corresponding configurations that use them as inputs are unavailable.

The analog signal present at VIN- is compared to the signal at VIN+ and the digital output of the comparator is adjusted accordingly.

The external reference is used when CREF = 0 (CMxCON<2>) and VIN+ is connected to the CxINA pin. When external reference voltages are used, the comparator module can be configured to have the reference sources externally. The reference signal must be between Vss and VDD, and can be applied to either pin of the comparator.

The comparator module also allows the selection of an internally generated reference voltage from the Comparator Voltage Reference (CVREF) module. This module is described in more detail in **Section 25.0 "Comparator Voltage Reference Module"**. The reference from the comparator reference voltage module is only available when CREF = 1. In this mode, the Internal Reference Voltage is applied to the comparator's VIN+ pin.

Note:	The comparator input pin, selected by
	CCH<1:0>, must be configured as an input
	by setting both the corresponding TRISF,
	TRISG or TRISH bit and the corresponding
	ANSELx bit in the ANCONx register.

## 24.5.2 COMPARATOR ENABLE AND OUTPUT SELECTION

The comparator outputs are read through the CMSTAT register. The CMSTAT<5> bit reads the Comparator 1 output, CMSTAT<6> reads Comparator 2 output and CMSTAT<7> reads Comparator 3 output. These bits are read-only.

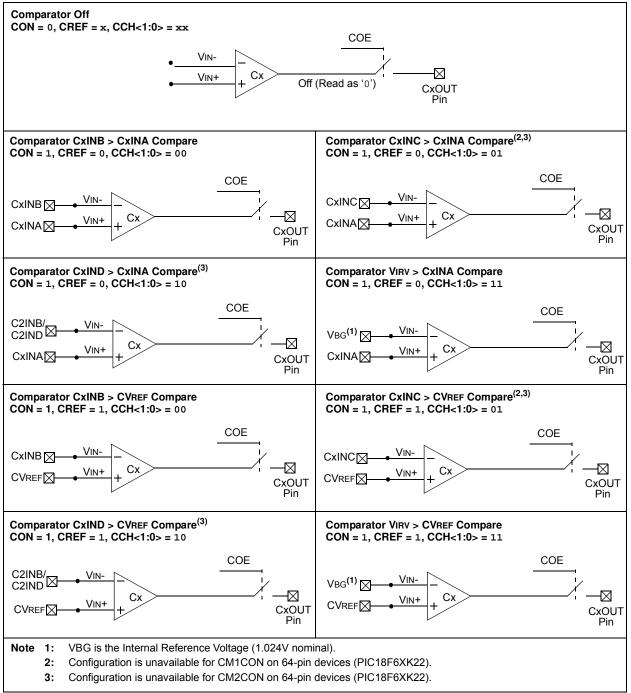
The comparator outputs may also be directly output to the RF2, RF1 and RG1 I/O pins by setting the COE bit (CMxCON<6>). When enabled, multiplexers in the output path of the pins switch to the output of the comparator. While in this mode, the TRISF<2:1> and TRISG<1> bits still function as the digital output enable bits for the RF2, RF1 and RG1 pins.

By default, the comparator's output is at logic high whenever the voltage on VIN+ is greater than on VIN-. The polarity of the comparator outputs can be inverted using the CPOL bit (CMxCON<5>).

The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications, as discussed in **Section 24.2 "Comparator Operation"**.

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FIGURE 24-4: COMPARATOR CONFIGURATIONS



### 24.6 Comparator Interrupts

The comparator interrupt flag is set whenever any of the following occurs:

- · Low-to-high transition of the comparator output
- High-to-low transition of the comparator output
- · Any change in the comparator output

The comparator interrupt selection is done by the EVPOL<1:0> bits in the CMxCON register (CMxCON<4:3>).

In order to provide maximum flexibility, the output of the comparator may be inverted using the CPOL bit in the CMxCON register (CMxCON<5>). This is functionally identical to reversing the inverting and non-inverting inputs of the comparator for a particular mode.

An interrupt is generated on the low-to-high or high-tolow transition of the comparator output. This mode of interrupt generation is dependent on EVPOL<1:0> in the CMxCON register. When EVPOL<1:0> = 01 or 10, the interrupt is generated on a low-to-high or high-tolow transition of the comparator output. Once the interrupt is generated, it is required to clear the interrupt flag by software. When EVPOL<1:0> = 11, the comparator interrupt flag is set whenever there is a change in the output value of either comparator. Software will need to maintain information about the status of the output bits, as read from CMSTAT<7:5>, to determine the actual change that occurred.

The CMPxIF bits (PIR6<2:0>) are the Comparator Interrupt Flags. The CMPxIF bits must be reset by clearing them. Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated. Table 24-2 shows the interrupt generation with respect to comparator input voltages and EVPOL bit settings.

Both the CMPxIE bits (PIE6<2:0>) and the PEIE bit (INTCON<6>) must be set to enable the interrupt. In addition, the GIE bit (INTCON<7>) must also be set. If any of these bits are clear, the interrupt is not enabled, though the CMPxIF bits will still be set if an interrupt condition occurs.

A simplified diagram of the interrupt section is shown in Figure 24-3.

**Note:** The CMPxIF bits will not be set when EVPOL<1:0> = 00.

CPOL	EVPOL<1:0>	Comparator Input Change	CxOUT Transition	Interrupt Generated
	0.0	VIN+ > VIN-	Low-to-High	No
	00	VIN+ < VIN-	High-to-Low	No
	01	VIN+ > VIN-	Low-to-High	Yes
0	UL	VIN+ < VIN-	High-to-Low	No
0	10	VIN+ > VIN-	Low-to-High	No
	10	Vin+ < Vin-	High-to-Low	Yes
	11	VIN+ > VIN-	Low-to-High	Yes
	11	VIN+ < VIN-	High-to-Low	Yes
		VIN+ > VIN-	High-to-Low	No
	00	VIN+ < VIN-	Low-to-High	No
	0.1	VIN+ > VIN-	High-to-Low	No
1	01	VIN+ < VIN-	Low-to-High	Yes
1	10	VIN+ > VIN-	High-to-Low	Yes
	10	VIN+ < VIN-	Low-to-High	No
	11	VIN+ > VIN-	High-to-Low	Yes
	11	VIN+ < VIN-	Low-to-High	Yes

 TABLE 24-2:
 COMPARATOR INTERRUPT GENERATION

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### 24.7 Comparator Operation During Sleep

When a comparator is active and the device is placed in Sleep mode, the comparator remains active and the interrupt is functional, if enabled. This interrupt will wake up the device from Sleep mode, when enabled. Each operational comparator will consume additional current. To minimize power consumption while in Sleep mode, turn off the comparators (CON = 0) before entering Sleep. If the device wakes up from Sleep, the contents of the CMxCON register are not affected.

## 24.8 Effects of a Reset

A device Reset forces the CMxCON registers to their Reset state. This forces both comparators and the voltage reference to the OFF state.

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	<b>INT0IF</b>	RBIF
	_	_	EEIF		CMP3IF	CMP2IF	CMP1IF
_	_	_	EEIE	_	CMP3IE	CMP2IE	CMP1IE
_	_	_	EEIP	_	CMP3IP	CMP2IP	CMP1IP
CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CVREN	CVROE	CVRSS	CVR4	CVR3	CVR2	CVR1	CVR0
CMP3OUT	CMP2OUT	CMP1OUT				_	
RF7	RF6	RF5	RF4	RF3	RF2	RF1	_
LATF7	LATF6	LATF5	LATF4	LATF3	LATF2	LATF1	_
TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	_
	_	RG5 ⁽¹⁾	RG4	RG3	RG2	RG1	RG0
_	_	_	LATG4	LATG3	LATG2	LATG1	LATG0
-	—	_	TRISG4	TRISG3	TRISG2	TRISG1	TRISG0
RH7	RH6	RH5	RH4	RH3	RH2	RH1	RH0
LATH7	LATH6	LATH5	LATH4	LATH3	LATH2	LATH1	LATH0
TRISH7	TRISH6	TRISH5	TRISH4	TRISH3	TRISH2	TRISH1	TRISH0
ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8
ANSEL23	ANSEL22	ANSEL21	ANSEL20	ANSEL19	ANSEL18	ANSEL17	ANSEL16
CCP3MD	CCP2MD	CCP1MD	UART2MD	UART1MD	SSP2MD	SSP1MD	ADCMD
	GIE/GIEH       —       —       —       CON       ANSEL15       ANSEL23       CCP3MD	GIE/GIEH         PEIE/GIEL           —         —           —         —           —         —           —         —           —         —           —         —           —         —           —         —           —         —           —         —           —         —           —         —           —         —           CON         COE           CMP30UT         CMP20UT           RF7         RF6           IATF6         —           —         —           —         —           —         —           —         —           IATH7         RH6	GIE/GIEH         PEIE/GIEL         TMR0IE           —         —         —           —         —         —           —         —         —           —         —         —           —         —         —           —         —         —           —         —         —           —         —         —           —         —         —           —         —         —           —         —         —           —         —         —           CON         COE         CPOL           CON         COE         CPOL           CON         COE         CPOL           CVREN         CVROE         CVRSS           CMP3OUT         CMP2OUT         CMP10UT           RF7         RF6         RF5           LATF7         LATF6         LATF5           TRISF7         TRISF6         TRISF5           —         —         —           —         —         —           —         —         —           —         —         —           —	GIE/GIEH         PEIE/GIEL         TMR0IE         INT0IE           —         —         —         EEIF           —         —         —         EEIP           CON         COE         CPOL         EVPOL1           CMRT         RH7         RF6         RF5         RF4           LATF7         LATF6         IATSF5         IATG4           —	GIE/GIEH         PEIE/GIEL         TMROIE         INTOIE         RBIE           —         —         —         EEIF         —           —         —         —         EEIP         —           —         —         —         EVPOL1         EVPOL0           CON         COE         CPOL         EVPOL1         EVPOL0           CON         COE         CPOL         EVPOL1         EVPOL0           CVREN         CVROE         CVRSS         CVR4         CVR3           CMP3OUT         CMP2OUT         CMP1OUT         —         —           RF7         RF6         RF5         RF4         RF3           LATF7         LATF6         LATF5         LATF4         LATF3           TRISF7         TRISF6         TRISF5         TRISF4         TRISG3           —         —         —         —         LATG4         LATG3	GIE/GIEH         PEIE/GIEL         TMROIE         INTOIE         RBIE         TMROIF           —         —         —         EEIF         —         CMP3IF           —         —         —         EEIF         —         CMP3IF           —         —         —         EEIF         —         CMP3IF           —         —         —         EEIP         —         CMP3IF           —         —         —         EEIP         —         CMP3IF           CON         COE         CPOL         EVPOL1         EVPOL0         CREF           CON         COE         CPOL         EVPOL1         EVPOL0         CREF           CVR         CVROE         CVRSS         CVR4         CVR3         CVR2           CMP30UT         CMP20UT         CMP10UT         —         —         —           RF7         RF6         RF5         RF4         RF3         RF2           LATF7         LATF6         LATF5         LATF4         LATF3         LATF2           TRISF7         TRISF6         TRISF3         TRISF4         RG3         RG2           —         —         —         —         —	GIE/GIEH         PEIE/GIEL         TMR0IE         INT0IE         RBIE         TMR0IF         INT0IF           —         —         —         EEIF         —         CMP3IF         CMP2IF           —         —         —         EEIF         —         CMP3IE         CMP2IF           —         —         —         EEIP         —         CMP3IP         CMP2IP           CON         COE         CPOL         EVPOL1         EVPOL0         CREF         CCH1           CVREN         CVROE         CVRSS         CVR4         EVPOL0         CREF         CCH1           CMP30UT         CMP20UT         CMP10UT         —         —         —         —         —           RF7         RF6         RF5         RF4         RF3

<b>TABLE 24-3:</b>	<b>REGISTERS ASSOCIATED WITH COMPARATOR MODULE</b>
--------------------	----------------------------------------------------

Legend: — = unimplemented, read as '0'.

Note 1: Bit is available when Master Clear is disabled (MCLRE = 0). When MCLRE is set, the bit is unimplemented.

## 25.0 COMPARATOR VOLTAGE REFERENCE MODULE

The comparator voltage reference is a 32-tap resistor ladder network that provides a selectable reference voltage. Although its primary purpose is to provide a reference for the analog comparators, it may also be used independently of them.

A block diagram of the module is shown in Figure 25-1. The resistor ladder is segmented to provide a range of CVREF values and has a power-down function to conserve power when the reference is not being used. The module's supply reference can be provided from either device VDD/VSS or an external voltage reference.

### 25.1 Configuring the Comparator Voltage Reference

The comparator voltage reference module is controlled through the CVRCON register (Register 25-1). The comparator voltage reference provides a range of output voltage with 32 levels.

The CVR<4:0> selection bits (CVRCON<4:0>) offer a range of output voltages. Equation 25-1 shows the how the comparator voltage reference is computed.

## EQUATION 25-1:

If CVRSS = 1:

 $CVREF = (VREF-) + (CVR < 4:0 > /32) \cdot (VREF+ - VREF-)$ 

<u>If CVRSS = 0:</u>

 $CVREF = (AVSS) + (CVR < 4:0 > /32) \cdot (AVDD - AVSS)$ 

The comparator reference supply voltage can come from either VDD and VSS, or the external VREF+ and VREF- that are multiplexed with RA3 and RA2. The voltage source is selected by the CVRSS bit (CVRCON<5>).

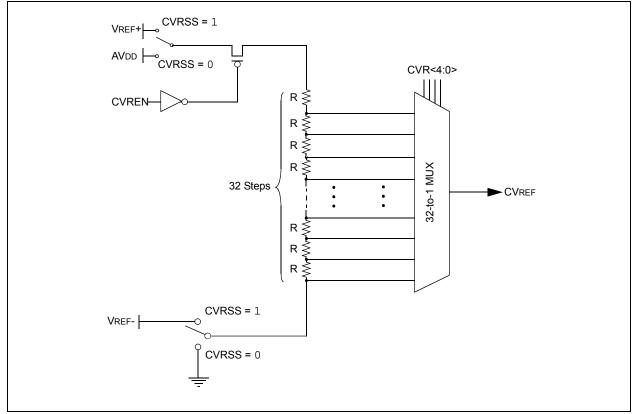
The settling time of the comparator voltage reference must be considered when changing the CVREF output (see Table 31-2 in Section 31.0 "Electrical Characteristics").

## REGISTER 25-1: CVRCON: COMPARATOR VOLTAGE REFERENCE CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CVREN	CVROE	CVRSS	CVR4	CVR3	CVR2	CVR1	CVR0
bit 7							bit (
Legend:							
R = Readab	ole bit	W = Writable b	oit	U = Unimplem	nented bit, rea	d as '0'	
-n = Value a	at POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown
bit 7 bit 6	1 = CVREF ci 0 = CVREF ci <b>CVROE:</b> Com 1 = CVREF vo 0 = CVREF vo	nparator Voltage rcuit is powered rcuit is powered nparator VREF O oltage level is ou oltage level is di	on down utput Enable utput on CVRE sconnected fr	bit EF pin Fom CVREF pin			
bit 5	1 = Compara	nparator VREF So ator reference so ator reference so	ource, CVRSR	c = Vref+ – Vr			
bit 4-0	<u>When CVRSS</u> CVREF = (VRE <u>When CVRSS</u>	EF-) + (CVR<4:0	>/32) • (VREF	+ – Vref-)	:0> ≤ 31 bits		

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### FIGURE 25-1: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM



## 25.2 Voltage Reference Accuracy/Error

The full range of voltage reference cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 25-1) keep CVREF from approaching the reference source rails. The voltage reference is derived from the reference source; therefore, the CVREF output changes with fluctuations in that source. The tested absolute accuracy of the voltage reference can be found in Section 31.0 "Electrical Characteristics".

## 25.3 Operation During Sleep

When the device wakes up from Sleep through an interrupt or a Watchdog Timer time-out, the contents of the CVRCON register are not affected. To minimize current consumption in Sleep mode, the voltage reference should be disabled.

## 25.4 Effects of a Reset

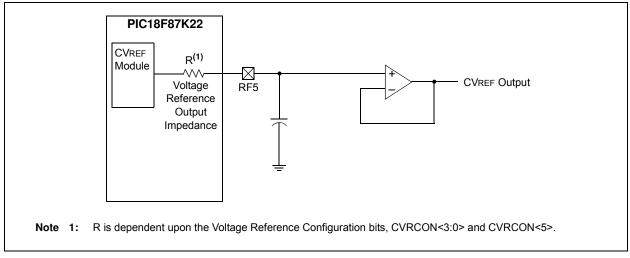
A device Reset disables the voltage reference by clearing bit, CVREN (CVRCON<7>). This Reset also disconnects the reference from the RF5 pin by clearing bit, CVROE (CVRCON<6>).

## 25.5 Connection Considerations

The voltage reference module operates independently of the comparator module. The output of the reference generator may be connected to the RF5 pin if the CVROE bit is set. Enabling the voltage reference output onto RF5 when it is configured as a digital input will increase current consumption. Connecting RF5 as a digital output, with CVRSS enabled, will also increase current consumption.

The RF5 pin can be used as a simple D/A output with limited drive capability. Due to the limited current drive capability, a buffer must be used on the voltage reference output for external connections to VREF. Figure 25-2 shows an example buffering technique.

### FIGURE 25-2: COMPARATOR VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE



### TABLE 25-1: REGISTERS ASSOCIATED WITH COMPARATOR VOLTAGE REFERENCE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CVRCON	CVREN	CVROE	CVRSS	CVR4	CVR3	CVR2	CVR1	CVR0
CM1CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CM2CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
CM3CON	CON	COE	CPOL	EVPOL1	EVPOL0	CREF	CCH1	CCH0
TRISF	TRISF7	TRISF6	TRISF5	TRISF4	TRISF3	TRISF2	TRISF1	—
TRISA	TRISA7	TRISA6	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0
ANCON1	ANSEL15	ANSEL14	ANSEL13	ANSEL12	ANSEL11	ANSEL10	ANSEL9	ANSEL8

Legend: — = unimplemented, read as '0'. Shaded cells are not used with the comparator voltage reference.

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NOTES:

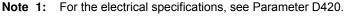
## 26.0 HIGH/LOW-VOLTAGE DETECT (HLVD)

The PIC18F87K22 family of devices has a High/Low-Voltage Detect module (HLVD). This is a programmable circuit that sets both a device voltage trip point and the direction of change from that point. If the device experiences an excursion past the trip point in that direction, an interrupt flag is set. If the interrupt is enabled, the program execution branches to the interrupt vector address and the software responds to the interrupt. The High/Low-Voltage Detect Control register (Register 26-1) completely controls the operation of the HLVD module. This allows the circuitry to be "turned off" by the user under software control, which minimizes the current consumption for the device.

The module's block diagram is shown in Figure 26-1.

## REGISTER 26-1: HLVDCON: HIGH/LOW-VOLTAGE DETECT CONTROL REGISTER

R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-1	R/W-0	R/W-0			
VDIRMAG	BGVST	IRVST	HLVDEN	HLVDL3 ⁽¹⁾	HLVDL2 ⁽¹⁾	HLVDL1 ⁽¹⁾	HLVDL0 ⁽¹⁾			
bit 7							bit 0			
Legend:										
R = Readable bit W = Writable bit U = Unimplemented bit, read as '0'										
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown			
bit 7		oltage Directio	•							
			0 1		oint (HLVDL<3:0 point (HLVDL<	,				
bit 6		d Gap Referen	0	•		5.0~)				
bit o		and gap voltag	•		ig bit					
		and gap voltag								
bit 5	IRVST: Intern	al Reference V	/oltage Stable	Flag bit						
	1 = Indicates	= Indicates that the voltage detect logic will generate the interrupt flag at the specified voltage range								
		that the voltaged the HLVD int		•	ate the interrup	t flag at the spe	ecified voltage			
bit 4	HLVDEN: Hig	gh/Low-Voltage	Detect Powe	r Enable bit						
	1 = HLVD is									
	0 = HLVD is disabled									
bit 3-0	HLVDL<3:0>: Voltage Detection Limit bits ⁽¹⁾									
		1111 = External analog input is used (input comes from the HLVDIN pin)								
	1110 <b>= Maxi</b> r	mum setting								
	•									
	0000 <b>= Minin</b>	num setting								
				D (00						



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The module is enabled by setting the HLVDEN bit (HLVDCON<4>). Each time the HLVD module is enabled, the circuitry requires some time to stabilize. The IRVST bit (HLVDCON<5>) is a read-only bit used to indicate when the circuit is stable. The module can only generate an interrupt after the circuit is stable and IRVST is set.

The VDIRMAG bit (HLVDCON<7>) determines the overall operation of the module. When VDIRMAG is cleared, the module monitors for drops in VDD below a predetermined set point. When the bit is set, the module monitors for rises in VDD above the set point.

## 26.1 Operation

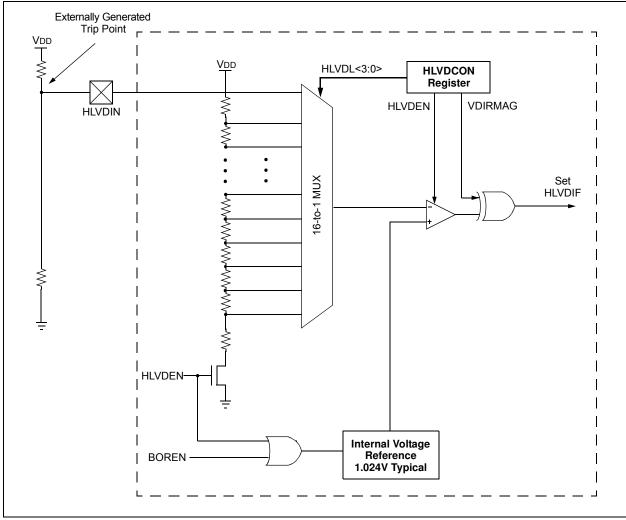
When the HLVD module is enabled, a comparator uses an internally generated voltage reference as the set point. The set point is compared with the trip point, where each node in the resistor divider represents a trip point voltage. The "trip point" voltage is the voltage level at which the device detects a high or low-voltage event, depending on the configuration of the module.

When the supply voltage is equal to the trip point, the voltage tapped off of the resistor array is equal to the internal reference voltage generated by the voltage reference module. The comparator then generates an interrupt signal by setting the HLVDIF bit.

The trip point voltage is software programmable to any of 16 values. The trip point is selected by programming the HLVDL<3:0> bits (HLVDCON<3:0>).

The HLVD module has an additional feature that allows the user to supply the trip voltage to the module from an external source. This mode is enabled when bits, HLVDL<3:0>, are set to '1111'. In this state, the comparator input is multiplexed from the external input pin, HLVDIN. This gives users the flexibility of configuring the High/Low-Voltage Detect interrupt to occur at any voltage in the valid operating range.





## 26.2 HLVD Setup

To set up the HLVD module:

- 1. Select the desired HLVD trip point by writing the value to the HLVDL<3:0> bits.
- 2. Set the VDIRMAG bit to detect high voltage (VDIRMAG = 1) or low voltage (VDIRMAG = 0).
- 3. Enable the HLVD module by setting the HLVDEN bit.
- 4. Clear the HLVD interrupt flag (PIR2<2>), which may have been set from a previous interrupt.
- 5. If interrupts are desired, enable the HLVD interrupt by setting the HLVDIE and GIE bits (PIE2<2> and INTCON<7>, respectively).

An interrupt will not be generated until the IRVST bit is set.

**Note:** Before changing any module settings (VDIRMAG, LVDL<3:0>), first disable the module (LVDEN = 0), make the changes and re-enable the module. This prevents the generation of false HLVD events.

## 26.3 Current Consumption

When the module is enabled, the HLVD comparator and voltage divider are enabled and consume static current. The total current consumption, when enabled, is specified in electrical specification Parameter D022B (Table 31-13).

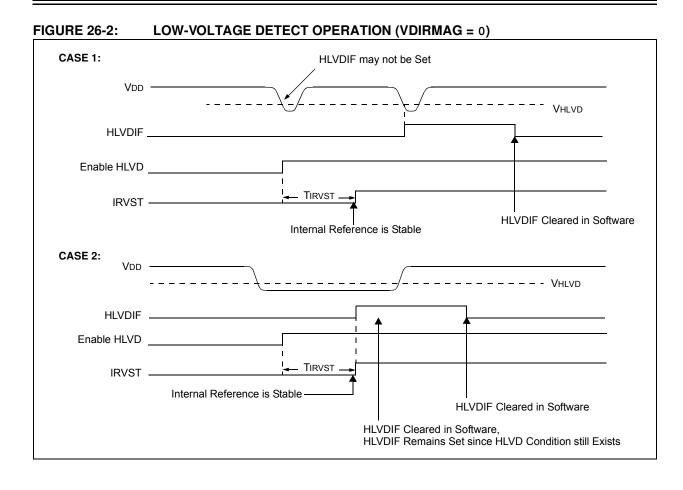
Depending on the application, the HLVD module does not need to operate constantly. To reduce current requirements, the HLVD circuitry may only need to be enabled for short periods where the voltage is checked. After such a check, the module could be disabled.

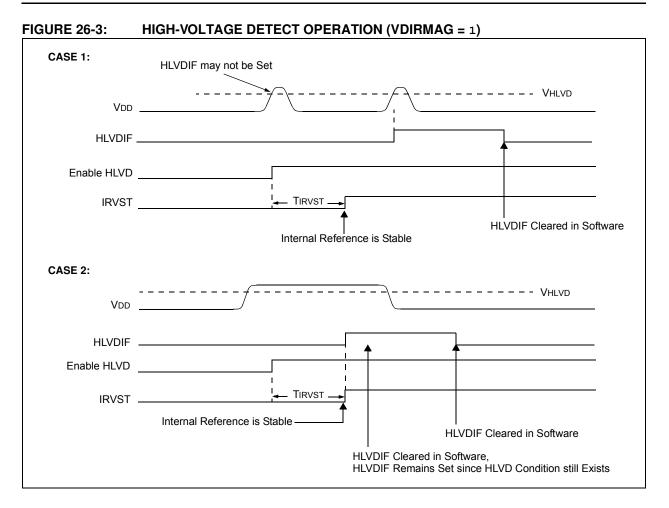
## 26.4 HLVD Start-up Time

The internal reference voltage of the HLVD module, specified in electrical specification Parameter 37 (Section 31.0 "Electrical Characteristics"), may be used by other internal circuitry, such as the programmable Brown-out Reset. If the HLVD or other circuits using the voltage reference are disabled to lower the device's current consumption, the reference voltage circuit will require time to become stable before a low or high-voltage condition can be reliably detected. This start-up time, TIRVST, is an interval that is independent of device clock speed. It is specified in electrical specification Parameter 37 (Table 31-13).

The HLVD interrupt flag is not enabled until TIRVST has expired and a stable reference voltage is reached. For this reason, brief excursions beyond the set point may not be detected during this interval (see Figure 26-2 or Figure 26-3).

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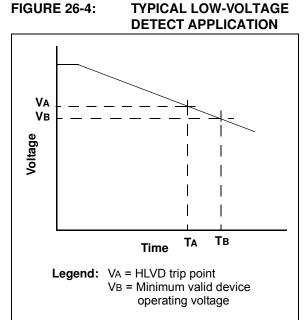




### 26.5 Applications

In many applications, it is desirable to detect a drop below, or rise above, a particular voltage threshold. For example, the HLVD module could be periodically enabled to detect Universal Serial Bus (USB) attach or detach. This assumes the device is powered by a lower voltage source than the USB when detached. An attach would indicate a High-Voltage Detect from, for example, 3.3V to 5V (the voltage on USB) and vice versa for a detach. This feature could save a design a few extra components and an attach signal (input pin).

For general battery applications, Figure 26-4 shows a possible voltage curve. Over time, the device voltage decreases. When the device voltage reaches voltage, VA, the HLVD logic generates an interrupt at time, TA. The interrupt could cause the execution of an Interrupt Service Routine (ISR), which would allow the application to perform "housekeeping tasks" and a controlled shutdown before the device voltage exits the valid operating range at TB. This would give the application a time window, represented by the difference between TA and TB, to safely exit.



## 26.6 Operation During Sleep

When enabled, the HLVD circuitry continues to operate during Sleep. If the device voltage crosses the trip point, the HLVDIF bit will be set and the device will wake up from Sleep. Device execution will continue from the interrupt vector address if interrupts have been globally enabled.

## 26.7 Effects of a Reset

A device Reset forces all registers to their Reset state. This forces the HLVD module to be turned off.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
HLVDCON	VDIRMAG	BGVST	IRVST	HLVDEN	HLVDL3	HLVDL2	HLVDL1	HLVDL0
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF
PIR2	OSCFIF	—	SSP2IF	BLC2IF	BCL1IF	HLVDIF	TMR3IF	TMR3GIF
PIE2	OSCFIE	_	SSP2IE	BLC2IE	BCL1IE	HLVDIE	TMR3IE	TMR3GIE
IPR2	OSCFIP	_	SSP2IP	BLC2IP	BCL1IP	HLVDIP	TMR3IP	TMR3GIP
TRISA	TRISA7 ⁽¹⁾	TRISA6 ⁽¹⁾	TRISA5	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0
ANCON0	ANSEL7	ANSEL6	ANSEL5	ANSEL4	ANSEL3	ANSEL2	ANSEL1	ANSEL0

#### TABLE 26-1: REGISTERS ASSOCIATED WITH HIGH/LOW-VOLTAGE DETECT MODULE

**Legend:** — = unimplemented, read as '0'. Shaded cells are unused by the HLVD module.

**Note 1:** PORTA<7:6> and their direction bits are individually configured as port pins based on various primary oscillator modes. When disabled, these bits read as '0'.

## 27.0 CHARGE TIME MEASUREMENT UNIT (CTMU)

The Charge Time Measurement Unit (CTMU) is a flexible analog module that provides accurate differential time measurement between pulse sources, as well as asynchronous pulse generation. By working with other on-chip analog modules, the CTMU can precisely measure time, capacitance and relative changes in capacitance or generate output pulses with a specific time delay. The CTMU is ideal for interfacing with capacitive-based sensors.

The module includes these key features:

- Up to 24 channels available for capacitive or time measurement input
- · On-chip precision current source
- · Four-edge input trigger sources
- Polarity control for each edge source
- Control of edge sequence

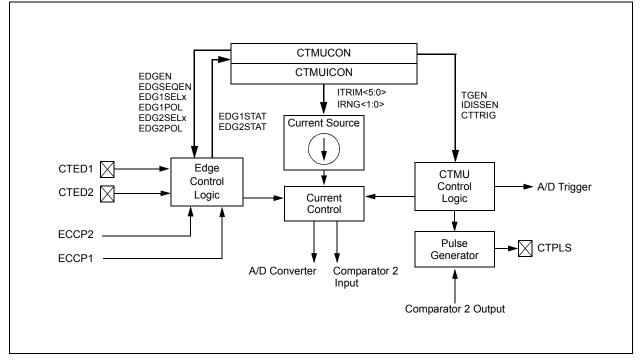


- Control of response to edges
- · Time measurement resolution of 1 nanosecond
- · High-precision time measurement
- Time delay of external or internal signal asynchronous to system clock
- Accurate current source suitable for capacitive measurement

The CTMU works in conjunction with the A/D Converter to provide up to 24 channels for time or charge measurement, depending on the specific device and the number of A/D channels available. When configured for time delay, the CTMU is connected to one of the analog comparators. The level-sensitive input edge sources can be selected from four sources: two external inputs or the ECCP1/ECCP2 Special Event Triggers.

The CTMU special event can trigger the Analog-to-Digital Converter module.

Figure 27-1 provides a block diagram of the CTMU.



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## 27.1 CTMU Registers

The control registers for the CTMU are:

- CTMUCONH
- CTMUCONL
- CTMUICON

The CTMUCONH and CTMUCONL registers (Register 27-1 and Register 27-2) contain control bits for configuring the CTMU module edge source selection, edge source polarity selection, edge sequencing, A/D trigger, analog circuit capacitor discharge and enables. The CTMUICON register (Register 27-3) has bits for selecting the current source range and current source trim.

### REGISTER 27-1: CTMUCONH: CTMU CONTROL HIGH REGISTER

R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
CTMUEN	—	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit	, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	CTMUEN: CTMU Enable bit
	<ul><li>1 = Module is enabled</li><li>0 = Module is disabled</li></ul>
bit 6	Unimplemented: Read as '0'
bit 5	CTMUSIDL: Stop in Idle Mode bit
	<ul><li>1 = Discontinue module operation when device enters Idle mode</li><li>0 = Continue module operation in Idle mode</li></ul>
bit 4	TGEN: Time Generation Enable bit
	<ul><li>1 = Enables edge delay generation</li><li>0 = Disables edge delay generation</li></ul>
bit 3	EDGEN: Edge Enable bit
	<ul><li>1 = Edges are not blocked</li><li>0 = Edges are blocked</li></ul>
bit 2	EDGSEQEN: Edge Sequence Enable bit
	<ul> <li>1 = Edge 1 event must occur before Edge 2 event can occur</li> <li>0 = No edge sequence is needed</li> </ul>
bit 1	IDISSEN: Analog Current Source Control bit
	<ul> <li>1 = Analog current source output is grounded</li> <li>0 = Analog current source output is not grounded</li> </ul>
bit 0	<b>CTTRIG:</b> Trigger Control bit 1 = Trigger output is enabled 0 = Trigger output is disabled

#### REGISTER 27-2: CTMUCONL: CTMU CONTROL LOW REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT		
bit 7		•		•			bit 0		
Legend:									
R = Readable	e bit	W = Writable I	bit	U = Unimplem	nented bit, read	l as '0'			
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	iown		
bit 7		dge 2 Polarity S	Soloct bit						
		programmed f		dae response					
		programmed f							
bit 6-5	EDG2SEL<1:	0>: Edge 2 So	urce Select bit	S					
	11 = CTED1 pin								
	10 = CTED2 pin 01 = ECCP1 Special Event Trigger								
		Special Event 1 Special Event 1							
bit 4		dge 1 Polarity S							
		programmed f		dge response					
		programmed f							
bit 3-2	EDG1SEL<1:	:0>: Edge 1 So	urce Select bit	S					
	11 = CTED1 pin								
	10 = CTED2 pin 01 = ECCP1 Special Event Trigger								
		Special Event 1							
bit 1		EDG2STAT: Edge 2 Status bit							
	1 = Edge 2 event has occurred								
	0 = Edge 2 e	vent has not oc	curred						
bit 0		Edge 1 Status b							
	0	vent has occur							
	0 = ⊨age 1 e	vent has not oc	currea						

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00 = Current Source Disabled

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0				
ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1	ITRIM0	IRNG1	IRNG0				
bit 7							bit 0				
<u> </u>											
Legend:											
R = Readable	e bit	W = Writable	bit	U = Unimplem	nented bit, read	1 as '0'					
-n = Value at	POR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkr	nown				
bit 7-2	ITRIM<5:0>:	Current Source	Trim bits								
	011111 <b>= M</b> a	aximum positive	change from	nominal current	t						
	011110										
	•										
	•										
	-	nimum nositive	change from	nominal current							
	000001 = Minimum positive change from nominal current 000000 = Nominal current output specified by IRNG<1:0>										
	111111 = Minimum negative change from nominal current										
	100010										
<b>h</b> #4.0	100001 = Maximum negative change from nominal current										
bit 1-0	IRNG<1:0>: Current Source Range Select bits										
	11 = 100 x Ba 10 = 10 x Bas										
		arrent Level (0.5	55 u A nominal	)							
				)							

### REGISTER 27-3: CTMUICON: CTMU CURRENT CONTROL REGISTER

## 27.2 CTMU Operation

The CTMU works by using a fixed current source to charge a circuit. The type of circuit depends on the type of measurement being made.

In the case of charge measurement, the current is fixed and the amount of time the current is applied to the circuit is fixed. The amount of voltage read by the A/D becomes a measurement of the circuit's capacitance.

In the case of time measurement, the current, as well as the capacitance of the circuit, is fixed. In this case, the voltage read by the A/D is representative of the amount of time elapsed from the time the current source starts and stops charging the circuit.

If the CTMU is being used as a time delay, both capacitance and current source are fixed, as well as the voltage supplied to the comparator circuit. The delay of a signal is determined by the amount of time it takes the voltage to charge to the comparator threshold voltage.

## 27.2.1 THEORY OF OPERATION

The operation of the CTMU is based on the equation for charge:

$$C = I \bullet \frac{dV}{dT}$$

More simply, the amount of charge measured in coulombs in a circuit is defined as current in amperes (I) multiplied by the amount of time in seconds that the current flows (t). Charge is also defined as the capacitance in farads (C) multiplied by the voltage of the circuit (V). It follows that:

 $I \bullet t = C \bullet V$ 

The CTMU module provides a constant, known current source. The A/D Converter is used to measure (V) in the equation, leaving two unknowns: capacitance (C) and time (t). The above equation can be used to calculate capacitance or time, by either the relationship using the known fixed capacitance of the circuit:

 $t = (C \bullet V)/I$ 

or by:

 $C = (I \bullet t)/V$ 

using a fixed time that the current source is applied to the circuit.

## 27.2.2 CURRENT SOURCE

At the heart of the CTMU is a precision current source, designed to provide a constant reference for measurements. The level of current is user-selectable across three ranges or a total of two orders of magnitude, with the ability to trim the output in  $\pm 2\%$  increments (nominal). The current range is selected by the IRNG<1:0> bits (CTMUICON<1:0>), with a value of '00' representing the lowest range.

Current trim is provided by the ITRIM<5:0> bits (CTMUICON<7:2>). These six bits allow trimming of the current source in steps of approximately 2% per step. Half of the range adjusts the current source positively and the other half reduces the current source. A value of '000000' is the neutral position (no change). A value of '100000' is the maximum negative adjustment (approximately -62%) and '011111' is the maximum positive adjustment (approximately +62%).

## 27.2.3 EDGE SELECTION AND CONTROL

CTMU measurements are controlled by edge events occurring on the module's two input channels. Each channel, referred to as Edge 1 and Edge 2, can be configured to receive input pulses from one of the edge input pins (CTED1 and CTED2) or CCPx Special Event Triggers. The input channels are level-sensitive, responding to the instantaneous level on the channel rather than a transition between levels. The inputs are selected using the EDG1SEL and EDG2SEL bit pairs (CTMUCONL<3:2, 6:5>).

In addition to source, each channel can be configured for event polarity using the EDGE2POL and EDGE1POL bits (CTMUCONL<7,4>). The input channels can also be filtered for an edge event sequence (Edge 1 occurring before Edge 2) by setting the EDGSEQEN bit (CTMUCONH<2>).

## 27.2.4 EDGE STATUS

The CTMUCON register also contains two status bits: EDG2STAT and EDG1STAT (CTMUCONL<1:0>). Their primary function is to show if an edge response has occurred on the corresponding channel. The CTMU automatically sets a particular bit when an edge response is detected on its channel. The level-sensitive nature of the input channels also means that the status bits become set immediately if the channel's configuration is changed and matches the channel's current state.

The module uses the edge status bits to control the current source output to external analog modules (such as the A/D Converter). Current is only supplied to external modules when only one (not both) of the status bits is set. Current is shut off when both bits are either set or cleared. This allows the CTMU to measure current only during the interval between edges. After both status bits are set, it is necessary to clear them before another measurement is taken. Both bits should be cleared simultaneously, if possible, to avoid re-enabling the CTMU current source.

In addition to being set by the CTMU hardware, the edge status bits can also be set by software. This permits a user application to manually enable or disable the current source. Setting either (but not both) of the bits enables the current source. Setting or clearing both bits at once disables the source.

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## 27.2.5 INTERRUPTS

The CTMU sets its interrupt flag (PIR3<3>) whenever the current source is enabled, then disabled. An interrupt is generated only if the corresponding interrupt enable bit (PIE3<3>) is also set. If edge sequencing is not enabled (i.e., Edge 1 must occur before Edge 2), it is necessary to monitor the edge status bits and determine which edge occurred last and caused the interrupt.

## 27.3 CTMU Module Initialization

The following sequence is a general guideline used to initialize the CTMU module:

- 1. Select the current source range using the IRNGx bits (CTMUICON<1:0>).
- 2. Adjust the current source trim using the ITRIMx bits (CTMUICON<7:2>).
- Configure the edge input sources for Edge 1 and Edge 2 by setting the EDG1SEL and EDG2SEL bits (CTMUCONL<3:2> and <6:5>, respectively).
- Configure the input polarities for the edge inputs using the EDG2POL and EDG1POL bits (CTMUCONL<7,4>).

The default configuration is for negative edge polarity (high-to-low transitions).

5. Enable edge sequencing using the EDGSEQEN bit (CTMUCONH<2>).

By default, edge sequencing is disabled.

6. Select the operating mode (Measurement or Time Delay) with the TGEN bit.

The default mode is Time/Capacitance Measurement.

 Configure the module to automatically trigger an A/D conversion when the second edge event has occurred, using the CTTRIG bit (CTMUCONH<0>).

The conversion trigger is disabled by default.

- 8. Discharge the connected circuit by setting the IDISSEN bit (CTMUCONH<1>).
- 9. After waiting a sufficient time for the circuit to discharge, clear IDISSEN.
- 10. Disable the module by clearing the CTMUEN bit (CTMUCONH<7>).
- 11. Clear the Edge Status bits, EDG2STAT and EDG1STAT (CTMUCONL<1:0>).
- 12. Enable both edge inputs by setting the EDGEN bit (CTMUCONH<3>).
- 13. Enable the module by setting the CTMUEN bit.

Depending on the type of measurement or pulse generation being performed, one or more additional modules may also need to be initialized and configured with the CTMU module:

- Edge Source Generation: In addition to the external edge input pins, CCPx Special Event Triggers can be used as edge sources for the CTMU.
- Capacitance or Time Measurement: The CTMU module uses the A/D Converter to measure the voltage across a capacitor that is connected to one of the analog input channels.
- Pulse Generation: When generating system clock independent, output pulses, the CTMU module uses Comparator 2 and the associated comparator voltage reference.

## 27.4 Calibrating the CTMU Module

The CTMU requires calibration for precise measurements of capacitance and time, as well as for accurate time delay. If the application only requires measurement of a relative change in capacitance or time, calibration is usually not necessary. An example of a less precise application is a capacitive touch switch, in which the touch circuit has a baseline capacitance and the added capacitance of the human body changes the overall capacitance of a circuit.

If actual capacitance or time measurement is required, two hardware calibrations must take place:

- The current source needs calibration to set it to a precise current.
- The circuit being measured needs calibration to measure or nullify any capacitance other than that to be measured.

### 27.4.1 CURRENT SOURCE CALIBRATION

The current source on board the CTMU module has a range of  $\pm 60\%$  nominal for each of three current ranges. For precise measurements, it is possible to measure and adjust this current source by placing a high-precision resistor, RCAL, onto an unused analog channel. An example circuit is shown in Figure 27-2.

To measure the current source:

- 1. Initialize the A/D Converter.
- 2. Initialize the CTMU.
- Enable the current source by setting EDG1STAT (CTMUCONL<0>).
- 4. Issue the settling time delay.
- 5. Perform the A/D conversion.
- 6. Calculate the current source current using  $I = V/R_{CAL}$ , where  $R_{CAL}$  is a high-precision resistance and V is measured by performing an A/D conversion.

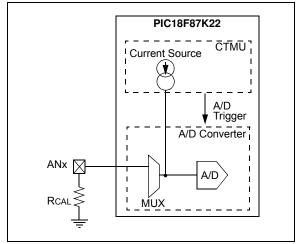
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The CTMU current source may be trimmed with the trim bits in CTMUICON using an iterative process to get the exact current desired. Alternatively, the nominal value without adjustment may be used. That value may be stored by software for use in all subsequent capacitive or time measurements.

To calculate the value for  $\ensuremath{\mathrm{RCAL}}$  , the nominal current must be chosen. Then, the resistance can be calculated.

For example, if the A/D Converter reference voltage is 3.3V, use 70% of full scale (or 2.31V) as the desired approximate voltage to be read by the A/D Converter. If the range of the CTMU current source is selected to be 0.55  $\mu$ A, the resistor value needed is calculated as RCAL = 2.31V/0.55  $\mu$ A, for a value of 4.2 MΩ. Similarly, if the current source is chosen to be 5.5  $\mu$ A, RCAL would be 420,000Ω, and 42,000Ω if the current source is set to 55  $\mu$ A.

#### FIGURE 27-2: CTMU CURRENT SOURCE CALIBRATION CIRCUIT



A value of 70% of full-scale voltage is chosen to make sure that the A/D Converter was in a range that is well above the noise floor. If an exact current is chosen to incorporate the trimming bits from CTMUICON, the resistor value of RCAL may need to be adjusted accordingly. RCAL may also be adjusted to allow for available resistor values. RCAL should be of the highest precision available, in light of the precision needed for the circuit that the CTMU will be measuring. A recommended minimum would be 0.1% tolerance.

The following examples show a typical method for performing a CTMU current calibration.

• Example 27-1 demonstrates how to initialize the A/D Converter and the CTMU.

This routine is typical for applications using both modules.

**Example 27-2** demonstrates one method for the actual calibration routine.

This method manually triggers the A/D Converter to demonstrate the entire step-wise process. It is also possible to automatically trigger the conversion by setting the CTMU's CTTRIG bit (CTMUCONH<0>).

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## EXAMPLE 27-1: SETUP FOR CTMU CALIBRATION ROUTINES

```
#include "pl8cxxx.h"
void setup(void)
{ //CTMUCON - CTMU Control register
 CTMUCONH = 0 \times 00;
 //make sure CTMU is disabled
 CTMUCONL = 0X90;
 //CTMU continues to run when emulator is stopped,CTMU continues
 //to run in idle mode, Time Generation mode disabled, Edges are blocked
 //No edge sequence order, Analog current source not grounded, trigger
 //output disabled, Edge2 polarity = positive level, Edge2 source =
 //source 0, Edgel polarity = positive level, Edgel source = source 0,
 // Set Edge status bits to zero
 //CTMUICON - CTMU Current Control Register
 CTMUICON = 0 \times 01;
 //0.55uA, Nominal - No Adjustment
//Setup AD converter;
TRISA=0x04;
 //set channel 2 as an input
 // Configured AN2 as an analog channel
 // ANCONO
 ANCON0 = 0 \times 04;
 // ANCON1
 ANCON1 = 0 XE0;
 // ADCON2
 // Resulst format 1= Right justified
 ADCON2bits.ADFM=1;
 // Acquition time 7 = 20TAD 2 = 4TAD 1=2TAD
 ADCON2bits.ACQT=1;
 ADCON2bits.ADCS=2;
 // Clock conversion bits 6= FOSC/64 2=FOSC/32
 // ADCON0
 // Vref+ = AVdd
 ADCON1bits.VCFG0 =0;
 // Vref+ = AVdd
 ADCON0bits.VCFG1 =0;
 ADCON0bits.VCFG = 0;
 // Vref- = AVss
 ADCON0bits.CHS=2;
 // Select ADC channel
 // Turn on ADC
 ADCON0bits.ADON=1;
ł
```

### EXAMPLE 27-2: CURRENT CALIBRATION ROUTINE

```
#include "pl8cxxx.h"
#define COUNT 500
 //@ 8MHz = 125uS.
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define RCAL .027
 //R value is 4200000 (4.2M)
 //scaled so that result is in
 //1/100th of uA
#define ADSCALE 1023
 //for unsigned conversion 10 sig bits
#define ADREF 3.3
 //Vdd connected to A/D Vr+
int main(void)
 int i;
 int j = 0; //index for loop
 unsigned int Vread = 0;
 double VTot = 0;
 float Vavg=0, Vcal=0, CTMUISrc = 0; //float values stored for calcs
//assume CTMU and A/D have been setup correctly
//see Example 25-1 for CTMU & A/D setup
setup();
CTMUCONHbits.CTMUEN = 1;
 //Enable the CTMU
 for(j=0;j<10;j++)</pre>
 {
 CTMUCONHbits.IDISSEN = 1;
 //drain charge on the circuit
 DELAY;
 //wait 125us
 CTMUCONHbits.IDISSEN = 0;
 //end drain of circuit
 CTMUCONLbits.EDG1STAT = 1;
 //Begin charging the circuit
 //using CTMU current source
 //wait for 125us
 DELAY;
 CTMUCONLbits.EDG1STAT = 0;
 //Stop charging circuit
 PIR1bits.ADIF = 0;
 //make sure A/D Int not set
 ADCON0bits.GO=1;
 //and begin A/D conv.
 while(!PIR1bits.ADIF);
 //Wait for A/D convert complete
 Vread = ADRES;
 //Get the value from the A/D
 PIR1bits.ADIF = 0;
 //Clear A/D Interrupt Flag
 VTot += Vread;
 //Add the reading to the total
 }
 Vavg = (float)(VTot/10.000);
 //Average of 10 readings
 Vcal = (float)(Vavg/ADSCALE*ADREF);
 CTMUISrc = Vcal/RCAL;
 //CTMUISrc is in 1/100ths of uA
```

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## 27.4.2 CAPACITANCE CALIBRATION

There is a small amount of capacitance from the internal A/D Converter sample capacitor, as well as stray capacitance from the circuit board traces and pads that affect the precision of capacitance measurements. A measurement of the stray capacitance can be taken by making sure the desired capacitance to be measured has been removed.

After removing the capacitance to be measured:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT (= 1).
- 3. Wait for a fixed delay of time, t.
- 4. Clear EDG1STAT.
- 5. Perform an A/D conversion.
- 6. Calculate the stray and A/D sample capacitances:

$$COFFSET = CSTRAY + CAD = (I \cdot t)/V$$

Where:

- I is known from the current source measurement step
- · t is a fixed delay
- V is measured by performing an A/D conversion

This measured value is then stored and used for calculations of time measurement or subtracted for capacitance measurement. For calibration, it is expected that the capacitance of CSTRAY + CAD is approximately known; CAD is approximately 4 pF.

An iterative process may be required to adjust the time, t, that the circuit is charged to obtain a reasonable voltage reading from the A/D Converter. The value of t may be determined by setting COFFSET to a theoretical value and solving for t. For example, if CSTRAY is theoretically calculated to be 11 pF, and V is expected to be 70% of VDD or 2.31V, t would be:

$$(4 \text{ pF} + 11 \text{ pF}) \cdot 2.31 \text{V}/0.55 \text{ }\mu\text{A}$$

or 63 µs.

See Example 27-3 for a typical routine for CTMU capacitance calibration.

### EXAMPLE 27-3: CAPACITANCE CALIBRATION ROUTINE

```
#include "pl8cxxx.h"
#define COUNT 25
 //@ 8MHz INTFRC = 62.5 us.
#define ETIME COUNT*2.5
 //time in uS
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define ADSCALE 1023
 //for unsigned conversion 10 sig bits
#define ADREF 3.3
 //Vdd connected to A/D Vr+
#define RCAL .027
 //R value is 4200000 (4.2M)
 //scaled so that result is in
 //1/100th of uA
int main(void)
{
 int i;
 int j = 0;
 //index for loop
 unsigned int Vread = 0;
 float CTMUISrc, CTMUCap, Vavg, VTot, Vcal;
//assume CTMU and A/D have been setup correctly
//see Example 25-1 for CTMU & A/D setup
setup();
CTMUCONHbits.CTMUEN = 1;
 //Enable the CTMU
 for(j=0;j<10;j++)</pre>
 {
 CTMUCONHbits.IDISSEN = 1;
 //drain charge on the circuit
 DELAY;
 //wait 125us
 CTMUCONHbits.IDISSEN = 0;
 //end drain of circuit
 CTMUCONLbits.EDG1STAT = 1;
 //Begin charging the circuit
 //using CTMU current source
 DELAY;
 //wait for 125us
 CTMUCONLbits.EDG1STAT = 0;
 //Stop charging circuit
 PIR1bits.ADIF = 0;
 //make sure A/D Int not set
 ADCON0bits.GO=1;
 //and begin A/D conv.
 while(!PIR1bits.ADIF);
 //Wait for A/D convert complete
 Vread = ADRES;
 //Get the value from the A/D
 PIR1bits.ADIF = 0;
 //Clear A/D Interrupt Flag
 VTot += Vread;
 //Add the reading to the total
 }
 Vavg = (float)(VTot/10.000);
 //Average of 10 readings
 Vcal = (float)(Vavg/ADSCALE*ADREF);
 //CTMUISrc is in 1/100ths of uA
 CTMUISrc = Vcal/RCAL;
 CTMUCap = (CTMUISrc*ETIME/Vcal)/100;
}
```

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## 27.5 Measuring Capacitance with the CTMU

There are two ways to measure capacitance with the CTMU. The absolute method measures the actual capacitance value. The relative method only measures for any change in the capacitance.

#### 27.5.1 ABSOLUTE CAPACITANCE MEASUREMENT

For absolute capacitance measurements, both the current and capacitance calibration steps found in **Section 27.4 "Calibrating the CTMU Module**" should be followed.

To perform these measurements:

- 1. Initialize the A/D Converter.
- 2. Initialize the CTMU.
- 3. Set EDG1STAT.
- 4. Wait for a fixed delay, T.
- 5. Clear EDG1STAT.
- 6. Perform an A/D conversion.
- 7. Calculate the total capacitance, CTOTAL = (I * T)/V, where:
  - I is known from the current source measurement step (Section 27.4.1 "Current Source Calibration")
  - · T is a fixed delay
  - V is measured by performing an A/D conversion
- Subtract the stray and A/D capacitance (COFFSET from Section 27.4.2 "Capacitance Calibration") from CTOTAL to determine the measured capacitance.

### 27.5.2 RELATIVE CHARGE MEASUREMENT

Not all applications require precise capacitance measurements. When detecting a valid press of a capacitance-based switch, only a relative change of capacitance needs to be detected.

In such an application, when the switch is open (or not touched), the total capacitance is the capacitance of the combination of the board traces, the A/D Converter and other elements. A larger voltage will be measured by the A/D Converter. When the switch is closed (or touched), the total capacitance is larger due to the addition of the capacitance of the human body to the above listed capacitances and a smaller voltage will be measured by the A/D Converter.

To detect capacitance changes simply:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT.
- 3. Wait for a fixed delay.
- 4. Clear EDG1STAT.
- 5. Perform an A/D conversion.

The voltage measured by performing the A/D conversion is an indication of the relative capacitance. In this case, no calibration of the current source or circuit capacitance measurement is needed. (For a sample software routine for a capacitive touch switch, see Example 27-4.)

#### EXAMPLE 27-4: ROUTINE FOR CAPACITIVE TOUCH SWITCH

```
#include "p18cxxx.h"
#define COUNT 500
 //@ 8MHz = 125uS.
#define DELAY for(i=0;i<COUNT;i++)</pre>
#define OPENSW 1000
 //Un-pressed switch value
#define TRIP 300
 //Difference between pressed
 //and un-pressed switch
#define HYST 65
 //amount to change
 //from pressed to un-pressed
#define PRESSED 1
#define UNPRESSED 0
int main(void)
ł
 unsigned int Vread;
 //storage for reading
 unsigned int switchState;
 int i;
 //assume CTMU and A/D have been setup correctly
 //see Example 25-1 for CTMU & A/D setup
 setup();
 CTMUCONHbits.CTMUEN = 1;
 //Enable the CTMU
 CTMUCONHbits.IDISSEN = 1;
 //drain charge on the circuit
 //wait 125us
 DELAY;
 CTMUCONHbits.IDISSEN = 0;
 //end drain of circuit
 CTMUCONLbits.EDG1STAT = 1;
 //Begin charging the circuit
 //using CTMU current source
 //wait for 125us
 DELAY;
 CTMUCONLbits.EDG1STAT = 0;
 //Stop charging circuit
 PIR1bits.ADIF = 0;
 //make sure A/D Int not set
 ADCON0bits.GO=1;
 //and begin A/D conv.
 while(!PIR1bits.ADIF);
 //Wait for A/D convert complete
 Vread = ADRES;
 //Get the value from the A/D
 if(Vread < OPENSW - TRIP)
 {
 switchState = PRESSED;
 else if(Vread > OPENSW - TRIP + HYST)
 {
 switchState = UNPRESSED;
 }
```

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# 27.6 Measuring Time with the CTMU Module

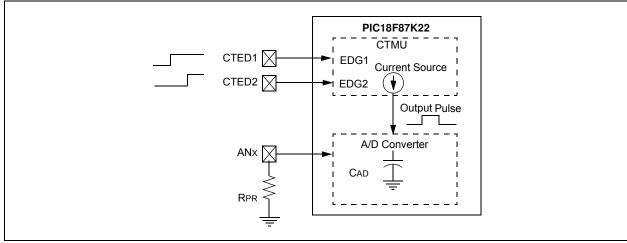
Time can be precisely measured after the ratio  $({\rm C/I})$  is measured from the current and capacitance calibration step. To do that:

- 1. Initialize the A/D Converter and the CTMU.
- 2. Set EDG1STAT.
- 3. Set EDG2STAT.
- 4. Perform an A/D conversion.
- 5. Calculate the time between edges as T = (C/I) * V, where:
  - I is calculated in the current calibration step (Section 27.4.1 "Current Source Calibration")
  - C is calculated in the capacitance calibration step (Section 27.4.2 "Capacitance Calibration")
  - V is measured by performing the A/D conversion

It is assumed that the time measured is small enough that the capacitance, COFFSET, provides a valid voltage to the A/D Converter. For the smallest time measurement, always set the A/D Channel Select register (AD1CHS) to an unused A/D channel, the corresponding pin for which is not connected to any circuit board trace. This minimizes added stray capacitance, keeping the total circuit capacitance close to that of the A/D Converter itself (25 pF).

To measure longer time intervals, an external capacitor may be connected to an A/D channel and that channel selected whenever making a time measurement.

# FIGURE 27-3: TYPICAL CONNECTIONS AND INTERNAL CONFIGURATION FOR TIME MEASUREMENT



#### 27.7 Creating a Delay with the CTMU Module

A unique feature on board the CTMU module is its ability to generate system clock independent output pulses, based on either an external voltage or an external capacitor value. When using an external voltage, this is accomplished using the CTDIN input pin as a trigger for the pulse delay. When using an external capacitor value, this is accomplished using the internal comparator voltage reference module and Comparator 2 input pin.The pulse is output onto the CTPLS pin. To enable this mode, set the TGEN bit.

See Figure 27-4 for an example circuit. When CTMUDS (ODCON3<0>) is cleared, the pulse delay is determined by the output of Comparator 2, and when it is set, the pulse delay is determined by the input of CTDIN. CDELAY is chosen by the user to determine the output pulse width on CTPLS. The pulse width is calculated by T = (CDELAY/I) * V, where I is known from the current source measurement step (Section 27.4.1 "Current Source Calibration") and V is the Internal Reference Voltage (CVREF).

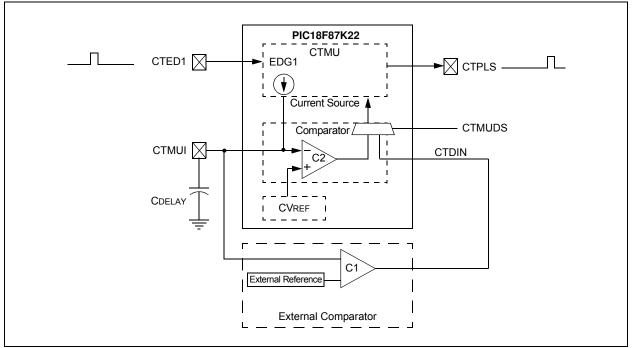
An example use of the external capacitor feature is interfacing with variable capacitive-based sensors, such as a humidity sensor. As the humidity varies, the pulse-width output on CTPLS will vary. An example use of the CTDIN feature is interfacing with a digital sensor. The CTPLS output pin can be connected to an input capture pin and the varying pulse width measured to determine the sensor's output in the application.

To use this feature:

- 1. If CTMUDS is cleared, initialize Comparator 2.
- 2. If CTMUDS is cleared, initialize the comparator voltage reference.
- 3. Initialize the CTMU and enable time delay generation by setting the TGEN bit.
- 4. Set EDG1STAT.

When CTMUDS is cleared, as soon as CDELAY charges to the value of the voltage reference trip point, an output pulse is generated on CTPLS. When CTMUDS is set, as soon as CTDIN is set, an output pulse is generated on CTPLS.

# FIGURE 27-4: TYPICAL CONNECTIONS AND INTERNAL CONFIGURATION FOR PULSE DELAY GENERATION



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# 27.8 Measuring Temperature with the CTMU Module

The CTMU, along with an internal diode, can be used to measure the temperature. The ADC can be connected to the internal diode and the CTMU module can source the current to the diode. The ADC reading will reflect the temperature. With the increase, the ADC readings will go low. This can be used for low-cost temperature measurement applications.

#### EXAMPLE 27-5: ROUTINE FOR TEMPERATURE MEASUREMENT USING INTERNAL DIODE

// Enable ADC and connect to Internal diode
// Right Justified
// Start conversion
<pre>// Read ADC results (inversely proportional to temperature)</pre>

Note: The temperature diode is not calibrated or standardized; the user must calibrate the diode to their application.

## 27.9 Operation During Sleep/Idle Modes

#### 27.9.1 SLEEP MODE

When the device enters any Sleep mode, the CTMU module current source is always disabled. If the CTMU is performing an operation that depends on the current source when Sleep mode is invoked, the operation may not terminate correctly. Capacitance and time measurements may return erroneous values.

#### 27.9.2 IDLE MODE

The behavior of the CTMU in Idle mode is determined by the CTMUSIDL bit (CTMUCONH<5>). If CTMUSIDL is cleared, the module will continue to operate in Idle mode. If CTMUSIDL is set, the module's current source is disabled when the device enters Idle mode. In this case, if the module is performing an operation when Idle mode is invoked, the results will be similar to those with Sleep mode.

## 27.10 Effects of a Reset on CTMU

Upon Reset, all registers of the CTMU are cleared. This disables the CTMU module, turns off its current source and returns all configuration options to their default settings. The module needs to be re-initialized following any Reset.

If the CTMU is in the process of taking a measurement at the time of Reset, the measurement will be lost. A partial charge may exist on the circuit that was being measured, which should be properly discharged before the CTMU makes subsequent attempts to make a measurement. The circuit is discharged by setting and clearing the IDISSEN bit (CTMUCONH<1>) while the A/D Converter is connected to the appropriate channel.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
CTMUCONH	CTMUEN	_	CTMUSIDL	TGEN	EDGEN	EDGSEQEN	IDISSEN	CTTRIG
CTMUCONL	EDG2POL	EDG2SEL1	EDG2SEL0	EDG1POL	EDG1SEL1	EDG1SEL0	EDG2STAT	EDG1STAT
CTMUICON	ITRIM5	ITRIM4	ITRIM3	ITRIM2	ITRIM1	ITRIM0	IRNG1	IRNG0
PIR3	TMR5GIF	-	RC2IF	TX2IF	CTMUIF	CCP2IF	CCP1IF	RTCCIF
PIE3	TMR5GIE		RC2IE	TX2IE	CTMUIE	CCP2IE	CCP1IE	RTCCIE
IPR3	TMR5GIP	_	RC2IP	TX2IP	CTMUIP	CCP2IP	CCP1IP	RTCCIP

#### TABLE 27-1: REGISTERS ASSOCIATED WITH CTMU MODULE

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used during ECCP operation.

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NOTES:

# 28.0 SPECIAL FEATURES OF THE CPU

The PIC18F87K22 family of devices includes several features intended to maximize reliability and minimize cost through elimination of external components. These include:

- Oscillator Selection
- Resets:
  - Power-on Reset (POR)
  - Power-up Timer (PWRT)
  - Oscillator Start-up Timer (OST)
- Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT) and On-chip Regulator
- · Fail-Safe Clock Monitor
- Two-Speed Start-up
- Code Protection
- ID Locations
- In-Circuit Serial Programming[™] (ICSP[™])

The oscillator can be configured for the application depending on frequency, power, accuracy and cost. All of the options are discussed in detail in **Section 3.0 "Oscillator Configurations"**.

A complete discussion of device Resets and interrupts is available in previous sections of this data sheet.

In addition to their Power-up and Oscillator Start-up Timers provided for Resets, the PIC18F87K22 family of devices has a Watchdog Timer, which is either permanently enabled via the Configuration bits or software controlled (if configured as disabled).

The inclusion of an internal RC oscillator (LF-INTOSC) also provides the additional benefits of a Fail-Safe Clock Monitor (FSCM) and Two-Speed Start-up. FSCM provides for background monitoring of the peripheral clock and automatic switchover in the event of its failure. Two-Speed Start-up enables code to be executed almost immediately on start-up, while the primary clock source completes its start-up delays.

All of these features are enabled and configured by setting the appropriate Configuration register bits.

# 28.1 Configuration Bits

The Configuration bits can be programmed (read as '0') or left unprogrammed (read as '1') to select various device configurations. These bits are mapped starting at program memory location, 300000h.

The user will note that address, 300000h, is beyond the user program memory space. In fact, it belongs to the configuration memory space (300000h-3FFFFh), which can only be accessed using table reads and table writes.

Software programming of the Configuration registers is done in a manner similar to programming the Flash memory. The WR bit in the EECON1 register starts a self-timed write to the Configuration register. In normal Operation mode, a TBLWT instruction, with the TBLPTR pointing to the Configuration register, sets up the address and the data for the Configuration register write. Setting the WR bit starts a long write to the Configuration register. The Configuration registers are written a byte at a time. To write or erase a configuration cell, a TBLWT instruction can write a '1' or a '0' into the cell. For additional details on Flash programming, refer to Section 7.5 "Writing to Flash Program Memory".

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File	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Default/ Unprogrammed Value
300000h	CONFIG1L	_	XINST	_	SOSCSEL1	SOSCSEL0	INTOSCSEL	_	RETEN	-1-1 11
300001h	CONFIG1H	IESO	FCMEN	_	PLLCFG	FOSC3	FOSC2	FOSC1	FOSC0	00-0 1000
300002h	CONFIG2L	-	BORPWR1	BORWPR0	BORV1	BORV0	BOREN1	BOREN0	PWRTEN	-111 1111
300003h	CONFIG2H	_	WDTPS4	WDTPS3	WDTPS2	WDTPS1	WDTPS0	WDTEN1	WDTEN0	-111 1111
300004h	CONFIG3L	WAIT ⁽²⁾	BW ⁽²⁾	ABW1 ⁽²⁾	ABW0 ⁽²⁾	EASHFT ⁽²⁾	—	_	RTCOSC	1
300005h	CONFIG3H	MCLRE	_	_	_	MSSPMSK	—	ECCPMX ⁽²⁾	CCP2MX	1 1-11
300006h	CONFIG4L	DEBUG	-		BBSIZ0		—		STVREN	111
300008h	CONFIG5L	CP7 ⁽¹⁾	CP6 ⁽¹⁾	CP5 ⁽¹⁾	CP4 ⁽¹⁾	CP3	CP2	CP1	CP0	1111 1111
300009h	CONFIG5H	CPD	CPB	_	_	_	—	_	_	11
30000Ah	CONFIG6L	WRT7 ⁽¹⁾	WRT6 ⁽¹⁾	WRT5 ⁽¹⁾	WRT4 ⁽¹⁾	WRT3	WRT2	WRT1	WRT0	1111 1111
30000Bh	CONFIG6H	WRTD	WRTB	WRTC	_	_	—	_	_	111
30000Ch	CONFIG7L	EBTR7 ⁽¹⁾	EBTR6 ⁽¹⁾	EBTR5 ⁽¹⁾	EBTR4 ⁽¹⁾	EBTR3	EBTR2	EBTR1	EBTR0	1111 1111
30000Dh	CONFIG7H	_	EBTRB	_	_	_	_	_	_	-1
3FFFFEh	DEVID1 ⁽³⁾	DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0	xxxx xxxx
3FFFFFh	DEVID2 ⁽³⁾	DEV10	DEV9	DEV8	DEV7	DEV6	DEV5	DEV4	DEV3	xxxx xxxx

#### TABLE 28-1: CONFIGURATION BITS AND DEVICE IDs

Legend:x = unknown, u = unchanged, - = unimplemented, q = value depends on condition. Shaded cells are unimplemented, read as '0'.Note1:Implemented only on the PIC18F67K22 and PIC18F87K22 devices.

2: Implemented only on the 80-pin devices (PIC18F8XK22).

3: See Register 28-14 for DEVID1 values. DEVID registers are read-only and cannot be programmed by the user.

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# REGISTER 28-1: CONFIG1L: CONFIGURATION REGISTER 1 LOW (BYTE ADDRESS 300000h)

U-0	R/P-1	U-0	R/P-1	R/P-1	R/P-1	U-0	R/P-1
—	XINST	_	SOSCSEL1	SOSCSEL0	INTOSCSEL		RETEN
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6	XINST: Extended Instruction Set Enable bit
	<ul> <li>1 = Instruction set extension and Indexed Addressing mode are enabled</li> <li>0 = Instruction set extension and Indexed Addressing mode are disabled (Legacy mode)</li> </ul>
bit 5	Unimplemented: Read as '0'
bit 4-3	SOSCSEL<1:0>: SOSC Power Selection and Mode Configuration bits
	<ul> <li>11 = High-power SOSC circuit is selected</li> <li>10 = Digital (SCLKI) mode; I/O port functionality of RC0 and RC1 is enabled</li> <li>01 = Low-power SOSC circuit is selected</li> <li>00 = Reserved</li> </ul>
bit 2	INTOSCSEL: LF-INTOSC Low-power Enable bit
	<ul> <li>1 = LF-INTOSC is in High-Power mode during Sleep</li> <li>0 = LF-INTOSC is in Low-Power mode during Sleep</li> </ul>
bit 1	Unimplemented: Read as '0'
bit 0	RETEN: VREG Sleep Enable bit
	<ul> <li>1 = Regulator power while in Sleep mode is controlled by VREGSLP (WDTCON&lt;7&gt;)</li> <li>0 = Regulator power while in Sleep mode is controlled by SRETEN (WDTCON&lt;4&gt;). Ultra low-power</li> </ul>

 Regulator power while in Sleep mode is controlled by SRETEN (WDTCON<4>). Ultra low-power regulator is enabled.

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					•		•					
R/P-0	) R/P-0	U-0	U-0	R/P-1	R/P-0	R/P-0	R/P-0					
IESO	FCMEN	—	PLLCFG ⁽¹⁾	FOSC3 ⁽²⁾	FOSC2 ⁽²⁾	FOSC1 ⁽²⁾	FOSC0 ⁽²⁾					
bit 7							bit (					
Legend:		P = Progran	nmable bit									
R = Read	able bit	W = Writable	e bit	U = Unimpler	nented bit, read	d as '0'						
-n = Value	e at POR	'1' = Bit is se	et	'0' = Bit is cle	ared	x = Bit is unkr	nown					
bit 7		al/External Os ed Start-up is	cillator Switchov enabled	ver bit								
		ed Start-up is										
bit 6	FCMEN: Fail	-Safe Clock N	lonitor Enable b	bit								
		Clock Monito Clock Monito										
bit 5	Unimplemer	nted: Read as	'0'									
bit 4	PLLCFG: 4x	PLLCFG: 4x PLL Enable bit ⁽¹⁾										
		r is multiplied										
bit 3-0	FOSC<3:0>:	FOSC<3:0>: Oscillator Selection bits ⁽²⁾										
	1100 = EC1I 1011 = EC2, 1010 = EC2I 1001 = INTIC 1000 = INTIC 0111 = RC, e 0110 = RCIC 0101 = EC3, 0100 = EC3I 0011 = HS1,	O, EC oscillat EC oscillator O, EC oscillat D1, internal RC D2, internal RC external RC os O, external RC EC oscillator O, EC oscillator HS oscillator scillator		T function on F er, 160 kHz-16 T function on F CLKOUT funct CKLOUT funct MHz-64 MHz T function on F er, 4 MHz-16 M	MHz) RA6 (medium p stion on RA6 ion on RA6 RA6 (high pow IHz)	bower, DC-160	kHz-16 MHz					
Note 1: 2:	Not valid for the IN INTIO + PLL can b			bit (OSCTUN	F<6>) Other P	II modes can b	e enabled b					

#### REGISTER 28-2: CONFIG1H: CONFIGURATION REGISTER 1 HIGH (BYTE ADDRESS 300001h)

2: INTIO + PLL can be enabled only by the PLLEN bit (OSCTUNE<6>). Other PLL modes can be enabled by either the PLLEN bit or the PLLCFG (CONFIG1H<4>) bit.

## REGISTER 28-3: CONFIG2L: CONFIGURATION REGISTER 2 LOW (BYTE ADDRESS 300002h)

U-0	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1
_	BORPWR1 ⁽¹⁾	BORPWR0 ⁽¹⁾	BORV1 ⁽¹⁾	BORV0 ⁽¹⁾	BOREN1 ⁽²⁾	BOREN0 ⁽²⁾	PWRTEN ⁽²⁾
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit, rea	id as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-5	BORPWR<1:0>: BORMV Power-Level bits ⁽¹⁾
	11 = ZPBORVMV instead of BORMV is selected
	10 = BORMV is set to a high-power level
	<ul> <li>01 = BORMV is set to a medium power level</li> <li>00 = BORMV is set to a low-power level</li> </ul>
h:+ 4 0	
bit 4-3	BORV<1:0>: Brown-out Reset Voltage bits ⁽¹⁾
	11 = VBORMV is set to 1.8V
	10 = VBORMV is set to 2.0V 01 = VBORMV is set to 2.7V
	0.1 = VBORMV is set to 2.7V 0.0 = VBORMV is set to 3.0V
bit 2-1	BOREN<1:0>: Brown-out Reset Enable bits ⁽²⁾
	11 = Brown-out Reset is enabled in hardware only (SBOREN is disabled)
	10 = Brown-out Reset is enabled in hardware only and disabled in Sleep mode (SBOREN is disabled)
	01 = Brown-out Reset is enabled and controlled by software (SBOREN is enabled)
	00 = Brown-out Reset is disabled in hardware and software
bit 0	<b>PWRTEN:</b> Power-up Timer Enable bit ⁽²⁾
	1 = PWRT is disabled
	0 = PWRT is enabled
Note 1:	For the specifications, see Section 31.1 "DC Characteristics: Supply Voltage PIC18F87K22 Family (Industrial/Extended)".

2: The Power-up Timer is decoupled from Brown-out Reset, allowing these features to be independently controlled.

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U-0	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1	R/P-1					
_	WDTPS4	WDTPS3	WDTPS2	WDTPS1	WDTPS0	WDTEN1	WDTEN0					
bit 7							bit					
Legend:		P = Programn	nable bit									
R = Readal	ble bit	W = Writable		U = Unimpler	nented bit, read	l as '0'						
-n = Value a	at POR	'1' = Bit is set		ʻ0' = Bit is cle		x = Bit is unkr	nown					
bit 7	Unimplemen	ted: Read as '	0'									
bit 6-2	Unimplemented: Read as '0' WDTPS<4:0>: Watchdog Timer Postscale Select bits											
		1 = Reserved										
		048,576 (4,194	.304s)									
		24,288 (2,097.1	,									
		62,144 (1,048.5										
	10001 = 1:13	31,072 (524.28	Bs)									
		10000 = 1:65,536 (262.144s)										
		01111 = 1:32,768 (131.072s)										
		110 = 1:16,384 (65.536s)										
		1:8,192 (32.768s)										
		4,096 (16.384s)										
	01011 = 1:2, 01010 = 1:1,											
	01010 - 1.1, 01001 = 1:51											
	01001 = 1.3	. ,										
	00111 = 1:12											
	00110 = 1:64	• •										
	00101 = 1:32	. ,										
	00100 = 1:16											
	00011 <b>= 1:8</b>	00011 = 1:8 (32  ms)										
	00010 <b>= 1</b> :4	00010 = 1.4 (16  ms)										
		00001 = 1:2 (8  ms)										
	00000 = 1:1	(4 ms)										
bit 1-0	WDTEN<1:0	: Watchdog Ti	mer Enable bi	ts								
	11 = WDT is	s enabled in ha	rdware; SWD]	TEN bit is disal	bled							
		s controlled by										
		•	while the dev	ice is active a	nd disabled in	Sleep mode; S	WDTEN bit i					
	disable	ed										
		s disabled in ha										

## REGISTER 28-4: CONFIG2H: CONFIGURATION REGISTER 2 HIGH (BYTE ADDRESS 300003h)

## REGISTER 28-5: CONFIG3L: CONFIGURATION REGISTER 3 LOW (BYTE ADDRESS 300004h)

U-1	U-1	U-1	U-1	U-1	U-0	U-0	R/P-1
WAIT ⁽¹⁾	BW ⁽¹⁾	ABW1 ⁽¹⁾	ABW0 ⁽¹⁾	EASHFT ⁽¹⁾	—		RTCOSC
bit 7							bit 0

Legend:	P = Programmable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	WAIT: External Bus Wait Enable bit ⁽¹⁾ 1 = Wait states on the external bus are disabled = Wait states on the external bus are enabled and selected by MEMCON <5:4>
bit 6	<ul> <li>0 = Wait states on the external bus are enabled and selected by MEMCON &lt;5:4&gt;</li> <li>BW: Data Bus Width Select bit⁽¹⁾</li> <li>1 = 16-Bit Data Width modes</li> <li>0 = 8-Bit Data Width modes</li> </ul>
bit 5-4	ABW<1:0>: External Memory Bus Configuration bits ⁽¹⁾ 11 = 8-Bit Address mode (Microcontroller mode) 10 = 12-Bit Address mode 01 = 16-Bit Address mode 00 = 20-Bit Address mode
bit 3	<b>EASHFT:</b> External Address Bus Shift Enable bit ⁽¹⁾ 1 = Address shifting is enabled; external address is shifted to start at 000000h 0 = Address shifting is disabled; external address bus reflects the PC value
bit 2-1	Unimplemented: Read as '0'
bit 0	<b>RTCOSC:</b> RTCC Reference Clock Select bit 1 = RTCC uses SOSC as the reference clock 0 = RTCC uses LF-INTOSC as the reference clock

Note 1: Unimplemented on 64-pin devices (PIC18F6XK22), read as '0'.

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R/P-1	U-0	U-0	U-0	R/P-1	U-0	R/P-1	R/P-1
MCLRE	_	—	_	MSSPMSK	—	ECCPMX ⁽¹⁾	CCP2MX
bit 7							bit 0
Legend:		P = Programm	able bit				
R = Readable b	oit	W = Writable I	oit	U = Unimpler	nented bit, read	l as '0'	
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkn	iown
bit 7	$1 = \overline{\text{MCLR}}$ pin	.R Pin Enable t is enabled; R0 t pin is enabled	3 <u>5 input</u> pin is				
bit 6-4	Unimplement	ted: Read as 'd	)'				
bit 3	1 = 7-Bit Addr	/ISSP V3 7-Bit ress Masking m	ode is enable	ed	able bit		
		ess Masking m		ed			
bit 2	•	ted: Read as '(	)'				
bit 1	1 = - ECCP1 (I	,			5, CCP6 onto R 3, CCP8 onto R		
	- ECCP1 (I	P1B/P1C) is mi P3B/P3C) is mi	ultiplexed ont ultiplexed ont	o RH7 and RH o RH5 and RH	6, CCP6 onto F 4, CCP8 onto F	RH7, and CCP7 RH5, and CCP9	⁽²⁾ onto RH6 ⁽²⁾ onto RH4
bit 0	CCP2MX: EC	CP2 MUX bit					
	0 = ECCP2 is	s multiplexed w s multiplexed v icrocontroller n	/ith RB3 in E	xtended Micro	controller mode	e; ECCP2 is mu	ultiplexed with
Note 1: Unin	nplemented on	64-pin devices	s (PIC18F6XI	<b>&lt;22), read as</b> '(	ט׳.		

#### REGISTER 28-6: CONFIG3H: CONFIGURATION REGISTER 3 HIGH (BYTE ADDRESS 300005h)

2: Not implemented on 32K devices (PIC18F65K22 and PIC18F85K22).

# REGISTER 28-7: CONFIG4L: CONFIGURATION REGISTER 4 LOW (BYTE ADDRESS 300006h)

R/P-1	U-0	U-0	R/P-0	U-0	U-0	U-0	R/P-1
DEBUG	_	_	BBSIZ0	-	—	_	STVREN
bit 7							bit 0

Legend:         P = Programmable bit				
R = Readable bit	W = Writable bit	U = Unimplemented bit	t, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

bit 7	<b>DEBUG:</b> Background Debugger Enable bit 1 = Background debugger is disabled, RB6 and RB7 are configured as general purpose I/O pins
	<ul> <li>0 = Background debugger is enabled, RB6 and RB7 are dedicated to In-Circuit Debug</li> </ul>
bit 6-5	Unimplemented: Read as '0'
bit 4	BBSIZ0: Boot Block Size Select bit
	1 = 2 kW boot block size 0 = 1 kW boot block size
bit 3-1	Unimplemented: Read as '0'
bit 0	STVREN: Stack Full/Underflow Reset Enable bit
	<ul> <li>1 = Stack full/underflow will cause Reset</li> <li>0 = Stack full/underflow will not cause Reset</li> </ul>

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R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1
CP7 ⁽¹⁾	CP6 ⁽¹⁾	CP5 ⁽¹⁾	CP4 ⁽¹⁾	CP3	CP2	CP1	CP0
bit 7							bit 0
Legend:		C = Clearable	bit				
R = Readabl	le bit	W = Writable	bit	U = Unimplei	mented bit, rea	ad as '0'	
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	CP7: Code	Protection bit ⁽¹⁾					
		is not code-prote is code-protecte					
bit 6	CP6: Code	Protection bit ⁽¹⁾					
		is not code-prote					
		is code-protecte	d ⁽²⁾				
bit 5		Protection bit ⁽¹⁾	(0)				
		is not code-prote is code-protecte					
bit 4		Protection bit ⁽¹⁾	u.,				
		is not code-prote	ected ⁽²⁾				
		is code-protecte					
bit 3	CP3: Code	Protection bit					
		is not code-prote					
		is code-protecte	d ⁽²⁾				
bit 2		Protection bit	(0)				
		is not code-prote is code-protecte					
bit 1		Protection bit	u				
bit i		is not code-prote	ected ⁽²⁾				
		is code-protecte					
bit 0	CP0: Code	Protection bit					
		is not code-prote is code-protecte					

#### REGISTER 28-8: CONFIG5L: CONFIGURATION REGISTER 5 LOW (BYTE ADDRESS 300008h)

- Note 1: This bit is available only on PIC18F67K22 and PIC18F87K22 devices.
  - 2: For the memory size of the blocks, see Figure 28-6.

#### REGISTER 28-9: CONFIG5H: CONFIGURATION REGISTER 5 HIGH (BYTE ADDRESS 300009h)

R/C-1	R/C-1	U-0	U-0	U-0	U-0	U-0	U-0		
CPD	CPB	—	_		—	—	—		
bit 7							bit 0		
Legend:		C = Clearable	bit						
R = Readable bit W = Writable bit				U = Unimplemented bit, read as '0'					
-n = Value at	POR	'1' = Bit is set		'0' = Bit is cleared x = Bit is unknown					
bit 7	CPD: Data E	EPROM Code I	Protection bit						
	1 = Data EEI	PROM is not co	de-protected						
	0 = Data EEI	PROM is code-p	protected						

- bit 6 **CPB:** Boot Block Code Protection bit 1 = Boot block is not code-protected⁽¹⁾
- 0 = Boot block is code-protected⁽¹⁾ bit 5-0 **Unimplemented:** Read as '0'

Note 1: For the memory size of the blocks, see Figure 28-6. The boot block size changes with BBSIZ0.

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R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1
WRT7 ⁽¹⁾	WRT6 ⁽¹⁾	WRT5 ⁽¹⁾	WRT4 ⁽¹⁾	WRT3	WRT2	WRT1	WRT0
bit 7			-		•		bit (
Legend:		C = Clearable	bit				
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, rea	ad as '0'	
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown
bit 7	WRT7: Write	Protection bit ⁽¹	)				
		s not write-prote s write-protecte					
bit 6	WRT6: Write	Protection bit ⁽¹	)				
		s not write-prote s write-protecte					
bit 5		Protection bit ⁽¹					
	1 = Block 5 is 0 = Block 5 is	s not write-prote s write-protecte	ected ⁽²⁾ d ⁽²⁾				
bit 4	WRT4: Write	Protection bit ⁽¹	)				
		not write-prote write-protecte					
bit 3	WRT3: Write	Protection bit					
		s not write-prote s write-protecte					
bit 2	WRT2: Write	Protection bit					
		s not write-prote s write-protecte					
bit 1	WRT1: Write	Protection bit					
		s not write-prote s write-protecte					
bit 0	WRT0: Write	Protection bit					
		s not write-prote s write-protecte					

#### REGISTER 28-10: CONFIG6L: CONFIGURATION REGISTER 6 LOW (BYTE ADDRESS 30000Ah)

- Note 1: This bit is available only on PIC18F67K22 and PIC18F87K22 devices.
  - 2: For the memory size of the blocks, see Figure 28-6.

#### REGISTER 28-11: CONFIG6H: CONFIGURATION REGISTER 6 HIGH (BYTE ADDRESS 30000Bh)

					•		
R/C-1	R/C-1	R-1	U-0	U-0	U-0	U-0	U-0
WRTD	WRTB	WRTC ⁽¹⁾	—		—	—	_
bit 7							bit 0
Legend:		C = Clearable	bit				
R = Readable	bit	W = Writable	bit	U = Unimplei	mented bit, read	l as '0'	
-n = Value at POR '1' = Bit is set				'0' = Bit is cleared x = Bit is unknown			
bit 7	WRTD: Data	EEPROM Write	e Protection b	it			
	1 = Data EEF	PROM is not wri	ite-protected				
	0 <b>= Data EE</b>	PROM is write-p	protected				
bit 6	WRTB: Boot	Block Write Pro	otection bit				
		ck is not write-pr					
	0 = Boot bloc	ck is write-prote	cted ⁽²⁾				
bit 5	WRTC: Conf	iguration Regist	er Write Prote	ection bit ⁽¹⁾			
	1 = Configura	ation registers a	re not write-n	rotected ⁽²⁾			

Configuration registers are not write-protected⁽²⁾
 Configuration registers are write-protected⁽²⁾

bit 4-0 Unimplemented: Read as '0'

**Note 1:** This bit is read-only in normal Execution mode; it can be written only in Program mode.

2: For the memory size of the blocks, see Figure 28-6.

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R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1
EBTR7 ^(1,3)	EBTR6 ^(1,3)	EBTR5 ^(1,3)	EBTR4 ^(1,3)	EBTR3 ⁽³⁾	EBTR2 ⁽³⁾	EBTR1 ⁽³⁾	EBTR0 ⁽³⁾
bit 7							bit 0
Legend:		C = Clearable	hit				
R = Readabl	e hit	W = Writable		II = I Inimpler	mented bit, read	l as 'O'	
-n = Value at		'1' = Bit is set		'0' = Bit is cle		x = Bit is unkr	
							IOWIT
bit 7	EBTR7: Table	e Read Protect	ion bit ^(1,3)				
		not protected protected from			ı other blocks ⁽²⁾ ıer blocks ⁽²⁾		
bit 6	EBTR6: Table	e Read Protect	ion bit ^(1,3)				
					$1 \text{ other blocks}^{(2)}$		
		protected from		executed in oth	ier blocks(2)		
bit 5		e Read Protect					
		not protected protected fron			ı other blocks ⁽²⁾ ıer blocks ⁽²⁾		
bit 4		e Read Protect					
		not protected protected from			other blocks ⁽²⁾		
bit 3		e Read Protect					
bit o				ds executed in	other blocks ⁽²⁾		
		protected from					
bit 2	EBTR2: Table	e Read Protect	ion bit ⁽³⁾				
		not protected protected from			ı other blocks ⁽²⁾ ıer blocks ⁽²⁾		
bit 1		e Read Protect					
	1 = Block 1 is		from table read		other blocks ⁽²⁾		
bit 0		e Read Protect					
Situ	1 = Block 0 is		from table rea		n other blocks ⁽² her blocks ⁽²⁾	)	
Note 1: T	his bit is available	e only on PIC1	8F67K22 and I	PIC18F87K22	devices.		
<b>2</b> : Fo	or the memory si	ze of the block	s. see Figure 2	28-6			

#### REGISTER 28-12: CONFIG7L: CONFIGURATION REGISTER 7 LOW (BYTE ADDRESS 30000Ch)

- 2: For the memory size of the blocks, see Figure 28-6
- 3: Enable the corresponding CPx bit to protect the block from external read operations.

#### REGISTER 28-13: CONFIG7H: CONFIGURATION REGISTER 7 HIGH (BYTE ADDRESS 30000Dh)

U-0	R/C-1	U-0	U-0	U-0	U-0	U-0	U-0
_	EBTRB ⁽²⁾	—	—	—	—	—	—
bit 7							bit 0

Legend:	C = Clearable bit		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	l as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 Unimplemented: Read as '0	0'
---------------------------------	----

bit 6 **EBTRB:** Boot Block Table Read Protection bit⁽²⁾

1 = Boot block is not protected from table reads executed in other blocks⁽¹⁾

0 = Boot block is protected from table reads executed in other blocks⁽¹⁾

bit 5-0 Unimplemented: Read as '0'

Note 1: For the memory size of the blocks, see Figure 28-6.

2: Enable the corresponding CPx bit to protect the block from external read operations.

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#### REGISTER 28-14: DEVID1: DEVICE ID REGISTER 1 FOR THE PIC18F87K22 FAMILY

R	R	R	R	R	R	R	R	
DEV2	DEV1	DEV0	REV4	REV3	REV2	REV1	REV0	
bit 7							bit 0	
Legend:								
R = Readable bit W = Writable bit			U = Unimpler	nented bit, read	as '0'			
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cleared x = Bit is unknown				
bit 7-5 <b>DEV&lt;2:0&gt;:</b> Device ID bits <u>Devices with DEV&lt;10:3&gt; of '0101_0010'</u> 010 = PIC18F65K22 000 = PIC18F66K22 101 = PIC18F85K22 011 = PIC18F86K22 <u>Devices with DEV&lt;10:3&gt; of '0101_0001'</u> 000 = PIC18F67K22 010 = PIC18F87K22 bit 4-0 <b>REV&lt;4:0&gt;:</b> Revision ID bits								
L:1 4 0	REV<4:0>: Revision ID bits							

These bits are used to indicate the device revision.

#### REGISTER 28-15: DEVID2: DEVICE ID REGISTER 2 FOR THE PIC18F87K22 FAMILY

R	R	R	R	R	R	R	R
DEV10 ⁽¹⁾	DEV9 ⁽¹⁾	DEV8 ⁽¹⁾	DEV7 ⁽¹⁾	DEV6 ⁽¹⁾	DEV5 ⁽¹⁾	DEV4 ⁽¹⁾	DEV3 ⁽¹⁾
bit 7							bit 0
Legend:							
R = Readable bit W = Writable bit			U = Unimplemented bit, read as '0'				
-n = Value at POR '1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown		

bit 7-0 **DEV<10:3>:** Device ID bits⁽¹⁾ These bits are used with the DEV<2:0> bits in the Device ID Register 1 to identify the part number. 0101 0010 = PIC18F65K22, PIC18F66K22, PIC18F85K22 and PIC18F86K22 0101 0001 = PIC18F67K22 and PIC18F87K22

**Note 1:** These values for DEV<10:3> may be shared with other devices. The specific device is always identified by using the entire DEV<10:0> bit sequence.

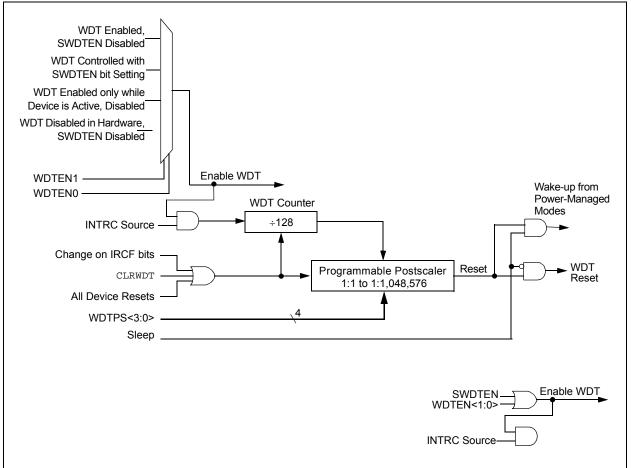
# 28.2 Watchdog Timer (WDT)

For the PIC18F87K22 family of devices, the WDT is driven by the LF-INTOSC source. When the WDT is enabled, the clock source is also enabled. The nominal WDT period is 4 ms and has the same stability as the LF-INTOSC oscillator.

The 4 ms period of the WDT is multiplied by a 16-bit postscaler. Any output of the WDT postscaler is selected by a multiplexer, controlled by bits in Configuration Register 2H. Available periods range from 4 ms to 4,194 seconds (about one hour). The WDT and postscaler are cleared when any of the following events occur: a SLEEP or CLRWDT instruction is executed, the IRCF bits (OSCCON<6:4>) are changed or a clock failure has occurred.

The WDT can be operated in one of four modes as determined by the WDTEN<1:0> (CONFIG2H<1:0> bits. The four modes are:

- WDT Enabled
- · WDT Disabled
- WDT under Software Control, SWDTEN (WDTCON<0>)
- WDT
  - Enabled during normal operation
  - Disabled during Sleep
  - Note 1: The CLRWDT and SLEEP instructions clear the WDT and postscaler counts when executed.
    - Changing the setting of the IRCF bits (OSCCON<6:4>) clears the WDT and postscaler counts.
    - **3:** When a CLRWDT instruction is executed, the postscaler count will be cleared.



#### FIGURE 28-1: WDT BLOCK DIAGRAM

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#### 28.2.1 CONTROL REGISTER

Register 28-16 shows the WDTCON register. This is a readable and writable register which contains a control bit that allows software to override the WDT Enable Configuration bit, but only if the Configuration bit has disabled the WDT.

#### REGISTER 28-16: WDTCON: WATCHDOG TIMER CONTROL REGISTER

R/W-0	U-0	R-x	R/W-0	U-0	R/W-0	R/W-0	R/W-0
REGSLF	D _	ULPLVL	SRETEN ⁽²⁾		ULPEN	ULPSINK	SWDTEN ⁽¹⁾
bit 7							bit (
Legend:							
R = Reada	ble bit	W = Writable	bit	U = Unimple	mented bit, read	d as '0'	
-n = Value	at POR	'1' = Bit is se	t	'0' = Bit is cle	eared	x = Bit is unk	nown
bit 7	REGSLP: F	Regulator Voltag	e Sleep Enable	e bit			
		or goes into Lov or stays in norm					
bit 6	Unimpleme	ented: Read as	ʻ0'				
bit 5	ULPLVL: U	Itra Low-Power	Wake-up Outp	ut bit			
	Not valid un	less ULPEN = 1	L				
	1 = Voltage	on RA0 pin > ~	· 0.5V				
	0 = Voltage	e on RA0 pin < ~	[,] 0.5V.				
bit 4	SRETEN: F	Regulator Voltage	e Sleep Disabl	e bit ⁽²⁾			
	1 = If RETE mode ii	EN (CONFIG1L< n Sleep	:0>) = 0 and the	e regulator is e	nabled, the dev	ice goes into Ul	tra Low-Powe
	0 = The re	gulator is on w ed by REGSLP	hen the device	e's Sleep mo	de is enabled a	and the Low-F	ower mode is
bit 3	Unimpleme	ented: Read as	'0'				
bit 2	ULPEN: Ult	ra Low-Power V	Vake-up Modul	e Enable bit			
		ow-Power Wake ow-Power Wake			VL bit indicate	s the comparate	or output
bit 1	ULPSINK:	Ultra Low-Powe	r Wake-up Cur	rent Sink Enal	ole bit		
		less ULPEN = 1	•				
		ow-Power Wake ow-Power Wake					
bit 0	SWDTEN: S	Software Contro	Iled Watchdog	Timer Enable	bit ⁽¹⁾		
		og Timer is on					
		og Timer is off					
Note 1	This hit has no e	ffaat if the Confi	auration hita M		are enabled		

**Note 1:** This bit has no effect if the Configuration bits, WDTEN<1:0>, are enabled.

**2:** This bit is available only when ENVREG = 1 and  $\overline{\text{RETEN}}$  = 0.

TABLE 28-2: SUMMARY OF WATCHDOG TIMER REGISTERS	3
-------------------------------------------------	---

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
RCON	IPEN	SBOREN	CM	RI	TO	PD	POR	BOR
WDTCON	REGSLP	_	ULPLVL	SRETEN		ULPEN	ULPSINK	SWDTEN

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by the Watchdog Timer.

## 28.3 On-Chip Voltage Regulator

All of the PIC18F87K22 family devices power their core digital logic at a nominal 3.3V. For designs that are required to operate at a higher typical voltage, such as 5V, all family devices incorporate two on-chip regulators that allow the device to run its core logic from VDD. Those regulators are:

- Normal On-Chip Regulator
- Ultra Low-Power On-Chip Regulator

The hardware configuration of these regulators is the same and is explained in Section 28.3.1 "Regulator Enable/Disable By Hardware". The regulators' only differences relate to when the device enters Sleep, as explained in Section 28.3.2 "Operation of Regulator in Sleep".

#### 28.3.1 REGULATOR ENABLE/DISABLE BY HARDWARE

The regulator can be enabled or disabled only by hardware. The regulator is controlled by the ENVREG pin and the VDDCORE/VCAP pin.

#### 28.3.1.1 Regulator Enable Mode

Tying VDD to the pin enables the regulator, which in turn, provides power to the core from the other VDD pins.

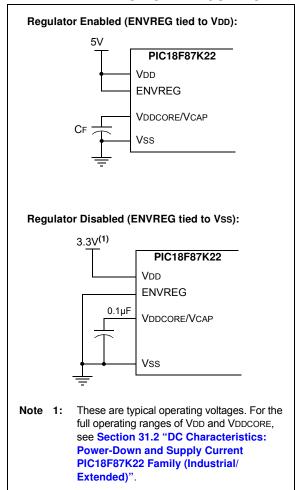
When the regulator is enabled, a low-ESR filter capacitor must be connected to the VDDCORE/VCAP pin (see Figure 28-2). This helps maintain the regulator's stability. The recommended value for the filter capacitor is given in Section 31.2 "DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/Extended)".

#### 28.3.1.2 Regulator Disable Mode

If the regulator is disabled by connecting Vss to the ENVREG pin, the power to the core is supplied directly by VDD. The voltage levels for VDD must not exceed the specified VDDCORE levels. In Regulator Disabled mode, a 0.1  $\mu$ F capacitor should be connected to the VDDCORE/VCAP pin (see Figure 28-2).

When the regulator is being used, the overall voltage budget is very tight. The regulator should operate the device down to 1.8V. When VDD drops below 3.3V, the regulator no longer regulates, but the output voltage follows the input until VDD reaches 1.8V. Below this voltage, the output of the regulator output may drop to 0V.

# FIGURE 28-2: CONNECTIONS FOR THE ON-CHIP REGULATOR



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# 28.3.2 OPERATION OF REGULATOR IN SLEEP

The difference in the two regulators' operation arises with Sleep mode. The ultra low-power regulator gives the device the lowest current in the Regulator Enabled mode.

The on-chip regulator can go into a lower power mode, when the device goes to Sleep, by setting the REGSLP bit (WDTCON<7>). This puts the regulator in a Standby mode so that the device consumes much less current.

The on-chip regulator can also go into the Ultra Low-Power mode, which consumes the lowest current possible with the regulator enabled. This mode is controlled by the RETEN bit (CONFIG1L<0>) and SRETEN bit (WDTCON<4>). The various modes of regulator operation are shown in Table 28-3.

When the ultra low-power regulator is in Sleep mode, the internal reference voltages in the chip will be shut off and any interrupts referring to the internal reference will not wake up the device. If the BOR or LVD is enabled, the regulator will keep the internal references on and the lowest possible current will not be achieved.

When using the ultra low-power regulator in Sleep mode, the device will take about 250  $\mu$ s, typical, to start executing the code after it wakes up.

Regulator	Power Mode	VREGSLP WDTCON<7>	SRETEN WDTCON<4>	RETEN CONFIG1L<0>			
Enabled	Normal Operation (Sleep)	0	x	1			
Enabled	Low-Power mode (Sleep)	1	x	1			
Enabled	Normal Operation (Sleep)	0	0	x			
Enabled	Low-Power mode (Sleep)	1	0	x			
Enabled	Ultra Low-Power mode (Sleep)	x	1	0			

### TABLE 28-3: SLEEP MODE REGULATOR SETTINGS⁽¹⁾

**Note 1:** x = Indicates that VIT status is invalid.

### 28.4 Two-Speed Start-up

The Two-Speed Start-up feature helps to minimize the latency period, from oscillator start-up to code execution, by allowing the microcontroller to use the INTOSC (LF-INTOSC, MF-INTOSC, HF-INTOSC) oscillator as a clock source until the primary clock source is available; it is enabled by setting the IESO Configuration bit.

Two-Speed Start-up should be enabled only if the primary oscillator mode is LP, XT or HS (Crystal-Based modes). Other sources do not require an OST start-up delay; for these, Two-Speed Start-up should be disabled.

When enabled, Resets and wake-ups from Sleep mode cause the device to configure itself to run from the internal oscillator block as the clock source, following the time-out of the Power-up Timer after a Power-on Reset is enabled. This allows almost immediate code execution while the primary oscillator starts and the OST is running. Once the OST times out, the device automatically switches to PRI_RUN mode.

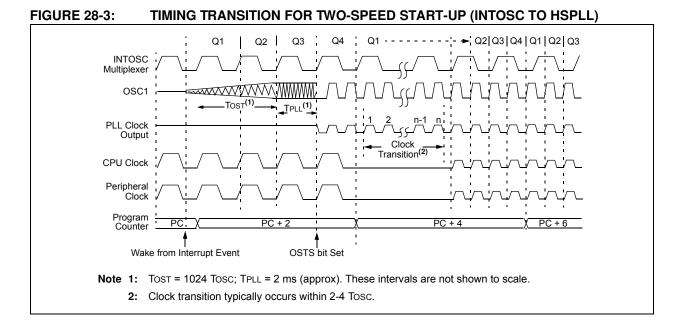
To use a higher clock speed on wake-up, the INTOSC or postscaler clock sources can be selected to provide a higher clock speed by setting bits, IRCF<2:0>, immediately after Reset. For wake-ups from Sleep, the INTOSC or postscaler clock sources can be selected by setting the IRCF<2:0> bits prior to entering Sleep mode.

In all other power-managed modes, Two-Speed Startup is not used. The device will be clocked by the currently selected clock source until the primary clock source becomes available. The setting of the IESO bit is ignored.

#### 28.4.1 SPECIAL CONSIDERATIONS FOR USING TWO-SPEED START-UP

While using the INTOSC oscillator in Two-Speed Startup, the device still obeys the normal command sequences for entering power-managed modes, including multiple SLEEP instructions (refer to Section 4.1.4 "Multiple Sleep Commands"). In practice, this means that user code can change the SCS<1:0> bit settings or issue SLEEP instructions before the OST times out. This would allow an application to briefly wake up, perform routine "housekeeping" tasks and return to Sleep before the device starts to operate from the primary oscillator.

User code can also check if the primary clock source is currently providing the device clocking by checking the status of the OSTS bit (OSCCON<3>). If the bit is set, the primary oscillator is providing the clock. Otherwise, the internal oscillator block is providing the clock during wake-up from Reset or Sleep mode.

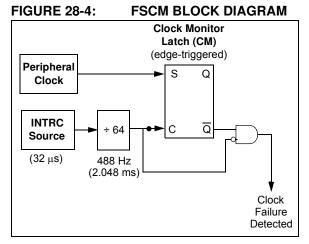


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### 28.5 Fail-Safe Clock Monitor

The Fail-Safe Clock Monitor (FSCM) allows the microcontroller to continue operation in the event of an external oscillator failure by automatically switching the device clock to the internal oscillator block. The FSCM function is enabled by setting the FCMEN Configuration bit.

When FSCM is enabled, the LF-INTOSC oscillator runs at all times to monitor clocks to peripherals and provide a backup clock in the event of a clock failure. Clock monitoring (shown in Figure 28-4) is accomplished by creating a sample clock signal, which is the output from the LF-INTOSC, divided by 64. This allows ample time between FSCM sample clocks for a peripheral clock edge to occur. The peripheral device clock and the sample clock are presented as inputs to the Clock Monitor (CM) latch. The CM is set on the falling edge of the device clock source, but cleared on the rising edge of the sample clock.



Clock failure is tested for on the falling edge of the sample clock. If a sample clock falling edge occurs while CM is still set, a clock failure has been detected (Figure 28-5). This causes the following:

- The FSCM generates an oscillator fail interrupt by setting bit, OSCFIF (PIR2<7>)
- The device clock source switches to the internal oscillator block (OSCCON is not updated to show the current clock source – this is the Fail-Safe condition)
- · The WDT is reset

During switchover, the postscaler frequency from the internal oscillator block may not be sufficiently stable for timing-sensitive applications. In these cases, it may be desirable to select another clock configuration and enter an alternate power-managed mode. This can be done to attempt a partial recovery or execute a controlled shutdown. See Section 4.1.4 "Multiple Sleep Commands" and Section 28.4.1 "Special Considerations for Using Two-Speed Start-up" for more details.

To use a higher clock speed on wake-up, the INTOSC or postscaler clock sources can be selected to provide a higher clock speed by setting bits, IRCF<2:0>, immediately after Reset. For wake-ups from Sleep, the INTOSC or postscaler clock sources can be selected by setting the IRCF<2:0> bits prior to entering Sleep mode.

The FSCM will detect only failures of the primary or secondary clock sources. If the internal oscillator block fails, no failure would be detected nor would any action be possible.

#### 28.5.1 FSCM AND THE WATCHDOG TIMER

Both the FSCM and the WDT are clocked by the INTOSC oscillator. Since the WDT operates with a separate divider and counter, disabling the WDT has no effect on the operation of the INTOSC oscillator when the FSCM is enabled.

As already noted, the clock source is switched to the INTOSC clock when a clock failure is detected. Depending on the frequency selected by the IRCF<2:0> bits, this may mean a substantial change in the speed of code execution. If the WDT is enabled with a small prescale value, a decrease in clock speed allows a WDT time-out to occur and a subsequent device Reset. For this reason, Fail-Safe Clock events also reset the WDT and postscaler, allowing it to start timing from when execution speed was changed, and decreasing the likelihood of an erroneous time-out.

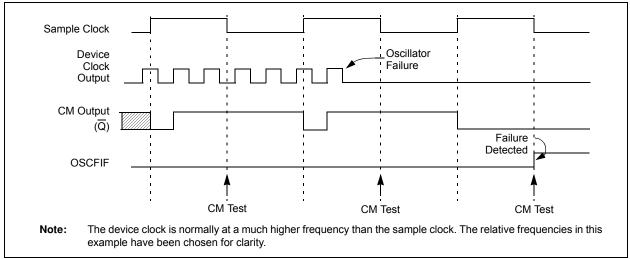
#### 28.5.2 EXITING FAIL-SAFE OPERATION

The Fail-Safe condition is terminated by either a device Reset or by entering a power-managed mode. On Reset, the controller starts the primary clock source specified in Configuration Register 1H (with any required start-up delays that are required for the Oscillator mode, such as the OST or PLL timer). The INTOSC multiplexer provides the device clock until the primary clock source becomes ready (similar to a Two-Speed Start-up). The clock source is then switched to the primary clock (indicated by the OSTS bit in the OSCCON register becoming set). The Fail-Safe Clock Monitor then resumes monitoring the peripheral clock.

The primary clock source may never become ready during start-up. In this case, operation is clocked by the INTOSC multiplexer. The OSCCON register will remain in its Reset state until a power-managed mode is entered.

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#### 28.5.3 FSCM INTERRUPTS IN POWER-MANAGED MODES

By entering a power-managed mode, the clock multiplexer selects the clock source selected by the OSCCON register. Fail-Safe Clock Monitoring of the power-managed clock source resumes in the power-managed mode.

If an oscillator failure occurs during power-managed operation, the subsequent events depend on whether or not the Oscillator Failure Interrupt Flag bit is enabled (OSCFIF = 1). If enabled, code execution will be clocked by the INTOSC multiplexer. An automatic transition back to the failed clock source will not occur.

If the interrupt is disabled, subsequent interrupts while in Idle mode will cause the CPU to begin executing instructions while being clocked by the INTOSC source.

#### 28.5.4 POR OR WAKE FROM SLEEP

The FSCM is designed to detect oscillator failure at any point after the device has exited Power-on Reset (POR) or low-power Sleep mode. When the primary device clock is EC, RC or INTRC modes, monitoring can begin immediately following these events. For Oscillator modes involving a crystal or resonator (HS, HSPLL, LP or XT), the situation is somewhat different. Since the oscillator may require a start-up time considerably longer than the FCSM sample clock time, a false clock failure may be detected. To prevent this, the internal oscillator block is automatically configured as the device clock and functions until the primary clock is stable (when the OST and PLL timers have timed out).

This is identical to Two-Speed Start-up mode. Once the primary clock is stable, the INTOSC returns to its role as the FSCM source.

Note: The same logic that prevents false oscillator failure interrupts on POR, or wake from Sleep, also prevents the detection of the oscillator's failure to start at all following these events. This can be avoided by monitoring the OSTS bit and using a timing routine to determine if the oscillator is taking too long to start. Even so, no oscillator failure interrupt will be flagged.

As noted in Section 28.4.1 "Special Considerations for Using Two-Speed Start-up", it is also possible to select another clock configuration and enter an alternate power-managed mode while waiting for the primary clock to become stable. When the new powermanaged mode is selected, the primary clock is disabled.

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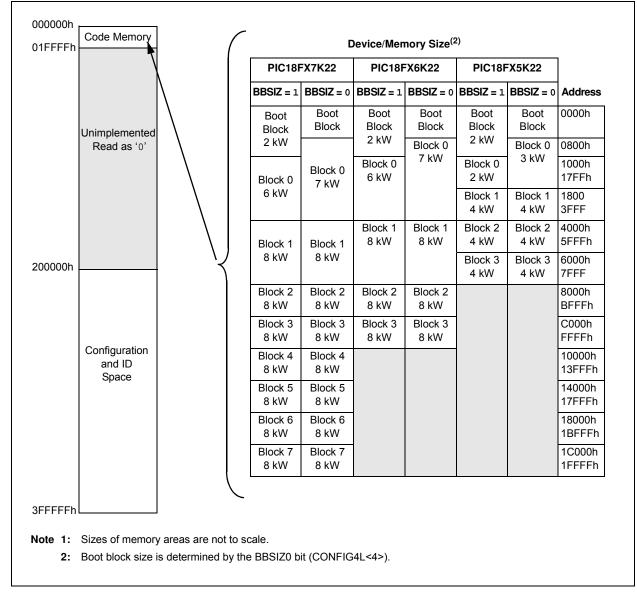
# 28.6 Program Verification and Code Protection

The user program memory is divided into four blocks for the PIC18FX5K22 and PIC18FX6K22 devices, and eight blocks for PIC18FX7K22 devices. One of these is a boot block of 1 or 2 Kbytes. The remainder of the memory is divided into blocks on binary boundaries. Each of the blocks has three code protection bits associated with them. They are:

- Code-Protect bit (CPx)
- Write-Protect bit (WRTx)
- External Block Table Read bit (EBTRx)

Figure 28-6 shows the program memory organization for 48, 64, 96 and 128 Kbyte devices and the specific code protection bit associated with each block. The actual locations of the bits are summarized in Table 28-4.

#### FIGURE 28-6: CODE-PROTECTED PROGRAM MEMORY FOR THE PIC18F87K22 FAMILY⁽¹⁾



		_				-	-	-	
File	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
300008h	CONFIG5L	CP7 ⁽¹⁾	CP6 ⁽¹⁾	CP5 ⁽¹⁾	CP4 ⁽¹⁾	CP3	CP2	CP1	CP0
300009h	CONFIG5H	CPD	CPB	_	_	_	_	_	—
30000Ah	CONFIG6L	WRT7 ⁽¹⁾	WRT6 ⁽¹⁾	WRT5 ⁽¹⁾	WRT4 ⁽¹⁾	WRT3	WRT2	WRT1	WRT0
30000Bh	CONFIG6H	WRTD	WRTB	WRTC		_	_	_	—
30000Ch	CONFIG7L	EBTR7 ⁽¹⁾	EBTR6 ⁽¹⁾	EBTR5 ⁽¹⁾	EBTR4 ⁽¹⁾	EBTR3	EBTR2	EBTR1	EBTR0
30000Dh	CONFIG7H	_	EBTRB		_		_	_	_
CCC DODI									

TABLE 28-4: SUMMARY OF CODE PROTECTION REGISTERS

**Legend:** Shaded cells are unimplemented.

Note 1: This bit is available only on the PIC18F67K22 and PIC18F87K22 devices.

#### 28.6.1 PROGRAM MEMORY CODE PROTECTION

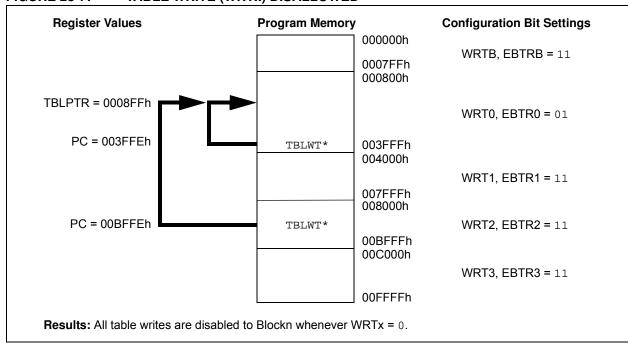
The program memory may be read to, or written from, any location using the table read and table write instructions. The Device ID may be read with table reads. The Configuration registers may be read and written with the table read and table write instructions.

In normal Execution mode, the CPx bits have no direct effect. CPx bits inhibit external reads and writes. A block of user memory may be protected from table writes if the WRTx Configuration bit is '0'.

The EBTRx bits control table reads. For a block of user memory, with the EBTRx bit set to '0', a table read instruction that executes from within that block is allowed

to read. A table read instruction that executes from a location outside of that block is not allowed to read and will result in reading '0's. Figures 28-7 through 28-9 illustrate table write and table read protection.

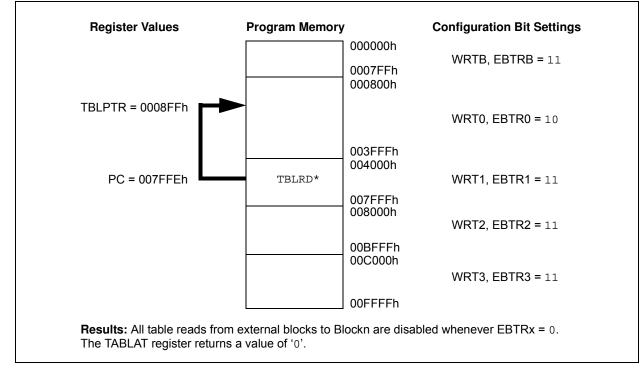
Note:	Code protection bits may only be written to a '0' from a '1' state. It is not possible to write a '1' to a bit in the '0' state. Code protection bits are only set to '1' by a full chip erase or block erase function. The full chip erase and block erase functions can only be initiated via ICSP or an external
	programmer. Refer to the device programming specification for more information.



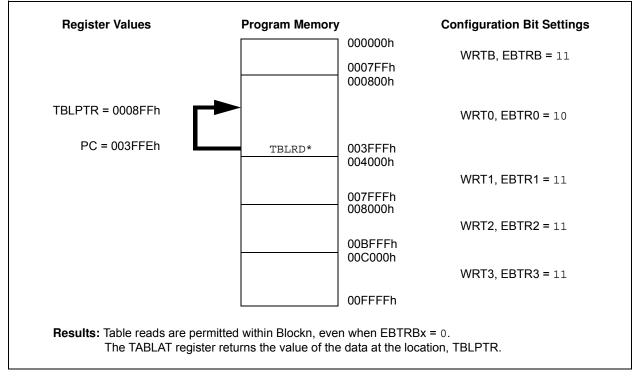
### FIGURE 28-7: TABLE WRITE (WRTx) DISALLOWED

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#### FIGURE 28-8: EXTERNAL BLOCK TABLE READ (EBTRx) DISALLOWED



#### FIGURE 28-9: EXTERNAL BLOCK TABLE READ (EBTRx) ALLOWED



#### 28.6.2 DATA EEPROM CODE PROTECTION

The entire data EEPROM is protected from external reads and writes by two bits: CPD and WRTD. CPD inhibits external reads and writes of data EEPROM. WRTD inhibits internal and external writes to data EEPROM. The CPU can always read data EEPROM under normal operation, regardless of the protection bit settings.

#### 28.6.3 CONFIGURATION REGISTER PROTECTION

The Configuration registers can be write-protected. The WRTC bit controls protection of the Configuration registers. In normal Execution mode, the WRTC bit is readable only. WRTC can only be written via ICSP or an external programmer.

### 28.7 ID Locations

Eight memory locations (20000h-200007h) are designated as ID locations, where the user can store checksum or other code identification numbers. These locations are both readable and writable, during normal execution, through the TBLRD and TBLWT instructions or during program/verify. The ID locations can be read when the device is code-protected.

# 28.8 In-Circuit Serial Programming

The PIC18F87K22 family of devices can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data, and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

For the various Programming modes, see the device programming specification.

# 28.9 In-Circuit Debugger

When the DEBUG Configuration bit is programmed to a '0', the In-Circuit Debugger (ICD) functionality is enabled. This function allows simple debugging functions when used with MPLAB[®] IDE. When the microcontroller has this feature enabled, some resources are not available for general use. Table 28-5 shows which resources are required by the background debugger.

I/O Pins:	RB6, RB7
Stack:	Two levels
Program Memory:	512 bytes
Data Memory:	10 bytes

TABLE 28-5: D	<b>EBUGGER RESOURCES</b>
---------------	--------------------------

To use the In-Circuit Debugger function of the microcontroller, the design must implement In-Circuit Serial Programming connections to MCLR/RG5/VPP, VDD, Vss, RB7 and RB6. This will interface to the In-Circuit Debugger module, available from Microchip or one of the third-party development tool companies.

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NOTES:

# 29.0 INSTRUCTION SET SUMMARY

The PIC18F87K22 family of devices incorporates the standard set of 75 PIC18 core instructions, as well as an extended set of 8 new instructions for the optimization of code that is recursive or that utilizes a software stack. The extended set is discussed later in this section.

## 29.1 Standard Instruction Set

The standard PIC18 MCU instruction set adds many enhancements to the previous PIC[®] MCU instruction sets, while maintaining an easy migration from these PIC MCU instruction sets. Most instructions are a single program memory word (16 bits), but there are four instructions that require two program memory locations.

Each single-word instruction is a 16-bit word divided into an opcode, which specifies the instruction type and one or more operands, which further specify the operation of the instruction.

The instruction set is highly orthogonal and is grouped into four basic categories:

- Byte-oriented operations
- Bit-oriented operations
- · Literal operations
- Control operations

The PIC18 instruction set summary in Table 29-2 lists **byte-oriented**, **bit-oriented**, **literal** and **control** operations. Table 29-1 shows the opcode field descriptions.

Most byte-oriented instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The destination of the result (specified by 'd')
- 3. The accessed memory (specified by 'a')

The file register designator, 'f', specifies which file register is to be used by the instruction. The destination designator, 'd', specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the WREG register. If 'd' is one, the result is placed in the file register specified in the instruction.

All **bit-oriented** instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The bit in the file register (specified by 'b')
- 3. The accessed memory (specified by 'a')

The bit field designator, 'b', selects the number of the bit affected by the operation, while the file register designator, 'f', represents the number of the file in which the bit is located.

The **literal** instructions may use some of the following operands:

- A literal value to be loaded into a file register (specified by 'k')
- The desired FSR register to load the literal value into (specified by 'f')
- No operand required (specified by '—')

The **control** instructions may use some of the following operands:

- A program memory address (specified by 'n')
- The mode of the CALL or RETURN instructions (specified by 's')
- The mode of the table read and table write instructions (specified by 'm')
- No operand required (specified by '—')

All instructions are a single word, except for four double-word instructions. These instructions were made double-word to contain the required information in 32 bits. In the second word, the 4 MSbs are '1's. If this second word is executed as an instruction (by itself), it will execute as a NOP.

All single-word instructions are executed in a single instruction cycle, unless a conditional test is true or the Program Counter is changed as a result of the instruction. In these cases, the execution takes two instruction cycles with the additional instruction cycle(s) executed as a NOP.

The double-word instructions execute in two instruction cycles.

One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1  $\mu$ s. If a conditional test is true, or the Program Counter is changed as a result of an instruction, the instruction execution time is 2  $\mu$ s. Two-word branch instructions (if true) would take 3  $\mu$ s.

Figure 29-1 shows the general formats that the instructions can have. All examples use the convention 'nnh' to represent a hexadecimal number.

The Instruction Set Summary, shown in Table 29-2, lists the standard instructions recognized by the Microchip MPASM[™] Assembler.

Section 29.1.1 "Standard Instruction Set" provides a description of each instruction.

### TABLE 29-1: OPCODE FIELD DESCRIPTIONS

RAM access bit:
a = 0: RAM location in Access RAM (BSR register is ignored)
a = 1: RAM bank is specified by BSR register
Bit address within an 8-bit file register (0 to 7).
Bank Select Register. Used to select the current RAM bank.
ALU Status bits: Carry, Digit Carry, Zero, Overflow, Negative.
Destination select bit: d = 0: store result in WREG d = 1: store result in file register f
Destination: either the WREG register or the specified register file location.
8-bit register file address (00h to FFh), or 2-bit FSR designator (0h to 3h).
12-bit register file address (000h to FFFh). This is the source address.
12-bit register file address (000h to FFFh). This is the destination address.
Global Interrupt Enable bit.
Literal field, constant data or label (may be either an 8-bit, 12-bit or a 20-bit value).
Label name.
The mode of the TBLPTR register for the table read and table write instructions. Only used with table read and table write instructions:
No Change to register (such as TBLPTR with table reads and writes)
Post-Increment register (such as TBLPTR with table reads and writes)
Post-Decrement register (such as TBLPTR with table reads and writes)
Pre-Increment register (such as TBLPTR with table reads and writes)
The relative address (2's complement number) for relative branch instructions or the direct address for Call/Branch and Return instructions.
Program Counter.
Program Counter Low Byte.
Program Counter High Byte.
Program Counter High Byte Latch.
Program Counter Upper Byte Latch.
Power-Down bit.
Product of Multiply High Byte.
Product of Multiply Low Byte.
Fast Call/Return mode select bit: s = 0: do not update into/from shadow registers s = 1: certain registers loaded into/from shadow registers (Fast mode)
21-bit Table Pointer (points to a Program Memory location).
8-bit Table Latch.
Time-out bit.
Top-of-Stack.
Unused or Unchanged.
Watchdog Timer.
Working register (accumulator).
Don't care ('0' or '1'). The assembler will generate code with $x = 0$ . It is the recommended form of use for compatibility with all Microchip software tools.
7-bit offset value for Indirect Addressing of register files (source).
7-bit offset value for Indirect Addressing of register files (destination).
Optional argument.
Indicates an Indexed Address.
The contents of text.
Specifies bit n of the register indicated by the pointer $expr$ .
Assigned to.
Register bit field.
In the set of.

Byte-oriented file register operations	Example Instruction
<u>15 10 9 8 7 0</u>	
OPCODE d a f (FILE #)	ADDWF MYREG, W, B
<ul> <li>d = 0 for result destination to be WREG register</li> <li>d = 1 for result destination to be file register (f)</li> <li>a = 0 to force Access Bank</li> <li>a = 1 for BSR to select bank</li> <li>f = 8-bit file register address</li> </ul>	
Byte to Byte move operations (2-word)	
15 12 11 0	
OPCODE f (Source FILE #)	MOVFF MYREG1, MYREG
<u>15 12 11 0</u>	
1111 f (Destination FILE #)	
f = 12-bit file register address	
Bit-oriented file register operations	
15 12 11 9 8 7 0	
OPCODE b (BIT #) a f (FILE #)	BSF MYREG, bit, B
<ul> <li>b = 3-bit position of bit in file register (f)</li> <li>a = 0 to force Access Bank</li> <li>a = 1 for BSR to select bank</li> <li>f = 8-bit file register address</li> </ul>	
Literal operations	
15 8 7 0	
OPCODE k (literal)	MOVLW 7Fh
k = 8-bit immediate value	
Control operations	
CALL, GOTO and Branch operations	
15 8 7 0	
OPCODE n<7:0> (literal)	GOTO Label
15 12 11 0	
1111 n<19:8> (literal)	
n = 20-bit immediate value	
15 8 7 0	
15 8 7 0 OPCODE S n<7:0> (literal)	CALL MYFUNC
	CALL MYFUNC
OPCODE         S         n<7:0> (literal)           15         12         11         0           1111         n<19:8> (literal)	CALL MYFUNC
OPCODE         S         n<7:0> (literal)           15         12         11         0	CALL MYFUNC
OPCODE         S         n<7:0> (literal)           15         12         11         0           1111         n<19:8> (literal)	CALL MYFUNC
OPCODE         S         n<7:0> (literal)           15         12         11         0           1111         n<19:8> (literal)         S = Fast bit	CALL MYFUNC BRA MYFUNC
OPCODE         S         n<7:0> (literal)           15         12         11         0           1111         n<19:8> (literal)         S = Fast bit           15         11         10         0	

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#### TABLE 29-2: PIC18F87K22 FAMILY INSTRUCTION SET

Mnem	onic,	Description	Qualas	16-E	Bit Insti	ruction V	Vord	Status	Natas
Opera		Description	Cycles	MSb			LSb	Affected	Notes
<b>BYTE-ORI</b>	ENTED (	OPERATIONS							
ADDWF	f, d, a	Add WREG and f	1	0010	01da	ffff	ffff	C, DC, Z, OV, N	1, 2
ADDWFC	f, d, a	Add WREG and Carry bit to f	1	0010	00da	ffff	ffff	C, DC, Z, OV, N	1, 2
ANDWF	f, d, a	AND WREG with f	1	0001	01da	ffff	ffff	Z, N	1, 2
CLRF	f, a	Clear f	1	0110	101a	ffff	ffff	Z	2
COMF	f, d, a	Complement f	1	0001	11da	ffff	ffff	Z, N	1, 2
CPFSEQ	f, a	Compare f with WREG, Skip =	1 (2 or 3)	0110	001a	ffff	ffff	None	4
CPFSGT	f, a	Compare f with WREG, Skip >	1 (2 or 3)	0110	010a	ffff	ffff	None	4
CPFSLT	f, a	Compare f with WREG, Skip <	1 (2 or 3)	0110	000a	ffff	ffff	None	1, 2
DECF	f, d, a	Decrement f	1	0000	01da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
DECFSZ	f, d, a	Decrement f, Skip if 0	1 (2 or 3)	0010	11da	ffff	ffff	None	1, 2, 3, 4
DCFSNZ	f, d, a	Decrement f, Skip if Not 0	1 (2 or 3)	0100	11da	ffff	ffff	None	1, 2
INCF	f, d, a	Increment f	1	0010	10da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
INCFSZ	f, d, a	Increment f, Skip if 0	1 (2 or 3)	0011	11da	ffff	ffff	None	4
INFSNZ	f, d, a	Increment f, Skip if Not 0	1 (2 or 3)	0100	10da	ffff	ffff	None	1, 2
IORWF	f, d, a	Inclusive OR WREG with f	1	0001	00da	ffff	ffff	Z, N	1, 2
MOVF	f, d, a	Move f	1	0101	00da	ffff	ffff	Z, N	1
MOVFF	f _s , f _d	Move f _s (source) to 1st word	2	1100	ffff	ffff	ffff	None	
	0 u	f _d (destination) 2nd word		1111	ffff	ffff	ffff		
MOVWF	f, a	Move WREG to f	1	0110	111a	ffff	ffff	None	
MULWF	f, a	Multiply WREG with f	1	0000	001a	ffff	ffff	None	1, 2
NEGF	f, a	Negate f	1	0110	110a	ffff	ffff	C, DC, Z, OV, N	-
RLCF	f, d, a	Rotate Left f through Carry	1	0011	01da	ffff	ffff	C, Z, N	1, 2
RLNCF	f, d, a	Rotate Left f (No Carry)	1	0100	01da	ffff	ffff	Z, N	-
RRCF	f, d, a	Rotate Right f through Carry	1	0011	00da	ffff	ffff	C, Z, N	
RRNCF	f, d, a	Rotate Right f (No Carry)	1	0100	00da	ffff	ffff	Z, N	
SETF	f, a	Set f	1	0110	100a	ffff	ffff	None	1, 2
SUBFWB	f, d, a	Subtract f from WREG with Borrow	1	0101	01da	ffff	ffff	C, DC, Z, OV, N	
SUBWF	f, d, a	Subtract WREG from f	1	0101	11da	ffff	ffff	C, DC, Z, OV, N	1, 2
SUBWFB	f, d, a	Subtract WREG from f with Borrow	1	0101		ffff	ffff	C, DC, Z, OV, N	
SWAPF	f, d, a	Swap Nibbles in f	1	0011	10da	ffff	ffff	None	4
TSTFSZ	, ,		1 (2  or  2)					None	
XORWF	f, a f d o	Test f, Skip if 0 Exclusive OR WREG with f	1 (2 or 3)		011a	ffff			1, 2
VORME	f, d, a		1		10da	ffff	ffff	∠, IN	

**Note 1:** When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

**3:** If the Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

Mnem	onic,	Description	Ovelas	16-E	Bit Instr	uction V	Vord	Status	Nata
Opera	ands	Description	Cycles	MSb			LSb	Affected	Notes
BIT-ORIE	NTED OP	ERATIONS							
BCF	f, b, a	Bit Clear f	1	1001	bbba	ffff	ffff	None	1, 2
BSF	f, b, a	Bit Set f	1		bbba	ffff	ffff	None	1, 2
BTFSC	f, b, a	Bit Test f, Skip if Clear	1 (2 or 3)			ffff	ffff	None	3, 4
BTFSS	f, b, a	Bit Test f, Skip if Set	1 (2 or 3)	1010	bbba	ffff	ffff	None	3, 4
BTG	f, b, a	Bit Toggle f	1	0111	bbba	ffff	ffff	None	1, 2
CONTROL		TIONS							
BC	n	Branch if Carry	1 (2)	1110	0010	nnnn	nnnn	None	
BN	n	Branch if Negative	1 (2)	1110	0110	nnnn	nnnn	None	
BNC	n	Branch if Not Carry	1 (2)	1110	0011	nnnn	nnnn	None	
BNN	n	Branch if Not Negative	1 (2)	1110	0111	nnnn	nnnn	None	
BNOV	n	Branch if Not Overflow	1 (2)	1110	0101	nnnn	nnnn	None	
BNZ	n	Branch if Not Zero	1 (2)	1110	0001	nnnn	nnnn	None	
BOV	n	Branch if Overflow	1 (2)	1110	0100	nnnn	nnnn	None	
BRA	n	Branch Unconditionally	2	1101	0nnn	nnnn	nnnn	None	
BZ	n	Branch if Zero	1 (2)	1110	0000	nnnn	nnnn	None	
CALL	n, s	Call Subroutine 1st word	2	1110	110s	kkkk	kkkk	None	
		2nd word			kkkk	kkkk	kkkk		
CLRWDT	—	Clear Watchdog Timer	1	0000	0000	0000		TO, PD	
DAW	_	Decimal Adjust WREG	1	0000	0000	0000		С	
GOTO	n	Go to Address 1st word	2	-	1111	kkkk	kkkk	None	
		2nd word			kkkk	kkkk	kkkk		
NOP	—	No Operation	1	0000	0000	0000		None	
NOP	—	No Operation	1	1111		XXXX		None	4
POP	—	Pop Top of Return Stack (TOS)	1	0000	0000	0000		None	
PUSH	—	Push Top of Return Stack (TOS)		0000	0000	0000		None	
RCALL	n	Relative Call	2	1101	1nnn	nnnn		None	
RESET		Software Device Reset	1	0000	0000	1111		All	
RETFIE	S	Return from Interrupt Enable	2	0000	0000	0001	000s	GIE/GIEH,	
RETLW	k	Return with Literal in WREG	2	0000	1100	kkkk	kkkk	PEIE/GIEL	
RETURN	S	Return from Subroutine	2	0000	0000	0001		None	
SLEEP	3	Go into Standby mode	2		0000	00001		TO, PD	

#### TABLE 29-2: PIC18F87K22 FAMILY INSTRUCTION SET (CONTINUED)

**Note 1:** When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

**3:** If the Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

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TABLE 29	-2. Г	IC18F8/K22 FAMILY INSTRU							
Mnemo	onic,	Description	Cycles	16-	Bit Inst	ruction	Word	Status	Notes
Opera	nds	Description					LSb	Affected	Notes
LITERAL C	OPERAT	IONS							
ADDLW	k	Add Literal and WREG	1	0000	1111	kkkk	kkkk	C, DC, Z, OV, N	
ANDLW	k	AND Literal with WREG	1	0000	1011	kkkk	kkkk	Z, N	
IORLW	k	Inclusive OR Literal with WREG	1	0000	1001	kkkk	kkkk	Z, N	
LFSR	f, k	Move literal (12-bit) 2nd word	2	1110	1110	00ff	kkkk	None	
		to FSR(f) 1st word		1111	0000	kkkk	kkkk		
MOVLB	k	Move Literal to BSR<3:0>	1	0000	0001	0000	kkkk	None	
MOVLW	k	Move Literal to WREG	1	0000	1110	kkkk	kkkk	None	
MULLW	k	Multiply Literal with WREG	1	0000	1101	kkkk	kkkk	None	
RETLW	k	Return with Literal in WREG	2	0000	1100	kkkk	kkkk	None	
SUBLW	k	Subtract WREG from Literal	1	0000	1000	kkkk	kkkk	C, DC, Z, OV, N	
XORLW	k	Exclusive OR Literal with WREG	1	0000	1010	kkkk	kkkk	Z, N	
DATA MEN	/IORY ←	PROGRAM MEMORY OPERATI	ONS						
TBLRD*		Table Read	2	0000	0000	0000	1000	None	
TBLRD*+		Table Read with Post-Increment		0000	0000	0000	1001	None	
TBLRD*-		Table Read with Post-Decrement		0000	0000	0000	1010	None	
TBLRD+*		Table Read with Pre-Increment		0000	0000	0000	1011	None	
TBLWT*		Table Write	2	0000	0000	0000	1100	None	
TBLWT*+		Table Write with Post-Increment		0000	0000	0000	1101	None	
TBLWT*-		Table Write with Post-Decrement		0000	0000	0000	1110	None	
TBLWT+*		Table Write with Pre-Increment		0000	0000	0000	1111	None	

#### TABLE 29-2: PIC18F87K22 FAMILY INSTRUCTION SET (CONTINUED)

**Note 1:** When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned.

**3:** If the Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

#### 29.1.1 STANDARD INSTRUCTION SET

ADDI	LW	ADD Litera	l to W		ADDWF	ADD W to	f		
Synta	IX:	ADDLW	k		Syntax:	ADDWF	f {,d {,a}}		
Opera	ands:	$0 \le k \le 255$			Operands:	$0 \leq f \leq 255$			
Opera	ation:	$(W) + k \rightarrow V$	N			d ∈ [0,1] a ∈ [0,1]			
Status	s Affected:	N, OV, C, D	)C, Z		Operation		doot		
Enco	ding:	0000	1111 kkł	ck kkkk	Operation: Status Affected:	$(W) + (f) \rightarrow$			
Desci	ription:		ts of W are ad k' and the res		Encoding: Description:	N, OV, C, DC, Z 0010 01da ffff ffff Add W to register 'f'. If 'd' is '0', the			
Word	s:	1				result is sto	ored in W. If 'd'	is '1', the	
Cycle	s:	1					ored back in re	•	
	cle Activity: Q1	Q2	Q3	Q4			he Access Bai he BSR is use		
	Decode	Read literal 'k'	Process Data	Write to W		lf 'a' is '0' a	nd the extend		
	Before Instruc W =	tion 10h	.5h			mode wher Section 29 Bit-Oriente	Literal Offset A never $f \le 95$ (5) <b>.2.3 "Byte-Or</b> ed Instruction set Mode" for	Fh). See iented and s in Indexed	
/	After Instructic W =	on 25h			Words:	1			
	vv –	2511			Cycles:	1			
					Q Cycle Activity:				
					Q1	Q2	Q3	Q4	
					Decode	Read register 'f'	Process Data	Write to destination	
					Example:	ADDWF	REG, 0, 0		
					Before Instruc W REG After Instructi W REG	= 17h = 0C2h			

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

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ADDWFC	ADD W an	d Carry	bit to f			
Syntax:	ADDWFC	f {,d {,	a}}			
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$					
Operation:	(W) + (f) +	$(C) \rightarrow de$	st			
Status Affected:	N,OV, C, D	C, Z				
Encoding:	0010	00da	ffff	ffff		
Description:	Add W, the Carry flag and data memor location 'f'. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed in data memory location 'f'. If 'a' is '0', the Access Bank is selecter If 'a' is '1', the BSR is used to select th					
	,	he BSR i				
	If 'a' is '0' a set is enab in Indexed mode wher Section 29 Bit-Oriente Literal Offe	led, this i Literal O never f ≤ 0.2.3 "By ed Instru	nstruction ffset Addr 95 (5Fh). te-Orient ctions in	operates essing See ed and Indexed		
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read register 'f'	Proce Data		Vrite to stination		
Example:	ADDWFC	REG,	0, 1			
Before Instruct Carry bit REG W After Instructi Carry bit REG W	= 1 = 02h = 4Dh					

ANDLW	AND Litera	al with W	1	
Syntax:	ANDLW	k		
Operands:	$0 \le k \le 255$			
Operation:	(W) .AND.	$k \rightarrow W$		
Status Affected:	N, Z			
Encoding:	0000	1011	kkkk	kkkk
Description:	The conten 8-bit literal			
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3	1	Q4
Decode	Read literal 'k'	Proce Data		Write to W
Example:	ANDLW	05Fh		
Before Instruc W After Instructic	= A3h			
	ווע			

ANDWF	AND W wit	th f		BC		Branch if C	Carry	
Syntax:	ANDWF	f {,d {,a}}		Synta	ax:	BC n		
Operands:	$0 \leq f \leq 255$			Oper	ands:	-128 ≤ n ≤ ′	127	
	$\begin{array}{l} d \in [0,1] \\ a \in [0,1] \end{array}$			Oper	ation:	if Carry bit i (PC) + 2 + 2		
Operation:	(W) .AND.	(f) $\rightarrow$ dest		Statu	s Affected:	None		
Status Affected:	N, Z			Enco	dina:	1110	0010 nn:	nn nnnn
Encoding:	0001	01da ff:	ff ffff		ription:	If the Carry	bit is '1', then	the program
Description:	The conten	ts of W are AN	Ded with	2000	inpuoli.	will branch.		ino program
	in W. If 'd' is in register ' If 'a' is '0', t	s '1', the result				added to the incremente instruction,	nplement num e PC. Since th d to fetch the the new addre n. This instruct	e PC will have next ess will be
	GPR bank.					two-cycle ir		
	If 'a' is '0' a	nd the extend	ed instruction	Word	ls:	1		
		led, this instrue	•	Cycle	es:	1(2)		
	mode wher	Literal Offset A never f ≤ 95 (5 .2.3 "Byte-Or	Fh). See		ycle Activity:			
		ed Instruction			Q1	Q2	Q3	Q4
		set Mode" for	details.		Decode	Read literal	Process	Write to
Words:	1				No	ʻn' No	Data	PC
Cycles:	1				No operation	operation	No operation	No operation
Q Cycle Activity	:			lf No	o Jump:	operation	oporation	opolation
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4
Decode	Read	Process	Write to		Decode	Read literal	Process	No
	register 'f'	Data	destination			'n'	Data	operation
Example:	ANDWF	REG, 0, 0		Exan	<u>nple:</u>	HERE	BC 5	
Before Instr					Before Instru	ction		
W REG	= 17h = C2h				PC		dress (HERE	)
After Instruc	•=				After Instructi			
W REG	= 02h = C2h				If Carry PC If Carry PC	= 0;	dress (here dress (here	

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BCF	Bit Clear f			BN		Branch if N	legative	
Syntax:	BCF f, b	{,a}		Synta	ax:	BN n		
Operands:	$0 \leq f \leq 255$			Oper	ands:	-128 ≤ n ≤ ′	127	
	0 ≤ b ≤ 7 a ∈ [0,1]			Oper	ation:	if Negative (PC) + 2 + 2	,	
Operation:	$0 \rightarrow f \le b >$			Statu	is Affected:	None		
Status Affected:	None			Enco	oding:	1110	0110 nn	nn nnnn
Encoding:	1001	bbba f	fff ffff		cription:	If the Nega	tive bit is '1', th	nen the
Description:	Bit 'b' in reg	gister 'f' is cle	ared.			program wi		
	,		ank is selected. ed to select the			added to the incremente	d to fetch the	e PC will have next
	set is enabl		ded instruction uction operates Addressing			,	the new addre n. This instruct istruction.	
		never f $\leq$ 95 (	,	Word	ds:	1		
		.2.3 "Byte-C	riented and ns in Indexed	Cycle	es:	1(2)		
	Literal Offs	set Mode" fo		Q C If Ju	ycle Activity: imp:			
Words:	1				Q1	Q2	Q3	Q4
Cycles:	1				Decode	Read literal	Process	Write to
Q Cycle Activity:						ʻn'	Data	PC
Q1	Q2	Q3	Q4		No operation	No operation	No operation	No operation
Decode	Read register 'f'	Process Data	Write register 'f'	lf No	o Jump:	operation	operation	operation
	- 5 - 1				Q1	Q2	Q3	Q4
Example:	BCF F	LAG_REG,	7, 0		Decode	Read literal	Process	No
Before Instruc	ction					'n'	Data	operation
FLAG_R After Instructio	EG = C7h			Exan	nole [.]	HERE	BN Jump	
	EG = 47h				Before Instru		T	
					PC After Instructi	= ad	dress (HERE)	
					If Negati PC If Negati	ve = 1; = ad	dress (Jump)	
					PC		dress (HERE	+ 2)

BNC		Branch if N	lot Carry		BNN
Synta	ax:	BNC n			Syntax
Oper	ands:	$-128 \le n \le 1$	127		Operan
Oper	ation:	if Carry bit i (PC) + 2 + 2			Operati
Statu	is Affected:	None			Status
Enco	oding:	1110	0011 nnr	nn nnnn	Encodi
Desc	cription:	If the Carry will branch.	Descrip		
		added to the incremented instruction,	nplement num e PC. Since the d to fetch the r the new addre n. This instruct istruction.	e PC will have next ess will be	
Word	ds:	1			Words:
Cycle	es:	1(2)	1(2)		
Q C If Ju	ycle Activity: imp:				Q Cyc If Jum
	Q1	Q2	Q3	Q4	
	Decode	Read literal	Process	Write to	
		ʻn'	Data	PC	
	No operation	No operation	No operation	No operation	
lf No	o Jump:				If No J
	Q1	Q2	Q3	Q4	
	Decode	Read literal	Process	No	] [
		'n'	Data	operation	
<u>Exar</u>	<u>nple:</u> Before Instruc	HERE	BNC Jump		<u>Exampl</u> Be
	PC		dress (HERE)	)	
	After Instruction				Af
	If Carry PC	= 0; = ad	dress (Jump	)	
	If Carry	= 1;	. 1		
	PC	= ad	dress (HERE	+ 2)	

BNN		Branch if I	Not Nega	tive				
Synta	ax:	BNN n						
Oper	ands:	-128 ≤ n ≤ 1	127					
Oper	ation:	if Negative (PC) + 2 +		:				
Statu	is Affected:	None						
Enco	oding:	1110	0111	nnnn	nnnn			
Desc	ription:	If the Nega program wi		,	the			
		added to th incremente instruction, PC + 2 + 2	The 2's complement number '2n' is added to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is then a two-cycle instruction.					
Word	ls:	1						
Cycle	es:	1(2)						
Q C If Ju	ycle Activity: imp:							
	Q1	Q2	Q3		Q4			
	Decode	Read literal 'n'	Proce Data		Write to PC			
	No	No	No		No			
	operation	operation	operat	ion c	operation			
If No	o Jump: Q1	Q2	Q3		04			
	Decode	Read literal	Proce	-	Q4 No			
	Decoue	'n'	Data		operation			
<u>Exan</u>	nple:	HERE	BNN	Jump				
	Before Instruc PC After Instructio	= ad	dress (1	HERE)				
	If Negativ PC If Negativ PC	= ad /e = 1;		Jump) HERE +	2)			
	.0	au		ا تدادم	- /			

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BNOV	Branch if N	lot Overflow							
Syntax:	BNOV n								
Operands:	-128 ≤ n ≤ 1	$-128 \le n \le 127$							
Operation:		if Overflow bit is '0', (PC) + 2 + 2n $\rightarrow$ PC							
Status Affected:	None	None							
Encoding:	1110	0101 nn	nn nnnn						
Description:	If the Overfl program wil	ow bit is '0', t I branch.	hen the						
	added to the incremented instruction,	d to fetch the the new addr n. This instruc	e PC will have next ess will be						
Words:	1								
Cycles: 1(2)									
Q Cycle Activity: If Jump:									
Q1	Q2	Q3	Q4						
Decode	Read literal 'n'	Process Data	Write to PC						
No operation	No operation	No operation	No operation						
If No Jump:			•						
Q1	Q2	Q3	Q4						
Decode	Read literal 'n'	Process Data	No operation						
Example:	HERE	BNOV Jump	)						
Before Instruc PC After Instructio	= ade	dress (HERE	)						
	v = 0; = address (Jump) v = 1; = address (HERE + 2)								

BNZ	Branch if N	NOT Zero	
Syntax:	BNZ n		
Operands:	-128 ≤ n ≤ ′	127	
Operation:	if Zero bit is (PC) + 2 + 2		
Status Affected:	None		
Encoding:	1110	0001 nn	nn nnnn
Description:	If the Zero will branch.	bit is '0', then	the program
	added to th incremente instruction,	nplement nun e PC. Since th d to fetch the the new addr n. This instruction.	ne PC will hav next ess will be
Words:	1		
Cycles:	1(2)		
Q Cycle Activity: If Jump:			
Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	Write to PC
No operation	No operation	No operation	No operation
If No Jump:	•		<u> </u>
Q1	Q2	Q3	Q4
Decode	Read literal 'n'	Process Data	No operation
Example:	HERE	BNZ Jump	)
Before Instrue PC		dress (HERE	)

=	address (HERE)
=	0;
=	address (Jump)
=	1;
=	address (HERE + 2)
	= = = =

BRA		Unconditio	nal Branch					
Synta	ax:	BRA n						
Oper	ands:	-1024 ≤ n ≤	$-1024 \le n \le 1023$					
Oper	ation:	(PC) + 2 + 2	$(PC) + 2 + 2n \rightarrow PC$					
Statu	s Affected:	None						
Enco	oding:	1101	0nnn nn	nn nnnn				
Desc	ription:	the PC. Sin incremented instruction,	ce the PC wil d to fetch the the new addr n. This instruc	next ess will be				
Word	ls:	1						
Cycle	es:	2	2					
QC	ycle Activity:							
	Q1	Q2	Q3	Q4				
	Decode	Read literal 'n'	Process Data	Write to PC				
	No operation	No operation	No operation	No operation				
<u>Exan</u>	nple: Before Instruc PC After Instructio PC	= ade	BRA Jump dress (HERE dress (Jump	:)				

BSF	Bit Set f			
Syntax:	BSF f, b {	[,a}		
Operands:	$0 \le f \le 255$ $0 \le b \le 7$ $a \in [0,1]$			
Operation:	$1 \rightarrow \text{f}$			
Status Affected:	None			
Encoding:	1000	bbba	ffff	ffff
Description:	Bit 'b' in reg	gister 'f' is	s set.	
	lf 'a' is '0', t If 'a' is '1', t GPR bank.			
	If 'a' is '0' a set is enabl in Indexed mode wher Section 29 Bit-Oriente Literal Offe	led, this in Literal Of never f ≤ 9 .2.3 "Byt ed Instru	nstruction fset Addre 95 (5Fh). te-Oriente ctions in	operates essing See ed and Indexed
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3		Q4
Decode	Read register 'f'	Proce Data		Write gister 'f'
Example:	BSF F	LAG_RE	G, 7, 1	
Before Instruc FLAG_R After Instructic FLAG_R	EG = 0A on			

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BTFSC	Bit Test File	, Skip if Clear		BTFSS		Bit Test File	, Skip if Set	
Syntax:	BTFSC f, b	{,a}		Syntax:		BTFSS f, b {	,a}	
Operands:	$0 \leq f \leq 255$			Operan	ids:	$0 \leq f \leq 255$		
	0 ≤ b ≤ 7 a ∈ [0,1]					0 ≤ b < 7 a ∈ [0,1]		
Operation:	skip if (f <b>)</b>	= 0		Operati	on:	skip if (f <b>)</b>	= 1	
Status Affected:	None			Status /	Affected:	None		
Encoding:	1011	bbba ff	ff ffff	Encodir	ng:	1010	bbba ff:	ff ffff
Description:	instruction is the next inst current instru and a NOP is	gister 'f' is '0', skipped. If bit ruction fetched uction execution executed inst rcle instruction	'b' is '0', then I during the on is discarded ead, making	Descrip	otion:	instruction is the next instruction current instruction and a NOP is	gister 'f' is '1', t skipped. If bit ruction fetched uction executio executed instruction.	'b' is '1', then during the n is discarded ead, making
	,	e Access Banl BSR is used to	k is selected. If a select the				e Access Bank BSR is used to	is selected. If select the
	is enabled, t Indexed Lite whenever f ≤ Section 29.3 Bit-Oriented	d the extended his instruction ral Offset Addr ≤ 95 (5Fh). See 2.3 "Byte-Orie d Instructions et Mode" for d	essing mode ented and in Indexed			set is enable Indexed Lite whenever f ≤ Section 29.2 Bit-Oriented	d the extended d, this instructi ral Offset Addr g 95 (5Fh). See 2.3 "Byte-Orie I Instructions et Mode" for d	on operates in essing mode nted and in Indexed
Words:	1			Words:		1		
Cycles:		cles if skip and 2-word instruc		Cycles:			ycles if skip an a 2-word instru	
Q Cycle Activity:	by a				le Activity:	by t		00011.
Q1	Q2	Q3	Q4	QOyo	Q1	Q2	Q3	Q4
Decode	Read	Process	No		Decode	Read	Process	No
	register 'f'	Data	operation			register 'f'	Data	operation
lf skip:				If skip:				
Q1	Q2	Q3	Q4	Г	Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation		No operation	No operation	No operation	No operation
If skip and followe			operation		1	d by 2-word ins		operation
Q1	Q2	Q3	Q4		Q1	Q2	Q3	Q4
No	No	No	No		No	No	No	No
operation	operation	operation	operation		operation	operation	operation	operation
No operation	No operation	No operation	No operation		No operation	No operation	No operation	No operation
Example:	HERE B FALSE : TRUE :	IFSC FLAG	;, l, O	Exampl	l <u>e:</u>	HERE BJ FALSE : TRUE :	TFSS FLAG	, 1, 0
Before Instruc PC After Instructi If FLAG	= add on <1> = 0;	ress (HERE)			efore Instruc PC ter Instructic If FLAG<	tion = add on = 0;	ress (HERE)	
PC If FLAG• PC	< <b>1&gt; =</b> 1;	ress (TRUE) ress (FALSE	)		PC If FLAG< PC	:1> = 1;	ress (FALSE)	)

BTG		Bit Toggle	f		BOV		Branch if C	Overflow			
Syntax:		BTG f, b {,a	}		Synt	ax:	BOV n				
Operands:		$0 \leq f \leq 255$			Oper	ands:	-128 ≤ n ≤ ′	$-128 \le n \le 127$			
		0 ≤ b < 7 a ∈ [0,1]			Oper	ation:	if Overflow (PC) + 2 + 2				
Operation:		$(\overline{f{<}b{>}}) \to f{<}b{>}$		Statu	is Affected:	None					
Status Affect	cted:	None			Enco	oding:	1110	0100 nn	inn nnnn		
Encoding:		0111	bbba f	fff ffff	Desc	cription:	If the Overf	low bit is '1', t	then the		
Description	:	Bit 'b' in data memory location, 'f', is inverted.				•	program wi	ll branch.			
		,		ank is selected. ed to select the			added to th incremente instruction,	d to fetch the the new addr	ne PC will have next ress will be		
				ded instruction			PC + 2 + 2n. This instruction is then two-cycle instruction.				
		set is enabled, this instruction operates in Indexed Literal Offset Addressing		Word	ds:	1					
			ever $f \le 95$ (	5Fh). See priented and	Cycl	es:	1(2)				
		<b>Bit-Oriente</b>		ns in Indexed		ycle Activity: imp:					
Words:		1				Q1	Q2	Q3	Q4		
Cycles:		1				Decode	Read literal	Process	Write to PC		
Q Cycle Ad	ctivity:					No	ʻn' No	Data No	No		
2	ຊ1 ໌	Q2	Q3	Q4		operation	operation	operation	operation		
Dec	code	Read	Process	Write	If N	o Jump:			<u>.</u>		
		register 'f'	Data	register 'f'		Q1	Q2	Q3	Q4		
Example:		BTG P	ORTC, 4,	0		Decode	Read literal 'n'	Process Data	No operation		
Р	e Instruc PORTC	= 0111 (	)101 <b>[75h]</b>		Exar	nple:	HERE	BOV Jum <u>r</u>	þ		
	nstructio PORTC		0101 <b>[65h]</b>			Before Instruct PC After Instruction	= ad	dress (here	:)		
						If Overflo PC If Overflo PC	= ad ow = 0;	dress (Jump dress (HERE			

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ΒZ		Branch if 2	Zero		
Synta	ax:	BZ n			
Oper	ands:	-128 ≤ n ≤ ′	127		
Oper	ation:	if Zero bit is (PC) + 2 + 2	,	;	
Statu	s Affected:	None			
Enco	ding:	1110	0000	nnnr	n nnnn
Desc	ription:	If the Zero will branch.		then the	e program
		The 2's cor added to th incremente instruction, PC + 2 + 2r two-cycle ir	e PC. Sir d to fetch the new n. This in	nce the n the ne addres istructio	PC will have ext s will be
Word	ls:	1			
Cycle	es:	1(2)			
Q C If Ju	ycle Activity:				
	Q1	Q2	Q3	3	Q4
	Decode	Read literal	Proce	ess	Write to
		'n'	Data	-	PC
	No	No	No		No
If No	operation Jump:	operation	operat	.1011	operation
	Q1	Q2	Q3	R	Q4
	Decode	Read literal	Proce		No
		'n'	Data	a	operation
Exan	nple:	HERE	ΒZ	Jump	
	Before Instruc PC After Instructio	= ad	dress (1	HERE)	
	If Zero PC If Zero	= 1;	dress (	Jump)	
	PC		dress ()		

	Subroutine	e Call			
Syntax:	CALL k {,s	s}			
Operands:	$0 \le k \le 104$ s $\in$ [0,1]	8575			
Operation:	$\begin{array}{l} (PC) + 4 \rightarrow \\ k \rightarrow PC < 20 \\ \text{if } s = 1 \\ (W) \rightarrow WS \\ (STATUS) \rightarrow \\ (BSR) \rightarrow B \end{array}$	):1>; , → STATI	JSS,		
Status Affected:	None				
Encoding: 1st word (k<7:0>) 2nd word(k<19:8>)	1110 1111	110s k ₁₉ kkk	k ₇ k} kkk		kkkk ₍ kkkk
	If 's' = 1, th registers ar respective	re also pu shadow i	ushed	into	their
Words:	s IATUSS a update occ is loaded in two-cycle in 2	nto PC<2	n, the 2 0:1>. (	20-bi	, no t value '
	update occ is loaded in two-cycle in	urs. Ther nto PC<2	n, the 2 0:1>. (	20-bi	, no t value '
Cycles:	update occ is loaded in two-cycle in 2	urs. Ther nto PC<2	n, the 2 0:1>. (	20-bi	, no t value '
	update occ is loaded in two-cycle in 2	urs. Ther nto PC<2	n, the 2 0:1>. ແ າ.	20-bi	, no t value '
Cycles: Q Cycle Activity:	update occ is loaded in two-cycle in 2 2	urs. Ther nto PC<2 nstruction	n, the 2 0:1>. ( n. <u>}</u> C to	20-bi CALI Rea 'k'<	, no t value ' is a
Cycles: Q Cycle Activity: Q1	update occ is loaded ir two-cycle ir 2 2 Q2 Read literal	urs. Ther nto PC<2 nstruction Q3 Push P	n, the 2 0:1>. ( n. C to k	20-bi CALI Rea 'k'< Writ	, no t value ' is a Q4 ad literal <19:8>,
Cycles: Q Cycle Activity: Q1 Decode No	update occ is loaded ir two-cycle ir 2 2 Q2 Read literal 'k'<7:0>, No	urs. Ther no PC<2 nstruction Q3 Push P stac No	n, the 2 0:1>. ( n. C to k	Rea 'k'< Op	, no t value ' is a <u>Q4</u> ad litera <19:8>, <u>te to PC</u> No eration
Cycles: Q Cycle Activity: Q1 Decode No operation	update occ is loaded in two-cycle in 2 2 2 Read literal 'k'<7:0>, No operation HERE	urs. Ther no PC<2 nstruction Push P stac No operat	n, the 2 0:1>. ( ). C to k iion	Rea 'k'< Op	, no t value ' is a <u>Q4</u> ad literal <19:8>, te to PC No eration

CLRF	Clear f	CLRWDT	Clear Watchdog Timer
Syntax:	CLRF f {,a}	Syntax:	CLRWDT
Operands:	$0 \leq f \leq 255$	Operands:	None
	a ∈ [0,1]	Operation:	000h $\rightarrow$ WDT,
Operation:	$\begin{array}{l} \text{000h} \rightarrow \text{f}, \\ 1 \rightarrow \text{Z} \end{array}$	·	$\begin{array}{l} \text{000h} \rightarrow \text{WDT postscaler,} \\ 1 \rightarrow \overline{\text{TO}}, \end{array}$
Status Affected:	$\downarrow \rightarrow Z$		$1 \rightarrow \frac{10}{PD}$
Encoding:	- 0110 101a ffff ffff	Status Affected:	TO, PD
Description:	Clears the contents of the specified	Encoding:	0000 0000 0000 0100
Description.	register.	Description:	CLRWDT instruction resets the
	If 'a' is '0', the Access Bank is selected.	•	Watchdog Timer. It also resets the post-
	If 'a' is '1', the BSR is used to select the GPR bank.		scaler of the WDT. Status bits, TO and PD, are set.
	If 'a' is '0' and the extended instruction	Words:	1
	set is enabled, this instruction operates	Cycles:	1
	in Indexed Literal Offset Addressing	Q Cycle Activity:	
	mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and	Q1	Q2 Q3 Q4
	Bit-Oriented Instructions in Indexed	Decode	No Process No
	Literal Offset Mode" for details.		operation Data operation
Words:	1		
Cycles:	1	Example:	CLRWDT
Q Cycle Activity:		Before Instruc WDT Cor	
Q1	Q2 Q3 Q4	After Instructio	
Decode	Read Process Write	WDT Cou	
	register 'f' Data register 'f'		
		TO PD	= 1 = 1
Example:	CLRF FLAG_REG,1	1 D	1
Before Instruc			
FLAG_R After Instructio			
FLAG_R			

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COMF	Compleme	ent f		CPFSEQ	Compare f	with W, Skip	if f = W
Syntax:	COMF f	{,d {,a}}		Syntax:	CPFSEQ	f {,a}	
Operands:	0 ≤ f ≤ 255			Operands:	$0 \leq f \leq 255$		
opolaliaol	d ∈ [0,1]				a ∈ [0,1]		
	a ∈ [0,1]			Operation:	(f) – (W),		
Operation:	$\overline{f} \rightarrow dest$				skip if (f) =	. ,	
Status Affected:	N, Z					comparison)	
Encoding:	0001	11da ff:	ff ffff	Status Affected:	None		
Description:		its of register 'f		Encoding:	0110	001a ff	
Description.		nted. If 'd' is '0'		Description:			f data memory
	•	/. If 'd' is '1', th				to the contents an unsigned s	
	stored back	k in register 'f'.				ien the fetched	
	lf 'a' is '0', t	the Access Ba	nk is selected.		,	and a NOP is e	
		the BSR is use	d to select the			aking this a two	
	GPR bank.				instruction.		2
		and the extend			<b>lf 'a' is</b> '0', f	the Access Ba	nk is selected.
		led, this instruct	•		,		ed to select the
		Literal Offset A never f ≤ 95 (5			GPR bank.		
		0.2.3 "Byte-Or	,			and the extend	
		ed Instruction					ction operates
	Literal Off	set Mode" for	details.			Literal Offset / never f ≤ 95 (5	•
Words:	1					0.2.3 "Byte-Or	,
Cycles:	1					ed Instruction	
Q Cycle Activity:					Literal Off	set Mode" for	details.
Q1	Q2	Q3	Q4	Words:	1		
Decode	Read	Process	Write to	Cycles:	1(2)		
	register 'f'	Data	destination			cles if skip and 2-word instru	
				Q Cycle Activity:	by c		
Example:	COMF	REG, 0, 0		Q Cycle Activity. Q1	Q2	Q3	Q4
Before Instruc	ction			Decode	Read	Process	No
REG	= 13h				register 'f'	Data	operation
After Instruction REG	on = 13h			If skip:			
W	= ECh			Q1	Q2	Q3	Q4
				No	No	No	No
				operation If skip and followe	operation	operation	operation
				Q1	Q2	Q3	Q4
				No	No	No	No
				operation	operation	operation	operation
				No	No	No	No
				operation	operation	operation	operation
				Example:	HERE	CPFSEQ REG	G, 0
					NEQUAL	:	
					EQUAL	:	
				Before Instru			
				PC Add W	ress = HE = ?	IRE	
				REG	= ?		
				After Instruct	<u></u>		

W; Address (EQUAL) W; Address (NEQUAL)

= = ≠

After Instruction If REG PC If REG PC

CPFSGT	Compare f w	vith W, Skip	if f > W	CPF	SLT	Compare f	with W, Skip	if f < W	
Syntax:	CPFSGT f	{,a}		Synt	ax:	CPFSLT 1	⁻ {,a}		
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]			Oper	ands:	0 ≤ f ≤ 255 a ∈ [0,1]			
Operation:	(f) – (W), skip if (f) > (V (unsigned co			Oper	ation:	(f) – (W), skip if (f) <	(f) – (W), skip if (f) < (W) (unsigned comparison)		
Status Affected:	None			Stati	s Affected:	None			
Encoding:	0110 0	010a ff:	ff ffff		ding:	0110	000a ff	ff ffff	
Description:	location 'f' to the contents of the W by performing an unsigned subtraction.			cription:	0110 000a ffff fff: Compares the contents of data memory location 'f' to the contents of W by performing an unsigned subtraction.				
	contents of V instruction is executed ins two-cycle ins	VREG, then t discarded ar tead, making truction.	nd a NOP is I this a			If the contents of instruction i	nts of 'f' are le W, then the fe s discarded a stead, making	ss than the etched nd a NOP is	
	lf 'a' is '1', the GPR bank.	e BSR is use	nk is selected. d to select the			lf 'a' is '0', t	he Access Ba	nk is selected. d to select the	
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed		Word Cycle			vcles if skip ar a 2-word instru			
	Literal Offse	et Mode" for	details.	QU	Q1	Q2	Q3	Q4	
Words:	1				Decode	Read	Process	No	
Cycles:		cles if skip a		lf sk	ip:	register 'f'	Data	operation	
	by a	2-word instr	uction.		Q1	Q2	Q3	Q4	
Q Cycle Activity:	02	03	04		No	No	No	No	
Q1 Decode	Q2 Read	Q3 Process	Q4 No		operation	operation	operation	operation	
Decode	register 'f'	Data	operation	lf sk	ip and followe	d by 2-word in	struction:		
lf skip:					Q1	Q2	Q3	Q4	
Q1	Q2	Q3	Q4		No	No	No	No	
No	No	No	No		operation	operation	operation	operation	
operation	operation	operation	operation		No operation	No operation	No operation	No operation	
If skip and followe Q1	Q2	Q3	Q4		operation	operation	operation	operation	
No	No	No	No	Exar	nnlo:	HERE (	CPFSLT REG	1	
operation	operation	operation	operation		<u>npie.</u>		грын кыз,	, 1	
No	No	No	No				:		
operation	operation	operation	operation		Before Instruc	tion			
					PC		dress (HERE	)	
<u>Example:</u>	HERE	CPFSGT RE	G, 0		W	= ?			
	NGREATER	:			After Instructio				
Dofere last	GREATER	:			lf REG PC	< W; = Ad	dress (LESS	)	
Before Instruc PC W		ress (here	)		If REG PC	≥ W;			
After Instruction If REG PC If REG PC	on > W; = Add ≤ W;	ress (GREA ress (NGRE							

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DAW	Decimal A	djust W Regis	ter	DECF	Decremen	Decrement f		
Syntax:	DAW			Syntax:	DECF f{,	d {,a}}		
Operands:	None			Operands:	$0 \leq f \leq 255$			
Operation:		> 9] or [DC = 1 6 $\rightarrow$ W<3:0>;	], then		$\begin{array}{l} d \in [0,1] \\ a \in [0,1] \end{array}$			
	else			Operation:	$(f) - 1 \rightarrow de$	est		
	$(W<3:0>) \rightarrow W<3:0>$		$(W<3:0>) \rightarrow W<3:0>$ State		Status Affected:	C, DC, N, 0	DV, Z	
	If [W<7:4> > 9] or [C = 1], then		then	Encoding:	0000	01da fi	ff ffff	
	C = 1; else	6 → W<7:4>;		Description:	result is sto	register, 'f'. I pred in W. If 'c pred back in r		
<b>.</b>	(W<7:4>) –	→ W<7:4>			lf 'a' is '0', t	he Access Ba	ank is selected.	
Status Affected: Encoding:	C 0000	0000 000	0 0111		lf 'a' is '1', t GPR bank.		ed to select the	
Description: Words:	resulting fro variables (e	ts the 8-bit valu om the earlier a each in packed es a correct pa	ddition of two BCD format)		set is enab in Indexed mode wher Section 29 Bit-Oriente	led, this instru Literal Offset never f ≤ 95 ( 2.2.3 "Byte-O ed Instructio	5Fh). See riented and ns in Indexed	
					Literal Off	set Mode" fo	r details.	
Cycles:	1			Words:	1			
Q Cycle Activity: Q1	Q2	Q3	Q4	Cycles:	1			
Decode	Read	Process	Write	Q Cycle Activity:				
Decode	register W	Data	W	Q1	Q2	Q3	Q4	
Example 1:	DAW			Decode	Read register 'f'	Process Data	Write to destination	
Before Instruc	ction			Evenale	5565	c)	0	
W C	= A5h = 0			Example:		CNT, 1,	0	
БС	= 0 = 0			Before Instru CNT	iction = 01h			
After Instruction				Z	= 0			
W C	= 05h = 1			After Instruct	tion			
БС	= 0			CNT Z	= 00h = 1			
Example 2:				L	- 1			
Before Instruc	ction							
W	= CEh							
C DC	= 0 = 0							
After Instructi	0							
W	= 34h							
C	= 1							
DC	= 0							

DECFSZ	Decrement	t f, Skip if 0		DCFSNZ
Syntax:	DECFSZ 1	f {,d {,a}}		Syntax:
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$			Operands:
Operation:	(f) – $1 \rightarrow de$ skip if resul			Operation:
Status Affected:	None			Status Affected:
Encoding:	0010	11da ff:	ff ffff	Encoding:
Description:	decremente placed in W	ts of register ' ed. If 'd' is '0', /. If 'd' is '1', th < in register 'f'	the result is ne result is	Description:
	which is alr and a NOP i	is '0', the nex eady fetched s executed in le instruction.		
	,		nk is selected. ed to select the	
	set is enabl in Indexed mode wher Section 29 Bit-Oriente	ed, this instru- Literal Offset / lever f ≤ 95 (5 <b>.2.3 "Byte-O</b> r	Fh). See riented and ns in Indexed	
Words:	1			
Cycles:	1(2)			Words:
- )	Note: 3 cy	cles if skip an		Cycles:
	by a	2-word instru	uction.	
Q Cycle Activity: Q1	Q2	Q3	Q4	Q Cycle Activity
Decode	Read	Process	Write to	Q1
	register 'f'	Data	destination	Decode
If skip:				
Q1	Q2	Q3	Q4	If skip: Q1
No operation	No operation	No operation	No operation	No
If skip and followe			oporation	operation
Q1	Q2	Q3	Q4	If skip and follow
No	No	No	No	Q1
operation	operation	operation	operation	No
No	No	No	No	operation
operation	operation	operation	operation	No operation
Example:	HERE	DECFSZ GOTO	CNT, 1, 1 LOOP	Example:
	CONTINUE			
Before Instruc PC				Before Instr
After Instructio	,	6 (HERE)		TEMP
CNT	= CNT – ²	1		After Instruc
If CNT	= 0;			TEMP
PC	= Address	G (CONTINUE	Ε)	If TEM

DCFS	SNZ	Decrement	f, Skip if Not	t <b>O</b>
Synta	IX:	DCFSNZ	f {,d {,a}}	
Dpera	ands:	$0 \le f \le 255$		
		d ∈ [0,1] a ∈ [0,1]		
Opera	ation:	(f) – $1 \rightarrow de$ skip if resul		
Statu	s Affected:	None		
Enco		0100	11da fff	f ffff
	ription:	The conten decremente placed in W	ts of register " ed. If 'd' is '0', /. If 'd' is '1', th < in register 'f'.	f' are the result is ne result is
		instruction v discarded a	is not '0', the which is alread and a NOP is ex king it a two-c	ly fetched is xecuted
			ne Access Bar ne BSR is use	
		set is enabl in Indexed I mode when Section 29 Bit-Oriente	nd the extended, this instruct Literal Offset $A$ lever $f \le 95$ (5) <b>.2.3 "Byte-Or</b> <b>rd Instruction</b> <b>set Mode</b> " for	ction operates Addressing Fh). See <b>iented and</b> <b>s in Indexed</b>
Vord	s:	1		
Cycle	es:		ycles if skip ar a 2-word instr	
Q Cy	cle Activity:			
Г	Q1	Q2	Q3	Q4
	Decode	Read register 'f'	Process Data	Write to destination
lf ski	p:	· egietei ·	2010	
	Q1	Q2	Q3	Q4
Ī	No	No	No	No
	operation	operation	operation	operation
lf ski	-	d by 2-word in		
г	Q1	Q2	Q3	Q4
	No	No	No	No
ŀ	operation No	operation No	operation No	operation No
	operation	operation	operation	operation
Exam	nple:	ZERO	DCFSNZ TEN :	1P, 1, 0
	Before Instruc TEMP After Instructic	=	?	
	TEMP If TEMP PC If TEMP PC	= = = ≠	TEMP – 1, 0; Address (2 0; Address (1	

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GOT	0	Unconditio	onal Brai	nch		
Synta	ax:	GOTO k				
Oper	ands:	$0 \le k \le 104$	8575			
Oper	ation:	$k \rightarrow PC<20$	0:1>			
Statu	s Affected:	None				
	ding: ord (k<7:0>) vord(k<19:8>)	1110 1111	1111 k ₁₉ kkk	k ₇ kl kkk		kkkk ₀ kkkk ₈
Description: GOTO allows an unconditional branch anywhere within entire 2-Mbyte memory range. The 20-bit value 'k' is loaded into PC<20:1>. GOTO is always a two-cycle instruction.						
Word	s:	2				
Cycle	es:	2				
QC	ycle Activity:					
	Q1	Q2	Q3			Q4
	Decode	Read literal 'k'<7:0>,	No operat		'k'	ad literal <19:8>, te to PC
	No operation	No operation	No operat		ор	No eration
Example: GOTO THERE After Instruction PC = Address (THERE)						

INCF	Increment	f		
Syntax:	INCF f{,c	d {,a}}		
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \\ a \in [0,1] \end{array}$			
Operation:	(f) + 1 $\rightarrow$ d	est		
Status Affected:	C, DC, N,	OV, Z		
Encoding:	0010	10da	ffff	ffff
Description:	The conten incremente placed in V placed bac	d. If 'd' is V. If 'd' is	'0', the '1', the	e result is
	,	he BSR i		to selected.
	set is enab in Indexed mode wher Section 29	led, this in Literal Of never f ≤ 9 0.2.3 "Byt ed Instru	nstructi fset Ad 95 (5Fh te-Orie ctions	n). See nted and in Indexed
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3		Q4
Decode	Read register 'f'	Proce Data		Write to destination
Example:	INCF	CNT,	1, 0	
Before Instruc CNT Z DC After Instructio CNT	= FFh = 0 = ? = ?			

INCFSZ	Increment	f, Skip if 0		INFSNZ	Increment	f, Skip if Not	0
Syntax:	INCFSZ f	{,d {,a}}		Syntax:	INFSNZ f	{,d {,a}}	
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$			Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$		
Operation:	(f) + 1 $\rightarrow$ de skip if resul			Operation:	(f) + 1 $\rightarrow$ description of the skip if results		
Status Affected:	None			Status Affected:	None		
Encodina:	0011	11da ff	ff ffff	Encoding:	0100	10da ff	ff ffff
Description:	The conten incremente placed in W	ts of register ' d. If 'd' is '0', t /. If 'd' is '1', th < in register 'f'	f' are he result is ne result is	Description:	incremente placed in W placed bac	ts of register ' d. If 'd' is '0', t /. If 'd' is '1', th k in register 'f'	he result is ne result is
	which is alr and a NOP i	is '0', the nex eady fetched s executed in le instruction.			instruction discarded a	is not '0', the which is alread and a NOP is e aking it a two-o	dy fetched is xecuted
			nk is selected. ed to select the		,		nk is selected ed to select the
	set is enabl in Indexed mode when Section 29 Bit-Oriente Literal Offs	ed, this instru Literal Offset lever f ≤ 95 (5 .2.3 "Byte-O	Fh). See riented and is in Indexed		set is enab in Indexed mode wher Section 29 Bit-Oriente	led, this instru Literal Offset <i>i</i> never f ≤ 95 (5 . <b>2.3 "Byte-O</b> r	Fh). See riented and is in Indexed
Words:	1			Words:	1		
Cycles:		cycles if skip a a 2-word inst		Cycles:		/cles if skip ar a 2-word instru	
Q Cycle Activity:	.,			Q Cycle Activity:	~)		
Q1	Q2	Q3	Q4	Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination	Decode	Read register 'f'	Process Data	Write to destination
lf skip:	register i	Dulu	destinution	lf skip:	register i	Data	destination
Q1	Q2	Q3	Q4	Q1	Q2	Q3	Q4
No	No	No	No	No	No	No	No
operation	operation	operation	operation	operation	operation	operation	operation
If skip and followe	•		<u></u>	If skip and follow			
Q1 No	Q2 No	Q3 No	Q4 No	Q1	Q2	Q3	Q4
operation	operation	operation	operation	No operation	No operation	No operation	No operation
No	No	No	No	No	No	No	No
operation	operation	operation	operation	operation	operation	operation	operation
Example:	NZERO	INCFSZ CI : :	NT, 1, 0	Example:	HERE ZERO NZERO	INFSNZ RE(	G, 1, 0
Before Instruc	ction			Before Instru	ction		
PC After Instruction	= Address on	G (HERE)		PC After Instruct	ion	S (HERE)	
CNT If CNT	= CNT + ⁻ = 0;			REG If REG	= REG + ≠ 0;		
PC If CNT	= Address ≠ 0;	S (ZERO)		PC If REG	= Addres: = 0:	S (NZERO)	

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IORLW	Inclusive	Inclusive OR Literal with W				
Syntax:	IORLW k					
Operands:	$0 \le k \le 255$	5				
Operation:	(W) .OR. k	$\rightarrow$ W				
Status Affected:	N, Z					
Encoding:	0000	1001	kkkk	kkkk		
Description:	The contents of W are ORed with the eight-bit literal 'k'. The result is placed in W.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	3	Q4		
Decode	Read literal 'k'	Proce Data		Vrite to W		
Example:	IORLW	35h				
Before Instruc W	Before Instruction					

After Instruction W = BFh

IORWF	Inclusive C	OR W with f				
Syntax:	IORWF f	{,d {,a}}				
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$					
Operation:	(W) .OR. (f)	$\rightarrow$ dest				
Status Affected:	N, Z					
Encoding:	0001	00da ffi	ff ffff			
Description:	'0', the resu	R W with regis ilt is placed in placed back i	W. If 'd' is '1',			
	,	he Access Bar he BSR is use	nk is selected. d to select the			
	set is enabl in Indexed I mode when Section 29 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	Q4			
Decode	Read register 'f'	Process Data	Write to destination			
Example: IORWF RESULT, 0, 1						

Before Instructi	on	
RESULT	=	13h
W	=	91h
After Instruction	n	
RESULT	=	13h
W	=	93h

LFSF	2	Load FSR				
Synta	ax:	LFSR f, k				
Oper	ands:	$\begin{array}{l} 0 \leq f \leq 2 \\ 0 \leq k \leq 409 \end{array}$	5			
Oper	ation:	$k\toFSRf$				
Statu	s Affected:	None				
Enco	ding:	1110 1111	1110 0000	00f k ₇ kk		k ₁₁ kkk kkkk
Desc	ription:	The 12-bit file select r				
Word	ls:	2				
Cycle	es:	2				
QC	ycle Activity:					
	Q1	Q2	Q3			Q4
	Decode	Read literal 'k' MSB	Proce Data		lite M	Write eral 'k' ISB to [:] SRfH
	Decode	Read literal 'k' LSB	Proce Data			te literal o FSRfL
<u>Exan</u>	nple: After Instructio FSR2H FSR2L	LFSR 2, on = 03 = AE	h			

мол	/F	Move f						
Synt	ax:	MOVF f	{,d {,a}}					
Ope	rands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$	i					
Ope	ration:	$f \to \text{dest}$						
Statu	us Affected:	N, Z						
Enco	oding:	0101	00da	ffff	ffff			
Description: The contents of register 'f' are mo a destination dependent upon the status of 'd'. If 'd' is '0', the result placed in W. If 'd' is '1', the result placed back in register 'f'. Locatio can be anywhere in the 256-byte bank.				the sult is sult is				
		lf 'a' is '0', lf 'a' is '1', GPR bank	the BSR					
		If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.						
Word	ds:	1	1					
Cycl	es:	1						
QC	cycle Activity:							
	Q1	Q2	Q3	3	Q4			
	Decode	Read register 'f'	Proce Data		Write W			
Exar	<u>mple:</u>		EG, 0,	0				
	Before Instruct REG W	= 22 = F	2h Fh					
	After Instructio REG W	= 22	2h 2h					

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моу	'FF	Move f to f	ŧ			
Synta	ax:	MOVFF fs	,f _d			
Oper	ands:	$0 \le f_s \le 409$ $0 \le f_d \le 409$				
Oper	ation:	$(f_s) \to f_d$				
Statu	is Affected:	None				
1st w	oding: /ord (source) word (destin.)	1100 1111	ffff ffff	ffff ffff	ffff _s ffff _d	
Desc	ription:	The conten moved to d Location of in the 4096 FFFh) and can also be FFFh.	estination source 'f -byte dat location	n register ^f s' can be a space (l of destina	'f _d '. anywhere 000h to tion 'f _d '	
		Either sour (a useful sp			an be W	
		transferring peripheral i	MOVFF is particularly useful for transferring a data memory location to a peripheral register (such as the transmit buffer or an I/O port).			
		The MOVFF PCL, TOSU destination	J, TOSH			
Word	ls:	2				
Cycle	es:	2				
QC	ycle Activity:					
	Q1	Q2	Q3	i	Q4	
	Decode	Read register 'f' (src)	Proce Data		No peration	
	Decode	No operation No dummy read	No operat	ion re	Write gister 'f' (dest)	
Exan	nple:	MOVFF	REG1, F	EG2		
	Before Instruction REG1 = 33h REG2 = 11h After Instruction					
	REG1 REG2	= 33 = 33				

MOVLB	Move Liter	al to Lov	w Nibł	ble i	n BSR
Syntax:	MOVLB k				
Operands:	$0 \le k \le 255$				
Operation:	$k \to BSR$				
Status Affected:	None				
Encoding:	0000	0001	kkk	k	kkkk
Description:	The eight-b Bank Selec of BSR<7:4 regardless	t Registe I> always	er (BSF s rema	R). T ains '	he value 0'
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3			Q4
Decode	Read literal 'k'	Proce Data			te literal to BSR
Example:	MOVLB	5			
Before Instruct	tion				

Before Instruction 02h BSR Register = After Instruction BSR Register = 05h

ΜΟν	'LW	Move Lite	eral to W					
Synta	ax:	MOVLW	k					
Oper	ands:	$0 \le k \le 25$	5					
Oper	ation:	$k\toW$						
Statu	s Affected:	None						
Encoding:		0000	1110	kkk	k	kkkk		
Description:		The eight-	The eight-bit literal 'k' is loaded into W.					
Words:		1						
Cycle	es:	1	1					
QC	ycle Activity:							
	Q1	Q2	Q3	3		Q4		
	Decode	Read literal 'k'	Proce Data		W	/rite to W		
<u>Exar</u>	nple:	MOVLW	5Ah					
	After Instructic W	n = 5Ah						

MOVWF	Move W to	) f					
Syntax:	MOVWF	f {,a}					
Operands:	$0 \leq f \leq 255$						
	a ∈ [0,1]						
Operation:	$(W) \to f$	$(W) \rightarrow f$					
Status Affected:	None						
Encoding:	0110	111a	ffff	f ffff			
Description:	Move data Location 'f' 256-byte b	can be a	•				
	,	he BSR i		to selected.			
	If 'a' is '0' and the extended instruction set is enabled, this instruction operate in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexee Literal Offset Mode" for details.						
Words:	1						
Cycles:	1						
Q Cycle Activity:							
Q1	Q2	Q3		Q4			
Decode	Read	Proce		Write			
	register 'f'	Data	1	register 'f'			
Example:	MOVWF	reg, 0					
Before Instruc W REG After Instructio	= 4Fh = FFh						
W REG	= 4Fh = 4Fh						

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MULLW	Multiply L	iteral with W		MULWF	Multiply W w	ith f	
Syntax:	MULLW	k		Syntax:	MULWF f{	,a}	
Operands:	$0 \le k \le 255$	5		Operands:	$0 \leq f \leq 255$		
Operation:	(W) x k $\rightarrow$	PRODH:PRO	DL		a ∈ [0,1]		
Status Affected:	None			Operation:	(W) x (f) $\rightarrow$ Pl	RODH:PROD	L
Encoding:	0000	1101 kk	kk kkkk	Status Affected:	None		
Description:	out betwee 8-bit literal placed in th pair. PROE	'k'. The 16-bit he PRODH:PF )H contains th	s of W and the result is RODL register	Encoding: Description:	An unsigned in between the of register file lo stored in the l	contents of W cation 'f'. The PRODH:PRO	is carried out and the 16-bit result is
	W is uncha	•	<i></i>		W and 'f' are		iigii byto. Dotii
		e Status flags			None of the S	itatus flags ar	e affected.
	possible in		w nor Carry is . A Zero result ted.		Note that neit possible in thi possible but r	s operation. A	nor Carry is A Zero result is
Words:	1				If 'a' is '0', the	Access Bank	k is selected. If
Cycles: Q Cycle Activity:	1				'a' is '1', the E GPR bank.	SR is used to	select the
Q1 Decode	Q2 Read literal 'k'	Q3 Process Data	Q4 Write registers PRODH: PRODL		If 'a' is '0' and is enabled, th Indexed Litera whenever f ≤ Section 29.2. Bit-Oriented Literal Offset	is instruction al Offset Addr 95 (5Fh). See 3 <b>"Byte-Orie</b> Instructions	essing mode nted and in Indexed
Example:	MULLW	0C4h		Words:	1		
Before Instruc W	ction = E2	0h		Cycles:	1		
PRODH	= ?	211		Q Cycle Activity	:		
PRODL After Instruction	= ? on			Q1	Q2	Q3	Q4
W PRODH PRODL	= E2 = AI = 08	Dh		Decode	Read register 'f'	Process Data	Write registers PRODH: PRODL
				Example:	MULWF	REG, 1	PRODL

Before Instruction

W REG PRODH PRODL

After Instruction

REG PRODH PRODL

W

C4h B5h ? ?

C4h B5h

8Ah 94h

= = =

=

= = =

NEGF	Negate f				
Syntax:	NEGF f	{,a}			
Operands:	$0 \le f \le 255$ $a \in [0,1]$	5			
Operation:	$(\overline{f}) + 1 \rightarrow f$				
Status Affected:	N, OV, C,	DC, Z			
Encoding:	0110	110a	ffff	ffff	
Description:	Location 'f' is negated using two's complement. The result is placed in the data memory location 'f'.				
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
	If 'a' is '0' and the extended instruct set is enabled, this instruction opera in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented an Bit-Oriented Instructions in Index Literal Offset Mode" for details.				
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	3	Q4	

Q1	Q2	Q3	Q4
Decode	Read	Process	Write
	register 'f'	Data	register 'f'

Example:	NEGF	REG,	1
----------	------	------	---

Before Instru	ction			
REG	=	0011	1010	[3Ah]
After Instructi	on			
REG	=	1100	0110	[C6h]

NOP	NOP No Operation							
Synta	ax:	NOP						
Operands: None								
Oper	ation:	No operati	on					
Statu	s Affected:	None						
Encoding:		0000	0000	000	00	0000		
		1111	XXXX	XXX	x	XXXX		
Desc	ription:	No operati	No operation.					
Word	ls:	1	1					
Cycle	es:	1						
QC	ycle Activity:							
	Q1	Q2	Q	3		Q4		
	Decode	No	No			No		
		operation	operat	lion	ор	eration		

Example:

None.

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POP Pop Top of Return Stack					
Syntax:	POP				
Operands:	None				
Operation:	$(\text{TOS}) \rightarrow \text{b}$	it bucket			
Status Affected:	None				
Encoding:	0000	0000 000	00 0110		
Description:	stack and is then becon was pushe This instruc the user to	alue is pulled or s discarded. The nes the previou d onto the retu- ction is provide properly mana corporate a soft	ne TOS value is value that rn stack. d to enable ge the return		
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	No operation	POP TOS value	No operation		
Example:	POP GOTO	NEW			
Before Instru TOS Stack (*	iction I level down)	= 0031A = 01433			
After Instruct TOS PC	ion	= 01433 = NEW	2h		

PUSH		Push Top o	of Ret	urn Stac	k	
Syntax:		PUSH				
Operands:		None				
Operation:		$(PC + 2) \rightarrow$	TOS			
Status Affected	d:	None				
Encoding:		0000	0000	000	0	0101
Description:		The PC + 2 the return s value is pus This instruc software sta then pushin	tack. ⁻ shed d tion al ack by	The prev lown on t llows imp modifyir	ious the s blem ng Tr	TOS stack. enting a OS and
Words:		1				
Cycles:		1				
Q Cycle Activ	ity:					
Q1		Q2		Q3		Q4
Decod	е	PUSH PC + 2 onto		No ration	on	No eration
		return stack	opo	auon	99	cration
Example:					0	
Example: Before In: TOS PC		return stack	=	345Ah 0124h		

RCA	LL	Relative Ca	all			
Synta	ax:	RCALL n				
Oper	ands:	-1024 ≤ n ≤	1023			
Oper	ation:		$(PC) + 2 \rightarrow TOS,$ (PC) + 2 + 2n $\rightarrow$ PC			
Statu	s Affected:	None				
Enco	ding:	1101	1nnn	nnni	n nnnn	
Desc	ription:	Subroutine from the cu address (PC stack. Then number '2n' have incren instruction, PC + 2 + 2r two-cycle in	rrent loca C + 2) is and the to the Penented to the new the new the new	ation. F pushed 2's co C. Sinc fetch addres structic	irst, return d onto the omplement ce the PC will the next cs will be	
		•				
Cycle		2				
QC	ycle Activity: Q1	Q2	Q3		Q4	
	Decode	Read literal 'n' PUSH PC to stack	Proce Data	SS	Write to PC	
	No operation	No operation	No operat	ion	No operation	

RES	ET	Reset				
Synta	ax:	RESET				
Operands: None						
Oper	ation:		Reset all registers and flags that are affected by a MCLR Reset.			
Statu	s Affected:	All				
Enco	ding:	0000	0000	1111 1111		
Desc	ription:	This instruction of the text of te				
Word	ls:	1				
Cycle	es:	1				
QC	ycle Activity:					
	Q1	Q2	Q3	8		Q4
	Decode	Start	No			No
		reset	operat	ion	ор	eration

Example:

After Instruction

Registers =	Reset Value
Flags* =	Reset Value

RESET

Example: HERE RCALL Jump

Before Instruction

PC = Address (HERE) After Instruction PC = Address (Jump) TOS = Address (HERE + 2)

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RET	FIE	Return from	Return from Interrupt				
Synta	ax:	RETFIE {	5}				
Oper	ands:	s ∈ [0,1]					
Oper	ation:	$1 \rightarrow GIE/G$ if s = 1, (WS) $\rightarrow$ W, (STATUSS) (BSRS) $\rightarrow$	$(TOS) \rightarrow PC,$ 1 $\rightarrow$ GIE/GIEH or PEIE/GIEL;				
Statu	is Affected:	GIE/GIEH,	PEIE/GIEL.				
Enco	oding:	0000	0000 00	01 000s			
Desc	ription:	and Top-of- the PC. Inte setting eithe Global Inter contents of STATUSS a their corres	Return from interrupt. Stack is popped and Top-of-Stack (TOS) is loaded into the PC. Interrupts are enabled by setting either the high or low-priority Global Interrupt Enable bit. If 's' = 1, the contents of the shadow registers WS, STATUSS and BSRS are loaded into their corresponding registers W, STATUS and BSR. If 's' = 0, no update				
Word	ls:	1	,				
Cycle	es:	2					
2	ycle Activity:						
	Q1	Q2	Q3	Q4			
	Decode	No operation	No operation	POP PC from stack Set GIEH or GIEL			
	No	No	No	No			
<u>Exan</u>	After Interrupt PC W BSR STATUS	operation RETFIE	operation = TOS = WS = BSRS = STATU = 1				

Syntax:		RETLW k				
Operan	ds:	$0 \le k \le 255$	$0 \le k \le 255$			
Operatio	on:	$k \rightarrow W$ , (TOS) $\rightarrow P$ PCLATU, P		are uncha	nged	
Status A	ffected:	None				
Encodin	ig:	0000	1100	kkkk	kkkk	
Descrip	tion:	W is loaded with the 8-bit literal 'k'. T Program Counter is loaded from the t of the stack (the return address). The high address latch (PCLATH) remain unchanged.			om the top ss). The	
Words:		1				
Cycles:		2				
-	e Activity:					
a e j e.	Q1	Q2	Q3		Q4	
	Decode	Read literal 'k'	Proce: Data	fro	OP PC m stack, ite to W	
	No	No	No		No	
c	operation	operation	operati	ion op	peration	
Example						
CA	LL TABLE	; W conta: ; offset v		le		
		; W now ha				
		; table va				
:						
TABLE						
	DWF PCL TLW k0		; W = offset ; Begin table			
RE :	TLW kl	; begin ta	adie			
:	TLW kn	; End of t	- abla			

Before	Instruction	
14/	_	

W	=	07h
After Instructi	on	
W	=	value of kn

RET	URN	Return from Subroutine				
Synta	ax:	RETURN	RETURN {s}			
Oper	ands:	$s \in [0,1]$				
Oper	ation:	$(TOS) \rightarrow PC;$ if s = 1, $(WS) \rightarrow W,$ $(STATUSS) \rightarrow STATUS,$ $(BSRS) \rightarrow BSR,$ PCLATU, PCLATH are unchanged				
Statu	s Affected:	None				
Enco	ding:	0000	0000	0001	001s	
Desc	ription:	popped and is loaded in 's'= 1, the c registers W loaded into registers W	Return from subroutine. The stack is popped and the top of the stack (TOS) is loaded into the Program Counter. If 's'= 1, the contents of the shadow registers WS, STATUSS and BSRS are loaded into their corresponding registers W, STATUS and BSR. If 's' = 0, no update of these registers			
Word	ls:	1				
Cycle	es:	2				
QC	ycle Activity:					
	Q1	Q2	Q	3	Q4	
	Decode	No operation	Proce Dat		POP PC rom stack	
	No operation	No operation	No No No			
<u>Exan</u>	nple:	RETURN				

After Instruction: PC = TOS

Syntax:	RLCF f{	.d {.a}}		
Operands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$			
Operation:	$(f < n >) \rightarrow dest < n + 1 >,$ $(f < 7 >) \rightarrow C,$ $(C) \rightarrow dest < 0 >$			
Status Affected:	C, N, Z			
Encoding:	0011 01da ffff fff			
Description:	The content one bit to th If 'd' is '0', tl is '1', the re 'f'.	e left throug he result is sult is store	gh the C placed i d back i	Carry flag. in W. If 'd' in register
	If 'a' is '0', tl If 'a' is '1', tl GPR bank.			
	If 'a' is '0' a			
	in Indexed I mode when Section 29. Bit-Oriente Literal Offs	Literal Offse ever f ≤ 95 2.3 "Byte- d Instruction et Mode" f	et Addre (5Fh). S Oriente ons in I or detai	ssing See d and ndexed
	mode when Section 29 Bit-Oriente	Literal Offse ever f ≤ 95 2.3 "Byte- d Instruction et Mode" f	et Addre (5Fh). S Oriente ons in I	ssing See d and ndexed
Words:	mode when Section 29 Bit-Oriente Literal Offs	Literal Offse ever f ≤ 95 2.3 "Byte- d Instruction et Mode" f	et Addre (5Fh). S Oriente ons in I or detai	ssing See d and ndexed
Words: Cycles:	mode when Section 29. Bit-Oriente Literal Offs	Literal Offse ever f ≤ 95 2.3 "Byte- d Instruction et Mode" f	et Addre (5Fh). S Oriente ons in I or detai	ssing See d and ndexed
	mode when Section 29. Bit-Oriente Literal Offs	Literal Offse ever f ≤ 95 2.3 "Byte- d Instruction et Mode" f	et Addre (5Fh). S Oriente ons in I or detai	ssing See d and ndexed
Cycles:	mode when Section 29. Bit-Oriente Literal Offs	Literal Offse ever f ≤ 95 2.3 "Byte- d Instruction et Mode" f	et Addre (5Fh). S Oriente ons in I or detai	ssing See d and ndexed
Cycles: Q Cycle Activity:	mode when Section 29. Bit-Oriente Literal Offs C 1	Literal Offse ever f ≤ 95 .2.3 "Byte- d Instruction eet Mode" f regi	et Addre (5Fh). S Oriente ons in I for detai ister f	ssing See d and ndexed Is.
Cycles: Q Cycle Activity: Q1 Decode	mode when Section 29. Bit-Oriente Literal Offs C 1 1 1 Q2 Read register 'f'	iteral Offse ever f ≤ 95 <b>.2.3 "Byte-</b> <b>d Instructi</b> <b>set Mode</b> " f <b>d reg</b> <b>Q3</b> Process Data	et Addre (5Fh). S Oriente ons in I for detai ister f	Ssing See d and ndexed ls. Q4 Vrite to
Cycles: Q Cycle Activity: Q1 Decode Example:	mode when Section 29. Bit-Oriente Literal Offs C 1 1 1 2 Read register 'f' RLCF	Literal Offse ever f ≤ 95 .2.3 "Byte- d Instruction at Mode" f regional Q3 Process	et Addre (5Fh). S Oriente ons in I for detai ister f	Ssing See d and ndexed ls. Q4 Vrite to
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct	mode when Section 29. Bit-Oriente Literal Offs C 1 1 1 2 Q2 Read register 'f' RLCF	Literal Offse ever f ≤ 95 .2.3 "Byte- d Instruction aet Mode" f regional Q3 Processe Data REG,	et Addre (5Fh). S Oriente ons in I for detai ister f	Ssing See d and ndexed ls. Q4 Vrite to
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct REG C	mode when Section 29. Bit-Oriente Literal Offs C 1 1 1 Q2 Read register 'f' RLCF ction = 1110 = 0	Literal Offse ever f ≤ 95 .2.3 "Byte- d Instruction aet Mode" f regional Q3 Processe Data REG,	et Addre (5Fh). S Oriente ons in I for detai ister f	Ssing See d and ndexed ls. Q4 Vrite to
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct REG C After Instruction	mode when Section 29. Bit-Oriente Literal Offs C 1 1 1 Q2 Read register 'f' RLCF ction = 1110 = 0	Literal Offse ever f ≤ 95 .2.3 "Byte- d Instruction aet Mode" f regi Q3 Process Data REG, 0110	et Addre (5Fh). S Oriente ons in I for detai ister f	Ssing See d and ndexed ls. Q4 Vrite to
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct REG C	mode when Section 29. Bit-Oriente Literal Offs C 1 1 1 Q2 Read register 'f' RLCF ction = 1110 = 0	Literal Offse ever f ≤ 95 .2.3 "Byte- d Instruction tet Mode" f regional Q3 Process Data REG, 0110 0110	et Addre (5Fh). S Oriente ons in I for detai ister f	Ssing See d and ndexed ls. Q4 Vrite to

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RLNCF	Rotate Let	ft f (No Carry)			
Syntax:	RLNCF	f {,d {,a}}			
Operands:	$0 \le f \le 255$ d $\in [0,1]$ a $\in [0,1]$				
Operation:	$(f \le n >) \rightarrow d$ $(f \le 7 >) \rightarrow d$	est <n +="" 1="">, est&lt;0&gt;</n>			
Status Affected:	N, Z				
Encoding:	0100	01da ff:	ff ffff		
Description:	one bit to t is placed ir	The contents of register 'f' are rota one bit to the left. If 'd' is '0', the resu is placed in W. If 'd' is '1', the resu stored back in register 'f'.			
	,	he Access Bai he BSR is use			
	set is enab in Indexed mode whe Section 29 Bit-Orient	and the extend led, this instruct Literal Offset <i>i</i> never f ≤ 95 (5 <b>0.2.3 "Byte-O</b> r <b>ed Instruction</b> <b>set Mode</b> " for	ction operates Addressing Fh). See <b>riented and</b> is in Indexed		
	◀	register f			
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	Read register 'f'	Process Data	Write to destination		
Example:	RLNCF	REG, 1,	0		
Before Instruc REG After Instructio	= 1010 1	.011			
REG	= 0101 0	111			

RRCF	Rotate Rig	ght f through	Carry	/
Syntax:	RRCF f{	,d {,a}}		
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$			
Operation:	$(f < n >) \rightarrow d$ $(f < 0 >) \rightarrow C$ $(C) \rightarrow dest$			
Status Affected:	C, N, Z			
Encoding:	0011	00da f	fff	ffff
Description:	one bit to ti flag. If 'd' is If 'd' is '1', register 'f'.	nts of register he right throu s '0', the resul the result is p	igh the t is pla blaced	e Carry iced in W. back in
		the Access B the BSR is us		
		led, this instr	uction	operates
	set is enab in Indexed mode when Section 29 Bit-Oriente		uction Addre 5Fh). Driente ons in	operates essing See ed and Indexed
	set is enab in Indexed mode when Section 29 Bit-Oriente	led, this instr Literal Offset never f ≤ 95 ( 0.2.3 "Byte-C ed Instructio	uction Addre 5Fh). Driente ons in or deta	operates essing See ed and Indexed
Words:	set is enab in Indexed mode when Section 29 Bit-Oriento Literal Off	bled, this instr Literal Offset never $f \le 95$ ( 0.2.3 "Byte-C ed Instruction set Mode" for	uction Addre 5Fh). Driente ons in or deta	operates essing See ed and Indexed
Words: Cycles:	set is enab in Indexed mode when Section 29 Bit-Oriente Literal Off	bled, this instr Literal Offset never $f \le 95$ ( 0.2.3 "Byte-C ed Instruction set Mode" for	uction Addre 5Fh). Driente ons in or deta	operates essing See ed and Indexed
	set is enab in Indexed mode when Section 29 Bit-Oriente Literal Off C	bled, this instr Literal Offset never $f \le 95$ ( 0.2.3 "Byte-C ed Instruction set Mode" for	uction Addre 5Fh). Driente ons in or deta	operates essing See ed and Indexed
Cycles:	set is enab in Indexed mode when Section 29 Bit-Oriente Literal Off C	bled, this instr Literal Offset never $f \le 95$ ( 0.2.3 "Byte-C ed Instruction set Mode" for	uction Addre 5Fh). Driente ons in or deta	operates essing See ed and Indexed
Cycles: Q Cycle Activity:	set is enab in Indexed mode when Section 25 Bit-Oriente Literal Off C 1 1 1 2 Q2 Read	led, this instr Literal Offset never f ≤ 95 ( 0.2.3 "Byte-C ed Instructio set Mode" fo regist	uction Addre 5Fh). Driente ons in or deta ter f	operates essing See ed and Indexed ils.
Cycles: Q Cycle Activity: Q1	set is enab in Indexed mode when Section 29 Bit-Oriente Literal Off C 1 1 2 2	led, this instr Literal Offset never f ≤ 95 ( 0.2.3 "Byte-C ed Instructio set Mode" fc regist Q3 Process	uction Addre 5Fh). Driente ons in or deta ter f	operates essing See ed and Indexed ils. Q4 //rite to
Cycles: Q Cycle Activity: Q1 Decode Example:	set is enab in Indexed mode when Section 25 Bit-Oriente Literal Off C 1 1 1 Q2 Read register 'f'	led, this instr Literal Offset never f ≤ 95 ( 0.2.3 "Byte-C ed Instructio set Mode" fc regist Q3 Process	vection Addre 5Fh). Driente ons in or deta er f	operates essing See ed and Indexed ils. Q4 //rite to
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct	set is enab in Indexed mode when Section 25 Bit-Oriente Literal Off C 1 1 1 2 Q2 Read register 'f' RRCF	Q3 REG, 0,	vection Addre 5Fh). Driente ons in or deta er f	operates essing See ed and Indexed ils. Q4 //rite to
Cycles: Q Cycle Activity: Q1 Decode Example:	set is enab in Indexed mode when Section 25 Bit-Oriente Literal Off C 1 1 1 Q2 Read register 'f'	Q3 REG, 0,	vection Addre 5Fh). Driente ons in or deta er f	operates essing See ed and Indexed ils. Q4 //rite to
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct REG	set is enab in Indexed mode when Section 25 Bit-Oriente Literal Off C 1 1 1 2 Read register 'f' RRCF ction = 1110 = 0	Q3 REG, 0,	vection Addre 5Fh). Driente ons in or deta er f	operates essing See ed and Indexed ils. Q4 //rite to
Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct REG C	set is enab in Indexed mode when Section 29 Bit-Oriente Literal Off $\frown$ C 1 1 1 Q2 Read register 'f' RRCF Stion = 1110 = 0	Q3 Process Data REG, 0, 0110	vection Addre 5Fh). Driente ons in or deta er f	operates essing See ed and Indexed ils. Q4 //rite to

RRN	ICF	Rotate Ri	gŀ	nt f (No	Carry	)
Synt	ax:	RRNCF	f	[,d {,a}}		
Ope	rands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$	5			
Ope	ration:	$(f < n >) \rightarrow (f < 0 >) \rightarrow (f <$			L>,	
Statu	us Affected:	N, Z				
Enco	oding:	0100		00da	fff	f ffff
Deso	cription:	one bit to	th in	e right. W. If 'd'	lf 'd' is is '1',	' are rotated '0', the resul the result is
			ov 1 tł	erriding ne bank	the BS	nk will be SR value. If 'a e selected as
		set is ena in Indexed mode whe <b>Section 2</b>	ble d L ene 9.	ed, this i iteral O ever f ≤ 2.3 "By d Instru	instruc ffset A 95 (5F te-Ori Iction	ented and s in Indexed
		Г	•	. re	egister	f
Wor	ds:	1				
Cycl	es:	1				
QC	cycle Activity:					
	Q1	Q2		Q	3	Q4
	Decode	Read		Proce		Write to
		register 'f'		Data	а	destination
Exar	mple 1:	RRNCF	R	EG, 1,	, 0	
	Before Instruc REG After Instructio	= 1101	0	111		
	REG		1	011		
<u>Exar</u>	mple 2:	RRNCF	R	EG, 0,	, 0	
	Before Instruc W REG	tion = ? = 1101	0	111		
	After Instructio W	n = 1110	1	011		
	REG	= 1101	0	111		

SETF	Set f	Set f				
Syntax:	SETF f{,;	a}				
Operands:	$0 \leq f \leq 255$					
	a ∈ [0,1]	a ∈ [0,1]				
Operation:	$FFh \rightarrow f$					
Status Affected:	None					
Encoding:	0110	100a	ffff	ffff		
Description:	The conten are set to F		specified	register		
	lf 'a' is '0', t lf 'a' is '1', t GPR bank.					
	set is enabl in Indexed mode when Section 29 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3		Q4		
Decode	Read register 'f'	Proces Data		Write gister 'f'		
Example:	SETF	REC	G,1			
Before Instruct REG	ion = 5A	h				
After Instructio						
REG	= FF	ĥ				

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SLEEP	Enter Slee	Enter Sleep Mode			
Syntax:	SLEEP				
Operands:	None				
Operation:					
Status Affected:	TO, PD				
Encoding:	0000	0000 00	00 0011		
Description:	cleared. This set. The	r-Down status ne Time-out st Watchdog Tir are cleared	atus bit (TO)		
		ssor is put into cillator stoppe			
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	No	Process	Go to		
	operation	Data	Sleep		
Example:	SLEEP				
Before Instruc TO = PD =	tion ? ?				
After Instruction TO = PD =	on 1 † 0				
† If WDT causes v	wake-up, this t	oit is cleared.			

SUB	SUBFWB Subtract f from W with Borrow				rrow		
Syntax:		SUBFWB f {,d {,a}}					
Ope	rands:	$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$					
Ope	ration:	$(W) - (f) - (\overline{C}) \rightarrow dest$					
Statu	us Affected:	N, OV, C, DC, Z					
Enco	oding:	0101 01da ffff ffff					
Desc	cription:	Subtract register 'f' and Carry flag (borrow) from W (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored in register 'f'.					
		If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.					
If 'a' is '0' and the extended instruct set is enabled, this instruction oper- Indexed Literal Offset Addressing i whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented an Bit-Oriented Instructions in Inde Literal Offset Mode" for details.				n operates in essing mode nted and n Indexed			
Wor	ds:	1					
Cycl	es:	1					
QC	cycle Activity:						
	Q1	Q2	Q		Q4		
	Decode	Read register 'f'	Proc Dat		Write to destination		
Evar	nnle 1 [.]	SUBFWB	REG,		dootination		
Example 1: Before Instruc			REG,	1, 0			
	REG W C	= 3 = 2 = 1					
	After Instructio	•					
	REG W	= FF					
	C	= 2 = 0					
	Z	= 0 = 1 ···	ecult is	negativ	A.		
Exar	nple 2:	= 1 ; result is negative SUBFWB REG, 0, 0					
	Before Instruc			-, -			
	REG W C	= 2 = 5 = 1					
	After Instructio	n					
	REG W	= 2 = 3					
	С	= 1					
	Z N	= 0 : 1	esult is	positive	ć		
Example 3:		SUBFWB		1, 0	-		
	Before Instruction						
	REG W C	= 1 = 2 = 0					
	After Instruction	= 0					
	W C	= 2 = 1					
	Z N	= 1 ; i = 0	esult is	zero			

SUBLW	s	Subtract W from Literal					
Syntax:	S	SUBLW	k	(			
Operands:		$0 \le k \le 255$					
Operation:		$k-(W)\toW$					
Status Affected:		N, OV, C, DC, Z					
Encoding:		0000 1000 kkkk kkkk					kkkk
Description:		W is subtracted from the eight-bit literal 'k'. The result is placed in W.					
Words:	1	1					
Cycles:	1	1					
Q Cycle Activity:							
Q1		Q2		Q3			Q4
Decode		Read eral 'k'		Proce: Data		۷	Vrite to W
Example 1:	S	UBLW	C	)2h			
Before Instruc	ction						
W	=	01h					
C After Instructi	=	?					
W	=	01h					
Ċ	=	1	;।	result is p	oositiv	/e	
Z N	=	0 0					
Example 2:	S	UBLW	C	)2h			
Before Instruc							
W	W = 02h						
С	=	?					
	After Instruction						
W C	=	= 00h = 1 ; result is zero					
Ž	=	1	, 10001110 2010				
N	=	0					
Example 3: SUBLW 02h							
Before Instruction							
W C	=	03h ?					
After Instruction	on						
W	= FFh ; (2's complement)						
C Z	=	0 0	0 ; result is negative				
Ň	=	0 1					

SUBWF	Subtract W	from f			
Syntax:	SUBWF f {,d {,a}}				
Operands:	$0 \leq f \leq 255$				
	$d \in [0,1]$				
Operation	$a \in [0,1]$				
Operation: Status Affected:	$(f) - (W) \rightarrow dest$				
	N, OV, C, DC, Z				
Encoding:					
Description:	Subtract W from register 'f' (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.				
	If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	Read register 'f'	Process Data	Write to destination		
Evenue 4:	•		destination		
Example 1: Before Instruc	SUBWF	REG, 1, 0			
REG	= 3				
W C	= 2 = ?				
After Instruction					
	on				
REG	= 1				
REG W C	= 1 = 2 = 1 ;r	esult is positiv	e		
REG W	= 1 = 2	result is positiv	e		
REG W C Z	= 1 = 2 = 1 ;r = 0	result is positive	e		
REG W C Z N <u>Example 2:</u> Before Instruc	= 1 = 2 = 1 ; r = 0 = 0 SUBWF		e		
REG W C Z N Example 2:	= 1 = 2 = 1 ; r = 0 = 0 SUBWF		e		
REG W C Z N Example 2: Before Instruct REG W C After Instruction	= 1 = 2 = 1 ; r = 0 = 0 SUBWF tion = 2 = 2 = ? on		e		
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG W	= 1 = 2 = 1 ; r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ?	REG, 0, 0	e		
REG W C Z N Before Instruct REG W C After Instruction REG W C	= 1 = 2 = 1 ; r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ?		e		
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG W	= 1 = 2 = 1;r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ? on = 2 = 0;r	REG, 0, 0	e		
REG W C Z N Before Instruct REG W C After Instructio REG W C Z N N Example 3:	= 1 = 2 = 1;r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ? on = 1;r = 0 SUBWF	REG, 0, 0	e		
REG W C Z N Before Instruct REG W C After Instructio REG W C Z N N Example 3: Before Instruct	= 1 = 2 = 1;r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ? on = 1;r = 0 SUBWF	REG, 0, 0	e		
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG W C Z N Example 3: Before Instruct REG W	= 1 = 2 = 1 ; r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ? on = 1 ; r = 1 = 0 SUBWF tion = 1 = 2	REG, 0, 0	e		
REG W C Z N Before Instruct REG W C After Instruction REG W C Z N Example 3: Before Instruct REG	= 1 = 2 = 1 ; r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ? on = 1 ; r = 0 SUBWF tion = 1 = 2 = ?	REG, 0, 0	e		
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG W C After Instruction REG W C After Instruction REG W C After Instruction REG	= 1 = 2 = 1;r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ? on = 1;r = 0 SUBWF tion = 1 = 2 = ? on = 1;r = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0	REG, 0, 0			
REG W C Z N Example 2: Before Instruct REG W C After Instruction REG W C Example 3: Before Instruct REG W C After Instruction REG W C After Instruction	= 1 = 2 = 1; r = 0 = 0 SUBWF tion = 2 = 2 = ? on = 2 = ? on = 1; r = 0 SUBWF tion = 1 = 0 SUBWF tion = 1 = 2 = ? on = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0 = 0	REG, 0, 0 result is zero REG, 1, 0	nt)		

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SUBWFB	S	Subtract W from f with Borrow					
Syntax:		JBWFB	f {,d {,a}	}			
Operands:		$0 \le f \le 255$ $d \in [0,1]$ $a \in [0,1]$					
Operation:		$(f) - (W) - (\overline{C}) \rightarrow dest$					
Status Affected:		N, OV, C, DC, Z					
Encoding:		0101 10da ffff ff			ffff		
Description:		Subtract W and the Carry flag (borrow) from register 'f' (2's complement method). If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.					
		If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.					
	se in m Se Bi	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \le 95$ (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.					
Words:	1						
Cycles:	1						
Q Cycle Activity:							
Q1		Q2	Q	-	Q4		
Decode		Read gister 'f'	Proce Data		Write to destination		
Example 1:		SUBWFB	REG, 1				
Before Instruc		SODMED	1010, 1	, 0			
REG W C	= = =	19h 0Dh 1	(000) (000)				
After Instruction	on	_					
REG W	=	0Ch 0Dh	(000)	) 1011 ) 1101	,		
ç	=	1	(000)	5 1101	- /		
Z N	=	0 0	; resu	lt is pos	sitive		
Example 2:		SUBWFB	REG, 0				
Before Instruc	tion						
REG W	=	1Bh 1Ah		1 1011 1 1010			
С	=	0	(000.	r TOT(	,		
After Instructio REG W	on = =	1Bh 00h	(000	1 1011	)		
C Z N	= = =	1 1 0	; resu	lt is zer	0		
Example 3:	1	SUBWFB	REG, 1	, 0			
Before Instruc REG W	tion = = =	03h 0Eh		0 0011 0 1101			
C After Instructio		1					
REG	=	F5h	; [2's (	1 0100 comp]			
W C Z N	= = =	0Eh 0 0 1		0 1101 It is neg			
IN	-	Ŧ	, iesu	it is Heg	Jalive		

SWAPF	Swap f					
Syntax:	SWAPF f	SWAPF f {,d {,a}}				
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$	$\begin{array}{l} 0 \leq f \leq 255 \\ d \in [0,1] \end{array}$				
Operation:	$(f<3:0>) \rightarrow$ $(f<7:4>) \rightarrow$	$(f<3:0>) \rightarrow dest<7:4>,$ $(f<7:4>) \rightarrow dest<3:0>$				
Status Affected:	None	None				
Encoding:	0011	ff ffff				
Description:	'f' are excha	anged. If 'd' is W. If 'd' is '1',				
		If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.				
	set is enabl in Indexed I mode when Section 29 Bit-Oriente	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	Q4			
Decode	Read register 'f'	Process Data	Write to destination			
Example: Before Instruc REG After Instructio REG	tion = 53h	EG, 1, 0				

TBLI	RD	Table Read					
Synta	ax:	TBLRD ( *; *	·+; *	-; +*)			
Oper	ands:	None					
Oper	ation:	if TBLRD *, (Prog Mem (TBLPTR)) $\rightarrow$ TABLAT; TBLPTR – No Change if TBLRD *+, (Prog Mem (TBLPTR)) $\rightarrow$ TABLAT; (TBLPTR) + 1 $\rightarrow$ TBLPTR if TBLRD *-, (Prog Mem (TBLPTR)) $\rightarrow$ TABLAT; (TBLPTR) – 1 $\rightarrow$ TBLPTR if TBLRD +*, (TBLPTR) + 1 $\rightarrow$ TBLPTR; (Prog Mem (TBLPTR)) $\rightarrow$ TABLAT					
Statu	s Affected:	None					
Enco	ding:	0000	0	000	000	00	10nn nn=0 * =1 *+ =2 *- =3 +*
Desc	ription:	This instruct of Program I program me Pointer (TBL	Men mor	nory (F y, a po	P.M.).	To ad	dress the
		The TBLPTF each byte in has a 2-Mby	the	progra	am me	mory	
		TBLPTR<		Ρ	rogran	n Mei	mory Word
		TBLPTR<	)> =				t Byte of mory Word
		The TBLRD i of TBLPTR a				nodify	the value
		no change	е				
		<ul> <li>post-incre</li> </ul>	mei	nt			
		<ul> <li>post-decr</li> </ul>	eme	ent			
<ul> <li>pre-increment</li> </ul>							
Words: 1							
Cycles: 2							
QC	ycle Activity						
	Q1	Q2		1	3		Q4
	Decode	No operation		N opera		ор	No eration

No operation (Write TABLAT)

No

operation

TBLRD	Table Re	ead	(Cor	itinued)
Example 1:	TBLRD	*+	;	
Before Instruction TABLAT TBLPTR MEMORY After Instruction TABLAT TBLPTR	(00A356h	)	= = =	55h 00A356h 34h 34h 00A357h
Example 2:	TBLRD	+*	;	
Before Instruction TABLAT TBLPTR MEMORY MEMORY After Instruction	(01A357h (01A358h		= = =	12h 34h
TABLAT TBLPTR			=	34h 01A358h

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No operation (Read Program

Memory)

No

operation

TBLWT	Table Wri	te			
Syntax:	TBLWT ( '	*; *+; *-; +*	f)		
Operands:	None				
Operation:	if TBLWT*, (TABLAT) → Holding Register; TBLPTR – No Change if TBLWT*+,				
	(TABLAT) (TBLPTR) if TBLWT* (TABLAT) (TBLPTR)	$() + 1 \rightarrow TE$ (-, -) $\rightarrow$ Holding	BLPTR g Register		
	if TBLWT (TBLPTR)				
Status Affected:	None		gitegister		
		0000	0000	11	
Encoding:	0000	0000	0000	11nn nn=0 *	
				=1 *+	
				=2 *-	
				=3 +*	
Description:		uction uses			
				T is written	
	to. The ho	•			
		0 0		am Memory	
	(P.M.). (Re				
	Organizat programm			details on	
		in the pro nas a 2-Mb of the TBL	gram men byte addre PTR selec	nory. ess range. ets which	
	byte of the access.	e program	memory I	ocation to	
	TBLPT			ficant Byte n Memory	
	TBLPT		lost Signif Program I Word	icant Byte of Memory	
	The TBLW value of T	⊤ instruct BLPTR as		odify the	
	<ul> <li>no char</li> </ul>	nge			
	• post-ind	crement			
	<ul> <li>post-de</li> </ul>	crement			
	<ul> <li>pre-inci</li> </ul>	rement			
Words:	1				
Cycles:	2				
Q Cycle Activity:					
	Q1	Q2	Q3	Q4	
	Decode	No	No	No	
	200000		operation	operation	
	No	No	No	No	
	operation	operation	operation	operation	
		(Read		(Write to	
		TABLAT)		Holding Register)	

Register)

			n
TBLWT	Table Write (C	ontin	ued)
Example 1:	TBLWT *+;		
Before Instru	ction		
TABLAT TBLPTF HOLDIN		= =	55h 00A356h
(00A350	6h)	=	FFh
After Instruct	ions (table write	comp	letion)
		=	55h 00A357h
(00A350		=	55h
Example 2:	TBLWT +*;		
Before Instru	ction		
		= =	34h 01389Ah
(01389/		=	FFh
(013896		=	FFh
After Instruct	ion (table write c	omple	etion)
TABLAT TBLPTF		= =	34h 01389Bh
(01389/		=	FFh
(013898		=	34h

TSTFSZ	Test f, Skip	o if O			
Syntax:	TSTFSZ f {	,a}			
Operands:	0 ≤ f ≤ 255 a ∈ [0,1]				
Operation:	skip if f = 0				
Status Affected:	None				
Encoding:	0110	011a fff	f ffff		
Description:	during the o	If 'f' = 0, the next instruction fetched during the current instruction execution is discarded and a NOP is executed, making this a two-cycle instruction.			
	,	he Access Bar he BSR is used			
	If 'a' is '0' and the extended instruction set is enabled, this instruction operate in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1				
Cycles:	•	vcles if skip and a 2-word instru			
Q Cycle Activity:					
Q1	Q2	Q3	Q4		
Decode	Read	Process	No		
lf skip:	register 'f'	Data	operation		
Q1	Q2	Q3	Q4		
No	No	No	No		
operation	operation	operation	operation		
If skip and followe	d by 2-word in	struction:			
Q1	Q2	Q3	Q4		
No	No	No	No		
operation	operation	operation	operation		
No operation	No operation	No operation	No operation		
Example:	NZERO	ISTFSZ CNT : :	, 1		
Before Instruction PC = Address (HERE) After Instruction					
If CNT PC If CNT PC	= 00 = Ad ≠ 00	dress (ZERO	,		

XORLW	Exclusive	e OR Liter	ral with \	N
Syntax:	XORLW	k		
Operands:	$0 \le k \le 25$	5		
Operation:	(W) .XOR	$k \to W$		
Status Affected:	N, Z			
Encoding:	0000	1010	kkkk	kkkk
Description:		ents of W a teral 'k'. T		
Words:	1			
Cycles:	1			
Q Cycle Activity:				
Q1	Q2	Q3		Q4
Decode	Read literal 'k'	Proces Data		Vrite to W
Example:	XORLW	0AFh		
Before Instruction W = B5h After Instruction W = 1Ah				

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XORWF	Exclusive OR W with f				
Syntax:	XORWF	f {,d {,a}}			
Operands:	$\begin{array}{l} 0 \leq f \leq 255 \\ d  \in  [0,1] \\ a  \in  [0,1] \end{array}$				
Operation:	(W) .XOR. (	(f) $\rightarrow$ dest			
Status Affected:	N, Z				
Encoding:	0001	10da	ffff	ffff	
Description:	Exclusive C register 'f'. I in W. If 'd' is in the regist	f 'd' is '0', t s '1', the res	he resul	t is stored	
	lf 'a' is '0', ti If 'a' is '1', ti GPR bank.				
	If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 29.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.				
Words:	1				
Cycles:	1				
Q Cycle Activity:					
Q1	Q2	Q3		Q4	
Decode	Read register 'f'	Process Data		Vrite to stination	
Example:	XORWF H	REG, 1,	0		
Before Instruct					
REG W	= AFh = B5h				
After Instruction REG W	n = 1Ah = B5h				

#### 29.2 Extended Instruction Set

In addition to the standard 75 instructions of the PIC18 instruction set, the PIC18F87K22 family of devices also provides an optional extension to the core CPU functionality. The added features include eight additional instructions that augment Indirect and Indexed Addressing operations and the implementation of Indexed Literal Offset Addressing for many of the standard PIC18 instructions.

The additional features of the extended instruction set are enabled by default on unprogrammed devices. Users must properly set or clear the XINST Configuration bit during programming to enable or disable these features.

The instructions in the extended set can all be classified as literal operations, which either manipulate the File Select Registers, or use them for Indexed Addressing. Two of the instructions, ADDFSR and SUBFSR, each have an additional special instantiation for using FSR2. These versions (ADDULNK and SUBULNK) allow for automatic return after execution.

The extended instructions are specifically implemented to optimize re-entrant program code (that is, code that is recursive or that uses a software stack) written in high-level languages, particularly C. Among other things, they allow users working in high-level languages to perform certain operations on data structures more efficiently. These include:

- Dynamic allocation and deallocation of software stack space when entering and leaving subroutines
- Function Pointer invocation
- Software Stack Pointer manipulation
- Manipulation of variables located in a software stack

A summary of the instructions in the extended instruction set is provided in Table 29-3. Detailed descriptions are provided in Section 29.2.2 "Extended Instruction Set". The opcode field descriptions in Table 29-1 (page 432) apply to both the standard and extended PIC18 instruction sets.

Note: The instruction set extension and the Indexed Literal Offset Addressing mode were designed for optimizing applications written in C; the user may likely never use these instructions directly in assembler. The syntax for these commands is provided as a reference for users who may be reviewing code that has been generated by a compiler.

#### 29.2.1 EXTENDED INSTRUCTION SYNTAX

Most of the extended instructions use indexed arguments, using one of the File Select Registers and some offset to specify a source or destination register. When an argument for an instruction serves as part of Indexed Addressing, it is enclosed in square brackets ("[]"). This is done to indicate that the argument is used as an index or offset. The MPASM[™] Assembler will flag an error if it determines that an index or offset value is not bracketed.

When the extended instruction set is enabled, brackets are also used to indicate index arguments in byte-oriented and bit-oriented instructions. This is in addition to other changes in their syntax. For more details, see Section 29.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands".

Note: In the past, square brackets have been used to denote optional arguments in the PIC18 and earlier instruction sets. In this text and going forward, optional arguments are denoted by braces ("{ }").

Mnemonic, Operands		Description	Cycles	16-Bit Instruction Word				Status
		Description	Cycles	MSb			LSb	Affected
ADDFSR	f, k	Add Literal to FSR	1	1110	1000	ffkk	kkkk	None
ADDULNK	k	Add Literal to FSR2 and Return	2	1110	1000	11kk	kkkk	None
CALLW		Call Subroutine using WREG	2	0000	0000	0001	0100	None
MOVSF	z _s , f _d	Move z _s (source) to 1st word	2	1110	1011	0zzz	ZZZZ	None
		f _d (destination) 2nd word		1111	ffff	ffff	ffff	
MOVSS	z _s , z _d	Move z _s (source) to 1st word	2	1110	1011	lzzz	ZZZZ	None
		z _d (destination) 2nd word		1111	xxxx	XZZZ	ZZZZ	
PUSHL	k	Store Literal at FSR2,	1	1110	1010	kkkk	kkkk	None
		Decrement FSR2						
SUBFSR	f, k	Subtract Literal from FSR	1	1110	1001	ffkk	kkkk	None
SUBULNK	k	Subtract Literal from FSR2 and	2	1110	1001	11kk	kkkk	None
		return						

#### TABLE 29-3: EXTENSIONS TO THE PIC18 INSTRUCTION SET

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#### 29.2.2 EXTENDED INSTRUCTION SET

ADD	FSR	Add Liter	al to FSF	ł			
Synta	ax:	ADDFSR	f, k				
Oper	ands:	$0 \le k \le 63$					
		f ∈ [ 0, 1,	2]				
Oper	ation:	FSR(f) + k	$x \rightarrow FSR($	f)			
Statu	s Affected:	None					
Enco	ding:	1110	1000	ffkk	kkkk		
Desc	ription:		The 6-bit literal 'k' is added to the contents of the FSR specified by 'f'.				
Word	ls:	1	1				
Cycle	es:	1					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read	Proces	s	Write to		
		literal 'k'	Data		FSR		

Example: ADDFSR 2, 23h

Before Instruction FSR2 = 03FFh After Instruction FSR2 = 0422h

ADD	ULNK	Add Liter	al to FSR2	2 and R	eturn		
Synta	ax:	ADDULN	K k				
Oper	ands:	$0 \le k \le 63$					
Oper	ation:	FSR2 + k	$\rightarrow$ FSR2,				
		$(TOS) \rightarrow I$	С				
Statu	s Affected:	None					
Enco	ding:	1110	1000	11kk	kkkk		
Desc	ription:	contents o	The 6-bit literal 'k' is added to the contents of FSR2. A RETURN is then executed by loading the PC with the TOS.				
		The instru execute; a the second	NOP is pe				
This may be thought of as a spec case of the ADDFSR instruction, where f = 3 (binary '11'); it operat only on FSR2.				ion,			
Word	ls:	1	1				
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Read literal 'k'	Process Data	s \	Write to FSR		
	No	No	No		No		
	Operation	Operation	Operatio	n O	peration		
<u>Exan</u>	nple:	ADDULNK 2	23h				

ampie:	AL	DULNK	2
Before Instruc	tion		
FSR2	=	03FFh	
PC	=	0100h	
After Instruction	n		
FSR2	=	0422h	
PC	=	(TOS)	

**Note:** All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction format then becomes: {label} instruction argument(s).

CALLW Subroutine Call Using WRE				VREG			
Synta	ax:	CALLW					
Oper	ands:	None					
Oper	ation:	(W) → PCL (PCLATH) -	$(PC + 2) \rightarrow TOS,$ $(W) \rightarrow PCL,$ $(PCLATH) \rightarrow PCH,$ $(PCLATU) \rightarrow PCU$				
Statu	is Affected:	None					
Enco	oding:	0000	0000 000	01 0100			
Description First, the return address (PC + 2) is pushed onto the return stack. Next, contents of W are written to PCL; th existing value is discarded. Then, th contents of PCLATH and PCLATU a latched into PCH and PCU, respectively. The second cycle is executed as a NOP instruction while new next instruction is fetched.				ack. Next, the to PCL; the d. Then, the PCLATU are U, cycle is ction while the			
		Unlike CAL: update W, S	Unlike CALL, there is no option to update W, STATUS or BSR.				
Word	ls:	1	1				
Cycle	es:	2	2				
QC	ycle Activity:						
	Q1	Q2	Q3	Q4			
	Decode	Read WREG	Push PC to stack	No operation			
	No	No	No	No			
	operation	operation	operation	operation			
$\begin{array}{ccccc} Example: & \text{HERE} & \text{CALLW} \\ \hline \\ Before Instruction \\ PC & = & address (HERE) \\ PCLATH & = & 10h \\ PCLATU & = & 00h \\ W & = & 06h \\ \hline \\ After Instruction \\ PC & = & 001006h \\ TOS & = & address (HERE + 2) \\ PCLATH & = & 10h \\ PCLATU & = & 00h \\ W & = & 06h \\ \end{array}$							

ΜΟν	SF	Move Inde	xed to f				
Synta	ax:	MOVSF [2	z _s ], f _d				
Operands:		•	$\begin{array}{l} 0 \leq z_{s} \leq 127 \\ 0 \leq f_{d} \leq 4095 \end{array}$				
Oper	ation:	((FSR2) + 2	$(z_s) \rightarrow f_d$				
Statu	s Affected:	None					
1st w	ding: vord (source) word (destin.)	1110 1111		zzz fff	zzzz _s ffff _d		
Desc	ription:	moved to d actual addr determined offset ' $z_s$ ', i of FSR2. TI register is s 'f _d ' in the se can be any space (000 The MOVSF	The contents of the source register are moved to destination register 'f _d '. The actual address of the source register is determined by adding the 7-bit literal offset ' $z_s$ ', in the first word, to the value of FSR2. The address of the destination register is specified by the 12-bit literal 'f _d ' in the second word. Both addresses can be anywhere in the 4096-byte data space (000h to FFFh). The MOVSF instruction cannot use the				
		destination	J, TOSH or 1 register. ant source a				
		an Indirect	Addressing in and will be 0	registe	•		
Word	ls:	2	2				
Cycle	es:	2					
QC	ycle Activity:						
	Q1	Q2	Q3		Q4		
	Decode	Determine source addr	Determine source add		Read urce reg		
	Decode	No operation No dummy read	No operation	re	Write gister 'f' (dest)		
<u>Exan</u>	nple:	MOVSF	[05h], RE0	32			
	Before Instruc FSR2 Contents of 85h REG2 After Instructio FSR2 Contents of 85h REG2	= 80 = 33 = 11 on = 80	h h h				

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MOVSS	Move Indexed to Indexed				
Syntax:	MOVSS [z _s ], [z _d ]				
Operands:	$\begin{array}{l} 0 \leq z_s \leq 127 \\ 0 \leq z_d \leq 127 \end{array}$				
Operation:	$((FSR2) + z_s) \rightarrow ((FSR2) + z_d)$				
Status Affected:	None				
Encoding: 1st word (source) 2nd word (dest.) Description	11101011 $1zzz$ $zzzz_s$ 1111 $xxxx$ $xzzz$ $zzzz_d$ The contents of the source register are moved to the destination register. The addresses of the source and destination registers are determined by adding the 7-bit literal offsets, ' $z_s$ ' or ' $z_d$ ', respectively, to the value of FSR2. Both registers can be located anywhere in the 4096-byte data memory space (000h to FFFh).				
	The MOVSS instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register.				
	If the resultant source address points t an Indirect Addressing register, the value returned will be 00h. If the resultant destination address points to an Indirect Addressing register, the instruction will execute as a NOP.				
Words:	2				
Cycles:	2				
Q Cycle Activity:					

PUSHL Store Literal at FSR2, Decrement FSR2					
Syntax:	PUSHL k	PUSHL k			
Operands:	$0 \le k \le 255$				
Operation:	$k \rightarrow (FSR2),$ FSR2 – 1 $\rightarrow$ FSR2				
Status Affected:	None				
Encoding:	1110	1010	kkkk	kkkk	
	memory add FSR2 is decr	•			
	operation. This instructi values onto a			to push	
Words:	This instructi			to push	
	This instructi values onto a			to push	
Words: Cycles: Q Cycle Activity:	This instructi values onto a 1			to push	
Cycles:	This instructi values onto a 1	a softwa		to push	

Example: PUSHL 08h

Before Instruction FSR2H:FSR2L Memory (01ECh)	= =	01ECh 00h
After Instruction FSR2H:FSR2L Memory (01ECh)	= =	01EBh 08h

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Determine	Determine	Read
	source addr	source addr	source reg
Decode	Determine dest addr	Determine dest addr	Write to dest reg

Example: MOVSS [05h], [06h]
-----------------------------

=	80h
=	33h
=	11h
=	80h
	0.01
=	33h
=	33h
	= = = =

SUBULNK k

 $FSR2 - k \rightarrow FSR2$ ,

 $0 \leq k \leq 63$ 

Subtract Literal from FSR2 and Return

SUBFSR	Subtract	Subtract Literal from FSR				
Syntax:	SUBFSR	SUBFSR f, k				
Operands:	$0 \le k \le 63$	$0 \le k \le 63$				
	f ∈ [ 0, 1,	2]				
Operation:	FSRf – k	$\rightarrow$ FSRf				
Status Affected:	None					
Encoding:	1110	1001	ffk}	2	kkkk	
Description:	The 6-bit l the conter by 'f'.					
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	Q3		Q4	
Decode	Read	Proce	SS	W	rite to	
	register 'f'	Data	1	dest	ination	
Example: SUBFSR 2, 23h Before Instruction						

03FFh

03DCh

FSR2

After Instruction FSR2

=

=

ope.		$(TOS) \rightarrow PC$				
Statu	s Affected:	None				
Enco	ding:	1110	1001	11kk	kkkk	
Desc	ription:	The 6-bit literal 'k' is subtracted from the contents of the FSR2. A RETURN is then executed by loading the PC with the TOS.				
		The instruction takes two cycles to execute; a NOP is performed during the second cycle.				
		This may b of the SUB (binary '11	FSR instru	ction, wh		
Word	Words:					
Cycle	Cycles:					
QC	Q Cycle Activity:					
	Q1	Q2	(	23	Q4	
	Decode	Read register		cess ata	Write to destination	
	No Operation	No Operatio		lo ration	No Operation	
	-	_		-		

Example: SUBULNK 23h

SUBULNK

Operands:

Operation:

Syntax:

Before Instruction					
FSR2	=	03FFh			
PC	=	0100h			
After Instruction					
FSR2	=	03DCh			
PC	=	(TOS)			

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#### 29.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note:	Enabling the PIC18 instruction set exten-					
	sion may cause legacy applications to					
	behave erratically or fail entirely.					

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing (Section 6.6.1 "Indexed Addressing with Literal Offset"). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank (a = 0) or in a GPR bank designated by the BSR (a = 1). When the extended instruction set is enabled and a = 0, however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bit-oriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between C and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see Section 29.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands").

Although the Indexed Literal Offset mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind, that when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

### 29.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument 'f' in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value 'k'. As already noted, this occurs only when 'f' is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets ("[]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within the brackets, will generate an error in the MPASM[™] Assembler.

If the index argument is properly bracketed for Indexed Literal Offset Addressing, the Access RAM argument is never specified; it will automatically be assumed to be '0'. This is in contrast to standard operation (extended instruction set disabled), when 'a' is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM Assembler.

The destination argument, 'd', functions as before.

In the latest versions of the MPASM Assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option,  $/_{Y}$ , or the PE directive in the source listing.

#### 29.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18F87K22 family, it is very important to consider the type of code. A large, re-entrant application that is written in C and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

ADDWF	ADD W to I (Indexed Li	ndexed iteral Offset r	node)			
Syntax:	ADDWF	[k] {,d}				
Operands:	$\begin{array}{l} 0 \leq k \leq 95 \\ d  \in  [0,1] \end{array}$					
Operation:	(W) + ((FSF	$(W) + ((FSR2) + k) \rightarrow dest$				
Status Affected:	N, OV, C, D	C, Z				
Encoding:	0010	01d0 kkł	k kkkk			
Description:	contents of	ts of W are ad the register in t by the value	dicated by			
	,	ne result is sto sult is stored l	red in W. If 'd' back in			
Words:	1					
Cycles:	1					
Q Cycle Activity:						
Q1	Q2	Q3	Q4			
Decode	Read 'k'	Process Data	Write to destination			
Example:	ADDWF [	[OFST],0				
Before Instructio W OFST FSR2 Contents of 0A2Ch After Instruction W Contents of 0A2Ch	= = =	17h 2Ch 0A00h 20h 37h 20h				

BSF	Bit Set Indexed (Indexed Literal Offset mode)								
Syntax:	BSF [k], b								
Operands:	$\begin{array}{l} 0 \leq f \leq 95 \\ 0 \leq b \leq 7 \end{array}$								
Operation:	$1 \rightarrow$ ((FSR)	2) + k) <b></b>							
Status Affected:	None								
Encoding:	1000 bbb0 kkkk kkkk								
Description:		Bit 'b' of the register indicated by FSR2 offset by the value 'k', is set.							
Words:	1								
Cycles:	1								
Q Cycle Activity:									
Q1	Q2	Q3	Q4						
Decode	Read register 'f'	Process Data	Write to destination						
Example:		FLAG_OFST]	, 7						
Before Instruct FLAG_OF FSR2		0Ah 0A00h							
Contents of 0A0Ah After Instructio	=	55h							
Contents									
of 0A0Ah	=	D5h							
SETF	Set Indexe (Indexed L	d iteral Offset ı	node)						
SETF Syntax:			node)						
	(Indexed L		node)						
Syntax:	(Indexed L SETF [k]	iteral Offset r	node)						
Syntax: Operands:	(Indexed L SETF [k] 0 ≤ k ≤ 95	iteral Offset r	node)						
Syntax: Operands: Operation:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS	iteral Offset r							
Syntax: Operands: Operation: Status Affected:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten	SR2) + k)	kk kkkk er indicated by						
Syntax: Operands: Operation: Status Affected: Encoding:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten	iteral Offset ( SR2) + k) 1000 kk	kk kkkk er indicated by						
Syntax: Operands: Operation: Status Affected: Encoding: Description:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset	SR2) + k)	kk kkkk er indicated by						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1	BR2) + k) 1000 kk ts of the regist et by 'k', are se	kk kkkk er indicated by et to FFh.						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1 2	iteral Offset r         SR2) + k)         1000       kk.         ts of the regist         et by 'k', are se         Q3	kk kkkk er indicated by et to FFh. Q4						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity:	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1	BR2) + k) 1000 kk ts of the regist et by 'k', are se	kk kkkk er indicated by et to FFh.						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1 Q2 Read 'k'	SR2) + k) 1000 kk ts of the regist et by 'k', are se Q3 Process	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k'	GR2) + k) 1000 kk ts of the registress by 'k', are se Q3 Process Data	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct OFST	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offse 1 1 Q2 Read 'k' SETF [ ion = 20	GR2) + k) 1000 kk ts of the regist et by 'k', are se Q3 Process Data CoFST] Ch	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct OFST FSR2 Contents	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k' SETF [ ion = 2C = 0A	iteral Offset r SR2) + k) 1000 kk: ts of the regist et by 'k', are se Q3 Process Data CoFST] Ch 000h	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct OFST FSR2 Contents of 0A2Ch	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k' SETF [ ion = 20 = 00	iteral Offset r SR2) + k) 1000 kk: ts of the regist et by 'k', are se Q3 Process Data CoFST] Ch 000h	kk kkkk er indicated by et to FFh. Q4 Write						
Syntax: Operands: Operation: Status Affected: Encoding: Description: Words: Cycles: Q Cycle Activity: Q1 Decode Example: Before Instruct OFST FSR2 Contents	(Indexed L SETF [k] $0 \le k \le 95$ FFh $\rightarrow$ ((FS None 0110 The conten FSR2, offset 1 1 Q2 Read 'k' SETF [ ion = 20 = 00	iteral Offset i SR2) + k) 1000 kk ts of the register ts of the register Q3 Process Data OFST] Ch 000h h	kk kkkk er indicated by et to FFh. Q4 Write						

### 29.2.5 SPECIAL CONSIDERATIONS WITH MICROCHIP MPLAB[®] IDE TOOLS

The latest versions of Microchip's software tools have been designed to fully support the extended instruction set for the PIC18F87K22 family. This includes the MPLAB C18 C Compiler, MPASM assembly language and MPLAB Integrated Development Environment (IDE).

When selecting a target device for software development, MPLAB IDE will automatically set default Configuration bits for that device. The default setting for the XINST Configuration bit is '0', disabling the extended instruction set and Indexed Literal Offset Addressing. For proper execution of applications developed to take advantage of the extended instruction set, XINST must be set during programming.

To develop software for the extended instruction set, the user must enable support for the instructions and the Indexed Addressing mode in their language tool(s). Depending on the environment being used, this may be done in several ways:

- A menu option or dialog box within the environment that allows the user to configure the language tool and its settings for the project
- A command line option
- · A directive in the source code

These options vary between different compilers, assemblers and development environments. Users are encouraged to review the documentation accompanying their development systems for the appropriate information.

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#### 30.0 DEVELOPMENT SUPPORT

The PIC[®] microcontrollers and dsPIC[®] digital signal controllers are supported with a full range of software and hardware development tools:

- Integrated Development Environment
- MPLAB[®] IDE Software
- Compilers/Assemblers/Linkers
  - MPLAB C Compiler for Various Device Families
  - HI-TECH C for Various Device Families
  - MPASM[™] Assembler
  - MPLINK[™] Object Linker/ MPLIB[™] Object Librarian
  - MPLAB Assembler/Linker/Librarian for Various Device Families
- Simulators
  - MPLAB SIM Software Simulator
- Emulators
  - MPLAB REAL ICE™ In-Circuit Emulator
- In-Circuit Debuggers
  - MPLAB ICD 3
  - PICkit™ 3 Debug Express
- Device Programmers
  - PICkit[™] 2 Programmer
  - MPLAB PM3 Device Programmer
- Low-Cost Demonstration/Development Boards, Evaluation Kits, and Starter Kits

#### 30.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16/32-bit microcontroller market. The MPLAB IDE is a Windows[®] operating system-based application that contains:

- A single graphical interface to all debugging tools
  - Simulator
  - Programmer (sold separately)
  - In-Circuit Emulator (sold separately)
  - In-Circuit Debugger (sold separately)
- · A full-featured editor with color-coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- · High-level source code debugging
- · Mouse over variable inspection
- Drag and drop variables from source to watch windows
- · Extensive on-line help
- Integration of select third party tools, such as IAR C Compilers

The MPLAB IDE allows you to:

- · Edit your source files (either C or assembly)
- One-touch compile or assemble, and download to emulator and simulator tools (automatically updates all project information)
- · Debug using:
  - Source files (C or assembly)
  - Mixed C and assembly
  - Machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost-effective simulators, through low-cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increased flexibility and power.

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#### 30.2 MPLAB C Compilers for Various Device Families

The MPLAB C Compiler code development systems are complete ANSI C compilers for Microchip's PIC18, PIC24 and PIC32 families of microcontrollers and the dsPIC30 and dsPIC33 families of digital signal controllers. These compilers provide powerful integration capabilities, superior code optimization and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

#### 30.3 HI-TECH C for Various Device Families

The HI-TECH C Compiler code development systems are complete ANSI C compilers for Microchip's PIC family of microcontrollers and the dsPIC family of digital signal controllers. These compilers provide powerful integration capabilities, omniscient code generation and ease of use.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

The compilers include a macro assembler, linker, preprocessor, and one-step driver, and can run on multiple platforms.

#### 30.4 MPASM Assembler

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel[®] standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM Assembler features include:

- · Integration into MPLAB IDE projects
- User-defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

#### 30.5 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK Object Linker combines relocatable objects created by the MPASM Assembler and the MPLAB C18 C Compiler. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB Object Librarian manages the creation and modification of library files of precompiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

#### 30.6 MPLAB Assembler, Linker and Librarian for Various Device Families

MPLAB Assembler produces relocatable machine code from symbolic assembly language for PIC24, PIC32 and dsPIC devices. MPLAB C Compiler uses the assembler to produce its object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- · Support for the entire device instruction set
- · Support for fixed-point and floating-point data
- Command line interface
- · Rich directive set
- Flexible macro language
- · MPLAB IDE compatibility

#### 30.7 MPLAB SIM Software Simulator

The MPLAB SIM Software Simulator allows code development in a PC-hosted environment by simulating the PIC MCUs and dsPIC[®] DSCs on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a comprehensive stimulus controller. Registers can be logged to files for further run-time analysis. The trace buffer and logic analyzer display extend the power of the simulator to record and track program execution, actions on I/O, most peripherals and internal registers.

The MPLAB SIM Software Simulator fully supports symbolic debugging using the MPLAB C Compilers, and the MPASM and MPLAB Assemblers. The software simulator offers the flexibility to develop and debug code outside of the hardware laboratory environment, making it an excellent, economical software development tool.

#### 30.8 MPLAB REAL ICE In-Circuit Emulator System

MPLAB REAL ICE In-Circuit Emulator System is Microchip's next generation high-speed emulator for Microchip Flash DSC and MCU devices. It debugs and programs PIC[®] Flash MCUs and dsPIC[®] Flash DSCs with the easy-to-use, powerful graphical user interface of the MPLAB Integrated Development Environment (IDE), included with each kit.

The emulator is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with either a connector compatible with incircuit debugger systems (RJ11) or with the new high-speed, noise tolerant, Low-Voltage Differential Signal (LVDS) interconnection (CAT5).

The emulator is field upgradable through future firmware downloads in MPLAB IDE. In upcoming releases of MPLAB IDE, new devices will be supported, and new features will be added. MPLAB REAL ICE offers significant advantages over competitive emulators including low-cost, full-speed emulation, run-time variable watches, trace analysis, complex breakpoints, a ruggedized probe interface and long (up to three meters) interconnection cables.

#### 30.9 MPLAB ICD 3 In-Circuit Debugger System

MPLAB ICD 3 In-Circuit Debugger System is Microchip's most cost effective high-speed hardware debugger/programmer for Microchip Flash Digital Signal Controller (DSC) and microcontroller (MCU) devices. It debugs and programs PIC[®] Flash microcontrollers and dsPIC[®] DSCs with the powerful, yet easyto-use graphical user interface of MPLAB Integrated Development Environment (IDE).

The MPLAB ICD 3 In-Circuit Debugger probe is connected to the design engineer's PC using a high-speed USB 2.0 interface and is connected to the target with a connector compatible with the MPLAB ICD 2 or MPLAB REAL ICE systems (RJ-11). MPLAB ICD 3 supports all MPLAB ICD 2 headers.

#### 30.10 PICkit 3 In-Circuit Debugger/ Programmer and PICkit 3 Debug Express

The MPLAB PICkit 3 allows debugging and programming of PIC[®] and dsPIC[®] Flash microcontrollers at a most affordable price point using the powerful graphical user interface of the MPLAB Integrated Development Environment (IDE). The MPLAB PICkit 3 is connected to the design engineer's PC using a full speed USB interface and can be connected to the target via an Microchip debug (RJ-11) connector (compatible with MPLAB ICD 3 and MPLAB REAL ICE). The connector uses two device I/O pins and the reset line to implement in-circuit debugging and In-Circuit Serial Programming[™].

The PICkit 3 Debug Express include the PICkit 3, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

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#### 30.11 PICkit 2 Development Programmer/Debugger and PICkit 2 Debug Express

The PICkit[™] 2 Development Programmer/Debugger is a low-cost development tool with an easy to use interface for programming and debugging Microchip's Flash families of microcontrollers. The full featured Windows[®] programming interface supports baseline (PIC10F, PIC12F5xx, PIC16F5xx), midrange (PIC12F6xx, PIC16F), PIC18F, PIC24, dsPIC30, dsPIC33, and PIC32 families of 8-bit, 16-bit, and 32-bit microcontrollers, and many Microchip Serial EEPROM products. With Microchip's powerful MPLAB Integrated Development Environment (IDE) the PICkit[™] 2 enables in-circuit debugging on most PIC[®] microcontrollers. In-Circuit-Debugging runs, halts and single steps the program while the PIC microcontroller is embedded in the application. When halted at a breakpoint, the file registers can be examined and modified.

The PICkit 2 Debug Express include the PICkit 2, demo board and microcontroller, hookup cables and CDROM with user's guide, lessons, tutorial, compiler and MPLAB IDE software.

#### 30.12 MPLAB PM3 Device Programmer

The MPLAB PM3 Device Programmer is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features a large LCD display (128 x 64) for menus and error messages and a modular, detachable socket assembly to support various package types. The ICSP™ cable assembly is included as a standard item. In Stand-Alone mode, the MPLAB PM3 Device Programmer can read, verify and program PIC devices without a PC connection. It can also set code protection in this mode. The MPLAB PM3 connects to the host PC via an RS-232 or USB cable. The MPLAB PM3 has high-speed communications and optimized algorithms for quick programming of large memory devices and incorporates an MMC card for file storage and data applications.

#### 30.13 Demonstration/Development Boards, Evaluation Kits, and Starter Kits

A wide variety of demonstration, development and evaluation boards for various PIC MCUs and dsPIC DSCs allows quick application development on fully functional systems. Most boards include prototyping areas for adding custom circuitry and provide application firmware and source code for examination and modification.

The boards support a variety of features, including LEDs, temperature sensors, switches, speakers, RS-232 interfaces, LCD displays, potentiometers and additional EEPROM memory.

The demonstration and development boards can be used in teaching environments, for prototyping custom circuits and for learning about various microcontroller applications.

In addition to the PICDEM[™] and dsPICDEM[™] demonstration/development board series of circuits, Microchip has a line of evaluation kits and demonstration software for analog filter design, KEELOQ[®] security ICs, CAN, IrDA[®], PowerSmart battery management, SEEVAL[®] evaluation system, Sigma-Delta ADC, flow rate sensing, plus many more.

Also available are starter kits that contain everything needed to experience the specified device. This usually includes a single application and debug capability, all on one board.

Check the Microchip web page (www.microchip.com) for the complete list of demonstration, development and evaluation kits.

### 31.0 ELECTRICAL CHARACTERISTICS

#### Absolute Maximum Ratings^(†)

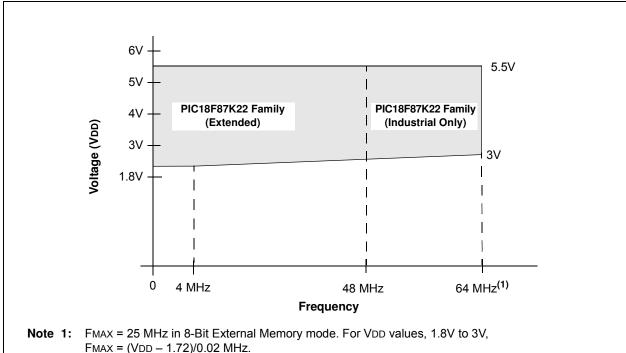
Ambient temperature under bias	40°C to +125°C
Storage temperature	65°C to +150°C
Voltage on any digital only I/O pin with respect to Vss (except VDD)	0.3V to 7.5V
Voltage on MCLR with respect to Vss	0.3V to 9.0V
Voltage on any combined digital and analog pin with respect to Vss (except VDD and MCLR)	0.3V to (VDD + 0.3V)
Voltage on VDD with respect to Vss (with regulator enabled)	0.3V to 5.5V
Voltage on VDD with respect to Vss (with regulator disabled)	0.3V to 3.6V
Total power dissipation (Note 1)	1W
Maximum current out of Vss pin	300 mA
Maximum current into VDD pin	250 mA
Input clamp current, liк (Vi < 0 or Vi > VDD)	±20 mA
Output clamp current, loк (Vo < 0 or Vo > VDD)	±20 mA
Maximum output current sunk by PORTA<7:6> and any PORTB and PORTC I/O pins	25 mA
Maximum output current sunk by any PORTD, PORTE and PORTJ I/O pins	8 mA
Maximum output current sunk by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins	2 mA
Maximum output current sourced by PORTA<7:6> and any PORTB and PORTC I/O pins	25 mA
Maximum output current sourced by any PORTD, PORTE and PORTJ I/O pins	8 mA
Maximum output current sourced by PORTA<5:0> and any PORTF, PORTG and PORTH I/O pins	32 mA
Maximum current sunk by all ports combined	200 mA

Note 1: Power dissipation is calculated as follows: Pdis = VDD x {IDD  $-\Sigma$  IOH} +  $\Sigma$  {(VDD - VOH) x IOH} +  $\Sigma$ (VOL x IOL)

**† NOTICE:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

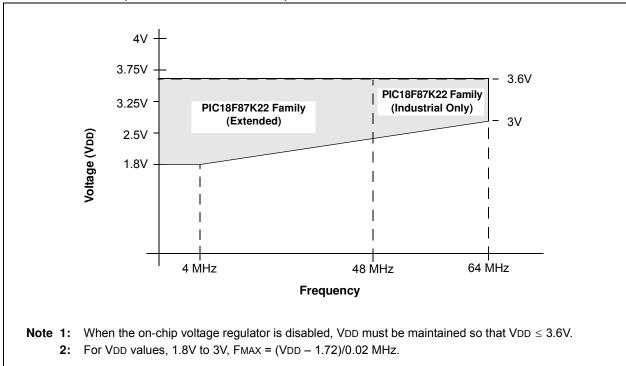
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2: FMAX = 64 MHz in all other modes. For VDD values, 1.8V to 3V, FMAX = (VDD - 1.72)/0.02 MHz.

### FIGURE 31-2: VOLTAGE-FREQUENCY GRAPH, REGULATOR DISABLED (INDUSTRIAL/EXTENDED)^(1,2)



#### 31.1 DC Characteristics: Supply Voltage PIC18F87K22 Family (Industrial/Extended)

	PIC18F87K22 Family (Industrial/Extended)			$\begin{array}{l} \mbox{Standard Operating Conditions (unless otherwise stated)}\\ \mbox{Operating temperature} & -40^\circ C \leq TA \leq +85^\circ C \mbox{ for industrial}\\ -40^\circ C \leq TA \leq +125^\circ C \mbox{ for extended} \end{array}$					
Param No.	Symbol	mbol Characteristic		Symbol Characteristic Min		Тур	Мах	Units	Conditions
D001	Vdd	Supply Voltage	1.8 1.8	_	3.6 5.5	V V	ENVREG tied to Vss ENVREG tied to VDD		
D001C	AVdd	Analog Supply Voltage	Vdd - 0.3	_	VDD + 0.3	V			
D001D	AVss	Analog Ground Potential	Vss – 0.3		Vss + 0.3	V			
D002	Vdr	RAM Data Retention Voltage ⁽¹⁾	1.5	_	—	V			
D003	VPOR	VDD Start Voltage to Ensure Internal Power-on Reset Signal	—	_	0.7	V	See Section 5.3 "Power-on Reset (POR)" for details		
D004	Svdd	VDD Rise Rate to Ensure Internal Power-on Reset Signal	0.05	_	-	V/ms	See Section 5.3 "Power-on Reset (POR)" for details		
D005	Bvdd	Brown-out Reset Voltage (High/Medium/ Low-Power mode) ⁽²⁾ BORV<1:0> = 11 ⁽³⁾ BORV<1:0> = 10 BORV<1:0> = 01 BORV<1:0> = 00	1.69 1.88 2.53 2.82	1.8 2.0 2.7 3.0	1.91 2.12 2.86 3.18				

Note 1: This is the limit to which VDD can be lowered in Sleep mode, or during a device Reset, without losing RAM data.

2: The following values are taken in HP-BOR mode.

3: The device will operate normally until Brown-out Reset occurs, even though VDD may be below VDDMIN.

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	PIC18F87K22 Family (Industrial/Extended)		$ \begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^\circ C \leq TA \leq +85^\circ C \mbox{ for industrial} \\ -40^\circ C \leq TA \leq +125^\circ C \mbox{ for extended} \end{array} $								
Param No.	Device	Тур	Max	Units	ns						
	Power-Down Current (IPD)	(1)									
	All devices	10	500	nA	-40°C						
		20	500	nA	+25°C	VDD = 1.8V ⁽⁴⁾					
		120	600	nA	+60°C	(Sleep mode)					
		630	1800	nA	+85°C	Regulator Disabled					
		4	9	μA	+125°C						
	All devices	50	700	nA	-40°C						
		60	700	nA	+25°C	$V_{DD} = 3.3V^{(4)}$					
		170	800	nA	+60°C	(Sleep mode)					
		700	2700	nA	+85°C	Regulator Disabled					
		5	11	μA	+125°C						
	All devices	350	1300	nA	-40°C						
		400	1400	nA	+25°C	VDD = 5V ⁽⁵⁾					
		550	1500	nA	+60°C	(Sleep mode)					
		1350	4000	nA	+85°C	Regulator Enabled					
		6	12	μA	+125°C						

**Note 1:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- MCLR = VDD; WDT enabled/disabled as specified.
- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).
- 6: 48 MHz, maximum frequency at +125°C.

PIC18F87K22 Family (Industrial/Extended)		$\begin{array}{llllllllllllllllllllllllllllllllllll$									
Param No.	Device	Typ Max Units Condition					ns				
	Supply Current (IDD) ^(2,3)										
	All devices	5.3	10	μA	-40°C						
		5.5	10	μA	+25°C	VDD = 1.8V ⁽⁴⁾					
		5.5	10	μA	+85°C	Regulator Disabled					
		12	24	μA	+125°C						
	All devices	10	15	μA	-40°C						
		10	16	μA	+25°C	VDD = 3.3V ⁽⁴⁾	Fosc = 31 kHz ( <b>RC RUN</b> mode,				
		11	17	μA	+85°C	Regulator Disabled	LF-INTOSC)				
		15	35	μA	+125°C		,				
	All devices	70	180	μA	-40°C						
		80	185	μA	+25°C	VDD = 5V ⁽⁵⁾					
		90	190	μA	+85°C	Regulator Enabled					
		200	500	μA	+125°C						
	All devices	410	850	μA	-40°C		Fosc = 1 MHz ( <b>RC RUN</b> mode,				
		410	800	μA	+25°C	VDD = 1.8V ⁽⁴⁾					
		410	830	μA	+85°C	Regulator Disabled					
		700	1500	μA	+125°C						
	All devices	680	990	μA	-40°C						
		680	960	μA	+25°C	VDD = 3.3V ⁽⁴⁾					
		670	950	μA	+85°C	Regulator Disabled	HF-INTOSC)				
		800	1700	μA	+125°C		,				
	All devices	760	1400	μA	-40°C						
		780	1400	μA	+25°C	VDD = 5V ⁽⁵⁾					
		800	1500	μA	+85°C	Regulator Enabled					
		1200	2400	μA	+125°C						
	All devices	760	1300	μA	-40°C						
		760	1400	μA	+25°C	VDD = 1.8V ⁽⁴⁾					
		770	1500	μA	+85°C	Regulator Disabled					
		800	1700	μA	+125°C						
	All devices	1.4	2.5	mA	-40°C		Fosc = 4 MHz				
		1.4	2.5	mA	+25°C	$V_{DD} = 3.3V^{(4)}$ Regulator Disabled	( <b>RC_RUN</b> mode,				
		1.4	2.5	mA	+85°C		HF-INTOSC)				
		1.5	3.0	mA	+125°C		,				
	All devices	1.5	2.7	mA	-40°C						
		1.5	2.7	mA	+25°C	$V_{DD} = 5V^{(5)}$					
		1.5	2.7	mA	+85°C	Regulator Enabled					
		1.6	3.3	mA	+125°C						

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

**6:** 48 MHz, maximum frequency at +125°C.

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#### 31.2 DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/Extended) (Continued)

PIC18F87K22 Family (Industrial/Extended)		Standard ( Operating f	•	•	tions (unless otherwise stated) -40°C $\leq$ TA $\leq$ +85°C for industrial -40°C $\leq$ TA $\leq$ +125°C for extended		
Param No.	Device	Тур	Max	Units		5	
	Supply Current (IDD) Cont.	(2,3)					
	All devices	2.1	5.5	μA	-40°C		
		2.1	5.7	μA	+25°C	VDD = 1.8V ⁽⁴⁾	
		2.2	6.0	μA	+85°C	Regulator Disabled	
		10	20	μA	+125°C		
	All devices	3.7	7.5	μA	-40°C		
		3.9	7.8	μA	+25°C	VDD = 3.3V ⁽⁴⁾	Fosc = 31 kHz ( <b>RC IDLE</b> mode,
		3.9	8.5	μA	+85°C	Regulator Disabled	LF-INTOSC)
		12	24	μA	+125°C		,
	All devices	70	180	μA	-40°C		
		80	190	μA	+25°C	VDD = 5V ⁽⁵⁾	
		80	200	μA	+85°C	Regulator Enabled	
		200	420	μA	+125°C		
	All devices	330	650	μA	-40°C		Fosc = 1 MHz ( <b>RC_IDLE</b> mode, HF-INTOSC)
		330	640	μA	+25°C	VDD = 1.8V ⁽⁴⁾	
		330	630	μA	+85°C	Regulator Disabled	
		500	850	μA	+125°C		
	All devices	520	850	μA	-40°C		
		520	900	μA	+25°C	$VDD = 3.3V^{(4)}$	
		520	850	μA	+85°C	Regulator Disabled	
		800	1200	μA	+125°C		
	All devices	590	940	μA	-40°C		
		600	960	μA	+25°C	$V_{DD} = 5V^{(5)}$	
		620 1000	990	μΑ	+85°C +125°C	Regulator Enabled	
	All devices	470	1400 770	μΑ μΑ	-40°C		
	Air devices	470	770	μΑ μΑ	-40 C +25°C	VDD = 1.8V ⁽⁴⁾	
		470	760	μΑ	+25 C +85°C	Regulator Disabled	
		700	1000	μA μA	+05 C +125°C		
	All devices	800	1400	μA	-40°C		
		800	1350	μA	+25°C	$V_{DD} = 3.3V^{(4)}$	Fosc = 4 MHz
		790	1300	μA μA	+25°C	Regulator Disabled	(RC_IDLE mode, internal HF-INTOSC)
		1100	1400	μA	+125°C		
	All devices	880	1600	μA	-40°C		
	,	890	1700	μA	+25°C	VDD = 5V ⁽⁵⁾	
		910	1800	μA	+85°C	Regulator Enabled	
		1200	2200	μA	+125°C		

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

**6:** 48 MHz, maximum frequency at +125°C.

PIC18F87K22 Family (Industrial/Extended)		$ \begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^\circ C \leq T A \leq +85^\circ C \mbox{ for industrial} \\ -40^\circ C \leq T A \leq +125^\circ C \mbox{ for extended} \end{array} $							
Param No.	Device	Тур	Max	Units		Conditions			
	Supply Current (IDD) Cont	(2,3)							
	All devices	130	390	μA	-40°C				
		130	390	μA	+25°C	VDD = 1.8V ⁽⁴⁾			
		130	390	μA	+85°C	Regulator Disabled			
		250	500	μA	+125°C				
	All devices	270	790	μA	-40°C				
		270	790	μA	+25°C	VDD = 3.3V ⁽⁴⁾	Fosc = 1 MHz ( <b>PRI_RUN</b> mode,		
		270	790	μA	+85°C	Regulator Disabled	EC oscillator)		
		400	900	μA	+125°C				
	All devices	430	990	μA	-40°C				
		450	980	μA	+25°C	VDD = 5V ⁽⁵⁾			
		460	980	μA	+85°C	Regulator Enabled			
		600	1300	μA	+125°C				
	All devices	430	860	μA	-40°C				
		530	900	μA	+25°C	VDD = 1.8V ⁽⁴⁾			
		490	880	μA	+85°C	Regulator Disabled			
		750	1600	μA	+125°C				
	All devices	850	1750	μA	-40°C		Fosc = 4 MHz		
		850	1700	μA	+25°C	VDD = 3.3V ⁽⁴⁾	$(\mathbf{PRI} \ \mathbf{RUN} \ mode)$		
		850	1800	μA	+85°C	Regulator Disabled	EC oscillator)		
		1150	2400	μA	+125°C				
	All devices	1.1	2.7	mA	-40°C				
		1.1	2.6	mA	+25°C	VDD = 5V ⁽⁵⁾			
		1.1	2.6	mA	+85°C	Regulator Enabled			
		2.0	4.0	mA	+125°C				
	All devices	12	19	mA	-40°C				
		12	19	mA	+25°C	VDD = 3.3V ⁽⁴⁾			
		12	19	mA	+85°C	Regulator Disabled	Fosc = 64 MHz		
		13	22	mA	+125°C ⁽⁶⁾		$(\mathbf{PRI} \ \mathbf{RUN} \ \mathrm{mode})$		
	All devices	13	20	mA	-40°C		EC oscillator)		
		13	20	mA	+25°C	VDD = 5V ⁽⁴⁾	EC Oscillator)		
		13	20	mA	+85°C	Regulator Enabled			
		14	23	mA	+125⁰C ⁽⁶⁾				

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

**3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

6: 48 MHz, maximum frequency at +125°C.

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PIC18F87K22 Family (Industrial/Extended)		$ \begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^{\circ}C \leq TA \leq +85^{\circ}C \mbox{ for industrial} \\ -40^{\circ}C \leq TA \leq +125^{\circ}C \mbox{ for extended} \end{array} $								
Param No.	Device	Тур	Max	Units		Condition	s			
	Supply Current (IDD) Cont	(2,3)								
	All devices	3.3	5.6	mA	-40°C					
		3.3	5.5	mA	+25°C	VDD = 3.3V ⁽⁴⁾				
		3.3	5.5	mA	+85°C	Regulator Disabled	Fosc = 16 MHz, ( <b>PRI_RUN</b> mode, 4 MHz EC oscillator with PLL)			
		3.6	6.0	mA	+125°C					
	All devices	3.5	5.9	mA	-40°C					
		3.5	5.8	mA	+25°C	Vdd = 5V ⁽⁵⁾				
		3.5	5.8	mA	+85°C	Regulator Enabled				
		3.8	7.0	mA	+125°C					
	All devices	12	18	mA	-40°C					
		12	18	mA	+25°C	VDD = 3.3V ⁽⁴⁾				
		12	18	mA	+85°C	Regulator Disabled				
		13	22	mA	+125°C ⁽⁶⁾		Fosc = 64 MHz, ( <b>PRI RUN</b> mode, 16 MHz			
	All devices	13	20	mA	-40°C	Vdd = 5V ⁽⁵⁾	EC oscillator with PLL)			
		13	20	mA	+25°C					
		13	20	mA	+85°C	Regulator Enabled				
		14	24	mA	+125°C ⁽⁶⁾					

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

- **3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).
- 6: 48 MHz, maximum frequency at +125°C.

PIC18F87K22 Family (Industrial/Extended)		Standard ( Operating f			-40°C ≤ TA ≤ -	<b>therwise stated)</b> +85°C for industrial +125°C for extended	
Param No.	Device	Тур	Max	Units	Conditions		
	Supply Current (IDD) Cont	(2,3)					
	All devices	42	73	μA	-40°C		
		42	73	μA	+25°C	VDD = 1.8V ⁽⁴⁾	
		43	74	μA	+85°C	Regulator Disabled	
		53	100	μA	+125°C		
	All devices	110	190	μA	-40°C		
		110	195	μA	+25°C	VDD = 3.3V ⁽⁴⁾	Fosc = 1 MHz
		110	195	μA	+85°C	Regulator Disabled	( <b>PRI_IDLE</b> mode, EC oscillator)
		130	250	μA	+125°C		
	All devices	280	450	μA	-40°C		
		290	440	μA	+25°C	VDD = 5V ⁽⁵⁾	
		300	460	μA	+85°C	Regulator Enabled	
		330	500	μA	+125°C		
	All devices	160	360	μA	-40°C		
		160	360	μA	+25°C	VDD = 1.8V ⁽⁴⁾	
		170	370	μA	+85°C	Regulator Disabled	
		200	400	μA	+125°C		
	All devices	330	650	μA	-40°C		
		340	660	μA	+25°C	$VDD = 3.3V^{(4)}$	Fosc = 4 MHz ( <b>PRI_IDLE</b> mode,
		340	660	μA	+85°C	Regulator Disabled	EC oscillator)
		370	700	μA	+125°C		
	All devices	510	900	μA	-40°C		
		520	950	μA	+25°C	VDD = 5V ⁽⁵⁾	
		540	990	μA	+85°C	Regulator Enabled	
		600	1200	μA	+125°C		
	All devices	4.7	9	mA	-40°C		
		4.8	9	mA	+25°C	VDD = 3.3V ⁽⁴⁾	
		4.8	10	mA	+85°C	Regulator Disabled	
		5.2	12	mA	+125°C ⁽⁶⁾		Fosc = 64 MHz
	All devices	5.1	11	mA	-40°C		( <b>PRI_IDLE</b> mode, EC oscillator)
		5.1	11	mA	+25°C	VDD = 5V ⁽⁵⁾	
		5.2	12	mA	+85°C	Regulator Enabled	
		5.7	14	mA	+125ºC ⁽⁶⁾		

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss,  $\overline{RETEN}$  (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

6: 48 MHz, maximum frequency at +125°C.

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PIC18F87K22 Family (Industrial/Extended)		$\begin{array}{ll} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^\circ C \leq TA \leq +85^\circ C \mbox{ for industrial} \\ -40^\circ C \leq TA \leq +125^\circ C \mbox{ for extended} \end{array}$							
Param No.	Device	Тур	Max	Units	Conditions				
	Supply Current (IDD) Cont	(2,3)							
	All devices	3.7	8.5	μA	-40°C				
		5.4	10	μA	+25°C	VDD = 1.8V ⁽⁴⁾			
		6.6	13	μA	+85°C	Regulator Disabled			
		13	30	μA	+125°C				
	All devices	8.7	18	μA	-40°C		$\Gamma_{000} = 20  t    - (3)$		
		10	20	μA	+25°C	VDD = 3.3V ⁽⁴⁾	Fosc = 32 kHz ⁽³⁾ ( <b>SEC_RUN</b> mode, SOSCSEL = 01)		
		12	23	μA	+85°C	Regulator Disabled			
		25	60	μA	+125°C				
	All devices	60	160	μA	-40°C				
		90	190	μA	+25°C	$V_{DD} = 5V^{(4)}$			
		100	240	μA	+85°C	Regulator Enabled			
		200	450	μA	+125°C				
	All devices		4	μA	-40°C				
		1.7	5	μA	+25°C	VDD = 1.8V ⁽⁴⁾			
		2.6	6	μA	+85°C	Regulator Disabled			
		9	20	μA	+125°C				
	All devices	-	7	μA	-40°C		Fosc = 32 kHz ⁽³⁾		
		2.8	9	μA	+25°C	VDD = 3.3V ⁽⁴⁾	(SEC_IDLE mode,		
		4.1	10	μA	+85°C	Regulator Disabled	SOSCSEL = 01)		
		17	40	μA	+125°C		/		
	All devices		150	μA	-40°C				
		80	180	μA	+25°C	VDD = 5V ⁽⁵⁾			
		100	240	μA	+85°C	Regulator Enabled			
		180	440	μA	+125°C				

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

**6:** 48 MHz, maximum frequency at +125°C.

	7K22 Family strial/Extended)	Standard ( Operating t	•	•	-40°C ≤ Ta ≤	otherwise stated) +85°C for industrial +125°C for extended	
Param No.	Device	Тур	Мах	Units		Conditions	
	Module Differential Currer	nts (AlwDT, A		IHLVD, Ale	OSCB, AIAD)		
	Watchdog Timer		,	,			
D022	All devices	0.3	1	μA	-40°C		
(∆IWDT)		0.3	1	μA	+25°C	VDD = 1.8V ⁽⁴⁾	
D022		0.3	1	μA	+85°C	Regulator Disabled	
(∆IWDT)		0.5	2	μA	+125°C		
	All devices	0.6	2	μA	-40°C		
		0.6	2	μA	+25°C	$V_{DD} = 3.3V^{(4)}$	
		0.7	2	μA	+85°C	Regulator Disabled	
		1	3	μA	+125°C		
	All Devices	0.6	2	μA	-40°C		
		0.6	2	μA	+25°C	VDD = 5V ⁽⁵⁾	
		0.7	2	μA	+85°C	Regulator Enabled	
		1.5	4	μA	+125°C		
D022A	Brown-out Reset						
( $\Delta$ IBOR)	All devices	4.6	19	μA	-40°C		
( $\Delta$ IBOR)		4.5	20	μA	+25°C	$V_{DD} = 3.3 V^{(4)}$	
		4.7	20	μA	+85°C	Regulator Disabled	High-Power BOR
		18	40	μA	+125°C		
	All devices	4.2	20	μA	-40°C		
		4.3	20	μA	+25°C	VDD = 5V ⁽⁵⁾	
		4.4	20	μA	+85°C	Regulator Enabled	High-Power BOR
		20	40	μA	+125°C		
D022B	High/Low-Voltage Detect						
(∆Ihlvd)	All devices	3.8	9	μA	-40°C		
		4.2	9	μA	+25°C	VDD = 1.8V ⁽⁴⁾	
		4.3	10	μA	+85°C	Regulator Disabled	
		4.5	12	μA	+125°C		
	All devices	4.5	11	μA	-40°C		
		4.8	12	μA	+25°C	VDD = 3.3V ⁽⁴⁾	
		4.8	12	μA	+85°C	Regulator Disabled	
		5.0	14	μA	+125°C		
	All devices	4.9	13	μA	-40°C	1	
		4.9	13	μA	+25°C	VDD = 5√ ⁽⁵⁾	
		4.9	13	μA	+85°C	Regulator Enabled	
		5.3	15	μA	+125°C	7	

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

MCLR = VDD; WDT enabled/disabled as specified.

**3:** Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).

5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).

**6:** 48 MHz, maximum frequency at +125°C.

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PIC18F87K22 Family (Industrial/Extended)		$\begin{array}{llllllllllllllllllllllllllllllllllll$								
Param No.	Device	Тур	Max	Units		Conditions	3			
D025	Real-Time Clock/Calendar	al-Time Clock/Calendar with SOSC Oscillator								
$(\Delta IRTCC)$	All devices	0.7	2.7	μA	-40°C					
		0.7	2.8	μA	+25°C	VDD = 1.8V ⁽⁴⁾				
		1.1	2.8	μA	+60°C	Regulator Disabled				
		1.1	2.9	μA	+85°C					
		2.2	4.4	μA	+125°C					
	All devices	1.2	2.9	μA	-40°C					
		1.1	2.8	μA	+25°C	$V_{DD} = 3.3 V^{(4)}$	22 769 14 1-			
		2	4.6	μA	+60°C	Regulator Disabled	32.768 kHz, SOSCRUN = 1			
		2	4.8	μA	+85°C	Regulator Disabled				
		4	6.5	μA	+125°C					
	All devices	1.5	4.4	μA	-40°C					
		1.5	4.4	μA	+25°C	VDD = 5V ⁽⁵⁾				
		1.7	4.7	μA	+60°C	- Regulator Enabled				
		1.7	4.7	μA	+85°C					
		3.5	6.9	μA	+125°C					

**Note 1:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in a high-impedance state and tied to VDD or Vss, and all features that add delta current are disabled (such as WDT, SOSC oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in Active Operation mode are:

OSC1 = External square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

- MCLR = VDD; WDT enabled/disabled as specified.
- 3: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.
- 4: Voltage regulator disabled (ENVREG = 0, tied to Vss, RETEN (CONFIG1L<0>) = 1).
- 5: Voltage regulator enabled (ENVREG = 1, tied to VDD, SRETEN (WDTCON<4>) = 1 and RETEN (CONFIG1L<0>) = 0).
- 6: 48 MHz, maximum frequency at +125°C.

#### 31.3 DC Characteristics: PIC18F87K22 Family (Industrial/Extended)

DC CHA	ARACTEI	RISTICS		nperature -4	l0°C ≤ 1	<b>5 (unless otherwise stated)</b> $TA \le +85^{\circ}C$ for industrial $TA \le +125^{\circ}C$ for extended
Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
	VIL	Input Low Voltage				
		All I/O Ports:				
D030		with TTL Buffer	Vss	0.15 VDD	V	VDD < 4.5V
D031		with Schmitt Trigger Buffer	—	0.8	V	$4.5V \leq V\text{DD} \leq 5.5V$
		RC3, RC4	Vss	0.2 VDD	V	VDD < 4.5
		RD5, RD6	Vss	1.5	V	$4.5V \le VDD \le 5.5V$
D031A		RC3, RC4	Vss	0.3 VDD	V	I ² C™ enabled
D031B		RD5, RD6	Vss	0.8	V	SMBus enabled
D032		MCLR	Vss	0.2 VDD	V	
D033		OSC1	Vss	0.2 VDD	V	LP, XT, HS modes
D033A		OSC1	Vss	0.2 VDD	V	EC modes
D034		SOSCI	Vss	0.3 VDD	V	
	Viн	Input High Voltage				
		I/O Ports with Analog Functions:				
D040		with TTL Buffer	0.25 VDD	Vdd	V	VDD < 4.5V
			2.0	Vdd	V	$4.5V \le VDD \le 5.5V$
D041		with Schmitt Trigger Buffer	0.8	Vdd	V	VDD < 4.5V
		RC3, RC4	0.7 Vdd	Vdd	V	VDD < 4.5V
		RD5, RD6	3V	5.5	V	$4.5V \le VDD \le 5.5V$
D041A		RC3, RC4	0.7 Vdd	Vdd	V	I ² C enabled
D041B		RD5, RD6	2.1	Vdd	V	SMBus enabled
D042		MCLR	0.8 VDD	Vdd	V	
D043		OSC1	0.9 Vdd	Vdd	V	RC mode
D043A		OSC1	0.7 VDD	Vdd	V	HS mode
D044		SOSCI	0.7 Vdd	Vdd	V	
	lı∟	Input Leakage Current ⁽¹⁾				
D060		I/O Ports	±50	±200	nA	$Vss \le VPIN \le VDD,$ Pin at high-impedance
D061		MCLR	—	±5	μA	$Vss \le VPIN \le VDD, +85^{\circ}C$
D063		OSC1	—	±5	μA	$Vss \leq V PIN \leq V DD$
	IPU	Weak Pull-up Current		1		
D070	IPURB	PORTB Weak Pull-up Current	50	400	μA	VDD = 3.3V, VPIN = VSS

Note 1: Negative current is defined as current sourced by the pin.

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#### 31.3 DC Characteristics: PIC18F87K22 Family (Industrial/Extended) (Continued)

DC CHA	RACTE	RISTICS	$\begin{array}{l} \mbox{Standard Operating Conditions (unless otherwise stated) \\ \mbox{Operating temperature} & -40^{\circ}C \leq TA \leq +85^{\circ}C \mbox{ for industrial} \\ & -40^{\circ}C \leq TA \leq +125^{\circ}C \mbox{ for extended} \end{array}$					
Param No.	Symbol	Characteristic	Min	Max	Units	Conditions		
	Vol	Output Low Voltage						
D080		I/O Ports:						
		PORTA, PORTB, PORTC	—	0.6	V	IOL = 8.5 mA, VDD = 4.5V, -40°C to +125°C		
		PORTD, PORTE, PORTF, PORTG, PORTH, PORTJ	—	0.6	V	IOL = 3.5 mA, VDD = 4.5V, -40°C to +125°C		
D083		OSC2/CLKO (EC modes)	—	0.6	V	IOL = 1.6 mA. VDD = 5.5V, -40°C to +125°C		
	Vон	Output High Voltage ⁽¹⁾						
D090		I/O Ports:			V			
		PORTA, PORTB, PORTC	Vdd - 0.7	—	V	IOH = -3 mA, VDD = 4.5V, -40°C to +125°C		
		PORTD, PORTE, PORTF, PORTG, PORTH, PORTJ	Vdd - 0.7	—	V	IOH = -2 mA, VDD = 4.5V, -40°C to +125°C		
D092		OSC2/CLKO (INTOSC, EC modes)	Vdd - 0.7	—	V	IOH = -1 mA, VDD = 5.5V, -40°С to +125°С		
		Capacitive Loading Specs on Output Pins						
D100	COSC2	OSC2 Pin	_	20	pF	In HS mode when external clock is used to drive OSC1		
D101	Сю	All I/O Pins and OSC2	—	50	pF	To meet the AC Timing Specifications		
D102	Св	SCLx, SDAx	—	400	pF	I ² C [™] Specification		

**Note 1:** Negative current is defined as current sourced by the pin.

#### 31.4 DC Characteristics: CTMU Current Source Specifications

DC CH	ARACT	ERISTICS	$\begin{array}{ll} \mbox{Standard Operating Conditions: 1.8V to 5.5V (unless otherwise stated)} \\ \mbox{Operating temperature} & -40^\circ C \leq TA \leq +85^\circ C \mbox{ for industrial} \\ -40^\circ C \leq TA \leq +125^\circ C \mbox{ for extended} \end{array}$					
Param No.	Sym Characteristic		Min	Typ ⁽¹⁾	Max	Units	Conditions	
	Ιουτ1	CTMU Current Source, Base Range	_	550	_	nA	CTMUICON<1:0> = 01	
	IOUT2	CTMU Current Source, 10x Range	—	5.5	—	μΑ	CTMUICON<1:0> = 10	
	Ιουτ3	CTMU Current Source, 100x Range		55	_	μA	CTMUICON<1:0> = 11	

<b>Note 1:</b> Nominal value at center point of current trim range (CTMUICON<7:2> = 0000
------------------------------------------------------------------------------------------

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DC CH	DC CHARACTERISTICS			$\begin{array}{l} \mbox{Standard Operating Conditions (unless otherwise stated)} \\ \mbox{Operating temperature } -40^{\circ}C \leq TA \leq +85^{\circ}C \mbox{ for industrial} \\ -40^{\circ}C \leq TA \leq +125^{\circ}C \mbox{ for extended} \end{array}$					
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions		
		Internal Program Memory Programming Specifications ⁽¹⁾							
D110	Vpp	Voltage on MCLR/VPP/RE5 pin	VDD + 1.5	—	10	V	(Note 3, Note 4)		
D113	IDDP	Supply Current during Programming	—	—	10	mA			
		Data EEPROM Memory					(Note 2)		
D120	Ed	Byte Endurance	100K	—	—	E/W	-40°C to +125°C		
D121	Vdrw	VDD for Read/Write	1.8	—	5.5	V	Using EECON to read/ write, ENVREG tied to VDD		
			1.8	—	3.6	V	Using EECON to read/ write, ENVREG tied to Vss		
D122	TDEW	Erase/Write Cycle Time	—	4	—	ms			
D123	Tretd	Characteristic Retention	40	—	—	Year	Provided no other specifications are violated		
D124	Tref	Number of Total Erase/Write Cycles before Refresh ⁽²⁾	1M	10M	—	E/W	-40°C to +125°C		
		Program Flash Memory							
D130	Еρ	Cell Endurance	10K	—	—	E/W	-40°C to +125°C		
D131	Vpr	VDD for Read	1.8	—	5.5	V	ENVREG tied to VDD		
			1.8	—	3.6	V	ENVREG tied to Vss		
D132B	VPEW	Voltage for Self-Timed Erase or Write Operations							
	_	VDD	1.8	_	5.5	V	ENVREG tied to VDD		
D133A	TIW	Self-Timed Write Cycle Time	-	2	_	ms			
D134	TRETD	Characteristic Retention	40	—	_	Year	Provided no other specifications are violated		
D135	Iddp	Supply Current during Programming	_	—	10	mA			
D140	TWE	Writes per Erase Cycle		_	1		For each physical address		

#### TABLE 31-1: MEMORY PROGRAMMING REQUIREMENTS

† Data in "Typ" column is at 3.3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

**Note 1:** These specifications are for programming the on-chip program memory through the use of table write instructions.

2: Refer to Section 9.8 "Using the Data EEPROM" for a more detailed discussion on data EEPROM endurance.

3: Required only if Single-Supply Programming is disabled.

4: The MPLAB[®] ICD 2 does not support variable VPP output. Circuitry to limit the MPLAB ICD 2 VPP voltage must be placed between the MPLAB ICD 2 and the target system when programming or debugging with the MPLAB ICD 2.

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#### TABLE 31-2: COMPARATOR SPECIFICATIONS

Operating	<b>Operating Conditions:</b> $1.8V \le VDD \le 5V$ , $-40^{\circ}C \le TA \le +125^{\circ}C$ (unless otherwise stated)									
Param No.	Sym	Characteristics	Min	Тур	Мах	Units	Comments			
D300	VIOFF	Input Offset Voltage	—	±5.0	40	mV				
D301	VICM	Input Common Mode Voltage	_		AVDD - 1.5	V				
D302	CMRR	Common Mode Rejection Ratio	55		—	dB				
D303	TRESP	Response Time ⁽¹⁾	—	150	400	ns				
D304	Тмс2оv	Comparator Mode Change to Output Valid*	—	—	10	μS				

**Note 1:** Response time measured with one comparator input at (AVDD – 1.5)/2, while the other input transitions from Vss to VDD.

#### TABLE 31-3: VOLTAGE REFERENCE SPECIFICATIONS

Operating	<b>Operating Conditions:</b> $1.8V \le VDD \le 5V$ , $-40^{\circ}C \le TA \le +125^{\circ}C$ (unless otherwise stated)									
Param No.	Sym	Characteristics	Min	Тур	Max	Units	Comments			
D310	VRES	Resolution	—	VDD/32		LSb				
D311	VRAA	Absolute Accuracy	_	—	1/2	LSb				
D312	VRur	Unit Resistor Value (R)		2k		Ω				
D313	TSET	Settling Time ⁽¹⁾	—	—	10	μS				

**Note 1:** Settling time measured while CVRR = 1 and CVR<3:0> transitions from '0000' to '1111'.

#### TABLE 31-4: INTERNAL VOLTAGE REGULATOR SPECIFICATIONS

Operatir	<b>Operating Conditions:</b> $-40^{\circ}C \le TA \le +125^{\circ}C$ (unless otherwise stated)									
Param No.	Sym	Characteristics	Min	Тур	Max	Units	Comments			
	Vrgout	Regulator Output Voltage		3.3		V				
	Cefc	External Filter Capacitor Value	4.7	10		μF	Capacitor must be low-ESR, a low series resistance (< 5Ω)			

#### 31.5 AC (Timing) Characteristics

#### 31.5.1 TIMING PARAMETER SYMBOLOGY

The timing parameter symbols have been created following one of the following formats:

1. TppS2ppS	3	3. TCC:ST	(I ² C specifications only)
2. TppS		4. Ts	(I ² C specifications only)
Т			
F	Frequency	Т	Time
Lowercase le	etters (pp) and their meanings:		
рр			
сс	CCP1	osc	OSC1
ck	CLKO	rd	RD
cs	CS	rw	RD or WR
di	SDI	sc	SCK
do	SDO	SS	SS
dt	Data in	tO	TOCKI
io	I/O port	t1	T1CKI
mc	MCLR	wr	WR
Uppercase le	etters and their meanings:		
S			
F	Fall	Р	Period
н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance
I ² C only			
AA	output access	High	High
BUF	Bus free	Low	Low
TCC:ST (I ² C s	specifications only)		
CC			
HD	Hold	SU	Setup
ST			
DAT	DATA input hold	STO	Stop condition
STA	Start condition		

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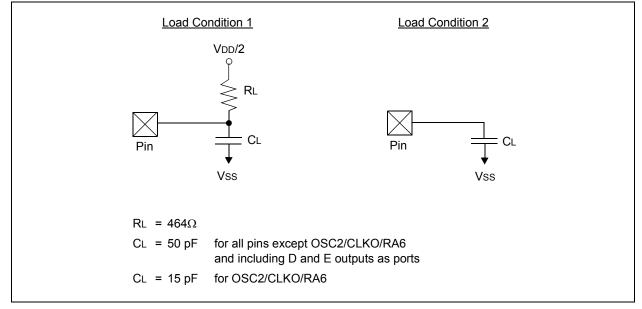
#### 31.5.2 TIMING CONDITIONS

The temperature and voltages specified in Table 31-5 apply to all timing specifications unless otherwise noted. Figure 31-3 specifies the load conditions for the timing specifications.

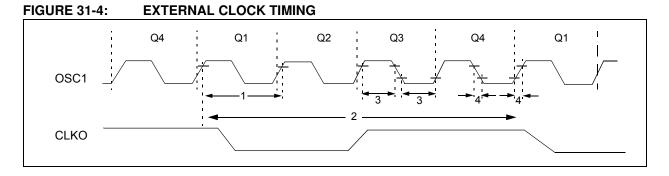
#### TABLE 31-5: TEMPERATURE AND VOLTAGE SPECIFICATIONS – AC

	Standard Operating Conditions (unless otherwise stated)					
AC CHARACTERISTICS	Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
AO ONANAO TENIS NOS	$-40^{\circ}C \le TA \le +125^{\circ}C$ for extended					
	Operating voltage VDD range as described in Section 31.1 and Section 31.3.					

#### FIGURE 31-3: LOAD CONDITIONS FOR DEVICE TIMING SPECIFICATIONS



### 31.5.3 TIMING DIAGRAMS AND SPECIFICATIONS



#### TABLE 31-6: EXTERNAL CLOCK TIMING REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
1A	Fosc	External CLKIN Frequency ⁽¹⁾	DC	64	MHz	EC, ECIO Oscillator mode -40°C ≤ TA ≤ +85°C
			DC	48	MHz	-40°C ≤ TA ≤ +125°C
		Oscillator Frequency ⁽¹⁾	DC	4	MHz	RC Oscillator mode
			0.1	4	MHz	XT Oscillator mode
			4	16	MHz	HS Oscillator mode
			4	16	MHz	HS + PLL Oscillator mode
			5	33	kHz	LP Oscillator mode
1	Tosc	External CLKIN Period ⁽¹⁾	15.6	—	ns	EC, ECIO Oscillator mode
		Oscillator Period ⁽¹⁾	250	—	ns	RC Oscillator mode
			250	10,000	ns	XT Oscillator mode
			40 62.5	250 250	ns ns	HS Oscillator mode HS + PLL Oscillator mode
			5	200	μs	LP Oscillator mode
2	Тсү	Instruction Cycle Time ⁽¹⁾	62.5	_	ns	Tcy = 4/Fosc
3	TosL,	External Clock in (OSC1)	30	_	ns	XT Oscillator mode
	TosH	High or Low Time	2.5	_	μs	LP Oscillator mode
			10	_	ns	HS Oscillator mode
4	TosR,	External Clock in (OSC1)		20	ns	XT Oscillator mode
	TosF	Rise or Fall Time	—	50	ns	LP Oscillator mode
				7.5	ns	HS Oscillator mode

**Note 1:** Instruction cycle period (TCY) equals four times the input oscillator time base period for all configurations except PLL. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1/CLKIN pin. When an external clock input is used, the "max." cycle time limit is "DC" (no clock) for all devices.

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Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions
F10	Fosc	Oscillator Frequency Range	4		5	MHz	VDD = 1.8-5.5V
			4	—	16	MHz	VDD = 3.0-5.5V, -40°C to +85°C
			4	—	12	MHz	VDD = 3.0-5.5V, -40°C to +125°C
F11	Fsys	On-Chip VCO System Frequency	16	—	20	MHz	VDD = 1.8-5.5V
			16	—	64	MHz	VDD = 3.0-5.5V, -40°C to +85°C
			16	—	48	MHz	VDD = 3.0-5.5V, -40°C to +125°C
F12	t _{rc}	PLL Start-up Time (Lock Time)	_		2	ms	
F13	$\Delta \text{CLK}$	CLKOUT Stability (Jitter)	-2		+2	%	

#### TABLE 31-7: PLL CLOCK TIMING SPECIFICATIONS (VDD = 1.8V TO 5.5V)

† Data in "Typ" column is at 3V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

#### TABLE 31-8: INTERNAL RC ACCURACY (INTOSC)

PIC18F87K22 Family		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +125^{\circ}C$					
Param No.		Min	Тур	Max	Units	Conditions	
OA1	HFINTOSC/MFINTOSC Accuracy @ Freq = 16 MHz, 8 MHz, 4 MHz, 2 MHz, 1 MHz, 500 kHz, 25						
		-2	—	2	%	+25°C	VDD = 3.0-5.5V
		-5		5	%		
		-5		5	%	-40°C to +85°C	VDD = 1.8-5.5V
		-10	_	10	%	-40°C to +125°C	VDD = 1.8-5.5V
OA2 LFINTOSC Accuracy @ Freq = 31 kHz							
		-15	_	15	%	-40°C to +125°C	VDD = 1.8-5.5V

Note 1: Frequency calibrated at 25°C. OSCTUNE register can be used to compensate for temperature drift.

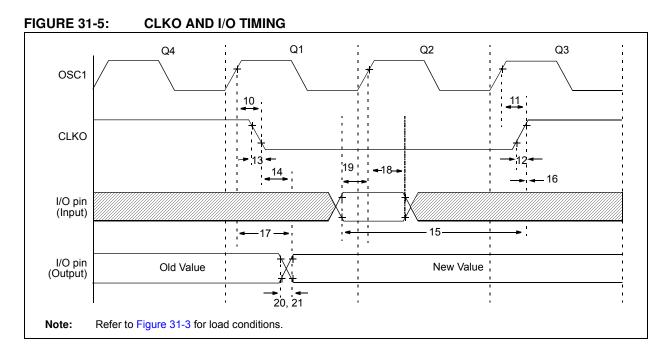


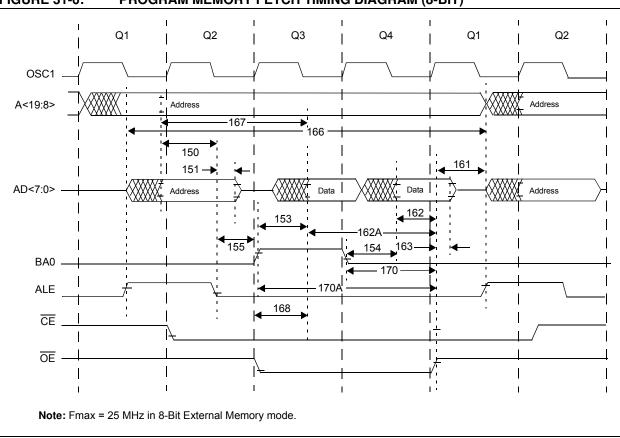
TABLE 31-9: CLKO AND I/O TIMING REQUIREMENTS
----------------------------------------------

Param No.	Symbol	Characteristic	Min	Тур	Max	Units	Conditions
10	TosH2cĸL	OSC1 $\uparrow$ to CLKO $\downarrow$	—	75	200	ns	(Note 1)
11	TosH2ckH	OSC1 ↑ to CLKO ↑	—	75	200	ns	(Note 1)
12	ТскR	CLKO Rise Time	—	15	30	ns	(Note 1)
13	ТскF	CLKO Fall Time	—	15	30	ns	(Note 1)
14	TckL2I0V	CLKO $\downarrow$ to Port Out Valid	_		0.5 Tcy + 20	ns	
15	TioV2скH	Port In Valid before CLKO ↑	0.25 Tcy + 25		—	ns	
16	TckH2iol	Port In Hold after CLKO ↑	0		_	ns	
17	TosH2IoV	OSC1 ↑ (Q1 cycle) to Port Out Valid	_	50	150	ns	
18	TosH2ıol	OSC1 ↑ (Q2 cycle) to Port Input Invalid (I/O in hold time)	100	_	—	ns	
19	TioV2osH	Port Input Valid to OSC1 ↑ (I/O in setup time)	0	_	_	ns	
20	TIOR	Port Output Rise Time		10	25	ns	
21	TIOF	Port Output Fall Time		10	25	ns	
22†	Tinp	INTx pin High or Low Time	20	_	—	ns	
23 <mark>†</mark>	Trbp	RB<7:4> Change INTx High or Low Time	Тсү		—	ns	

† These parameters are asynchronous events not related to any internal clock edges.

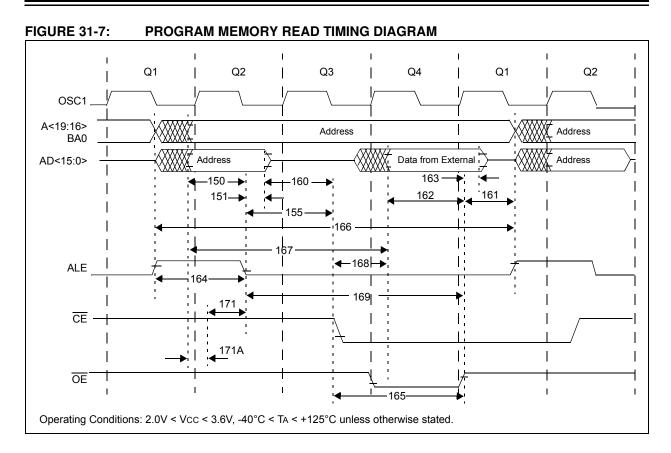
**Note 1:** Measurements are taken in EC mode, where CLKO output is 4 x Tosc.

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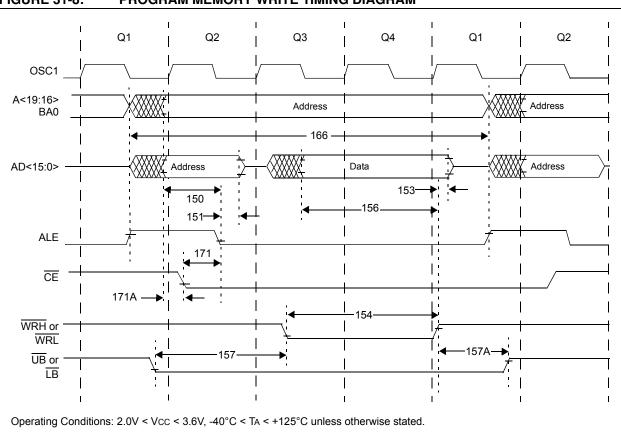
### FIGURE 31-6: PROGRAM MEMORY FETCH TIMING DIAGRAM (8-BIT)

Param No	Symbol	Characteristics	Min	Тур	Мах	Units
150	TadV2alL	Address Out Valid to ALE $\downarrow$ (address setup time)	0.25 Tcy – 10		—	ns
151	TalL2adl	ALE $\downarrow$ to Address Out Invalid (address hold time)	5		—	ns
153	BA01	BA0 $\uparrow$ to Most Significant Data Valid	0.125 TCY		—	ns
154	BA02	BA0 $\downarrow$ to Least Significant Data Valid	0.125 TCY		—	ns
155	TalL2oeL	ALE $\downarrow$ to $\overline{OE} \downarrow$	0.125 TCY		—	ns
161	ToeH2adD	OE ↑ to A/D Driven	0.125 Tcy – 5	_	—	ns
162	TadV2oeH	Least Significant Data Valid Before $\overline{OE}$ $\uparrow$ (data setup time)	20	—	—	ns
162A	TadV2oeH	Most Significant Data Valid Before OE ↑ (data setup time)	0.25 Tcy + 20	—	_	ns
163	ToeH2adI	OE ↑ to Data in Invalid (data hold time)	0		_	ns
166	TalH2alH	ALE $\uparrow$ to ALE $\uparrow$ (cycle time)	—	TCY	—	ns
167	TACC	Address Valid to Data Valid	0.5 Tcy – 10		—	ns
168	Тое	$\overline{OE} \downarrow$ to Data Valid	—	_	0.125 Tcy + 5	ns
170	TubH2oeH	BA0 = 0 Valid Before OE ↑	0.25 TCY	_	_	ns
170A	TubL2oeH	BA0 = 1 Valid Before $\overline{OE}$ $\uparrow$	0.5 TCY	_	_	ns



Param. No	Symbol	Characteristics	Min	Тур	Max	Units
150	TadV2alL	Address Out Valid to ALE ↓ (address setup time)	0.25 Tcy – 10	_		ns
151	TalL2adl	ALE $\downarrow$ to Address Out Invalid (address hold time)	5	_	—	ns
155	TalL2oeL	ALE $\downarrow$ to $\overline{OE} \downarrow$	10	0.125 Tcy	—	ns
160	TadZ2oeL	AD High-Z to $\overline{OE} \downarrow$ (bus release to $\overline{OE}$ )	0		_	ns
161	ToeH2adD	OE ↑ to AD Driven	0.125 Tcy – 5	_	—	ns
162	TadV2oeH	LS Data Valid before $\overline{OE}$ $\uparrow$ (data setup time)	20	_	—	ns
163	ToeH2adl	$\overline{OE}$ $\uparrow$ to Data In Invalid (data hold time)	0	_	_	ns
164	TalH2alL	ALE Pulse Width	—	0.25 TCY	—	ns
165	ToeL2oeH	OE Pulse Width	0.5 Tcy – 5	0.5 TCY	_	ns
166	TalH2alH	ALE $\uparrow$ to ALE $\uparrow$ (cycle time)	—	Тсү	—	ns
167	Tacc	Address Valid to Data Valid	0.75 Tcy – 25		—	ns
168	Тое	$\overline{OE}\downarrow$ to Data Valid			0.5 TCY – 25	ns
169	TalL2oeH	ALE ↓ to OE ↑	0.625 Tcy – 10		0.625 Tcy + 10	ns
171	TalH2csL	Chip Enable Active to ALE $\downarrow$	0.25 Tcy - 20	_	—	ns
171A	TubL2oeH	AD Valid to Chip Enable Active		_	10	ns

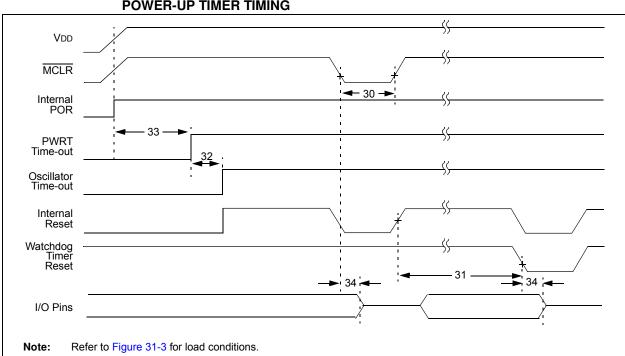
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### FIGURE 31-8: PROGRAM MEMORY WRITE TIMING DIAGRAM

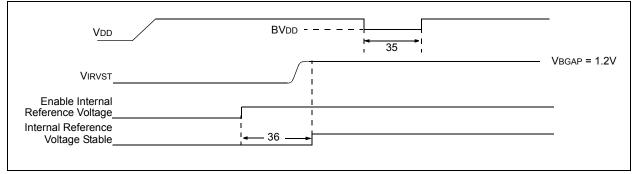
TABLE 31-12: PROGRAM MEMORY WRITE TIMING REQUIREMENT
------------------------------------------------------

Param. No	Symbol	Characteristics	Min	Тур	Max	Units
150	TadV2alL	Address Out Valid to ALE $\downarrow$ (address setup time)	0.25 Tcy – 10	—	_	ns
151	TalL2adl	ALE $\downarrow$ to Address Out Invalid (address hold time)	5	—		ns
153	TwrH2adl	$\overline{WRn}$ $\uparrow$ to Data Out Invalid (data hold time)	5	—	_	ns
154	TwrL	WRn Pulse Width	0.5 Tcy – 5	0.5 TCY	_	ns
156	TadV2wrH	Data Valid before $\overline{WRn}$ $\uparrow$ (data setup time)	0.5 Tcy – 10	—	_	ns
157	TbsV2wrL	Byte Select Valid before WRn ↓ (byte select setup time)	0.25 TCY	—	_	ns
157A	TwrH2bsl	WRn ↑ to Byte Select Invalid (byte select hold time)	0.125 Tcy – 5	_	_	ns
166	TalH2alH	ALE $\uparrow$ to ALE $\uparrow$ (cycle time)	—	Тсү	_	ns
171	TalH2csL	Chip Enable Active to ALE $\downarrow$	0.25 TCY – 20	—	_	ns
171A	TubL2oeH	AD Valid to Chip Enable Active		_	10	ns



## FIGURE 31-9: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

### FIGURE 31-10: BROWN-OUT RESET TIMING

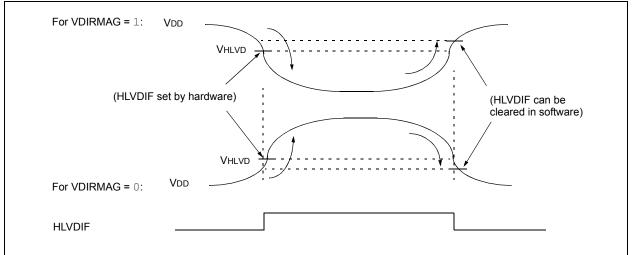


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## TABLE 31-13: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER AND BROWN-OUT RESET REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Тур	Мах	Units	Conditions
30	TmcL	MCLR Pulse Width (low)	2		_	μS	
31	Twdt	Watchdog Timer Time-out Period (no postscaler)	—	4.00	_	ms	
32	Tost	Oscillation Start-up Timer Period	1024 Tosc		1024 Tosc	_	Tosc = OSC1 period
33	TPWRT	Power-up Timer Period	_	65.5	140	ms	
34	Tioz	I/O High-Impedance from MCLR Low or Watchdog Timer Reset	—	2	—	μS	
35	TBOR	Brown-out Reset Pulse Width	200		_	μS	$VDD \le BVDD$ (see D005)
36	TIRVST	Time for Internal Reference Voltage to become Stable	—	25	—	μS	
37	Thlvd	High/Low-Voltage Detect Pulse Width	200		_	μS	$V D D \leq V H L V D$
38	TCSD	CPU Start-up Time	5	_	10	μS	
39	TIOBST	Time for INTOSC to Stabilize	_	1	_	μS	





### TABLE 31-14: HIGH/LOW-VOLTAGE DETECT CHARACTERISTICS

	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial $-40^{\circ}C \le TA \le +125^{\circ}C$ for extended									
Param No.	Sym	Characteristic		Min	Тур	Max	Units	Conditions		
D420		HLVD Voltage on VDD		1.80	1.86	1.90	V			
		Transition High-to-Low	HLVDL<3:0> = 0001	2.03	2.12	2.13	V			
			HLVDL<3:0> = 0010	2.24	2.33	2.35	V			
			HLVDL<3:0> = 0011	2.40	2.49	2.53	V			
			HLVDL<3:0> = 0100	2.50	2.59	2.62	V			
			HLVDL<3:0> = 0101	2.70	2.75	2.84	V			
			HLVDL<3:0> = 0110	2.82	2.93	2.97	V			
			HLVDL<3:0> = 0111	2.95	3.07	3.10	V			
			HLVDL<3:0> = 1000	3.24	3.30	3.41	V			
			HLVDL<3:0> = 1001	3.42	3.48	3.59	V			
			HLVDL<3:0> = 1010	3.61	3.67	3.79	V			
			HLVDL<3:0> = 1011	3.82	3.87	4.01	V			
			HLVDL<3:0> = 1100	4.06	4.21	4.26	V			
			HLVDL<3:0> = 1101	4.33	4.42	4.55	V			
			HLVDL<3:0> = 1110	4.64	4.77	4.87	V			

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### FIGURE 31-12: TIMER0 AND TIMER1 EXTERNAL CLOCK TIMINGS

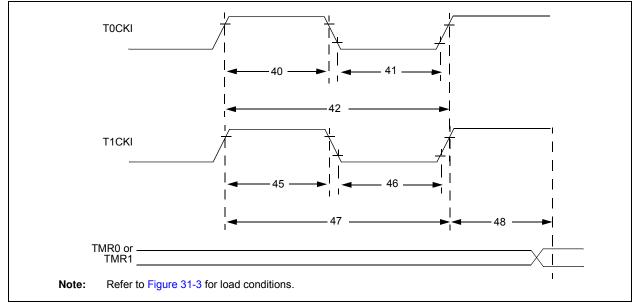


TABLE 31-15:	TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS
--------------	-----------------------------------------------

Param No.	Symbol		Characteristic		Min	Max	Units	Conditions
40	T⊤0H	T0CKI High F	T0CKI High Pulse Width No prescaler 0.5 Tcy + 20 -		—	ns		
				With prescaler	10	_	ns	
41	T⊤0L	T0CKI Low P	ulse Width	No prescaler	0.5 Tcy + 20	_	ns	
				With prescaler	10	_	ns	
42	T⊤0P	T0CKI Period	Period No pre		Tcy + 10	_	ns	
				With prescaler	Greater of: 20 ns or (TcY + 40)/N	_	ns	N = prescale value (1, 2, 4,, 256)
45	T⊤1H	1H T1CKI High Time	Synchronous, no prescaler		0.5 Tcy + 20	_	ns	
			Synchronous, with prescaler		10	—	ns	
			Asynchronous		30	_	ns	
46	T⊤1L	T1CKI Low	Synchronous, n	o prescaler	0.5 Tcy + 5	_	ns	
		Time	Synchronous, with prescaler		10	—	ns	
			Asynchronous		30	_	ns	
47	TT1P	T1CKI Input Period	Synchronous		Greater of: 20 ns or (Tcy + 40)/N	_	ns	N = prescale value (1, 2, 4, 8)
			Asynchronous		60		ns	
	F⊤1	T1CKI Oscilla	tor Input Frequency Range		DC	50	kHz	
48	TCKE2TMRI	Delay from Ex Timer Increm	kternal T1CKI Clo ent	ock Edge to	2 Tosc	7 Tosc	_	

### FIGURE 31-13: CAPTURE/COMPARE/PWM TIMINGS (ECCP1, ECCP2 MODULES)

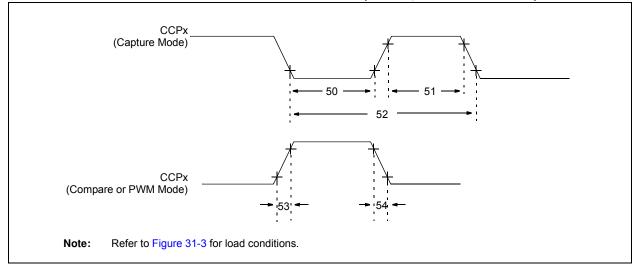
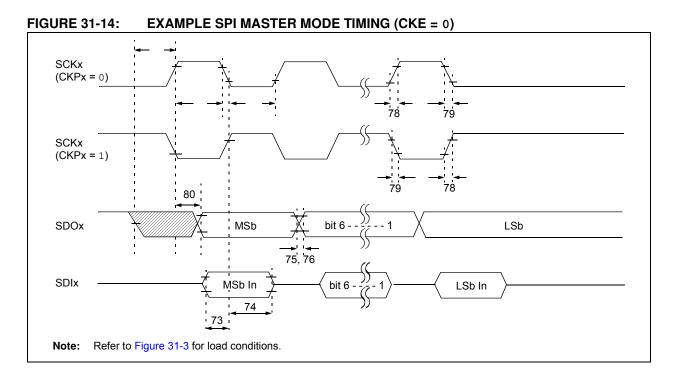


TABLE 31-16: CAPTURE/COMPARE/PWM REQUIREMENTS (ECCP1, ECCP2 MODULES)

Param No.	Symbol	с	haracteristic	Min	Max	Units	Conditions
50	TccL	CCPx Input Low	No prescaler	0.5 Tcy + 20		ns	
		Time	With prescaler	10	_	ns	
51	ТссН	CCPx Input	No prescaler	0.5 Tcy + 20	_	ns	
		High Time	With prescaler	10	-	ns	
52	TCCP	CCPx Input Perio	CCPx Input Period		_	ns	N = prescale value (1, 4 or 16)
53	TccR	CCPx Output Fal	ll Time	—	25	ns	
54	TccF	CCPx Output Fal	II Time	—	25	ns	

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Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
73	TDIV2scH, TDIV2scL	Setup Time of SDIx Data Input to SCKx Edge	20	—	ns	
73A	Тв2в	Last Clock Edge of Byte 1 to the 1st Clock Edge of Byte 2	1.5 Tcy + 40	_	ns	
74	TscH2DIL, TscL2DIL	Hold Time of SDIx Data Input to SCKx Edge	40		ns	
75	TDOR	SDOx Data Output Rise Time	—	25	ns	
76	TDOF	SDOx Data Output Fall Time	—	25	ns	
78	TscR	SCKx Output Rise Time (Master mode)	—	25	ns	
79	TscF	SCKx Output Fall Time (Master mode)	—	25	ns	
80	TscH2doV, TscL2doV	SDOx Data Output Valid after SCKx Edge	—	50	ns	

25

25

25

25

50

___

____

TCY

ns

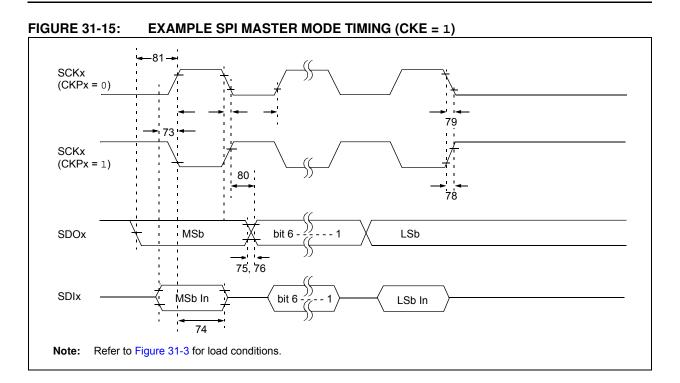
ns

ns

ns

ns

ns



#### Param. Symbol Min Units Conditions Characteristic Max No. 73 Setup Time of SDIx Data Input to SCKx Edge TDIV2scH, 20 ns TDIV2scL 73A Тв2в Last Clock Edge of Byte 1 to the 1st Clock Edge 1.5 TCY + 40 ns _ of Byte 2 74 TscH2DIL, Hold Time of SDIx Data Input to SCKx Edge 40 ns TscL2DIL

#### TABLE 31-18: EXAMPLE SPI MODE REQUIREMENTS (MASTER MODE, CKE = 1)

SDOx Data Output Rise Time

SDOx Data Output Fall Time

SCKx Output Rise Time (Master mode)

SCKx Output Fall Time (Master mode)

SDOx Data Output Valid after SCKx Edge

SDOx Data Output Setup to SCKx Edge

75

76

78

79

80

81

TDOR

TDOF

TscR

TscF

TSCH2DOV,

TscL2doV

TDOV2scH,

TDOV2scL

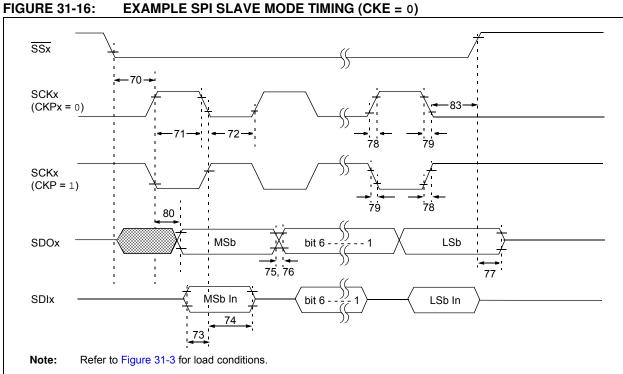
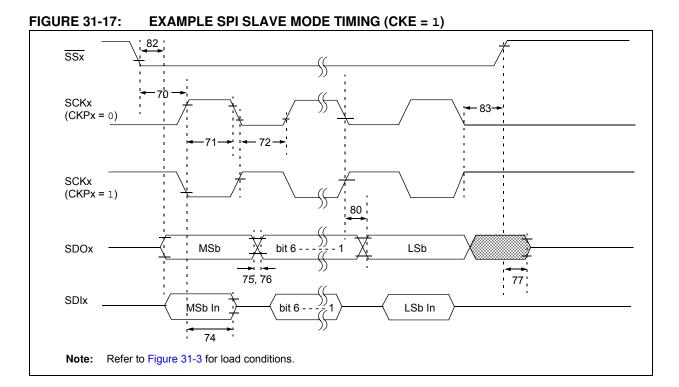


TABLE 31-	19: EXAMPLE SPI MODE REQUIREMENTS (SLAVE MODE TIMING, CKE = 0
Note:	Refer to Figure 31-3 for load conditions.

Param No.	Symbol	Characteristic	Characteristic		Max	Units	Conditions
70	TssL2scH, TssL2scL	$\overline{\operatorname{SSx}}\downarrow$ to SCKx $\downarrow$ or SCKx $\uparrow$ Input	↓ to SCKx ↓ or SCKx ↑ Input		—	ns	
70A	TssL2WB	SSx to Write to SSPBUF		3 Tcy	_	ns	
71	TscH	SCKx Input High Time	Continuous	1.25 Tcy + 30	—	ns	
71A		(Slave mode)	Single Byte	40	—	ns	(Note 1)
72	TscL	SCKx Input Low Time	Continuous	1.25 Tcy + 30	_	ns	
72A		(Slave mode)	Single Byte	40		ns	(Note 1)
73	TDIV2scH, TDIV2scL	Setup Time of SDIx Data Input to SCKx	20	—	ns		
73A	Тв2в	Last Clock Edge of Byte 1 to the First Clo	ck Edge of Byte 2	1.5 Tcy + 40	—	ns	(Note 2)
74	TscH2diL, TscL2diL	Hold Time of SDIx Data Input to SCKx	Hold Time of SDIx Data Input to SCKx Edge			ns	
75	TDOR	SDOx Data Output Rise Time			25	ns	
76	TDOF	SDOx Data Output Fall Time			25	ns	
77	TssH2doZ	SSx ↑ to SDOx Output High-Impedance	e	10	50	ns	
78	TscR	SCKx Output Rise Time (Master mode)			25	ns	
79	TscF	SCKx Output Fall Time (Master mode)			25	ns	
80	TscH2doV, TscL2doV	SDOx Data Output Valid after SCKx Edge		—	50	ns	
83	TscH2ssH, TscL2ssH	SSx	1.5 Tcy + 40	_	ns		

Note 1: Requires the use of Parameter #73A.

2: Only if Parameter #71A and #72A are used.



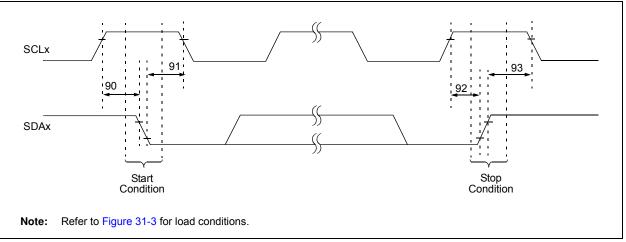
Param No.	Symbol	Characteristic	Min	Max	Units	Conditions	
70	TssL2scH, TssL2scL	$\overline{\text{SSx}} \downarrow$ to SCKx $\downarrow$ or SCKx $\uparrow$ Input		3 Тсү		ns	
70A	TssL2WB	SSx to Write to SSPBUF		3 TCY	_	ns	
71	TscH	SCKx Input High Time	Continuous	1.25 Tcy + 30	—	ns	
71A		(Slave mode)	Single Byte	40	_	ns	(Note 1)
72	TscL	SCKx Input Low Time	Continuous	1.25 Tcy + 30	—	ns	
72A		(Slave mode)	Single Byte	40		ns	(Note 1)
73A	Тв2в	Last Clock Edge of Byte 1 to the First	Clock Edge of Byte 2	1.5 Tcy + 40		ns	(Note 2)
74	TscH2diL, TscL2diL	Hold Time of SDIx Data Input to SC	40	—	ns		
75	TDOR	SDOx Data Output Rise Time			25	ns	
76	TDOF	SDOx Data Output Fall Time		—	25	ns	
77	TssH2doZ	SSx ↑ to SDOx Output High-Impeda	ance	10	50	ns	
78	TscR	SCKx Output Rise Time (Master mo	de)		25	ns	
79	TscF	SCKx Output Fall Time (Master mod	le)		25	ns	
80	TscH2doV, TscL2doV	SDOx Data Output Valid after SCKx	—	50	ns		
82	TssL2DoV	SDOx Data Output Valid after SSx ↓	_	50	ns		
83	TscH2ssH, TscL2ssH	SSx ↑ after SCKx Edge	1.5 Tcy + 40	_	ns		

Note 1: Requires the use of Parameter #73A.

**2:** Only if Parameter #71A and #72A are used.

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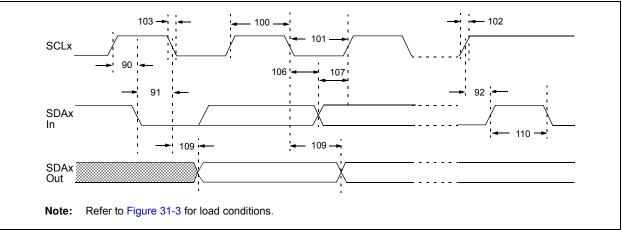
### FIGURE 31-18: I²C[™] BUS START/STOP BITS TIMING



## TABLE 31-21: I²C[™] BUS START/STOP BITS REQUIREMENTS (SLAVE MODE)

Param. No.	Symbol	Characte	ristic	Min	Max	Units	Conditions
90	TSU:STA	Start Condition	100 kHz mode	4700	_	ns	Only relevant for Repeated
		Setup Time	400 kHz mode	600	_		Start condition
91	THD:STA	Start Condition	100 kHz mode	4000	_	ns	After this period, the first
		Hold Time	400 kHz mode	600	_		clock pulse is generated
92	Tsu:sto	Stop Condition	100 kHz mode	4700	—	ns	
		Setup Time	400 kHz mode	600	_		
93	THD:STO	Stop Condition	100 kHz mode	4000	_	ns	
		Hold Time	400 kHz mode	600	_		





Param. No.	Symbol	Characteris	tic	Min	Max	Units	Conditions
100	Тнідн	Clock High Time	100 kHz mode	4.0		μS	
			400 kHz mode	0.6	—	μS	
			MSSP module	1.5 TCY	—		
101	TLOW	Clock Low Time	100 kHz mode	4.7	—	μS	
			400 kHz mode	1.3	—	μS	
			MSSP module	1.5 TCY	—		
102	TR	SDAx and SCLx Rise Time	100 kHz mode	—	1000	ns	
			400 kHz mode	20 + 0.1 CB	300	ns	CB is specified to be from 10 to 400 pF
103	TF	SDAx and SCLx Fall Time	100 kHz mode	_	300	ns	
			400 kHz mode	20 + 0.1 CB	300	ns	CB is specified to be from 10 to 400 pF
90	TSU:STA	Start Condition Setup Time	100 kHz mode	4.7	—	μS	Only relevant for Repeated
			400 kHz mode	0.6	—	μS	Start condition
91	THD:STA	Start Condition Hold Time	100 kHz mode	4.0	—	μS	After this period, the first clock
			400 kHz mode	0.6	—	μS	pulse is generated
106	THD:DAT	Data Input Hold Time	100 kHz mode	0	—	ns	
			400 kHz mode	0	0.9	μS	
107	TSU:DAT	Data Input Setup Time	100 kHz mode	250	—	ns	(Note 2)
			400 kHz mode	100	—	ns	
92	Tsu:sto	Stop Condition Setup Time	100 kHz mode	4.7	—	μS	
			400 kHz mode	0.6	—	μS	
109	ΤΑΑ	Output Valid from Clock	100 kHz mode	—	3500	ns	(Note 1)
			400 kHz mode		—	ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7	_	μS	Time the bus must be free before
			400 kHz mode	1.3	-	μS	a new transmission can start
D102	Св	Bus Capacitive Loading			400	pF	

### TABLE 31-22: I²C[™] BUS DATA REQUIREMENTS (SLAVE MODE)

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCLx to avoid unintended generation of Start or Stop conditions.

2: A Fast mode I²C[™] bus device can be used in a Standard mode I²C bus system, but the requirement, TSU:DAT ≥ 250 ns, must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCLx signal. If such a device does stretch the LOW period of the SCLx signal, it must output the next data bit to the SDAx line, TR max. + TSU:DAT = 1000 + 250 = 1250 ns (according to the Standard mode I²C bus specification), before the SCLx line is released.

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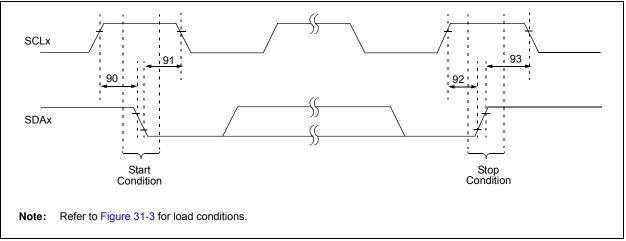
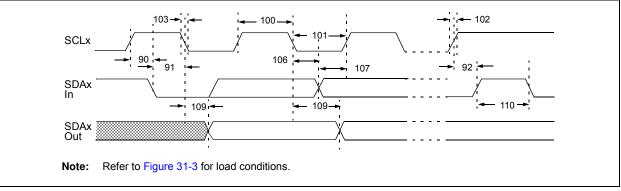


TABLE 31-23:	MSSP I ² C [™] BUS START/STOP BITS REQUIREMENTS
--------------	---------------------------------------------------------------------

Param. No.	Symbol	Characte	eristic	Min	Max	Units	Conditions
90	TSU:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	Only relevant for
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_		Repeated Start
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_		condition
91	THD:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	After this period, the
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_		first clock pulse is
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_		generated
92	Tsu:sto	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_		
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)			
93	THD:STO	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	ns	
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_	]	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	]	

**Note 1:** Maximum pin capacitance = 10 pF for all  $I^2C^{TM}$  pins.





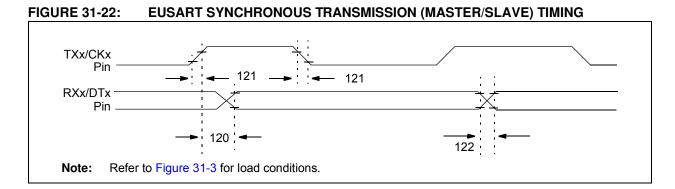
Param. No.	Symbol	Charac	teristic	Min	Max	Units	Conditions
100	Тнідн	Clock High	100 kHz mode	2(Tosc)(BRG + 1)		_	
		Time	400 kHz mode	2(Tosc)(BRG + 1)		_	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)		_	
101	TLOW	Clock Low Time	100 kHz mode	2(Tosc)(BRG + 1)	—		
			400 kHz mode	2(Tosc)(BRG + 1)		_	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	_	
102	TR	SDAx and	100 kHz mode	—	1000	ns	CB is specified to be from
		SCLx Rise	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF
		Time	1 MHz mode ⁽¹⁾	_	300	ns	
103	TF	SDAx and	100 kHz mode	_	300	ns	CB is specified to be from
		SCLx Fall Time	400 kHz mode	20 + 0.1 Св	300	ns	10 to 400 pF
			1 MHz mode ⁽¹⁾	_	100	ns	
90	TSU:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	—	Only relevant for Repeated
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_	—	Start condition
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	—	
91	THD:STA	Start Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	—	After this period, the first
		Hold Time	400 kHz mode	2(Tosc)(BRG + 1)	_	—	clock pulse is generated
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	—	
106	THD:DAT	Data Input	100 kHz mode	0	_	ns	
		Hold Time	400 kHz mode	0	0.9	μS	
			1 MHz mode ⁽¹⁾	_	_	ns	
107	TSU:DAT	Data Input	100 kHz mode	250	_	ns	(Note 2)
		Setup Time	400 kHz mode	100	_	ns	
			1 MHz mode ⁽¹⁾	_	_	ns	
92	Tsu:sto	Stop Condition	100 kHz mode	2(Tosc)(BRG + 1)	_	—	
		Setup Time	400 kHz mode	2(Tosc)(BRG + 1)	_	—	
			1 MHz mode ⁽¹⁾	2(Tosc)(BRG + 1)	_	—	
109	ΤΑΑ	Output Valid	100 kHz mode	_	3500	ns	
		from Clock	400 kHz mode		1000	ns	
			1 MHz mode ⁽¹⁾		_	ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7	—	μS	Time the bus must be free
			400 kHz mode	1.3	—	μS	before a new transmission
			1 MHz mode ⁽¹⁾	—	—	μS	can start
D102	Св	Bus Capacitive L	oading	_	400	pF	

TABLE 31-24:	MSSP I ² C [™] BUS DATA REQUIREMENTS
--------------	----------------------------------------------------------

**Note 1:** Maximum pin capacitance = 10 pF for all  $I^2C^{TM}$  pins.

2: A Fast mode I²C bus device can be used in a Standard mode I²C bus system, but Parameter #107 ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCLx signal. If such a device does stretch the LOW period of the SCLx signal, it must output the next data bit to the SDAx line, Parameter #102 + Parameter #107 = 1000 + 250 = 1250 ns (for 100 kHz mode), before the SCLx line is released.

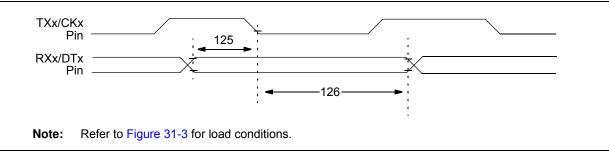
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### TABLE 31-25: EUSART/AUSART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
120	TCKH2DTV	SYNC XMIT (MASTER and SLAVE) Clock High to Data Out Valid	_	40	ns	
121	TCKRF	Clock Out Rise Time and Fall Time (Master mode)	—	20	ns	
122	TDTRF	Data Out Rise Time and Fall Time	_	20	ns	

### FIGURE 31-23: EUSART/AUSART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING



### TABLE 31-26: EUSART/AUSART SYNCHRONOUS RECEIVE REQUIREMENTS

Param. No.	Symbol	Characteristic	Min	Max	Units	Conditions
125	TDTV2CKL	<u>SYNC RCV (MASTER and SLAVE)</u> Data Hold before CKx $\downarrow$ (DTx hold time)	10		ns	
126	TCKL2DTL	Data Hold after CKx $\downarrow$ (DTx hold time)	15		ns	

Param No.	Sym	Characteristic	Min	Тур	Max	Units	Conditions
A01	NR	Resolution	—	_	12	bit	$\Delta VREF \ge 5.0V$
A03	EIL	Integral Linearity Error	—	±1	±6.0	LSB	ΔVREF = 5.0V
A04	Edl	Differential Linearity Error	_	±1	+3.0/-1.0	LSB	ΔVREF = 5.0V
A06	EOFF	Offset Error	—	±1	±9.0	LSB	ΔVREF = 5.0V
A07	Egn	Gain Error	_	±1	±8.0	LSB	ΔVREF = 5.0V
A10	—	Monotonicity ⁽¹⁾	_	_		_	$VSS \le VAIN \le VREF$
A20	$\Delta VREF$	Reference Voltage Range (VREFH – VREFL)	3	—	Vdd - Vss	V	
A21	Vrefh	Reference Voltage High	Vss + 3.0V	_	VDD + 0.3V	V	
A22	Vrefl	Reference Voltage Low	Vss – 0.3V	_	VDD - 3.0V	V	
A25	VAIN	Analog Input Voltage	VREFL	_	VREFH	V	
A30	Zain	Recommended Impedance of Analog Voltage Source	_	_	2.5	kΩ	
A50	IREF	VREF Input Current ⁽²⁾	—	—	5 150	μΑ μΑ	During VAIN acquisition During A/D conversion cycle

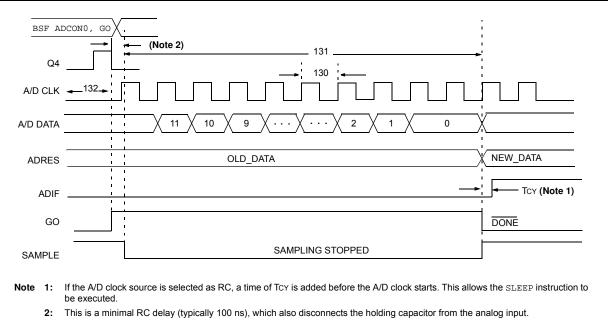
### TABLE 31-27: A/D CONVERTER CHARACTERISTICS: PIC18F87K22 FAMILY (INDUSTRIAL)

Note 1: The A/D conversion result doesn't decrease with an increase in the input voltage.

2: VREFH current is from the RA3/AN3/VREF+ pin or VDD, whichever is selected as the VREFH source. VREFL current is from the RA2/AN2/VREF- pin or VSs, whichever is selected as the VREFL source.

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Param No.	Symbol	Characteristic	Min	Max	Units	Conditions
130	Tad	A/D Clock Period	0.8	12.5 ⁽¹⁾	μs	Tosc-based, VREF $\geq$ 3.0V
			1.4	25 ⁽¹⁾	μs	VDD = 3.0V; Tosc-based, VREF full range
			_	1	μs	A/D RC mode
			_	3	μs	VDD = 3.0V; A/D RC mode
131	TCNV	Conversion Time (not including acquisition time) ⁽²⁾	14	15	Tad	
132	TACQ	Acquisition Time ⁽³⁾	1.4		μs	-40°C to +125°C
135	Tswc	Switching Time from Convert $\rightarrow$ Sample		(Note 4)		
137	TDIS	Discharge Time	0.2	—	μs	-40°C to +125°C

### TABLE 31-28: A/D CONVERSION REQUIREMENTS

Note 1: The time of the A/D clock period is dependent on the device frequency and the TAD clock divider.

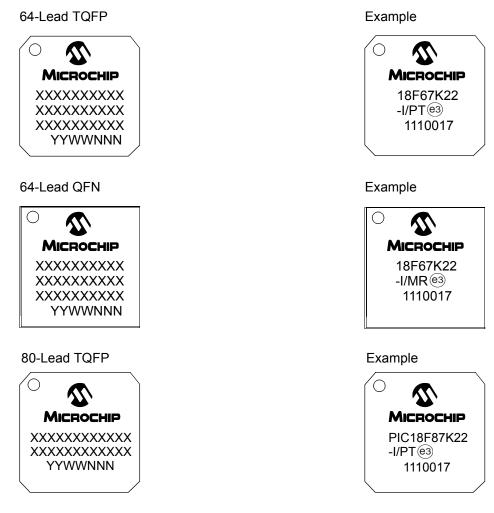
2: ADRES registers may be read on the following TCY cycle.

**3:** The time for the holding capacitor to acquire the "New" input voltage when the voltage changes full scale after the conversion (VDD to Vss or Vss to VDD). The source impedance (Rs) on the input channels is 50 $\Omega$ .

4: On the following cycle of the device clock.

## 32.0 PACKAGING INFORMATION

### 32.1 Package Marking Information



Legend	: XXX Y YY WW NNN (©3) *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.
Note:	be carried	nt the full Microchip part number cannot be marked on one line, it will d over to the next line, thus limiting the number of available s for customer-specific information.

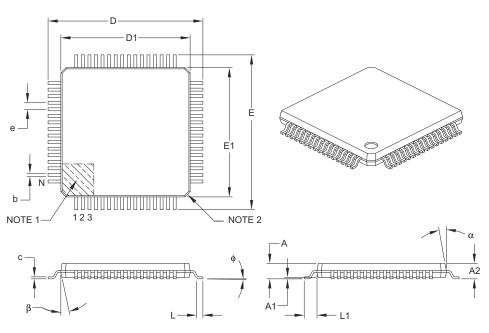
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### 32.2 Package Details

The following sections give the technical details of the packages.

### 64-Lead Plastic Thin Quad Flatpack (PT) – 10x10x1 mm Body, 2.00 mm [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS		
Di	mension Limits	MIN	NOM	MAX
Number of Leads	N		64	
Lead Pitch	е		0.50 BSC	
Overall Height	А	_	-	1.20
Molded Package Thickness	A2	0.95	1.00	1.05
Standoff	A1	0.05	-	0.15
Foot Length	L	0.45	0.60	0.75
Footprint	L1	1.00 REF		
Foot Angle	ф	0°	3.5°	7°
Overall Width	E		12.00 BSC	
Overall Length	D		12.00 BSC	
Molded Package Width	E1		10.00 BSC	
Molded Package Length	D1	10.00 BSC		
Lead Thickness	С	0.09	-	0.20
Lead Width	b	0.17	0.22	0.27
Mold Draft Angle Top	α	11°	12°	13°
Mold Draft Angle Bottom	β	11°	12°	13°

#### Notes:

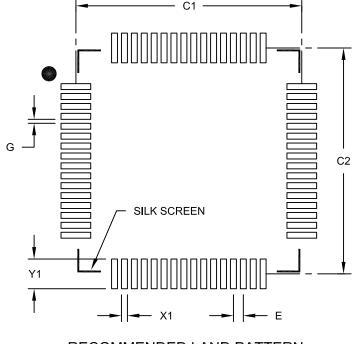
- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Chamfers at corners are optional; size may vary.
- 3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-085B

64-Lead Plastic Thin Quad Flatpack (PT) 10x10x1 mm Body, 2.00 mm Footprint [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	MILLIMETERS			
Dimensio	on Limits	MIN	NOM	MAX
Contact Pitch	E	0.50 BSC		
Contact Pad Spacing	C1		11.40	
Contact Pad Spacing	C2		11.40	
Contact Pad Width (X64)	X1			0.30
Contact Pad Length (X64)	Y1			1.50
Distance Between Pads	G	0.20		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

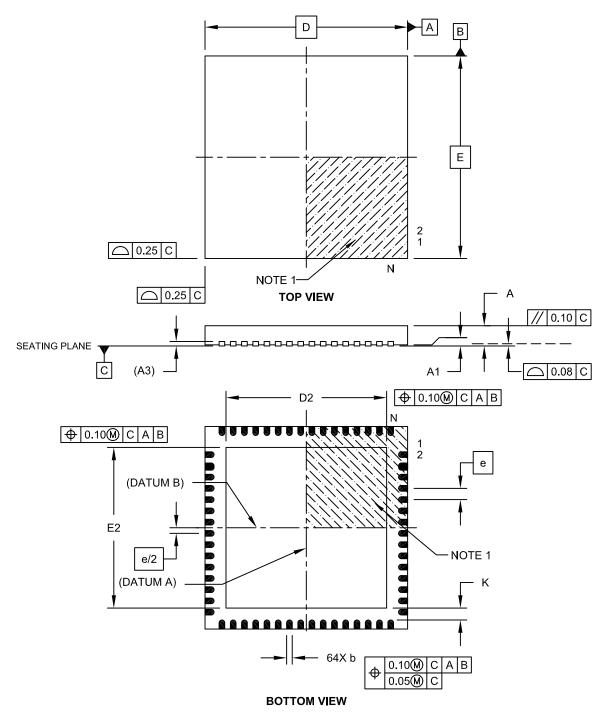
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

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### 64-Lead Plastic Quad Flat, No Lead Package (MR) – 9x9x0.9 mm Body [QFN]

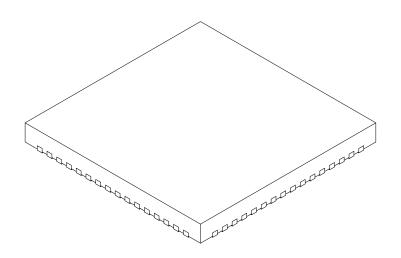
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-149B Sheet 1 of 2

### 64-Lead Plastic Quad Flat, No Lead Package (MR) – 9x9x0.9 mm Body [QFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS			
Dimensio	on Limits	MIN	NOM	MAX
Number of Pins	N		64	
Pitch	е		0.50 BSC	
Overall Height	Α	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3	0.20 REF		
Overall Width	E		9.00 BSC	
Exposed Pad Width	E2	7.05	7.15	7.50
Overall Length	D		9.00 BSC	
Exposed Pad Length	D2	7.05	7.15	7.50
Contact Width	b	0.18	0.25	0.30
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	-	-

#### Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Package is saw singulated.

3. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

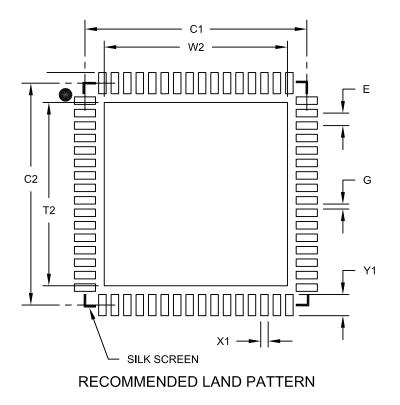
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-149B Sheet 2 of 2

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## 64-Lead Plastic Quad Flat, No Lead Package (MR) – 9x9x0.9 mm Body [QFN] With 0.40 mm Contact Length

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS				
Dimension	n Limits	MIN	NOM	MAX	
Contact Pitch	Е	0.50 BSC			
Optional Center Pad Width	W2			7.35	
Optional Center Pad Length	T2			7.35	
Contact Pad Spacing	C1		8.90		
Contact Pad Spacing	C2		8.90		
Contact Pad Width (X64)	X1			0.30	
Contact Pad Length (X64)	Y1			0.85	
Distance Between Pads	G	0.20			

#### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

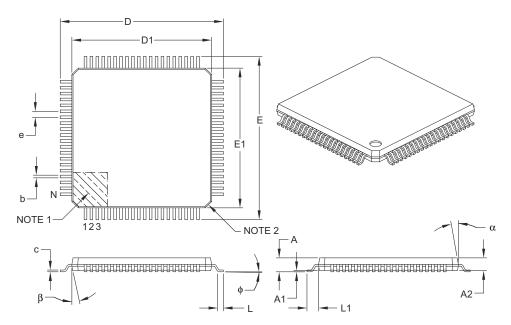
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2149A

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### 80-Lead Plastic Thin Quad Flatpack (PT) – 12x12x1 mm Body, 2.00 mm [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	nits MILLIMETERS		
	Dimension Limits	MIN	NOM	MAX
Number of Leads	N		80	
Lead Pitch	e		0.50 BSC	
Overall Height	A	-	-	1.20
Molded Package Thickness	A2	0.95	1.00	1.05
Standoff	A1	0.05	-	0.15
Foot Length	L	0.45	0.60	0.75
Footprint	L1	1.00 REF		
Foot Angle	ф	0°	3.5°	7°
Overall Width	E		14.00 BSC	
Overall Length	D		14.00 BSC	
Molded Package Width	E1		12.00 BSC	
Molded Package Length	D1		12.00 BSC	
Lead Thickness	С	0.09	-	0.20
Lead Width	b	0.17	0.22	0.27
Mold Draft Angle Top	α	11°	12°	13°
Mold Draft Angle Bottom	β	11°	12°	13°

#### Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

2. Chamfers at corners are optional; size may vary.

3. Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25 mm per side.

- 4. Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.

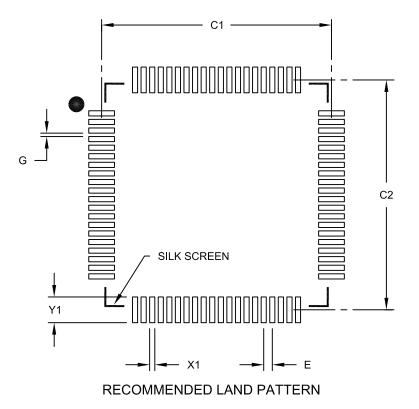
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-092B

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80-Lead Plastic Thin Quad Flatpack (PT)-12x12x1mm Body, 2.00 mm Footprint [TQFP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	N	<b>MILLIMETER</b>	S
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		0.50 BSC	
Contact Pad Spacing	C1		13.40	
Contact Pad Spacing	C2		13.40	
Contact Pad Width (X80)	X1			0.30
Contact Pad Length (X80)	Y1			1.50
Distance Between Pads	G	0.20		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2092B

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## APPENDIX A: REVISION HISTORY

### **Revision A (November 2009)**

Original data sheet for PIC18F87K22 family devices.

### Revision B (May 2010)

Minor edits to text throughout document. Replaced all TBDs with the correct value.

### **Revision C (March 2011)**

Section 2.4 "Voltage Regulator Pins (ENVREG and VCAP/VDDCORE)" has been replaced with a new and more detailed description and the 80-pin TQFP diagrams have been updated. Minor text edits throughout document.

### Revision D (June 2011)

Updated Section 31.2 "DC Characteristics: Power-Down and Supply Current PIC18F87K22 Family (Industrial/Extended)" and Table 31-8 with new electrical specification information. The extended temperature information has been included in this revision.

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### APPENDIX B: MIGRATION FROM PIC18F87J11 AND PIC18F8722 TO PIC18F87K22

Devices in the PIC18F87K22, PIC18F87J11 and PIC18F8722 families are similar in their functions and features. Code can be migrated from the other families to the PIC18F87K22 without many changes. The differences between the device families are listed in Table B-1.

## TABLE B-1:NOTABLE DIFFERENCES BETWEEN PIC18F87K22, PIC18F87J11 AND PIC18F8722<br/>FAMILIES

Characteristic	PIC18F87K22 Family	18F87J11 Family	PIC18F8722 Family
Max Operating Frequency	64 MHz	48 MHz	40 MHz
Max Program Memory	128 Kbytes	128 Kbytes	128 Kbytes
Data Memory	3,862 bytes	3,930 bytes	3,930 bytes
Program Memory Endurance	10,000 Write/Erase (minimum)	10,000 Write/Erase (minimum)	10,000 Write/Erase (minimum)
Single Word Write for Flash	No	Yes	No
Oscillator Options	PLL can be used with INTOSC	PLL can be used with INTOSC	PLL can be used with INTOSC
СТМИ	Yes	No	No
RTCC	Yes	No	No
SOSC Oscillator Options TICKI Clock	Low-Power Oscillator Option for SOSC T1CKI can be used as a Clock without Enabling the SOSC Oscillator	No	No
INTOSC	Up to 16 MHz	Up to 8 MHz	Up to 8 MHz
SPI/I ² C™	2	2	2
Timers	11	5	5
ECCP	3	3	
CCP	7	2	2
Data EEPROM	Yes	No	Yes
Programmable BOR	Multiple Level BOR	One Level BOR	Multiple Level BOR
WDT Prescale Options	22	16	16
5V Operation	Yes	No (3.3V)	Yes
nanoWatt XLP	Yes	No	No
Regulator	Yes	Yes	No
Low-Power BOR	Yes	No	No
A/D Converter	12-Bit Resolution, 24 Input Channels, Differential	10-Bit Resolution, 15 Input Channels, Non-Differential	10-Bit Resolution, 16 Input Channels, Non-Differential
Internal Temperature Sensor	Yes	No	No
Programmable HLVD	Yes	No	Yes
EUSART	2 EUSARTs	2 EUSARTs	2 EUSARTs
Comparators	3	2	2
Oscillator options	14 Options by Fosc<3:0>	8 Options by Fosc<3:0>	12 Options by Fosc<3:0>
Ultra-Low-Power Wake-up (ULPW)	Yes	No	No
Power-up Timer	Yes	Yes	Yes
MCLR Pin as Input Port	Yes	No	Yes

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BNZ BOR. See Brown-out Reset.	
BOV	
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PART NO.	X /XX XXX T Temperature Package Pattern Range	<ul> <li>Examples:</li> <li>a) PIC18F87K22-I/PT 301 = Industrial temperature, TQFP package, QTP pattern #301.</li> <li>b) PIC18F87K22T-I/PT = Tape and reel, Industrial temperature, TQFP package</li> </ul>
Device ^(1,2)	PIC18F65K22, PIC18F65K22T PIC18F66K22, PIC18F66K22T PIC18F67K22, PIC18F67K22T PIC18F85K22, PIC18F85K22T PIC18F86K22, PIC18F86K22T PIC18F87K22, PIC18F87K22T	<ul> <li>C) PIC18F87K22T-E/PT = Tape and reel, Extended temperature, TQFP package</li> </ul>
Temperature Range	I = -40°C to +85°C (Industrial) E = -40°C to +125°C (Extended)	
Package	PT = TQFP (Plastic Thin Quad Flatpack) MR = QFN (Plastic Quad Flat)	Note 1: F = Standard Voltage Range 2: T = In tape and reel PLCC and
Pattern	QTP, SQTP, Code or Special Requirements (blank otherwise)	TQFP packages only 3: RSL = Silicon Revision A3

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